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(54) SINE SATURATION TRANSFORM

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- (63) Continuation of application No. 11/048,232, filed on Feb. 1, 2005, which is a continuation of application No. 10/184,032, filed on Jun. 26, 2002, now Pat. No. 6.850,787.
- (60) Provisional application No. 60/302,438, filed on Jun. 29, 2001.
- (51) **Int. Cl.**

A61B 5/00 (2006.01)

(56) References Cited

U.S. PATENT DOCUMENTS

4,960,128 A	10/1990	Gordon et al.
4,964,408 A	10/1990	Hink et al.
5,041,187 A	8/1991	Hink et al.
5,069,213 A	12/1991	Polczynski
5,163,438 A	11/1992	Gordon et al.
5,337,744 A	8/1994	Branigan

5,341,805	A	8/1994	Stavridi et al.
D353,195	S	12/1994	Savage et al.
D353,196	S	12/1994	Savage et al.
5,377,676	A	1/1995	Vari et al.
D359,546	S	6/1995	Savage et al.
5,431,170	A	7/1995	Mathews
D361,840	S	8/1995	Savage et al.

(Continued)

FOREIGN PATENT DOCUMENTS

WO WO 98/42250 10/1998

(Continued)

OTHER PUBLICATIONS

U.S.P.T.O. Office Action for U.S. Appl. No. 10/184,032, dated May 5, 2004, in 13 pages.

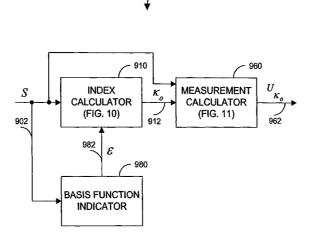
(Continued)

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(57) ABSTRACT

A transform for determining a physiological measurement is disclosed. The transform determines a basis function index from a physiological signal obtained through a physiological sensor. A basis function waveform is generated based on basis function index. The basis function waveform is then used to determine an optimized basis function waveform. The optimized basis function waveform is used to calculate a physiological measurement.

20 Claims, 14 Drawing Sheets



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U.S. PATENT	DOCUMENTS	6,368,283 B1	4/2002	Xu et al.
		6,371,921 B1	4/2002	Caro et al.
	Savage et al.	6,377,829 B1	4/2002	Al-Ali
	Branigan et al.	6,388,240 B2	5/2002	Schulz et al.
	Savage et al.	6,397,091 B2	5/2002	Diab et al.
5,456,252 A 10/1995	Vari et al.	6,430,525 B1	8/2002	Weber et al.
5,482,036 A 1/1996	Diab et al.	6,463,311 B1	10/2002	
5,490,505 A 2/1996	Diab et al.	6,470,199 B1		Kopotic et al.
5,494,043 A 2/1996	O'Sullivan et al.	6,501,975 B2		Diab et al.
5,533,511 A 7/1996	Kaspari et al.	6,505,059 B1		Kollias et al.
	Savage et al.	6,515,273 B2	2/2003	
5,562,002 A 10/1996		6,519,486 B1 *		Edgar et al 600/336
	Caro et al.			
	Durand et al.	6,519,487 B1	2/2003	
	Diab et al.	6,525,386 B1		Mills et al.
/ /	Kiani-Azarbayjany et al.	6,526,300 B1		Kiani et al.
	Diab et al.	6,541,756 B2		Schulz et al.
		6,542,764 B1		Al-Ali et al.
	Tobler et al.	6,580,086 B1		Schulz et al.
	Diab et al.	6,584,336 B1		Ali et al.
	Tobler et al.	6,595,316 B2	7/2003	Cybulski et al.
	Lepper, Jr. et al.	6,597,933 B2	7/2003	Kiani et al.
	Diab et al.	6,606,511 B1	8/2003	Ali et al.
	Lepper, Jr. et al.	6,632,181 B2	10/2003	Flaherty et al.
	Diab et al.	6,639,668 B1	10/2003	Trepagnier
5,782,757 A 7/1998	Diab et al.	6,640,116 B2	10/2003	Diab
5,785,659 A 7/1998	Caro et al.	6,643,530 B2		Diab et al.
5,791,347 A 8/1998	Flaherty et al.	6,650,917 B2		Diab et al.
5,810,734 A 9/1998	Caro et al.	6,654,624 B2		Diab et al.
5,823,950 A 10/1998	Diab et al.	6,658,276 B2		Kianl et al.
	Caro et al.	6,661,161 B1		Lanzo et al.
	Caro et al.	6,671,531 B2		Al-Ali et al.
	Baker et al 600/300			Diab et al.
	Kiani-Azarbayjany et al.	6,678,543 B2		
	Mills et al.	6,684,090 B2		Ali et al.
	Wohltmann et al.	6,684,091 B2	1/2004	
		6,697,656 B1	2/2004	
5,919,134 A 7/1999		6,697,657 B1		Shehada et al.
	Tobler et al.	6,697,658 B2	2/2004	
	Lepper, Jr. et al.	RE38,476 E	3/2004	Diab et al.
	Kiani et al.	6,699,194 B1	3/2004	Diab et al.
5,995,858 A 11/1999		6,714,804 B2	3/2004	Al-Ali et al.
	Mills et al.	RE38,492 E	4/2004	Diab et al.
	Diab et al.	6,721,582 B2	4/2004	Trepagnier et al.
	Diab et al.	6,721,585 B1	4/2004	
6,027,452 A 2/2000	Flaherty et al.	6,725,075 B2	4/2004	Al-Ali
6,036,642 A 3/2000	Diab et al.	6,735,459 B2	5/2004	
6,045,509 A 4/2000	Caro et al.	6,745,060 B2		Diab et al.
6,067,462 A 5/2000	Diab et al.	6,760,607 B2	7/2004	
	Diab et al.	6,770,028 B1		Ali et al.
	Diab et al.	6,771,994 B2		Kiani et al.
	Lepper, Jr. et al.	6,792,300 B1		Diab et al.
	Kaestle et al.	6,813,511 B2		Diab et al.
	Shehada			
6,144,868 A 11/2000		6,816,741 B2	11/2004	
	Kiani-Azarbayjany et al.	6,822,564 B2	11/2004	
	Gerhardt et al.	6,826,419 B2		Diab et al.
		6,830,711 B2		Mills et al.
	Diab et al.	6,850,787 B2		Weber et al.
	Mills et al.	6,850,788 B2	2/2005	Al-Ali
	Coffin, IV et al.	6,852,083 B2		Caro et al.
	Diab et al.	6,861,639 B2	3/2005	Al-Ali
	Diab et al.	6,898,452 B2	5/2005	Al-Ali et al.
	Snyder et al.	6,920,345 B2	7/2005	Al-Ali et al.
	Diab et al.	6,931,268 B1	8/2005	Kiani-Azarbayjany et al.
6,241,683 B1 6/2001	Macklem et al.	6,934,570 B2		Kiani et al.
	Diab et al.	6,939,305 B2		Flaherty et al.
6,263,222 B1 7/2001	Diab et al.	6,943,348 B1		Coffin, IV
	Lepper, Jr. et al.	6,950,687 B2	9/2005	
	Tobler et al.	6,961,598 B2	11/2005	
	Tobler et al.	6,970,792 B1	11/2005	
6,321,100 B1 11/2001				
	Al-Ali et al.	6,979,812 B2	1/2005	
		6,985,764 B2		Mason et al.
6,343,224 B1 1/2002		6,993,371 B2		Kiani et al.
	Kiani et al.	6,996,427 B2		Ali et al.
6,360,114 B1 3/2002	Diab et al.	6,999,904 B2	2/2006	Weber et al.

