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(54) **RECEIVE CIRCUIT FOR ULTRASOUND IMAGING**

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128/916; 73/625-626; 367/103, 123

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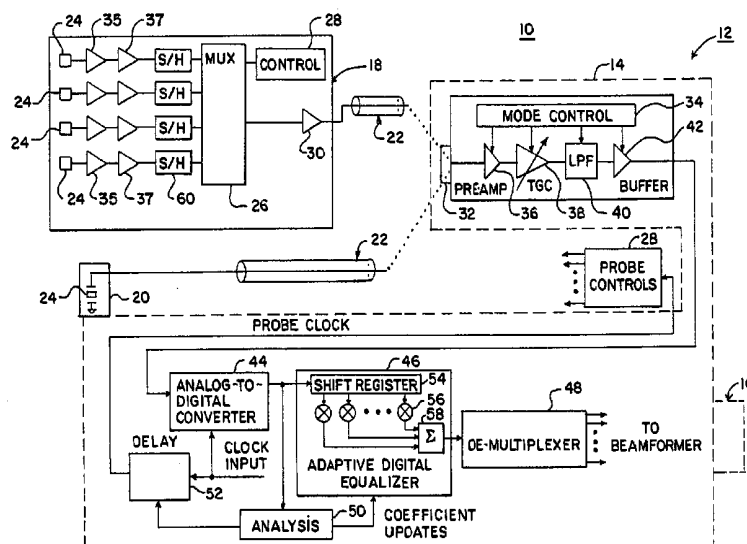
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*Primary Examiner*—Francis J. Jaworski

(57) **ABSTRACT**

Methods and systems for receiving different types of signal formats from different ultrasound transducers are provided. A base unit of an ultrasound system includes a connector and receiver circuit for connecting with one of multiple different types of transducers. For example, a conventional transducer providing analog information associated with a single element on one receive channel is connected with the connector and receiver circuit. Alternatively, a transducer outputting time division multiplex or other multiplex information representing multiple transducer elements is connected with the connector and receiver circuit. The receiver circuit processes the received information differently depending on the data format. For example, the preamplifier impedance or gain is different for single element signals versus time division multiplex signals. As another example, a low pass filter bandwidth is larger for time division multiplex signals than for signals representing a single element. Receive circuits with an analog-to-digital converter positioned prior to a digital demultiplexer are also included for use with multiplexed data formats. Time division multiplex or other multiplex signals from a transducer array are converted to a digital format prior to demultiplexing. Methods and systems for adapting multiplexing or processing of multiplexed signals are also provided. A test signal or ultrasound signals are transmitted from a transducer to the receiver circuit and analyzed. The multiplexer of the transducer is synchronized with an analog digital converter with selectable delays responsive to the analysis. Coefficients or other filter characteristics of a digital equalizer are selected as a function of the analysis.

**29 Claims, 4 Drawing Sheets**



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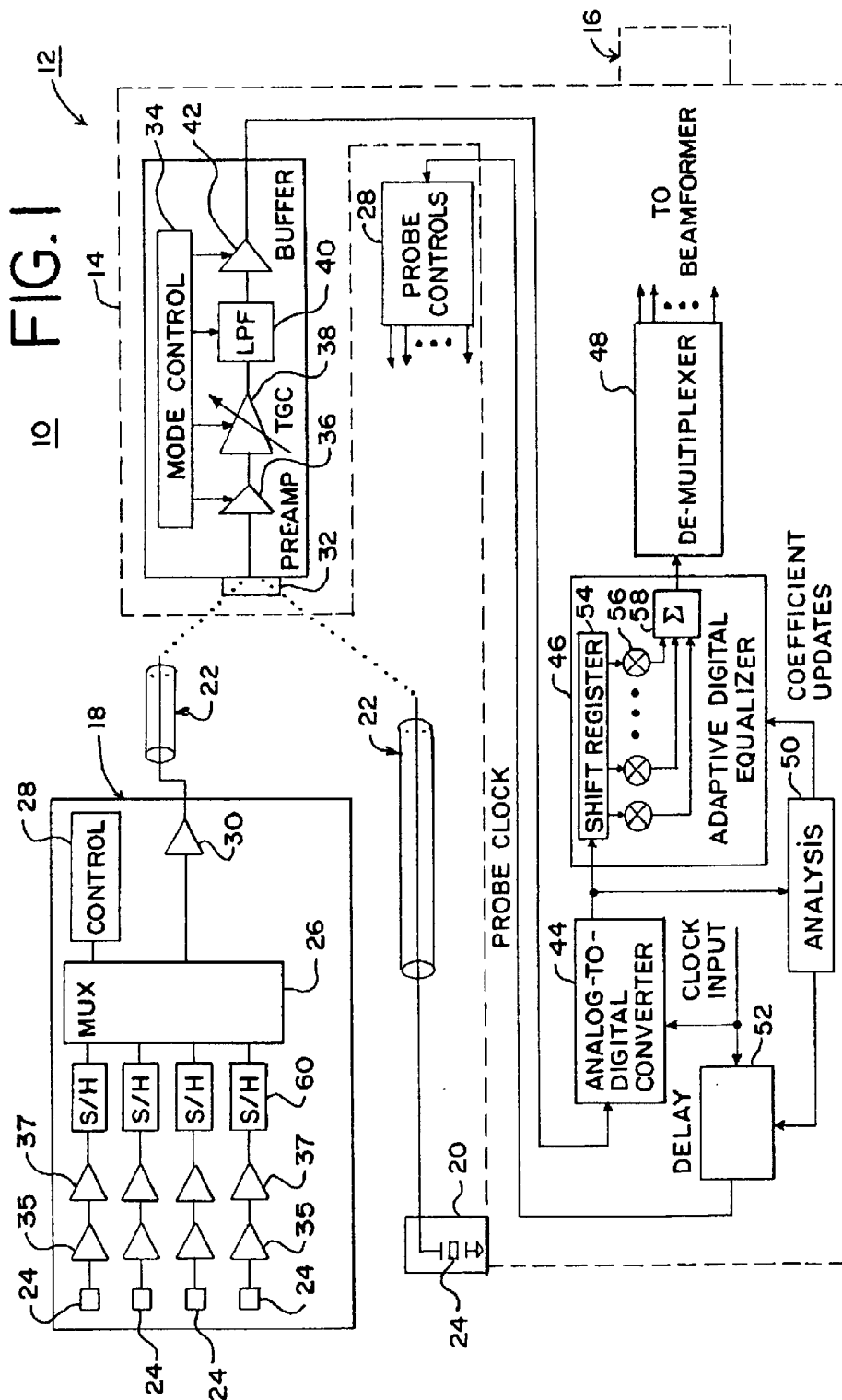


FIG. 2

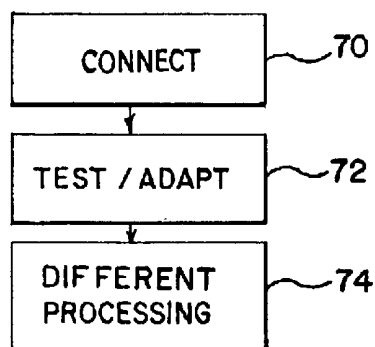


FIG. 3

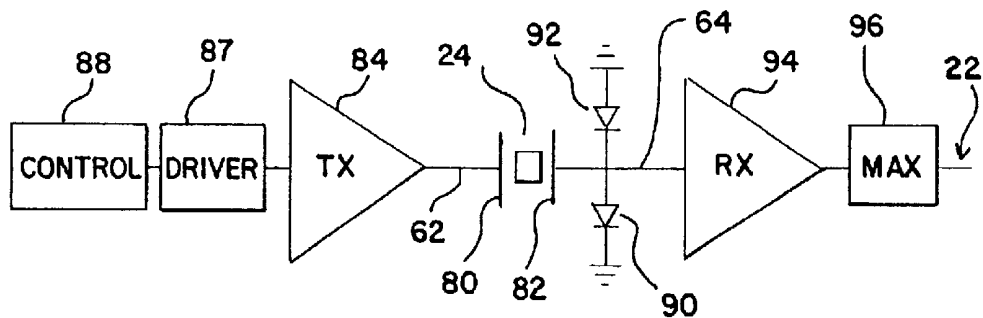


FIG. 4

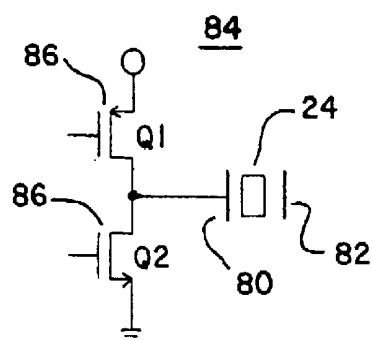


FIG. 5

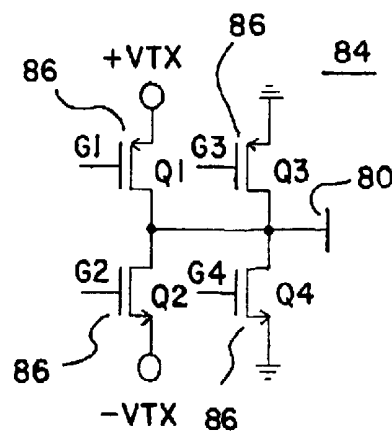


FIG. 6

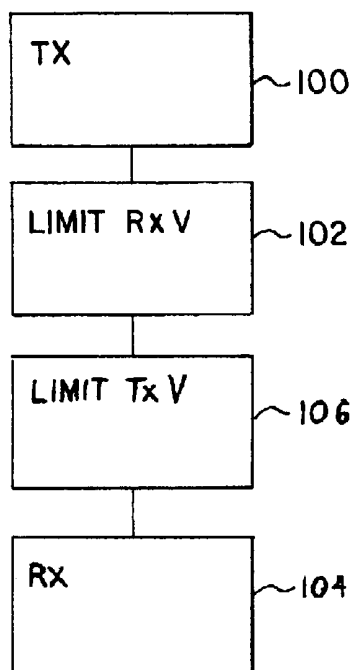


FIG. 7

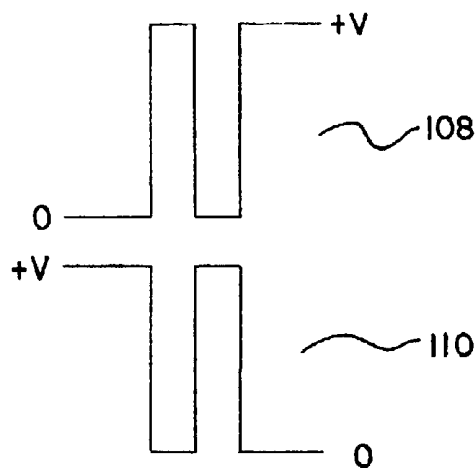
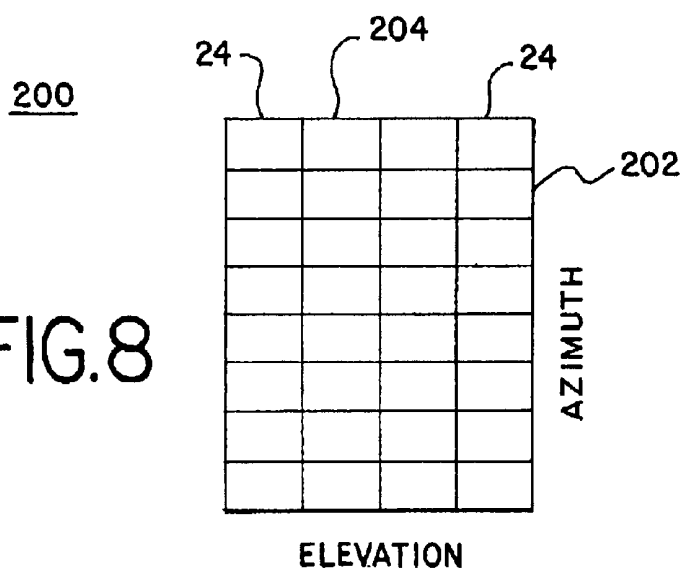
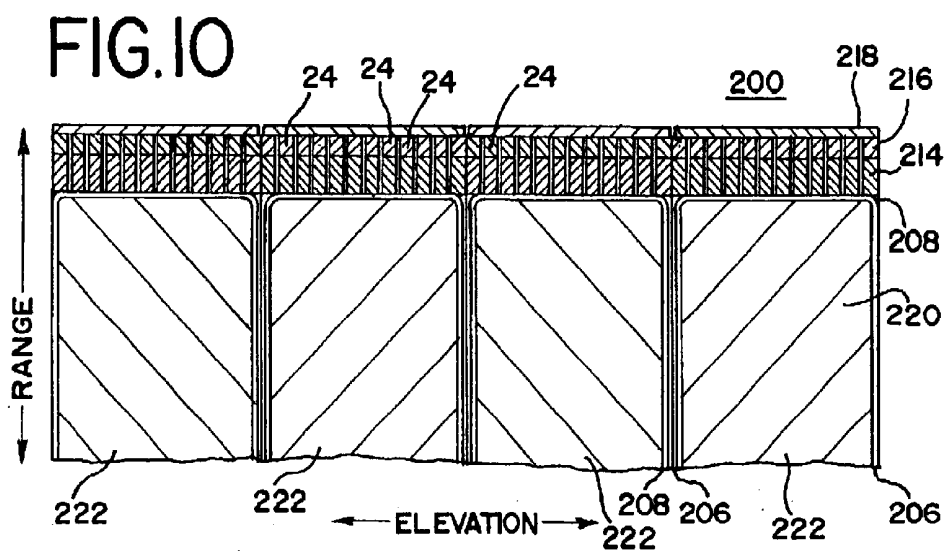
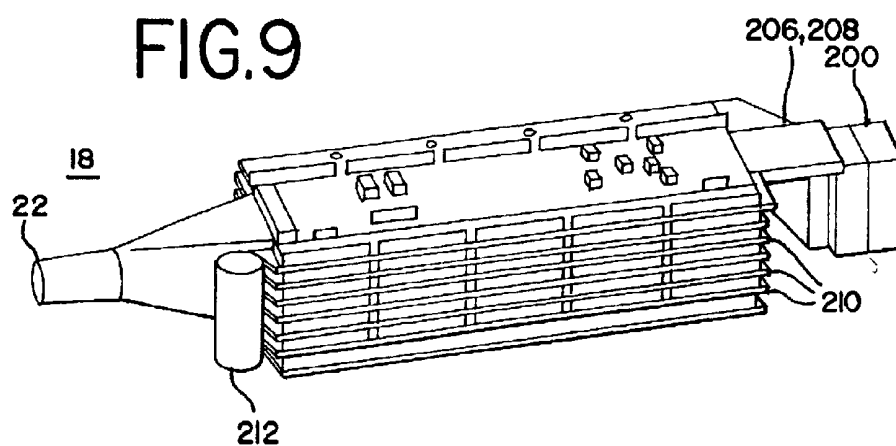
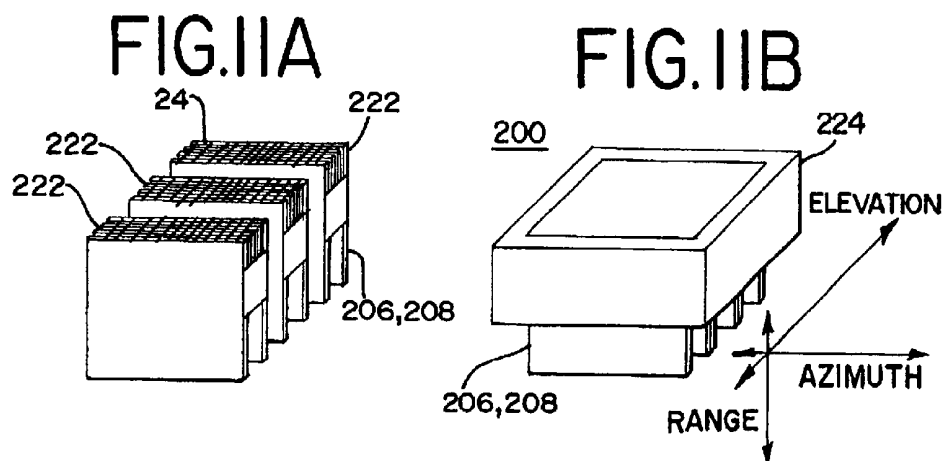


FIG. 8





## RECEIVE CIRCUIT FOR ULTRASOUND IMAGING

### BACKGROUND

The present invention relates to receive circuits for ultrasound imaging. In particular, receive circuits for use with different transducers are provided.

Ultrasound imaging for echocardiography applications requires transducers with high volume-per-second rates for scanning. For real-time imaging of moving structures, 20 or more, such as 35, two or three-dimensional representations are generated each second. Large amounts of information are communicated from an ultrasound probe to an ultrasound system base unit.

Various transducers and associated beamformers have been provided for three-dimensional ultrasound imaging. Currently, mostly mechanical transducers are used. However, the associated imaging is not provided in real time and typically requires ECG gating. Two-dimensional transducer arrays for faster electronic/electronic steering and volume acquisition also have been provided. For example, sparse two-dimensional arrays or fully sampled two-dimensional arrays have been used. Sparse arrays provide poor contrast resolution.

Fully sampled two-dimensional arrays use expensive additional beamforming hardware. Two-dimensional arrays repetitively generate transmit beams and responsive receive beams. The beams are electronically steered within the three-dimensional volume. Electronic steering requires a system channel for each of the elements used. Since the number of elements in a two-dimensional array is high, the number of channels required is high. More channels require a greater number of cables. Providing beamforming or partial beamforming within the probe of the transducer array may reduce the number of cables required, but the required number of channels and hardware for sampling the two-dimensional array is still high. Furthermore, analog delays used for beamforming in the probe are expensive and large, and the beamformer in the probe may have limited programmability.

Transducer arrays include elements with a ground electrode and a signal electrode switchably connected to separate transmit and receive system channels. With beamforming capabilities built into the probe, high voltage transistors or diodes operating as switches to isolate the transmit channels from the receive channels are also included within the probe. These high voltage devices are not easily integrated with the beamforming circuitry, so require additional space.

In one system disclosed in U.S. Pat. No. 5,622,177, the number of system channels and cables is reduced by using time division multiplexing. Data from a plurality of elements is multiplexed onto one signal line. However, time division multiplexed data has different characteristics than conventional data representing the signal from a single transducer element. Receive circuitry designed for use with conventional data may improperly introduce noise or errors in time division multiplexed data.

### BRIEF SUMMARY

By way of introduction, the preferred embodiments described below include methods and systems for receiving different types of signal formats from different ultrasound transducers. A base unit of an ultrasound system includes a

connector and receiver circuit for connecting with one of multiple different types of transducers. For example, a conventional transducer providing analog information associated with a single element on one receive channel is connected with the connector and receiver circuit. Alternatively, a transducer outputting time division multiplex or other multiplex information representing multiple transducer elements is connected with the connector and receiver circuit. The receiver circuit processes the received information differently depending on the data format. For example, the preamplifier impedance or gain is different for single element signals versus time division multiplex signals. As another example, a low pass filter bandwidth is larger for time division multiplex signals than for signals representing a single element.

The embodiments also include receive circuits with an analog-to-digital converter positioned prior to a digital demultiplexer. Time division multiplex or other multiplex signals from a transducer array are converted to a digital format prior to demultiplexing.

Other embodiments include methods and systems for adapting multiplexing or processing of multiplexed signals. A test signal or ultrasound signals are transmitted from a transducer to the receiver circuit and analyzed. The multiplexer of the transducer is synchronized with an analog digital converter with selectable delays responsive to the analysis. In another embodiment, coefficients or other filter characteristics of a digital equalizer are selected as a function of the analysis.

The present invention is defined by the following claims, and nothing in this section should be taken as a limitation on those claims. Further aspects and advantages of the invention are discussed below in conjunction with the preferred embodiments.

### BRIEF DESCRIPTION OF SEVERAL VIEWS OF THE DRAWINGS

The components and figures are not necessarily to scale, emphasis instead being placed upon illustrating the principles of the invention. Moreover, in the figures, like reference numerals designate corresponding parts throughout the different views.

FIG. 1 is a block diagram of one embodiment of an ultrasound system for receiving different types of signals from different transducer probes.