US 7,377,899 B2

Page 3

7,003,338	B2	2/2006	Weber et al.	7,254,431	B2	8/2007	Al-Ali
7,003,339	B2	2/2006	Diab et al.	7,254,433	B2	8/2007	Diab et al.
7,015,451	B2	3/2006	Dalke et al.	7,254,434	B2	8/2007	Schulz et al.
7,024,233	B2	4/2006	Al et al.	7,272,425	B2	9/2007	Al-Ali
7,027,849	B2	4/2006	Al-Ali	7,274,955	B2	9/2007	Kiani et al.
7,030,749	B2	4/2006	Al-Ali	D554,263	S	10/2007	Al-Ali
7,039,449	B2	5/2006	Al-Ali	7,280,858	B2	10/2007	Al-Ali et al.
7,041,060	B2	5/2006	Flaherty et al.	7,289,835	B2	10/2007	Mansfield et al.
7,044,918	B2	5/2006	Diab	7,292,883	B2	11/2007	De Felice et al.
7,067,893	B2	6/2006	Mills et al.	7,295,866	B2	11/2007	Al-Ali
7,096,052	B2	8/2006	Mason et al.				
7,096,054	B2	8/2006	Abdul-Hafiz et al.	FC	REIG	N PATE	NT DOCUMENTS
7,132,641	B2	11/2006	Schulz et al.				
7,142,901	B2	11/2006	Kiani et al.	WO W	O 02/4	7582 A3	6/2002
7,149,561	B2	12/2006	Diab				
7,186,966	B2	3/2007	Al-Ali		OT	HER PU	BLICATIONS
7,190,261	B2	3/2007	Al-Ali		-		
7,215,984	B2	5/2007	Diab	U.S.P.T.O. Not	ice of	Allowance	e for U.S. Appl. No. 10/184,032,
7,215,986	B2	5/2007	Diab	dated Oct. 5, 2	004, in	7 pages.	
7,221,971	B2	5/2007	Diab	U.S.P.T.O. Office	e Acti	on for U.S	6. Appl. No. 11/048,232, dated Jun.
7,225,006	B2	5/2007	Al-Ali et al.	6, 2007, in 8 p	ages.		
7,225,007	B2	5/2007	Al-Ali	U.S.P.T.O. Office	e Acti	on for U.S	S. Appl. No. 11/417,914, dated Jun.
RE39,672	Ε	6/2007	Shehada et al.	25, 2007, in 7	oages.		, ,
7,239,905	B2	7/2007	Kiani-Azarbayjany et al.				
7,245,953	В1	7/2007	Parker	* cited by ex	amine		

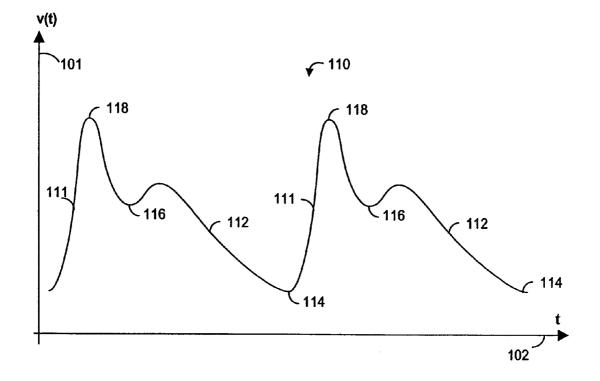


FIG. 1A

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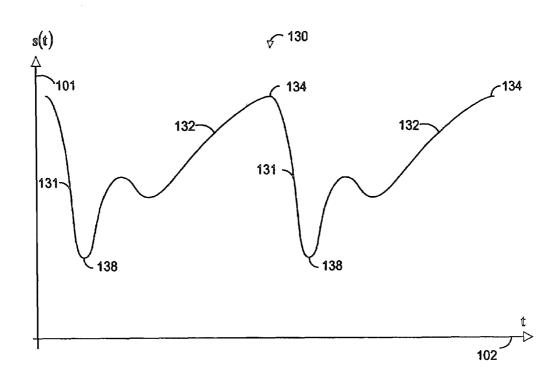


FIG. 1B

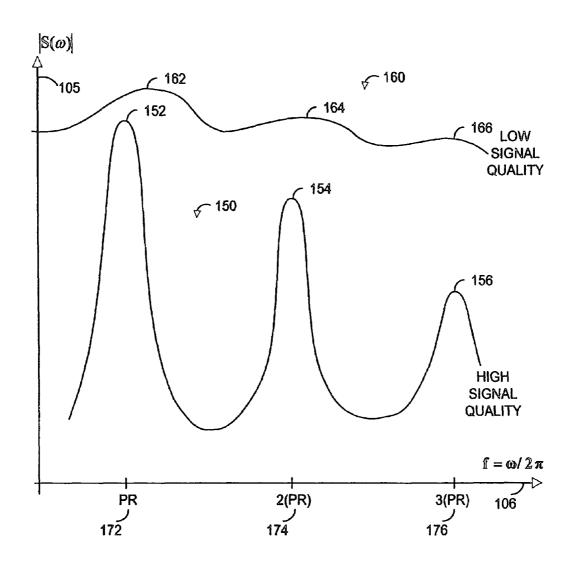


FIG. 1C

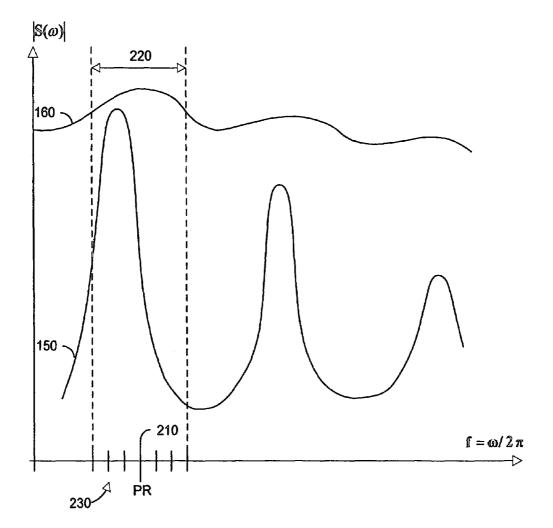


FIG. 2

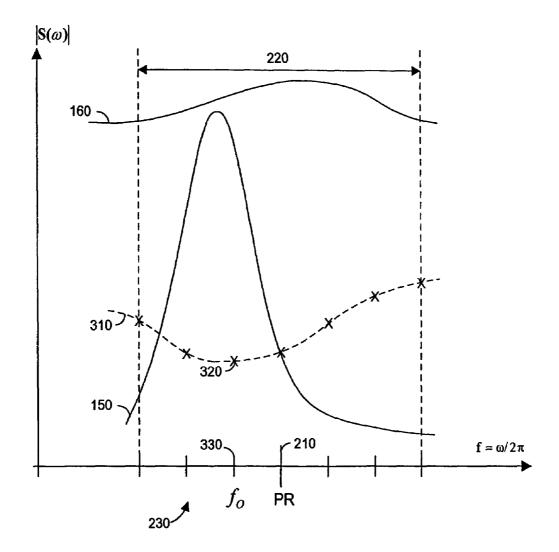


FIG. 3

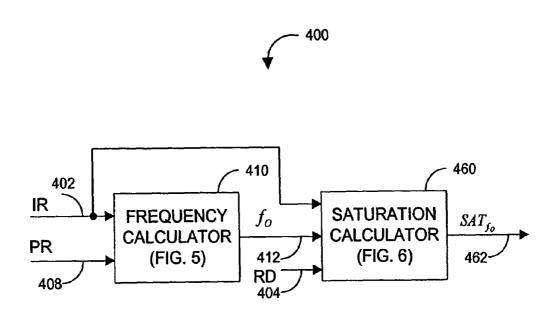


FIG. 4

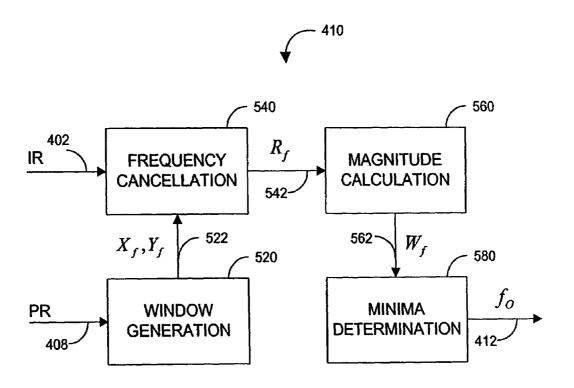


FIG. 5