FIG. 2 is a flow chart diagram of one embodiment of a method for receiving data associated with a plurality of transducer elements on a single cable.

FIG. 3 is a block diagram of one embodiment of a transducer with isolated transmit and receive channels.

FIG. 4 is a circuit diagram of one embodiment of a transmitter.

FIG. 5 is a circuit diagram of an alternative embodiment of a transmitter.

FIG. 6 is a flow chart diagram of one embodiment representing use of the isolated transmit and receive channels of FIG. 5 to transmit and receive acoustic information.

FIG. 7 is a graphical representation of unipolar pulses with opposite phases.

FIG. 8 is a graphical representation of a multi-dimensional transducer array.

FIG. 9 is a perspective view of one embodiment of an interior of a probe including a multi-dimensional transducer array connected with circuit boards.

FIG. 10 is a cross sectional diagram of one embodiment of a multi-dimensional array assembled from modules.

FIGS. 11A and 11B are graphical representations of steps performed for manufacturing a multi-dimensional array using pre-diced modules.

#### DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

Faster or more complex two-dimensional and three-dimensional ultrasound imaging is provided by using multiplexing. A multiplexer is provided within a probe so that information from multiple transducer elements are multiplexed onto one signal channel for transmission to a base unit or ultrasound system for further processing. To avoid having different systems for different types of transducers, receive circuitry of an ultrasound system is operable in different modes based on the format of signals provided by the transducer. To further minimize the number of channels connecting a probe to an ultrasound system without adversely affecting the size of the probe, a transmit channel is separated from the receive channel by a transducer element. This separation isolates the transmit channel while minimizing integration of high voltage devices within the probe. To allow the element to isolate the transmit and receive channels, the transducer array is manufactured from separately diced modules, each module including signal traces to opposite sides of each element.

The developments discussed above for multiplexing may be used independent of the multiplexing or other features. These independent developments or features are described in three general sections below. Receive circuitry for receiving information associated with different signal formats or for receiving just multiplexed format is described first. Isolation of the transmit path from the receive path using a transducer element and associated methods of use are described second. Finally, transducer arrays and methods of manufacture are described.

##### Receive Circuitry:

FIG. 1 shows a block diagram of one embodiment of an ultrasound system 10. The system 10 includes a base unit 12 with receive circuitry 14 and an image processor 16. The receive circuitry 14 is operable to connect with different types of transducer probes 18, 20 via a cable 22. A plurality of receive circuits 14 are electrically connectable with the probes 18, 20 for processing signals from an array of elements 24. Additional, different or fewer components may be provided in the system 10, such as providing only one type of transducer probe 18, 20.

One transducer probe 20 comprises an array of piezoelectric or microelectromechanical elements 24 for transducing between acoustic and electrical energies. The probe 20 includes a single element, a linear array of elements or a multi-dimensional array of elements. The probe 20 also includes a housing covering the array. The housing is shaped to be a hand-held device or may be shaped for insertion into cavities or cardiovascular system of a patient. The probe 20 connects to the receive circuitry 14 using a cable 22 for each element 24 of the array. Each cable 22 transmits an analog signal representing the acoustic energy received at a single element 24. The signaling provided on the cable 22 from the probe 20 are conventional signals free of multiplexing or other intermediate circuits between the element 24 and the connector 32. The probe 20 provides signals or other information formatted differently than the signals from probe 18.

The probe 18 includes a linear or multi-dimensional array of elements 24 connected with a multiplexer 26. In one embodiment, 1,536 elements 24 are configured as a two-dimensional or multi-dimensional array. The probe 18 also includes a housing covering the array. The housing is shaped

to be a hand-held device or may be shaped for insertion into cavities or cardiovascular system of a patient. In one embodiment, the transducer probe 18 comprises multi-dimensional transducer probe manufactured as discussed below using modules, but other linear or multi-dimensional arrays using a ground plane or with separate signaling made from one PZT component or modules may be used.

The probe 18 includes preamplifiers 35 and time gain controls 37 as a receive channel 64 prior to multiplexing. The receive channel 64 connects with the element 24. The preamplified and time gain controlled information are provided to sample and hold circuits 60. The sample and hold circuit 60 comprise analog delays for multiplexing analog information from multiple elements 24 onto one output. In the preferred embodiment, no sample and hold function exists. Analog waveforms are interleaved in time with no "hold" and no "analog delay" operation. Use of a sample and hold is not a requirement but it is one possible alternative.

In one embodiment, the receive circuits in the probe 18 dissipate less than 5 watts. In one embodiment, one multiplexer 26 is provided for every eight elements 24, but a single multiplexer may be provided for all elements or for a different number of elements. The multiplexer 26 comprises an analog or digital network of switches responsive to a probe control 28. In one embodiment, the multiplexer 26 combines signals from a plurality of elements 24 using time division multiplexing. In alternative embodiments, frequency multiplexing or other multiplexing schemes now known or later developed may be used. The probe control 28 controls the multiplexer 26 in response to a clock signal so that analog signals from each of the elements are assigned a specific time slot within a frame of time division multiplex information. In one embodiment, the probe 18 and associated multiplexer 26 comprise the time division multiplexing probe discussed in U.S. Pat. No. 5,622,177, the disclosure of which is incorporated herein by reference. Additional, different or fewer components may be provided in a probe 18, such as providing additional amplifiers or filters in the probe 18 or a probe free of the preamplifiers or time gain controls.

The multiplexer 26 outputs time division multiplex or other formatted data to a line driver 30. The line driver 30 comprises an amplifier or other device integrated with or separate from the multiplexer 26 for transmitting the multiplexed information over the cable 22. Separate cables 22 may be provided for additional multiplexers 26, such as 192 or 256 cables 22.

The base unit 12 comprises an ultrasound imaging system, such as a hand-held, cart based or other system for generating a two-dimensional or three-dimensional representation of a patient. The receiver circuitry 14 receives information from one or more transducer probes 18, 20 for beamformation, detection and other ultrasound image processing by the image processor 16.

The receive circuit 14 includes a connector 32, a mode control processor 34, a preamplifier 36, a time gain control circuit 38, a low pass filter 40, a buffer 42, an analog-to-digital converter 44, a digital equalizer 46, a digital demultiplexer 48, an analysis processor 50, and a selectable delay 52. Additional, different or fewer components may be provided. The receiver circuit 14 comprises one or various combinations of two or more of the components described above. For example, the receiver circuitry comprises just the preamplifier 36 or just the low pass filter 40. The receive circuit 14 is operable with the transducer probe 20 where the signals from elements 24 may or may not be amplified and/or processed before transmission to the base unit 12. A second mode of operation allows transmission of time



division or other multiplexed signals representing a group of elements along one signal line or cable 22. The receive circuitry 14 comprises a single receive channel within the base unit 12. Multiple receive channels for association with different cables 22 and different elements 24 are provided.

The connector 32 comprises a female or male latch with electrical contacts for connecting with a bundle of cables 22. The connector 32 is operable to connect with different transducer probes 18, 20. For example, a probe with time division multiplexing capabilities is connected to the connector 32. As another example, the probe 18 is disconnected from the connector 32 and the other probe 20 is connected to the connector 32. The connector 32 releasably maintains physical and electrical contact with the bundle of cables 22. In alternative embodiments, a separate connector 32 is provided for different probes 18, 20. The same base unit 12 and receive circuit 14 may be used for receiving and processing information from different types of transducer probes 18, 20. For example, the connector 32 connects with the probe 18 for imaging using a fully populated two-dimensional or 1.5 dimensional array. Time division multiplexing allows for steering in two spatial dimensions for two-dimensional or three-dimensional imaging while minimizing the number of cables 22 for communicating signals to the base unit 12. The same connector 32 connects with the other transducer probe 20 for ultrasound imaging using signals free of multiplexing. In one embodiment, multiple connectors 32 are provided with relay or solid-state switching into the common receive circuit 14 to provide rapid access to a selection of transducers. Each individual connector 32 may accept either multiplexed transducers 18 or conventional transducers 20.

The mode control processor 34 comprises a control processor, general processor, application specific integrated circuit or other analog or digital device for controlling components of the receive circuit 14, such as the preamplifier 36 and low-pass filter 40. In response to a configuration entered by the user, in response to control signals provided by the probe control 28, in response to a detection by the connector 32 of a type of probe, or in response to analysis of signals received from the ultrasound probe 18, 20, the mode control processor 34 configures one or more components of the receive circuitry 14 for processing in accordance with the type of data or data format provided by the probe 18, 20. The characteristics of the receive circuit are configured as a function of the data format.

The preamplifier 36 comprises transistors or other analog or digital devices for providing a low noise, wide band matched receiver. The preamplifier 36 is programmable or responsive to the mode control processor 34 for programming characteristics of the preamplifier. For operation with the transducer probe 20 or operation with signals representing a single transducer element 24, the preamplifier 36 is programmed to have an impedance characteristic similar to or at the impedance of the element 24 and the cable 22, such as 1 kOhm impedance. The impedance matches a generalization based on expected variations in cable impedances for different types of probes 20. The preamplifier 36 may alternatively be programmable for specifically matching specific types of probes 20 with different cables 22, cable lengths or elements 22. Preamplifier input impedance, gain and frequency response may be controlled either by selectable switched components or by altering preamplifier bias current. In practice, both methods may be employed simultaneously within an integrated circuit realization. For operation with multiplexed signals, the preamplifier 36 is programmed for an impedance match to the line driver 30 or

other output circuitry of the probe 18. For example, the preamplifier 36 is programmed to provide an approximate 50 ohms impedance match. In alternative embodiments, different preamplifiers 36 are selected by the mode control processor.

In another embodiment, the gain characteristic of the preamplifier 36 is selected as a function of the format of signals or type of probe 18, 20. Multiplexed transducers 18 may require lower preamplifier gain than conventional transducers 20 because signals are already preamplified within the transducer prior to multiplexing. Also, noise performance of the system preamplifier 36 is not as stringent for multiplexed transducers 18 with integral preamplifiers 36, so a degraded noise preamplifier might be desirable to save power or otherwise optimize input impedance, gain, and frequency response.