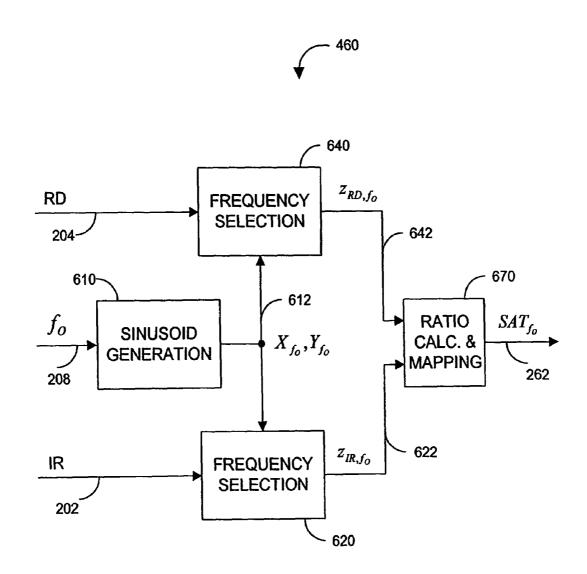


FIG. 6



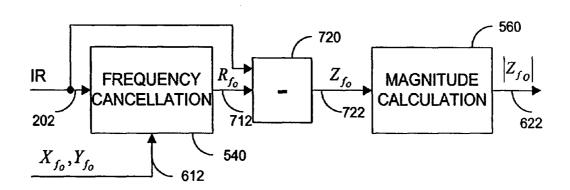


FIG. 7

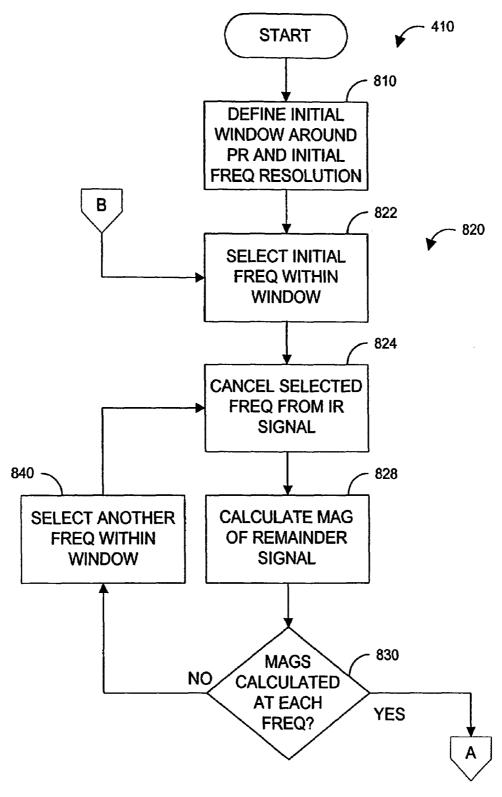
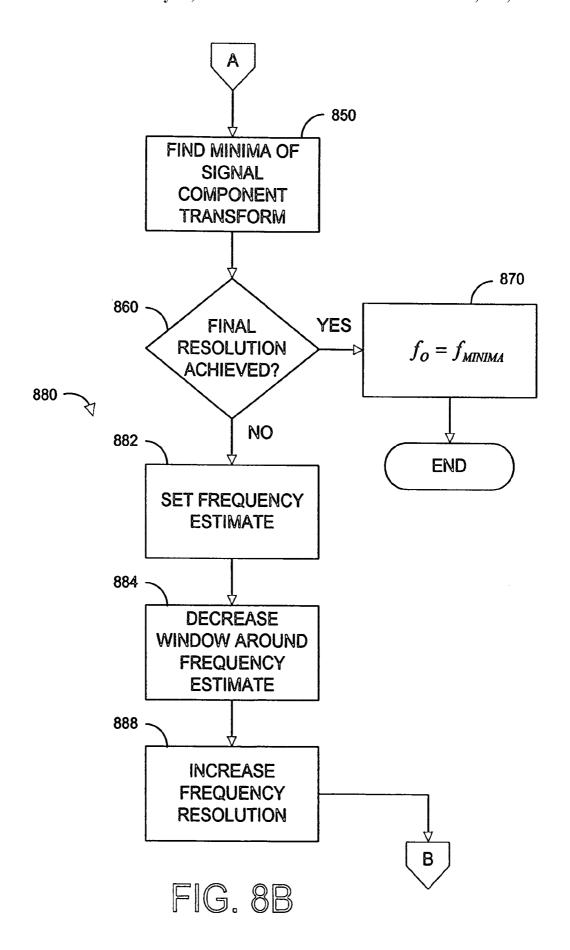


FIG. 8A



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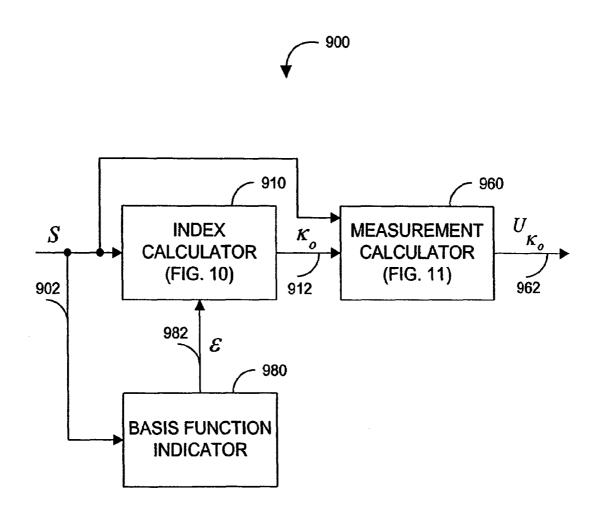


FIG. 9

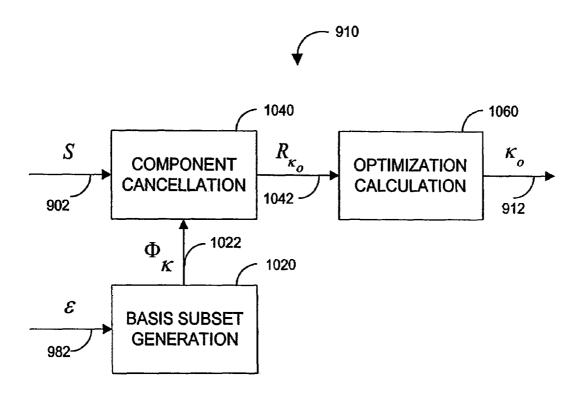


FIG. 10

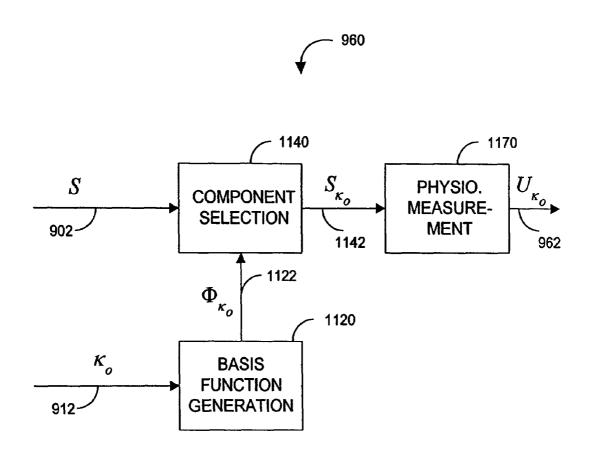


FIG. 11

SINE SATURATION TRANSFORM

CROSS-REFERENCE TO RELATED **APPLICATIONS**

The present application claims priority benefit under 35 U.S.C. §120 to, and is a continuation of, U.S. patent application Ser. No. 11/048,232, filed Feb. 1, 2005, entitled Signal Component Processor," which is a continuation of U.S. patent application Ser. No. 10/184,032, filed Jun. 26, 10 2002, entitled "Signal Component Processor," now U.S. Pat. No. 6,850,787, which claims priority benefit under 35 U.S.C. §119(e) from U.S. Provisional Application No. 60/302,438, filed Jun. 29, 2001, entitled "Signal Component foregoing disclosures herein by reference.

BACKGROUND OF THE INVENTION

Early detection of low blood oxygen is critical in the 20 medical field, for example in critical care and surgical applications, because an insufficient supply of oxygen can result in brain damage and death in a matter of minutes. Pulse oximetry is a widely accepted noninvasive procedure for measuring the oxygen saturation level of arterial blood, 25 an indicator of oxygen supply. A pulse oximeter typically provides a numerical readout of the patient's oxygen saturation and pulse rate. A pulse oximetry system consists of a sensor attached to a patient, a monitor, and a cable connecting the sensor and monitor. Conventionally, a pulse oximetry 30 sensor has both red (RD) and infrared (IR) light-emitting diode (LED) emitters and a photodiode detector. The pulse oximeter measurements are based upon the absorption by arterial blood of the two wavelengths emitted by the sensor. The pulse oximeter alternately activates the RD and IR 35 sensor emitters and reads the resulting RD and IR sensor signals, i.e. the current generated by the photodiode in proportion to the detected RD and IR light intensity, in order to derive an arterial oxygen saturation value, as is wellknown in the art. A pulse oximeter contains circuitry for 40 controlling the sensor, processing the sensor signals and displaying the patient's oxygen saturation and pulse rate.

SUMMARY OF THE INVENTION

FIG. 1A illustrates a plethysmograph waveform 110, which is a display of blood volume, shown along the ordinate 101, over time, shown along the abscissa 102. The shape of the plethysmograph waveform 110 is a function of heart stroke volume, pressure gradient, arterial elasticity and 50 peripheral resistance. Ideally, the waveform 110 displays a short, steep inflow phase 111 during ventricular systole followed by a typically three to four times longer outflow phase 112 during diastole. A dicrotic notch 116 is generally attributed to closure of the aortic valve at the end of 55 ventricular systole.

FIG. 1B illustrates a corresponding RD or IR sensor signal s(t) 130, such as described above. The typical plethysmograph waveform 110 (FIG. 1A), being a function of blood volume, also provides a light absorption profile. A pulse 60 oximeter, however, does not directly detect light absorption and, hence, does not directly measure the plethysmograph waveform 110. However, IR or RD sensor signals are 180° out-of-phase versions of the waveform 110. That is, peak detected intensity 134 occurs at minimum absorption 114 and minimum detected intensity 138 occurs at maximum absorption 118.

FIG. 1C illustrates the corresponding spectrum of s(t), which is a display of signal spectral magnitude $|S(\omega)|$, shown along the ordinate 105, versus frequency, shown along the abscissa 106. The plethysmograph spectrum is depicted under both high signal quality 150 and low signal quality 160 conditions. Low signal quality can result when a pulse oximeter sensor signal is distorted by motion-artifact and noise. Signal processing technologies such as described in U.S. Pat. No. 5,632,272, assigned to the assignee of the present invention and incorporated by reference herein, allow pulse oximetry to function through patient motion and other low signal quality conditions.