Another programmable characteristic is the bandwidth of the preamplifier 36. For multiplexed information, the preamplifier 36 is not band limited or operates over a wide band, such as passing frequencies having a symbol rate of more than twice the center frequency of the transducer array (e.g., more than 5 MHz, 30 MHz or 100 MHz or more) for time division multiplexing. For information free of multiplexing, the bandwidth may be 2–15 MHz, such as associated with ultrasound frequencies or the frequency band of the transducer. Other characteristics of the preamplifier 36 may be adapted or altered as a function of the data format provided from the transducer probe 18, 20.

Signal conditioning blocks may be included in the multiplexer 26 or with the preamplifier 36 to provide pre- and post-equalization for frequency dependent losses in the cable 22. In alternative embodiments, the digital equalizer 46 provides post-equalization. The equalization may minimize inter-symbol interference. For example, pre-emphasis or high-frequency boost could be applied prior to driving the cable to compensate for frequency-dependent cable losses. An all-pass phase correction filter could also be implemented in the system receiver 14 to further reduce inter-symbol interference prior to the ADC.

The time gain control 38 (i.e. depth gain control) comprises an adjustable gain amplifier for variably amplifying analog signals. For signals representing a single element 24, the variable gain comprises a 40 to 80 dB range, but other gains may be used to account for the approximately one dB per MHz per centimeter of depth attenuation of ultrasound signals. The time gain control 38 operates the same or differently for multiplexed signals. Where a time gain control 38 is provided in the probe 18, the time gain control 38 of the receive circuit 14 provides less or no variable gain for multiplexed signals. Where the time gain control 38 applies a variable gain, the application of the gain accounts for the time division multiplexing by applying a same gain within each frame of signals from multiple elements 24.

The low-pass filter 40 comprises an anti-aliasing filter implemented as a finite impulse response or infinite impulse response filter. The low-pass filter 40 band limits signals so signals greater than  $\frac{1}{2}$  the digital sampling rate do not alias into the signal spectrum. By lowering the bandwidth of the low-pass filter, a greater signal-to-noise ratio is provided as long as signals of interest are not removed or reduced. Signals of interest provided by the probe 20 or representing a single element 24 are provided in a 2–15 MHz frequency range. The low-pass filter 40 is programmed with a 6 dB down or other cutoff frequency of 30 MHz, 15 MHz less or other frequency. The bandwidth may be programmed as a function of the type of imaging or type of probe 20 used. For multiplexed signals, such as time division multiplex

information, the bandwidth is greater to pass multiplexed signals while minimizing inter-symbol interference. For example, the bandwidth is 30 MHz or greater, such as 50 or 100 MHz, to provide a Nyquist channel shape or a linear-phase low-pass filter with the following magnitude response symmetry:  $|H(f)|=1-|H(F_{\text{sample}}-f)|$ , for  $0 < f < F_{\text{sample}}$ , where  $F_{\text{sample}}$  is the multiplexed sample rate (e.g. 96 MHz). In practice,  $H(f)$  is an approximation to a Nyquist channel and errors are corrected via the digital equalizer 46.

The buffer 42 comprises an amplifier or other analog components for buffering signals input to the analog-to-digital converter 44. The buffer 42 provides the same characteristics regardless of the type of data or data format used, but may provide programmable characteristics that differ as a function of data format. For example, faster slew rate may be required from 42 for multiplexed data. A programmable slew rate limit could be used to conserve power in non-multiplexed modes.

The analog-to-digital converter 44 samples the analog signals and outputs digital representations in any one of various now known or later developed codes. For data representing a single element 24, the analog-to-digital converter 44 samples the data in response to a clock input but without reference to other timing information. For time division multiplex data, the analog-to-digital converter clock input is synchronized with the multiplexer 26. The synchronization allows proper separation of signals from each of the different elements 24 with minimized cross signal interface.

The digitized samples are provided to an adaptive digital equalizer 46. The digital equalizer 46 comprises a programmable finite impulse response filter, such as implemented using a shift register 54, multipliers 56 and a summer 58. In alternative embodiments, a processor or other device is used to implement the equalizer 46. The digital equalizer 46 filters time division multiplex information to remove inter-symbol interference. The filter coefficients applied to the multipliers 56 are based on a transfer function or generation of inter-symbol interference from the element 24 through various stages or components of the receive circuit 14 that operate on the analog signal. In one embodiment, the filter coefficients are programmable to allow for adaptations or variations in the transfer function. The coefficients are selected in response to a test signal or other data processing accounting for detected differences in the transfer function, such as caused by different probes 18, different processing characteristics of analog components of the receive circuit 14 or changes due to time and temperature. For signals representing a single element 24 or signals free of multiplexing, the digital equalizer 46 passes the signals, such as providing no delay in a single tap with a multiplier coefficient of one.

The demultiplexer 48 comprises a digital demultiplexer, such as a network of switches for separating signals from various time slots in a frame of time division multiplex information. The demultiplexer 48 operates as a conditional demultiplexer. The receive signals are digitally demultiplexed. For example, the demultiplexer outputs signals from different elements 24 on different outputs for beam formation and other image processing by the image processor 16. For conventional signals or signals free of multiplexing, the demultiplexer 48 passes the information to the image processor 16 for beam formation.

The optional analysis processor 50 comprises a digital signal processor, a general processor, an application specific integrated circuit, analog components, digital components and combinations thereof for synchronizing the analog-to-digital converter 44 with the multiplexer 26 or selecting

coefficients for the digital equalizer 46. The analysis processor 50 operates on a test signal. The probe control 28 causes the multiplexer 26 to transmit a known or predetermined digital or analog test signal through the cable 22 and receive circuit 14 to the analysis processor 50.

The test signal is transmitted as part of a calibration function, such as in response to user input or connection of the probe 18 to the connector 32. The base unit 12 commands or the probe control 28 automatically generate the test signals. In alternative embodiments, test signals are transmitted periodically. For example, a test signal is transmitted in a preamble or header for each frame of time division multiplexed information. One or both of synchronization and adaptive equalization are provided in response to periodic transmission of the test signals. For stability, some phase sensitive acquisition sequences, such as acquisition for Doppler processing, minimize or do not provide any adaptation or changes in phasing through synchronization or the equalization.

One or both of multiplexing or processing of the receive signals is adapted in response to the analysis of the test signal. For example, the operation of the multiplexer 26 is adapted to the operation of the analog-to-digital converter 44 by synchronizing clock signals. The analysis processor 50 selects a selectable delay 52 for phasing the clock signal provided to the multiplexer 26 in reference to the analog digital converter 44. Fixed delays in clocking circuitry, variable delays due to clock signal path lengths, multiplexer circuit delays, multiplex signal path length, group delays and amplifiers and digitization of delays cause misalignment, resulting in mixing signals from different elements 24 by the analog-to-digital converter 44. These misalignments may vary as a function of the probe 18, the receive circuit configuration, time, temperature and processes. The analysis processor 50 determines the beginning of each frame by detecting a known pattern or the test signal. Using the selectable delay 52, the phase of the clocking signals applied to the analog-to-digital converter 44 and the multiplexer 26 are synchronized. In alternative embodiments, the analog-to-digital converter clock signal is phased relative to the clock signal provided to the multiplexer 26, or a group or subgroup of receive circuits 14 are used to determine the phase of a clock signal common to more than one multiplexer 26 relative to another clock signal common to more than one analog-to-digital converter 44. The adaptive clock adjustments simplify the multiplexing control circuitry and interface between the receive circuit 14 and the probe 18. One clock line or cable 22 is provided without additional and separate phasing information. In alternative embodiments, separate clock and phasing signals are provided to the probe controls 28.

In one embodiment, the processing by the receive circuit 14 is altered or adaptive as a function of the test signal by the analysis processor 52. For example, the analysis processor 50 selects coefficients from a lookup table or calculates coefficients for use by the digital equalizer 46. The digital equalizer provides symbol alignment or removal of inter-symbol interference. The analysis processor 50 compares a known or stored test signal to the received test signal. Differences between the received test signal and the stored test signal are used to select coefficients. The coefficients are selected so that the receive signals are undistorted or inter-symbol interference removed or diminished. In alternative embodiments, results from more than one analysis processor 50 are used to select coefficients for use by the digital equalizer 46.

In one embodiment, the receive circuit 14 includes a transmit receive switch. In alternative embodiments discussed below, no transmit and receive switch is provided.

FIG. 2 represents a flow chart of one embodiment of operation of the system 10 of FIG. 1. In act 70, one of various possible probes 18, 20 are connected with a base unit 12. One of the probes 18, 20 is selected and attached to the connector 32. For example, a user desires three-dimensional cardiac imaging, so a two-dimensional array of elements in the probe 18 associated with time division multiplexing is connected.

For probes associated with multiplexing, a test signal is transmitted in act 72. Multiplexing or processing are adapted in response to the test signal. For data free of multiplexing, act 72 is optional or not provided. The test signal is transmitted in response to connection of the probe 18, response to control signals from the receive circuitry 14, in response to user input, automatically, or periodically. For example, a test signal is transmitted as part of an initial calibration process or is transmitted periodically in the header of first time slot or other slot of each frame of time division multiplex information. The received test signal is compared to an expected test signal. In response to comparison, equalization coefficients or other processing of the receive circuit is adapted or altered. Additionally or alternatively, the timing of the test signal is identified and selectable delays determined for synchronizing the analog-to-digital converter 44 with the multiplexer 26.

In act 74, the receive circuitry 14 is configured to have different characteristics as a function of the type of probe or format of the data received from the probe 18, 20 connected with the receive circuitry 14. Where the data format corresponds to multiple elements, such as time division multiplexed data, the information is processed in response to different impedance, gain, filtering, equalization, analog to digital conversion or other processes than for data associated with a single element or free of intervening circuitry in the probe 20. Any one or combination of two or more of the various characteristics may be altered as a function of the data format. Additional or different characteristics may also or alternatively be altered. Act 74 may be performed before or after act 72.

The analog information is then digitized. For time division multiplex information, the analog-to-digital converter 44 is synchronized with the multiplexed information. The multiplexed information is then demultiplexed for beamformation and other imaging processes.

#### Transmit and Receive Isolation:

A transducer element 24 may be used to isolate the transmit channel from the receive channel in either of the probes 18, 20 discussed above or another probe for use with different receive circuits. While useful for single element transducers, linear arrays, or arrays with limited or no transmit or receive circuitry within the probe, using a transducer element 24 to isolate the transmit and receive channels is particularly useful for multi-dimensional transducer arrays with at least part of transmit, and/or receive circuitry incorporated within the probe, such as discussed above for the time division multiplexing probe 18. A fully populated multi-dimensional transducer array requires a large number of transmit and receive channels. By placing transmit or receive circuitry within the probe and providing multiplexing, the number of cables 22 or channels from the probe 18 to the base unit 12 are minimized. However, the transmit and receive circuitry then coexists in a small space, making isolation of the receive circuits from the high voltages of the transmit circuits difficult. High voltage switches, such as switches able to withstand 200 volts of reverse voltage, are difficult to integrate with other receive circuits, such as a multiplexer. High voltage transmit and

receive switching is replaced with the transducer element for isolating the transmit channel from the receive channel.

FIG. 3 shows a transducer element 24 isolating or separating a transmit path 62 and a receive path 64. Direct connection between the transmit path 62 and the receive path 64 is avoided. The element 24 isolates the paths 62, 64 to allow high voltage transmission without subjecting the receive path 64 to the high voltage. High voltage devices are provided as part of the transmit path 62 but not as part of the receive path 64 in one embodiment. In alternative embodiments, high voltage devices are provided on the receive path 64.

The element 24 comprises one of a plurality of elements in a multi-dimensional or linear array. 1.5 dimensional and 2-dimensional arrays may be represented as multi-dimensional arrays of a  $N \times M$  grid of elements where both  $N$  and  $M$  are greater than 1. For multi-dimensional arrays, the elements may be small and have a high impedance as compared to elements 24 of a linear array. Parasitic loading associated with a cable 22 is also absent or reduced for use with a multiplexer and the probe 18. A smaller transmit pulser and very low power receive preamplifier may be used given the high element impedance than for a lower impedance.

The element 24 includes two electrodes 80 and 82. The electrodes 80 and 82 are on opposite of the element 24, such as being on a top and bottom of the element on a range dimension. The electrode 80 is free of an electrical connection with the electrode 82. Separate signal traces comprise or connect with each of the electrodes 80 and 82. Each element 24 is associated with two or more separate signal traces for associated separate electrodes 80, 82. In alternative embodiments, two or more electrodes share a same signal trace. One electrode 80 connects to the transmit path 62, and the other electrode 82 connects to the receive path 64. The element 24 is free of an electrical connection directly to ground, such as provided by an electrode connected directly to ground.

The transmit path 62 connects with the electrode 80 for applying a transmit waveform to the element 24. The transmit path 62 comprises at least one signal trace to element 24 within the probe 18. In other embodiments, additional transmit circuitry, such as a waveform generator 84, a switch driver 87, and a controller 88 are incorporated within the transmit path 62 and within the probe 18. In alternative embodiments, the controller 88, the driver 87, the waveform generator 84 or combinations thereof are positioned external to the probe 18, such as within the base unit 12.

The waveform generator 84 comprises one or more high voltage transistors, such as FET transistors, for generating unipolar, bipolar or sinusoidal waveforms. One embodiment of a transmit waveform generator 84 for generating a unipolar waveform is shown in FIG. 4. Two high voltage transistors 86, such as CMOS FET transistors with at least a withstand of 200 volts connect in series between a voltage source and ground. In one embodiment, one transistor comprises a PFET, and the other transistor comprises an NFET. The transistors 86 provide high voltage and ground driving of a unipolar waveform at the electrode 80. Since the transmit waveform generator 84 comprises a switch mode device, power dissipation is minimal. This circuitry for each element 24 uses about 0.2 millimeters<sup>2</sup> of die area. For a 2-dimensional array of 1,536 elements, about 307 millimeter<sup>2</sup> of die area is used. Other integration formats may be provided, such as providing groups of high voltage FET transistors in smaller application specific integrated circuits.

In alternative embodiments, other devices, such as digital-to-analog converters, are used for waveform generation.

FIG. 5 shows a network of transistors **86** for generating a bipolar waveform. Four transistors **86** allow generation of a bipolar waveform ending with a positive voltage, negative voltage, or zero voltage. Three transistors **86** may be used if the bipolar waveform is capable of ending at only one polarity, such as a positive voltage. Of the transistors, **Q1** and **Q2** of FIGS. 4 and 5 may have an integral reverse diode from the drain to the source, but transistors **Q3** and **Q4** avoid the reverse diode configuration to avoid conducting through the diodes. Other configurations and networks of transistors **86** may be used.

Each of the transistors **86** connects to a reference voltage, such as a positive voltage, a negative voltage or ground. As shown in FIG. 4, one transistor **86** connects to ground and the other transistor **86** connects to a positive or negative voltage. As shown in FIG. 5, two transistors **86** connect to ground, one transistor connects to a positive voltage, and another transistor connects to a negative voltage.

The driver **87** comprises a transistor or FET driver for controlling operation of the waveform generator **84**. In alternative embodiments, other drivers may be used. The driver **87** is integrated as part of an application specific integrated circuit, but may have separate devices or comprise a general processor. The driver **87** is operable to provide voltage changes for operating the transistors **86**. For example, the transistor **Q2** of FIG. 4 is controlled by application of a 10 volt or 0 volt signal from the driver **87**. The transistor **Q1** is controlled by application of a 200 volt or 190 volt signal from the driver **87**.

The controller **88** comprises a general processor, analog components, digital components, application specific integrated circuit, or combinations thereof for controlling one or more drivers **87** associated with one or more elements **24**. In one embodiment, the controller **88** is integrated on the same application specific integrated circuit as the driver **87**, but may be a separate device. The controller **88** outputs binary signals to control the operation of the driver **87** and waveform generator **84**. The controller **88** in one embodiment extrapolates or selects transmit configurations or waveform parameters for an entire array or sub-array based on simple control signals provided from external to the probe **18**. In alternative embodiments, the controller **88** is located external to the probe.

The receive path **64** comprises at least a single signal trace connected with the electrode **82** on an opposite side of the element **24** from the transmit path **62**. In other embodiments, the receive path **64** includes one or more of diodes **90**, **92**, preamplifier **94** and a multiplexer **96**. Additional, different or fewer circuits may be provided as part of the receive path **64**, such as a filter. The electronics may not contain an explicit filter in the probe where the transducer element itself may be sufficient and/or the natural low-pass response of the amplifier is sufficient to filter the receive signal. The receive path is included within the probe **18** with the element **24**. In alternative embodiments, a multiplexer is not provided and the preamplifier **94** is provided in a base unit **12** separate from the probe **18** or in the probe **18**. A cable **22** connects the receive path **64** to the base unit **12**.

The diodes **90** and **92** comprises Schottky diodes or other high current, low voltage diode devices. In one embodiment, the diodes **90** and **92** are free of quiescent power dissipation. Each of the diodes **90** and **92** connects to ground with an opposite or different polarity. The diodes **90** and **92** comprise a diode clamp to limit voltage swings on the receive path **64** at the electrode **82**. For example, the diodes **90** and **92** limit

voltage transitions to between plus or minus 0.2 to 0.7 volts. In alternative embodiments, transistors or other devices are used for limiting the voltage at the electrode **82**.

In one embodiment, the diodes **90** and **92** are integrated in an application specific integrated circuit with the preamplifier and multiplexer circuits **94** and **96**. Other integration formats may be provided, such as providing discrete diode arrays and separate preamplifier/multiplexer circuits in smaller application specific integrated circuits.

The preamplifier **94** comprises one or more transistors for amplifying a signal from the electrode **82**. For example, a differential BJT pair with current outputs are provided using a 7 volt BiCMOS process or other transistor process. Using 20 low  $\mu\text{A}$  per channel with a 5 volt supply allows a consumption of 0.1 milliwatts per channel. Other preamplifiers with different power consumptions and associated components and characteristics may be used. The preamplifier **94** may alternatively or additionally include a time or depth gain control amplifier or a filter. For a time gain control amplifier integrated within the probe **18**, a low power device for providing some but not all of the time gain compensation may be used. In alternative embodiments, a larger, more power consuming variable amplifier is provided.

The multiplexer **96** comprises a network of switches, such as transistors and analog sample and hold circuits for multiplexing the signals of a plurality of transmit paths **64** onto one cable **22**. For example, the multiplexer **96** comprises an 8 to 1 multiplexer for multiplexing signals from 8 different elements **24** into one frame of analog information. In one embodiment, the multiplexer **96** is operable to provide 12 MSPS for each receive path **64** for a total of 96 MSPS for 8 receive paths **64**. The circuitry of the receive path **64** is free of high voltage devices and may be integrated into one application specific integrated circuit or other general circuit in a small space within the probe **18**.

Connecting the transmit and receive path **62** and **64** to opposite electrodes **80** and **82**, respectively isolates the high voltages and high voltage devices of the transmit path **62** from the low voltage devices of the receive path **64**. FIG. 6 shows a flow chart of one embodiment for transmitting and receiving using the element **24** of FIG. 3. In act **100**, a high voltage transmit waveform is provided to the transducer element **24**, and the voltage in the receive path **64** is limited in act **102**. Subsequently, the voltage on the transmit path **62** is limited in act **106** and echo signals are received on the receive path **64** in act **104**.

The transmit and receive operation of the element **24** is free of switches to select between the transmit and receive path **62** and **64**. In response to control signals from the controller **88**, the driver circuit **86** causes the waveform generator **84** to generate a high voltage (e.g., 200 volt) transmit waveform in act **100**. Where the waveform generator **84** is positioned within the probe **18**, the transmit waveform is generated within the probe **18**. The transmit waveform is applied to one electrode **80** of the element **24**. The voltage of the other electrode is limited, effectively acting as a ground or D.C. reference, in act **102**. The diodes **90** and **92** clamp the voltage of the receive path **64** connected to the electrode **82** to within a small voltage range as compared to the high voltage of the transmit waveform. In response, the element **24** generates an acoustic signal due to the potential difference across the electrodes **80** and **82**. The element **24** also isolates the transmit path **62** from the receive path **64**, preventing damage to receive circuitry without high voltage switching.