Ideally, plethysmograph energy is concentrated at the Processor." The present application also incorporates the 15 pulse rate frequency 172 and associated harmonics 174, 176. Accordingly, motion-artifact and noise may be reduced and pulse oximetry measurements improved by filtering out sensor signal frequencies that are not related to the pulse rate. Under low signal quality conditions, however, the frequency spectrum is corrupted and the pulse rate fundamental 152 and harmonics 154, 156 can be obscured or masked, resulting in errors in the computed pulse rate. In addition, a pulse rate, physiologically, is dynamic, potentially varying significantly between different measurement periods. Hence, maximum plethysmograph energy may not correspond to the computed pulse rate except under high signal quality conditions and stable pulse rates. Further, an oxygen saturation value calculated from an optical density ratio, such as a normalized red over infrared ratio, at the pulse rate frequency can be sensitive to computed pulse rate errors. In order to increase the robustness of oxygen saturation measurements, therefore, it is desirable to improve pulse rate based measurements by identifying sensor signal components that correspond to an optimization, such as maximum signal energy.

> One aspect of a signal component processor comprises a physiological signal, a basis function index determined from the signal, a basis function waveform generated according to the index, a component derived from the sensor signal and the waveform, and a physiological measurement responsive to the component. In one embodiment, the component is responsive to the inner product of the sensor signal and the waveform. In another embodiment, the index is a frequency and the waveform is a sinusoid at the frequency. In that embodiment, the signal processor may further comprise a pulse rate estimate derived from the signal wherein the frequency is selected from a window including the pulse rate estimate. The physiological measurement may be an oxygen saturation value responsive to a magnitude of the compo-

> Another aspect of a signal component processor comprises a signal input, a basis function indicator derived from the signal input, a plurality of basis functions generated according to the indicator, a plurality of characteristics of the signal input corresponding to the basis functions and an optimization of the characteristics so as to identify at least one of said basis functions. In one embodiment, the indicator is a pulse rate estimate and the processor further comprises a window configured to include the pulse rate estimate, and a plurality of frequencies selected from within the window. In another embodiment, the characteristic comprises a plurality of signal remainders corresponding to the basis functions and a plurality of magnitudes of the signal remainders. In that embodiment, the optimization comprises a minima of the magnitudes. In a further embodiment, the characteristic comprises a plurality of components corresponding to the

basis functions and a plurality of magnitudes of the components. In this embodiment, the optimization comprises a maxima of the magnitudes.

An aspect of a signal component processing method comprises the steps of receiving a sensor signal, calculating an estimated pulse rate, determining an optimization of the sensor signal proximate the estimated pulse rate, defining a frequency corresponding to the optimization, and outputting a physiological measurement responsive to a component of the sensor signal at the frequency. In one embodiment the determining step comprises the substeps of transforming the sensor signal to a frequency spectrum encompassing the estimated pulse rate and detecting an extrema of the spectrum indicative of the frequency. The transforming step may 15 comprise the substeps of defining a window including the estimated pulse rate, defining a plurality of selected frequencies within the window, canceling the selected frequencies, individually, from the sensor signal to generate a plurality of remainder signals and calculating a plurality of magnitudes 20 of a frequency selection; of the remainder signals. The detecting step may comprise the substep of locating a minima of the magnitudes.

In another embodiment, the outputting step comprises the substeps of inputting a red (RD) portion and an infrared (IR) portion of the sensor signal, deriving a RD component of the 25 RD portion and an IR component of the IR portion corresponding to the frequency and computing an oxygen saturation based upon a magnitude ratio of the RD component and the IR component. The deriving step may comprise the substeps of generating a sinusoidal waveform at the frequency and selecting the RD component and the IR component utilizing the waveform. The selecting step may comprise the substep of calculating the inner product between the waveform and the RD portion and the inner product between the waveform and the IR portion. The 35 selecting step may comprise the substeps of canceling the waveform from the RD portion and the IR portion, leaving a RD remainder and an IR remainder, and subtracting the RD remainder from the RD portion and the IR remainder from the IR portion.

A further aspect of a signal component processor comprises a first calculator means for deriving an optimization frequency from a pulse rate estimate input and a sensor signal, and a second calculator means for deriving a physiological measurement responsive to a sensor signal compo- 45 nent at the frequency. In one embodiment, the first calculator means comprises a signal component transform means for determining a plurality of signal values corresponding to a plurality of selected frequencies within a window including the pulse rate estimate, and a detection means for determining a particular one of the selected frequencies corresponding to an optimization of the sensor signal. The second calculator means may comprise a waveform means for generating a sinusoidal signal at the frequency, a frequency selection means for determining a component of the sensor 55 signal from the sinusoidal signal and a calculator means for deriving a ratio responsive to the component.

BRIEF DESCRIPTION OF THE DRAWINGS

FIGS. 1A-C are graphical representations of a pulse oximetry sensor signal;

FIG. 1A is a typical plethysmograph illustrating blood volume versus time;

FIG. 1B is a pulse oximetry sensor signal illustrating detected light intensity versus time;

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FIG. 1C is a pulse oximetry sensor signal spectrum illustrating both high signal quality and low signal quality conditions:

FIGS. 2-3 are magnitude versus frequency graphs for a pulse oximetry sensor signal illustrating an example of signal component processing;

FIG. 2 illustrates a frequency window around an estimated pulse rate; and

FIG. 3 illustrates an associated signal component transo form:

FIGS. 4-7 are functional block diagrams of one embodiment of a signal component processor;

FIG. 4 is a top-level functional block diagram of a signal component processor;

FIG. 5 is a functional block diagram of a frequency calculator:

FIG. 6 is a functional block diagram of a saturation calculator; and

FIG. 7 is a functional block diagram of one embodiment of a frequency selection;

FIGS. **8**A-B are flowcharts of an iterative embodiment of a frequency calculator; and

FIGS. **9-11** are functional block diagrams of another embodiment of a signal component processor;

FIG. 9 is a top-level functional block diagram of a signal component processor;

FIG. 10 is a functional block diagram of an index calculator; and

FIG. 11 is a functional block diagram of a measurement 30 calculator.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

FIGS. 2 and 3 provide graphical illustration examples of signal component processing. Advantageously, signal component processing provides a direct method for the calculation of saturation based on pulse rate. For example, it is not necessary to compute a frequency transform, such as an FFT, which derives an entire frequency spectrum. Rather, signal component processing singles out specific signal components, as described in more detail below. Further, signal component processing advantageously provides a method of refinement for the calculation of saturation based on pulse rate.

FIG. 2 illustrates high and low signal quality sensor signal spectrums 150, 160 as described with respect to FIG. 1C, above. A frequency window 220 is created, including a pulse rate estimate PR 210. A pulse rate estimate can be calculated as disclosed in U.S. Pat. No. 6,002,952, entitled "Signal Processing Apparatus and Method," assigned to the assignee of the present invention and incorporated by reference herein. A search is conducted within this window 220 for a component frequency f_Q at an optimization. In particular, selected frequencies 230, which include PR, are defined within the window 220. The components of a signals(t) at each of these frequencies 230 are then examined for an optimization indicative of an extrema of energy, power or other signal characteristic. In an alternative embodiment, the 60 components of the signals(t) are examined for an optimization over a continuous range of frequencies within the window 220.

FIG. 3 illustrates an expanded portion of the graph described with respect to FIG. 2, above. Superimposed on the high signal quality 150 and low signal quality 160 spectrums is a signal component transform 310. In one embodiment, a signal component transform 310 is indicative

of sensor signal energy and is calculated at selected signal frequencies 230 within the window 220. A signal component transform 310 has an extrema 320 that indicates, in this embodiment, energy optimization at a particular one 330 of the selected frequencies 230. The extrema 320 can be, for example, a maxima, minima or inflection point. In the embodiment illustrated, each point of the transform 310 is the magnitude of the signal remaining after canceling a sensor signal component at one of the selected frequencies. The extrema 320 is a minima, which indicates that canceling the corresponding frequency 330 removes the most signal energy. In an alternative embodiment, not illustrated, the transform 310 is calculated as the magnitude of signal components at each of the selected frequencies 230. In that embodiment, the extrema is a maxima, which indicates the largest energy signal at the corresponding frequency. The result of a signal component transform 310 is identification of a frequency f_Q 330 determined from the frequency of a signal component transform extrema 320. Frequency f_O 330 is then used to calculate an oxygen saturation. A signal component transform and corresponding oxygen saturation calculations are described in additional detail with respect to FIGS. 4-8, below. Although signal component processing is described above with respect to identifying a particular frequency within a window including a pulse rate estimate PR, a similar procedure could be performed on 2PR, 3PR etc. resulting in the identification of multiple frequencies f_{O1} , f_{O2} , etc., which could be used for the calculation of oxygen saturation as well.

Advantageously, a signal component transform 310 is calculated over any set of selected frequencies, unrestricted by the number or spacing of these frequencies. In this manner, a signal component transform 310 differs from a FFT or other standard frequency transforms. For example, a FFT is limited to N evenly-distributed frequencies spaced at a resolution of f_s/N, where N is the number of signal samples and f_s is the sampling frequency. That is, for a FFT, a relatively high sampling rate or a relatively large record length or both are needed to achieve a relatively high resolution in frequency. Signal component processing, as described herein, is not so limited. Further, a signal component transform 310 is advantageously calculated only over a range of frequencies of interest. A FFT or similar frequency transformation may be computationally more burdensome than signal component processing, in part because such a transform is computed over all frequencies within a range determined by the sampling frequency, f_s .

FIGS. 4-7 illustrate one embodiment of a signal component processor. FIG. 4 is a top-level functional block diagram of a signal component processor 400. The signal component processor 400 has a frequency calculator 410 and a saturation calculator 460. The frequency calculator 410 has an IR signal input 402, a pulse rate estimate signal PR input 408 and a component frequency f_O output 412. The frequency calculator 410 performs a signal component transform based upon the PR input 408 and determines the f_O output 412, as described with respect to FIGS. 2-3, above. The frequency calculator 410 is described in further detail with respect to FIG. 5, below.

In an alternative embodiment, the frequency calculator **410** determines $f_{\mathcal{O}}$ **412** based upon a RD signal input substituted for, or in addition to, the IR signal input **402**. Similarly, one of ordinary skill in the art will recognize that $f_{\mathcal{O}}$ can be determined by the frequency calculator **410** based upon one or more inputs responsive to a variety of sensor wavelengths.

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The saturation calculator 460 has an IR signal input 402, a RD signal input 404, a component frequency f_O input 412 and an oxygen saturation output, SAT_f462. The saturation calculator 460 determines values of the IR signal input 402 and the RD signal input 404 at the component frequency f_O 412 and computes a ratio of those values to determine SAT_f462, as described with respect to FIG. 6, below. The IR signal input 402 and RD signal input 404 can be expressed as:

$$IR = \begin{bmatrix} IR_0 \\ \vdots \\ IR_{N-1} \end{bmatrix};$$

$$RD = \begin{bmatrix} RD_0 \\ \vdots \\ RD_{N-1} \end{bmatrix}$$
(1)

where N is the number of samples of each signal input.

FIG. 5 shows one embodiment of the frequency calculator 410. In this particular embodiment, the frequency calculator functions are window generation 520, frequency cancellation 540, magnitude calculation 560 and minima determination 580. Window generation 520, frequency cancellation 540 and magnitude calculation 560 combine to create a signal component transform 310 (FIG. 3), as described with respect to FIG. 3, above. Minima determination 580 locates the signal component transform extrema 320 (FIG. 3), which identifies f_O 412, also described with respect to FIG. 3, above.