For a subsequent receive operation of act **104**, the voltage at the transmit path **62** is limited. In one embodiment, a

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transistor **86** of the waveform generator **84** connects a ground or reference voltage to the electrode **80**. For example, Q2 of the waveform generator **84** shown in FIG. **4** is switched "on" to ground the electrode **80**. In an alternative embodiment, another reference voltage, such as a positive voltage applied through Q1 is connected to the electrode **80** to limit the voltage swing or change of the electrode **80**. While the voltage of the transmit path and associated electrode is limited in act **106**, electrical signals are generated at the electrode **82** in response to acoustic echo signals received by the element **24** in act **104**. Since the electrical signals received are small, such as less than 0.2 volts, the diodes **90** and **92** avoid introducing noise within or clipping the receive signal. The receive signal is amplified, filtered, multiplexed, or otherwise processed for transmission over the cable **22** to the base unit **12**. For example, the amplifier **94** preamplifies the signals and adjusts the gain of the electrical signals as a function of time. The multiplexer **96** multiplexes the electrical signals with other electrical signals responsive to different transducer elements **24**. The same process is repeated for receive channels **64** associated with other elements **24**. The transmit and receive operations are performed free of selecting between transmit and receive paths for connection with an electrode. Each of the transmit and receive paths **62** and **64** act to ground or otherwise maintain an electrode **80**, **82** at a reference voltage during reception and transmission, respectively.

Using the waveform generator **84** shown in FIG. **4**, unipolar waveforms may be generated ending either with zero voltage or a positive voltage. The unipolar waveform generator **84** is capable of ending on a positive or zero voltage state without damage to the circuit. An alternative embodiment would allow unipolar waveform generation between zero and a negative voltage by swapping the NMOS and PMOS devices and using a negative power supply. In either case a low impedance condition is provided whether the unipolar transmit waveform ends at a 0 voltage or other voltage.

FIG. **7** shows two mirror symmetric unipolar waveforms **108** and **110**. The first unipolar waveform **108** begins at a low state or zero voltage level, includes a positive voltage pulse, returns to a 0 voltage level and then ends at a high state or positive voltage level. The subsequent unipolar waveform **110** begins at a high state or positive voltage and ends at a low state or zero voltage. Since one waveform begins at the higher voltage and ends at the lower voltage and the other waveform **108** begins at the lower voltage and ends at the higher voltage with the same number of cycles, the two waveforms sum, to substantially a zero value. Substantially accounts for differences in rise and fall path times of the transistors **86** and other differences in performance using transmit waveforms beginning at different voltages. In alternative embodiments, the high state is zero volts and the low state is a negative voltage.

The mirror symmetric capability of the unipolar waveform generator **84** allows for tissue harmonic or other harmonic imaging using phase inversion with unipolar transmit waveforms. As acoustic energy responsive to the transmit waves propagates and scatters within tissue, energy at second harmonics or other harmonics of the fundamental transmit frequency is generated. The receive signals responsive to each of the unipolar waveforms include information at the fundamental frequencies as well as the harmonic frequencies. When the receive signals responsive to the phase inverted transmit unipolar waveforms are combined or added, information at the fundamental frequencies cancels, leaving information at harmonic frequencies.

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Harmonic imaging in response to phase inversion of transmit waveforms is provided using simple unipolar waveforms. The transistors **86** used for generating the unipolar waveform are designed to avoid rise time and fall time mismatches, minimizing the amount of harmonic information introduced by the waveform generator **84**. The material of the element **24** has a high poling voltage in one embodiment to minimize differences in operation or receive mismatches due to initial generation at two different DC bias points (e.g. 0 and +V). Transmission of a phase inverted unipolar pulses may be used with systems having a transmit channel within the base unit or within the probe, and with systems using transmit and receive switching.

#### Multi-Dimensional Transducer:

Various transducers can be used with any of the transmit and receive paths, probes and receive circuits discussed above. Some such multi-dimensional transducer arrays for fully sampled use with time division multiplexing and element based isolation of transmit and receive paths is shown in FIGS. **8–11**. Time division multiplexing reduces the channel count or number of cables **22** without limiting the beamforming performed by the base unit **12**. Separate signal traces or connection of opposite electrodes **80** and **82** to transmit and receive path allows integration of transmit and receive circuitry in the probe **18** without power consuming transmit and receive switching. Various aspects of the multi-dimensional transducer may be used independent of other aspects of the embodiments described herein, such as using a particular element spacing without time division multiplexing or other integration of circuitry within the probe **18**.

FIG. **8** shows a 2-dimensional array **200** of elements **24**. The elements **24** are spaced in a grid along the elevation and azimuth dimensions. A different or same number of elements **24** may be provided along the elevation dimension than along the azimuth dimension. A plurality of elements **24** are provided in columns **204** along the azimuth dimension. The elements **24** have a pitch or spacing along the azimuth dimension. In one embodiment, a  $\frac{1}{2}$  wavelength pitch is used. From the center of one element to the center of an adjacent element **24** along the azimuth dimension, a distance of  $\frac{1}{2}$  of a wavelength is provided. For example, in an array designed for operation at 2.5 MHz, the pitch is 300 micrometers. Other spacings may be used.

The elements **24** are provided in rows **202** along the elevation dimension. The pitch or spacing along the elevation dimension is greater than the pitch or spacing along the azimuth dimension. In one embodiment, the pitch along the azimuth dimension is  $\frac{2}{3}$  or less, such as  $\frac{1}{2}$ , than the pitch along the elevation dimension. For the 2.5 MHz center frequency array example given above, the pitch in elevation is 600 micrometers or one wavelength. For large pitches, each individual element may be sub-diced for proper operation or to maintain a desired ratio of the width to a thickness of the element **24**. In the example provided above, the elements **24** are sub-diced along the elevation dimension, such as providing a dicing cut extending through about 90 percent of PZT material at the center of each array, but not sub-diced along the azimuth dimension. Other sub-dicing depths may be used.

FIG. **8** shows thirty-two elements **24**. In alternative embodiments, different numbers of elements are provided, such as 1,536 elements in 64 azimuthally spaced rows **202** and 24 elevational spaced columns **204**, or 2,048 elements in 64 azimuthally spaced rows **202** and 32 elevation spaced columns **204**.

FIG. **9** shows a probe **18** integrating the array **200**. The probe **18** includes the array **200**, flexible circuit materials or

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signal traces **206**, **208**, a plurality of circuit boards **210**, a capacitor **212** and a bundle of cables **22**. These components are housed within a plastic or other ergonomically shaped probe cover or housing. Different, fewer or additional components may be included in the probe **18**.

The flexible circuits **206**, **208** comprise Kapton or other flexible, thin, electrical insulating material with deposited signal traces on one or two sides. Flexible circuit is used herein to describe any flexible or non-rigid material with one or more electrical conductors. In one embodiment, the flexible circuit material is 50  $\mu\text{m}$  thick. Separate flexible circuit materials **206** and **208** are provided for separate transmit and receive paths. For example, one flexible circuit **206** provides electrodes and traces from one side of the elements **24** of the array **200**, and the other flexible circuit **208** comprises electrodes and traces from an opposite or different side of the elements **204** of the array **200**.

FIG. **10** shows an elevation cross-section of the array **200** and the associated connections of the two flexible circuits **206** and **208**. The array **200** is subdivided along the elevation dimension into four modules **222**. Additionally, the array **200** may be subdivided along the elevation direction into different or fewer modules **222**. For example, only one, two, three, or more modules may be used. Each module has an associated pair of flexible circuits **206** and **208**. Each module **222** includes a plurality of layers along the range dimension, such as a first matching layer **218**, a first electrode layer on the top of the element **24** formed from the first flexible circuit **208**, a second matching layer **216**, an element or piezoelectric (PZT) layer **214**, a second electrode on a bottom side of the piezoelectric layer **214** formed from the second flexible circuit **208** and a backing material **220**. Additional, different or fewer layers may be provided in one, more or all of the modules **222**. For example, only one or three or more matching layers **216**, **218** are used, or both matching layers **216** and **218** are on a top side of the top electrode and flexible circuit **208**.

The two different flexible circuits **208** and **206** are folded along one or two sides of the modules from the PZT material or layer **214** towards and along the backing material **220**. Separate signal traces are provided to each of the elements **24** on both sides or top and bottom of the elements **24**. Separate signal traces are provided on the flexible circuit **206** for each of the elements **24**, and separate signal traces are provided on the flexible circuit **208** for each of the elements **24**. Each of the elements **24** independently connects with the separate signal traces on the top and bottom along the range dimension of the element **24**. Separate signal traces allow for element based isolation of the transmit and receive paths. In alternative embodiments, a common ground connects with a plurality of elements **24**.

The backing material **220** of each module **222** is separated from the other backing material **220** of another module **222** by two or four layers of flexible circuit **206**, **208**. The PZT layer **214** of one module **222** is separated by one or two flexible circuit layers **208** from the PZT layer **214** of another module **222**. The width of the PZT layer **214** is greater than the width of the backing material **220** to account for the different thicknesses due to the different number of flexible circuits **206**, **208**. By having a thin flexible circuit material, adverse acoustic effects are avoided by minimizing the separation between elements **24** of different modules **222**.

Referring again to FIG. **9**, the flexible circuits **208** and **206** are shown as having an increasing width away from the array of elements **200**. Increasing the width allows for greater separation of the signal traces from individual elements **24**. The greater separation provides for less capacitive coupling between signal traces.

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FIG. **9** shows a plurality of printed circuit boards **210**, such as seven circuit boards **210**. In alternative embodiments, a single circuit board **210**, a different number of circuit boards **210** or no circuit board are provided in the probe **18**. In one embodiment, six circuit boards **210** include transmit and receive circuitry, such as the probe integrated circuitry discussed above. Each transmit and receive circuit board **210** connects with one of six elevationally spaced modules **222**. In alternative embodiments, one circuit board **210** connects with elements **24** in different modules **222**, or elements **24** in a same module **222** connect with different circuit boards **210**. A seventh circuit board comprises a control logic circuit board. The control logic circuit board interfaces with the base unit **12** for operating the transmit and receive circuitry. The printed circuit boards **210** and other components of the probe **18** are sized to fit within the handle of the probe **18**. The probe **18** is designed for ergonomic handling by a user, such as being less than four inches in diameter or providing a handheld grip.

In one embodiment, the circuit boards include one or more multiplexers. For example, a plurality of eight to one multiplexers are provided for multiplexing signals from the elements **24** onto **192** system channels or cables **22**. In other embodiments, fewer or more multiplexers for use with fewer or more cables **22** or system channels are provided. For example, an array **200** of 64 rows **202** and 32 columns **204** is provided with multiplexers for transmitting time division multiplexed information on 256 cables **22**. Providing the multiplexer in the probe **18** with the array **200**, fewer cables **22** and associated system channels or signal lines are provided than the number of elements **24** of the array **200**. For example, the product of the number of elements along the elevation dimension and the number of elements along the azimuth dimension is greater than the number of cables **22**.

The circuit boards **210** connect with the flexible circuits **206** and **208** using any now-known or later developed connectors or connections. Using two or more separate signal traces for each element **24** provides electrical connections for twice the number of elements **24**. The connectors are attached to the flexible circuits **206**, **208** prior to array fabrication. In one embodiment, a ball grid array (BGA) or other matrix of bumps or other structures for soldering to the traces on the flexible circuits **206** and **208** are provided. Small pitch matrix type BGA connectors may be used. For example, the BGA connects the receive path signal traces to the multiplexer, and the multiplexer is then connected to the printed circuit boards, reducing the number of connections to the printed circuit boards. In another embodiment, transmit or receive circuitry are deposited or otherwise formed on the flexible circuit, resulting in the need for fewer connections from the flexible circuits **206**, **208** to the printed circuit boards **210**. In yet another embodiment, a direct attachment, such as wire bond jumping or other interconnections, is provided between the flexible circuit and the printed circuit boards.

FIGS. **11A** and **B** represent acts in a process for manufacturing the array **200**. FIG. **11A** shows three modules of elements **24**. Each module **222** includes at least two rows and two columns of elements **24** in an N by M array. The PZT layer **214** of each module **222** and the associated flexible circuits **208**, **206** are diced independently for each module **222**. The dicing includes one or both of dicing along the azimuth or elevation dimensions to form the elements **24**. By dicing the electrodes or flexible circuits **208**, **206** separately for each module **222**, each module **222** may be tested separately. Separate testing allows for disposal of a defective module **222** before final assembly. For example,

capacitance tests or acoustic tests are performed for each element **24** of each of the modules **222**.

Each of the separately diced modules **222** is formed as shown in FIG. **10**. Any of various manufacturing processes may be used, and different orders of assembly provided. In one embodiment, the first matching layer **216**, a slab of piezoelectric layer **214** and a flexible circuit **206** positioned on the bottom of the piezoelectric layer **214** are stacked on top of the backing layer **220**. Precision tooling with pins and associated holes or templates are used for aligning these layers. The bottom flexible circuit **206** has signal traces on both sides for connecting with different elements **24**. The aligned layers are then bonded or glued together.

After bonding, the bottom layer of flexible circuit material **206** is folded along the sides of the backing layer **220** below the layer of piezoelectric layer **214**. The width of the backing layer **220** is narrower than the width of the piezoelectric layer **214** by about the width of one or two layers of the flexible circuit **206**. In one embodiment, the flexible circuit **206** is folded on two sides of the backing layer **220**, but may be folded on just one side. The bottom flexible circuit **206** is tightly bonded to the backing material by placing the partial module **222** through a Teflon coated or other frame with bonding material or glue. Alternatively, the bottom flexible circuit **206** is bonded to the sides of the module **222** during a later act of bonding the top flexible circuit **208**.

The first matching layer **216** and piezoelectric layer **214** are diced along the azimuth dimension. For example, six major dicing kerfs are formed that extend into, but not through, the flexible circuit material **206**. Minor dicing kerfs may also be formed along the azimuth dimension. The minor dicing kerfs extend about 90% into the piezoelectric layer **214**. Other dicing depths may be used. Epoxy, silicone or other material is used to fill the diced kerfs. A kerf-filling material with a higher acoustic impedance may be used since only the piezoelectric layer **214** and the first matching layer **216** are diced in azimuth. In alternative embodiments, other layer are diced in azimuth and a lower acoustic impedance kerf-filling material is used. In alternative embodiments, no kerf-filling material is used.

The surface of the first matching layer **216** is ground or otherwise processed to remove any excess kerf-filling material if necessary. The top flexible circuit **208** and the second matching layer **218** are aligned and bonded together using pins and holes or templates. The bonded top flexible circuit **208** is then bonded to bottom matching layer **216**. In alternative embodiments, the top flex **208** and top matching layer **218** are aligned and bonded to the bottom matching layer **216** on the module **222** as one operation associated with filling the kerfs.

The position of the top flex circuit **208** and associated signal traces relative to the bottom flex circuit **206** and associated signal traces is within a tolerance sufficient to allow separate signal traces for each element **24**. For example, a tolerance of plus or minus 50  $\mu\text{m}$  allows a dicing area of 100  $\mu\text{m}$  between each of the elements **24** along the elevation dimension without adversely cutting a signal trace. Other tolerances and distances are possible. The alignment is performed using precision-tooled pins and holes, template or optical alignment. By providing signal traces on flex circuits **206**, **208** on both sides of the module **222**, less dense signal traces are provided, allowing larger dicing windows. In alternative embodiments, a greater density of signal traces is provided and the flexible circuits **208** and **206** are provided on one side of the module **222**.

The top flexible circuit **208** is folded along one or two sides of the piezoelectric layer **214** and backing layer **220**.

The flexible circuit **208** extends from the piezoelectric layer **214** towards the backing material **220**. Where signal traces are provided on a top side or outward facing side of the bottom flexible circuit **206** and on a bottom or inward facing side of the top flexible circuit **208**, an insulation layer is added between the two flexible circuits **206** and **208**. For example, a 25  $\mu\text{m}$  or other thickness of Teflon or electrically non-conductive material is applied to one or both of the flexible circuit layers **206** prior to assembly or during assembly. The upper flexible circuit **208** is then bonded to the sides of the modules **222** by passing through a frame with a Teflon coating or other coating. Both flexible circuits and the associated electrodes are bonded to the module **222**.

The module **222** is then diced in the elevation dimension, such as dicing to form 64 columns **204** of elements **24**. The dicing extends through both flexible circuits **206** and **208** and the piezoelectric layer **214** into the backing layer **220**. In one embodiment, no minor dicing kerfs are provided, but minor dicing kerfs may be used. In one embodiment, the top flexible circuit **208** is examined through a microscope for optically aligning the dicing saw. The elevational dices in combination with the earlier azimuthal dices define the elements **24**. The elevational dices may be provided for each of the modules **222** at a same time or at different times. The dicing results in top and bottom separate electrodes and associated signal traces for each of the elements **24** without a grounding plane common to all of the elements. In alternative embodiments, a grounding plane is used with only one separate signal trace for each element **24**.

The separately diced modules **222** are aligned as shown in FIG. **11B**. The modules **222** are positioned adjacent to each other along the elevation or azimuth dimension to form a larger array **200** of elements **24**. Each of the modules **222** is separated from another module **222** by one or more of the flexible circuits **206**, **208**. In one embodiment, each of the modules **222** represents 64 azimuthally-spaced rows **202** and four or six elevationally-spaced columns **204** of elements **24**. By aligning four or six modules **222** in the elevation and azimuth dimensions, a 64 by 24 grid of elements **24** is provided. Other number of modules, sizes and number of elements grids may be used with or without separation of modules **222** by flexible circuits **206**, **208**.

The top flexible circuit **208** has signal traces formed on a bottom side so that the flexible circuit **208** electrically insulates the signal traces of one module **222** from the signal traces of another module **222**. In alternative embodiments, an insulator material, such as additional Kapton or other material, is positioned between the two modules **200** for electrical isolation of the signal traces.

Prior to aligning, each of the modules **222** is pressed through a Teflon-covered frame or other frame with glue or other bonding material. The pressing tightly fits the flexible circuits **206** and **208** along the sides of the modules **222** for minimizing any separation between modules.

The modules **222** are positioned within a frame **224**. The frame comprises a graphite material, another conductive material, or other non-conductive material. The four modules **222** either press-fit within the frame **224** or are positionable within the frame **224**. When the modules **222** are positioned within the frame **224**, the spacing between the PZT layer **214** of the modules **222** is 50–150 micrometers, but other spacing may be used. The spacing is the result of the flexible circuit material between the piezoelectric layers **214** of each module **222**. A 50–150 micrometer spacing is either 0–100 micrometers larger than a normal kerf width. Other relative widths may be used. Minimizing the separation between modules **222** minimizes the beam width in the



elevation dimension or the elevation point spread function. The frame 224 aligns the modules 222 in both dimensions but may provide less tolerance within an azimuth dimension. Higher tolerance alignment may be provided through manual optical alignment, pin and hole alignment or precise machining of the frame 224 as a template.

After the modules 222 are aligned within the frame 224, the kerfs from the separate dicing are filled with silicone or other kerf-filling material. The kerf-filling material also acts to bond the modules 222 to each other and the frame 224. In alternative embodiments, the kerfs of the modules 222 are filled prior to alignment. In alternative embodiments, no kerf filling is used. A protective layer of lens material or other focusing or non-focusing acoustically transparent material is positioned over or around the array 200. For example, high temperature or room temperature vulcanized silicon is formed over the array 208. Where the array 200 is fully sampled, the additional protective layer provides for no focus or limited focus.

The flexible circuits 206, 208 and associated signal traces are connected to the printed circuit boards or multiplexers. The output of the multiplexers are connected to cables 22. The cables electrically connect the elements 24 of the array 200 to the base unit 12.

In alternative embodiments, different multi-dimensional arrays are provided with a multiplexer integrated within the probe 18 and/or isolation of transmit and receive paths by the transducer element 24. Multiplexing allows multiplexing of multiple channels onto a single channel, such as through time division multiplexing. The amount of multiplexing, the bandwidth desired, the center frequency, and the clock rate determine the amount of multiplexing used. For example, a system with a 40 MHz clock rate may use up to a 25 MHz center frequency transducer assuming Nyquist sampling rate up to 1.6 times the center frequency. With multiplexing, the center frequency may be reduced to reduce the number of system channels or cables 22. In the example above, a 2:1 multiplexer allows use of up to a 12.5 MHz center frequency transducer with a 120% bandwidth, but doubles the number of elements 24 using one cable 22. A 3:1 multiplexer allows use of up to a 8.3 MHz center frequency transducer. 4:1 allows 6.3 MHz, 5:1 allows 5.0 MHz, 6:1 allows 4.2 MHz, 7:1 allows 3.6 MHz and 8:1 allows 2.5 MHz. Higher clock rates allow either more multiplexing or higher center frequency transducers.

Some multi-dimensional arrays provide a plurality of transducer elements arranged with N elements along a first dimension where N is greater than one and with M elements along a second dimension where M is greater than one and not equal to N. For example, a multi-PZT layer linear array, a 1.5D, I-beam, +-beam or other arrays of elements 24 have different distributions of elements 24. A probe houses the array 200 of elements 24. A multiplexer within the probe and connected to at least two of the plurality of transducer elements 24 allows for a greater number of elements 24 with a fewer number of system channels or cables 22 connected to the base unit 12.

Multiplexing allows higher resolution use of 1.5 dimensional transducer arrays, such as arrays with two or more elevation rows of 96 elements 24 in the azimuth dimension. For example, with 2:1 time domain multiplexing, a 1.5D array with three or four rows of 96 elements uses 192 system channels or cables 22 at up to 12.5 MHz. With 7:1 multiplexing of 7 segments or rows of 96 elements 24, the array may operate at up to 3.6 MHz with 192 system channels or cables 22 in a 40 MHz clock rate system.

A plano-concave transducer with isolated left and right elevation aperture spaced elements 24 may also benefit from

multiplexing. For example, see the arrays described in U.S. Pat. No. 6,043,589, the disclosure of which is incorporated herein by reference. Two or three segmented arrays operate at a higher center frequency and/or with more elements by multiplexing signals from one or more elements with signals from another element.

Transducers configured as two or more separate or intersecting linear or curved linear arrays may also benefit from multiplexing. A first linear array is positioned along one dimension and a second linear array is positioned along the second dimension or not parallel to the first array. For example, the various I-beam, +-beam or other arrays disclosed in U.S. Pat. No. 6,014,473, the disclosure of which is incorporated herein by reference, use multiplexing to allow for a greater number of elements with the same or fewer cables 22. In this example, one linear array is used for imaging and one or more other orthogonal arrays provide tracking information. By multiplexing, image resolution is sacrificed less by using system channels or cables for tracking arrays. For example, one imaging and two tracking arrays each use 192 elements 24 with 3:1 multiplexing to 192 cables 22. Other distributions of elements 24 within the arrays may be used.

Bi-layer or multiple layer transducer arrays may also benefit from multiplexing. Two or more layers of PZT within a linear or other array of elements 24 are used for harmonic imaging. One or more one dimensional arrays of elements 24 along the azimuth dimension have layers of elements 24 or PZT along the range dimension. For example, the arrays disclosed in U.S. Pat. No. 6,673,016 (Ser. No. 10/076,688, filed Feb. 14, 2002) or U.S. Pat. No. 5,957,851 use multiple layers of elements 24 separated by electrodes. Multiplexing allows for a greater number of separately addressable PZT layers and/or elements 24. The relative phasing of one layer to another layer provides for either fundamental or harmonic operations.

A square grid of elements as a two-dimensional array or a single linear array may also benefit from multiplexing. Multiplexing allows for more elements with fewer system channels or cables 22. Multiplexing provides higher resolution and/or faster scanning for two or three dimensional imaging.

While the invention has been described above by reference to various embodiments, it should be understood that many changes and modifications can be made without departing from the scope of the invention. It is therefore intended that the foregoing detailed description be understood as an illustration of the presently preferred embodiment of the invention, and not as a definition of the invention. It is only the following claims, including all equivalents, that are intended to define the scope of this invention.

What is claimed is:

1. A system for receiving different types of signals from different ultrasound transducers, the system comprising:
  - a connector operable to connect with first and second ultrasound transducer probes;
  - a receiver circuit operable to process time division multiplexed information if the connector is connected with the first ultrasound transducer probe and operable to process other information if the connector is connected with the second ultrasound transducer probe, the other information free of time division multiplexing, the processing of time division multiplexed information different than the processing of other information.
2. The system of claim 1 wherein the receiver circuit comprises a pre-amplifier.



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3. The system of claim 2 wherein the pre-amplifier has a programmable impedance characteristic, a first impedance characteristic of the programmable impedance characteristic corresponding to a transducer element impedance and a second impedance characteristic of the programmable impedance characteristic corresponding to a linear driver impedance.

4. The system of claim 2 wherein the pre-amplifier has a programmable gain, a first gain for time division multiplexed information and a second gain for other information, the first gain different than the second gain.

5. The system of claim 1 wherein the receiver circuit comprises a programmable low-pass filter, a first bandwidth for other information and a second bandwidth for time division multiplexed information, the second bandwidth greater than the first bandwidth.

6. The system of claim 1 wherein the receive circuit comprises a digital equalizer, the digital equalizer operable to filter the time division multiplexed information and operable to pass the other information.

7. The system of claim 1 wherein the receive circuit comprises an analog-to-digital converter and a digital de-multiplexer.

8. The system of claim 1 wherein the receive circuit comprises an analog-to-digital converter; and further comprising:

a selectable delay operatively connected with a clock signal, the selectable delay operable to synchronize the analog-to-digital converter with the time division multiplexed information.

9. A method for receiving different types of signals from different ultrasound transducers, the method comprising:

- (a) connecting a transducer to an ultrasound system;
- (b) outputting from the transducer on one signal line information in a data format corresponding to one of: multiple elements or a single element;
- (c) processing the information with a first characteristic if the data format corresponds to multiple elements; and
- (d) processing the information with a second characteristic if the data format corresponds to a single element, the first characteristic different than the second characteristic.

10. The method of claim 9 wherein (b) comprises outputting the information in a time division multiplexed data format wherein at least two divisions of time corresponds to information from two different elements received at a substantially same time.

11. The method of claim 9 wherein (c) comprises setting a first impedance to match a linear driver impedance and (d) comprises setting a second impedance to match a transducer element impedance.

12. The method of claim 9 wherein (c) comprises setting a first pre-amplifier gain and (d) comprises setting second pre-amplifier gain, the second pre-amplifier gain different than the first pre-amplifier gain.

13. The method of claim 9 wherein (c) comprises setting a low-pass filter passband at greater than for (d).

14. The method of claim 9 wherein (c) comprises digitally equalizing time division multiplexed data and (d) comprises passing the information.

15. The method of claim 9 wherein the information comprises analog time division multiplexed information and further comprising:

- (e) converting the analog time division multiplexed information to digital information; and
- (f) de-multiplexing the digital information.

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16. The method of claim 9 wherein the information comprises analog time division multiplexed information, further comprising:

- (e) synchronizing an analog-to-digital converter with the analog time division multiplexed information.

17. An ultrasound system for receiving data associated with a plurality of elements on a single cable, the system comprising:

an ultrasound transducer including a plurality of elements and a multiplexer connected with the plurality of elements;

an analog-to-digital converter operative to receive analog ultrasound signals from the multiplexer; and

a digital de-multiplexer operative to receive digital information from the analog-to-digital converter.

18. The system of claim 17 further comprising a receive circuit operable in a first mode for time division multiplexed signals from the multiplexer and operable in a second mode for signals free of multiplexing, wherein the digital de-multiplexer is operable in a pass mode for the signals free of multiplexing.

19. The system of claim 17 further comprising a digital equalizer connected between the analog-to-digital converter and the digital de-multiplexer.

20. The system of claim 19 wherein the digital equalizer comprises a programmable finite impulse response filter, the finite impulse response programmable as a function of a test signal from the multiplexer.

21. The system of claim 17 wherein the multiplexer and analog-to-digital converter each include a clock input; and further comprising:

a selectable delay connected between the multiplexer clock input and the analog-to-digital clock input, the selectable delay operable to synchronize the multiplexer and the analog-to-digital converter.

22. An ultrasound method for receiving data associated with a plurality of elements on a single cable, the method comprising:

- (a) receiving a test signal from a transducer;
- (b) processing the test signal;
- (c) multiplexing ultrasound information in the transducer;
- (d) processing the time division multiplexed ultrasound information from the transducer; and
- (e) adapting one of (c) and (d) as a function of (b).

23. The method of claim 22 wherein (e) comprises determining equalizer coefficients as a function of (b).

24. The method of claim 23 wherein (b) comprises comparing the test signal to a stored signal; and (e) comprises determining in response to the comparison results.

25. The method of claim 22 wherein (e) comprises determining a clock signal adjustment as a function of (b).

26. The method of claim 25 wherein (b) comprises identifying a time signal in the test signal, and (e) comprises delaying a clock signal as a function of the time signal.

27. The method of claim 22 wherein (d) comprises converting the time division multiplexed ultrasound information from an analog format to a digital format and (e) comprises synchronizing the time division multiplexed ultrasound information from the transducer with the conversion of (d).

28. The method of claim 22 wherein (a) occurs interleaved with each frame multiplexed information of (c).

29. The method of claim 22 wherein (a) occurs prior to (c).

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外部链接	<a href="#">Espacenet</a> <a href="#">USPTO</a>		

#### 摘要(译)

提供了用于从不同超声换能器接收不同类型的信号格式的方法和系统。超声系统的基本单元包括连接器和接收器电路，用于与多种不同类型的换能器中的一种连接。例如，在一个接收通道上提供与单个元件相关的模拟信息的传统换能器与连接器和接收器电路连接。或者，输出表示多个换能器元件的时分多路复用或其他多路复用信息的换能器与连接器和接收器电路连接。接收器电路根据数据格式不同地处理接收的信息。例如，前置放大器阻抗或增益对于单个元件信号与时分多路复用信号是不同的。另一个例如，时分多路复用信号的低通滤波器带宽大于表示单个元件的信号。还包括具有位于数字多路分解器之前的模数转换器的接收电路，用于多路复用数据格式。来自换能器阵列的时分多路复用或其他多路复用信号在多路分解之前被转换成数字格式。还提供了用于适应多路复用信号的多路复用或处理的方法和系统。测试信号或超声信号从换能器传输到接收器电路并进行分析。换能器的多路复用器与模拟数字转换器同步，具有响应于分析的可选择的延迟。选择数字均衡器的系数或其他滤波器特性作为其函数分析。