As shown in FIG. 5, window generation 520 has a PR input 408 and defines a window 220 (FIG. 3) about PR 210 (FIG. 3) including a set of selected frequencies 230 (FIG. 3)

$$\{f_m; m = 0, \dots M - 1\};$$

$$f = \begin{bmatrix} f_0 \\ \vdots \\ f_{M-1} \end{bmatrix}$$
(2)

where M is the number of selected frequencies 230 (FIG. 3) within the window 220 (FIG. 3). Window generation 520 has a sinusoidal output X_f , Y_f 522, which is a set of sinusoidal waveforms $x_{n,f}$, $y_{n,f}$ each corresponding to one of the set of selected frequencies 230 (FIG. 3). Specifically

$$X_{f} = \begin{bmatrix} x_{0,f} \\ \vdots \\ x_{N-1,f} \end{bmatrix}; \quad Y_{f} = \begin{bmatrix} y_{0,f} \\ \vdots \\ y_{N-1,f} \end{bmatrix}$$
(3a)

$$x_{n,f} = \sin(2\pi f n); \quad y_{n,f} = \cos(2\pi f n)$$
 (3b)

Also shown in FIG. 5, the frequency cancellation 540 has IR 402 and X_f , Y_f 522 inputs and a remainder output R_f 542, which is a set of remainder signals $r_{n,r}$ each corresponding to one of the sinusoidal waveforms $x_{n,f}$, $y_{n,f}$. For each selected frequency f 230 (FIG. 3), frequency cancellation 540 cancels that frequency component from the input signal IR 402 to generate a remainder signal $r_{n,f}$. In particular, frequency cancellation 540 generates a remainder R_f 542

$$R_{f} = \begin{bmatrix} r_{0,f} \\ \vdots \\ r_{N-1,f} \end{bmatrix}$$

$$R_{f} = IR - \frac{IR \cdot X_{f}}{|X_{f}|^{2}} X_{f} - \frac{IR \cdot Y_{f}}{|Y_{f}|^{2}} Y_{f}$$

$$(4a)$$

Additionally, as shown in FIG. 5, the magnitude calculation 560 has a remainder input R_f 542 and generates a magnitude output W_c 562, where

$$W_f = |R_f| = \sqrt{\sum_{n=0}^{N-1} r_{n,f}^2}$$
 (5)

Further shown in FIG. 5, the minima determination 580 20 has the magnitude values W_f 562 as inputs and generates a component frequency f_O output. Frequency f_O is the particular frequency associated with the minimum magnitude value

$$W_{fo} = \min \{W_f\} \tag{6}$$

FIG. 6 shows that the saturation calculator 460 functions are sinusoid generation 610, frequency selection 620, 640 and ratio calculation 670. Sinusoid generation has a component frequency f_O input 208 and a sinusoidal waveform X_{f_O} , Y_{f_O} output 612, which has a frequency of f_O . Frequency selection 620, 640 has a sensor signal input, which is either an IR signal 202 or a RD signal 204 and a sinusoid waveform X_{f_O} , Y_{f_O} input 612. Frequency selection 620, 640 provides magnitude outputs Z_{IR,f_O} 622 and Z_{RD,f_O} 642 which are the frequency components of the IR 202 and RD 204 sensor signals at the f_O frequency. Specifically, from equation 3(a)

$$X_{fo} = \begin{bmatrix} x_{0,fo} \\ \vdots \\ x_{N-1,fo} \end{bmatrix}; \quad Y_{fo} = \begin{bmatrix} y_{0,fo} \\ \vdots \\ y_{N-1,fo} \end{bmatrix}$$

$$(7)$$

Then, referring to equation 1

$$\begin{split} z_{IR,fo} &= \left| \frac{IR \cdot X_{fo}}{|X_{fo}|^2} X_{fo} + \frac{IR \cdot Y_{fo}}{|Y_{fo}|^2} Y_{fo} \right| \\ &= \sqrt{\frac{(IR \cdot Y_{fo})^2}{|X_{fo}|^2} + \frac{(IR \cdot Y_{fo})^2}{|Y_{fo}|^2}} \\ z_{RD,fo} &= \left| \frac{RD \cdot X_{fo}}{|X_{fo}|^2} X_{fo} + \frac{RD \cdot Y_{fo}}{|Y_{fo}|^2} Y_{fo} \right| \\ &= \sqrt{\frac{(RD \cdot Y_{fo})^2}{|X_{fo}|^2} + \frac{(RD \cdot Y_{fo})^2}{|Y_{fo}|^2}} \end{split}$$
(8b)

For simplicity of illustration, EQS. 8a-b assume that the cross-product of X_{f_o} and Y_{f_o} is zero, although generally this is not the case. The ratio calculation and mapping **670** has Z_{IR,f_o} **622** and Z_{RD,f_o} **642** as inputs and provides SAT_{f_o} **262** as an output. That is

$$SAT_{fo} = g\{z_{RD,fo}/z_{IR,fo}\}$$

$$(9)$$

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where g is a mapping of the red-over-IR ratio to oxygen saturation, which may be an empirically derived lookup table, for example.

FIG. 7 illustrates an alternative embodiment of frequency selection 620 (FIG. 6), as described above. In this embodiment, frequency cancellation 540 and magnitude calculation 560, as described with respect to FIG. 5, can also be used, advantageously, to perform frequency selection. Specifically, frequency cancellation 540 has IR 202 and X_{f_0} , Y_{f_0} 612 as inputs and generates a remainder signal R_{f_0} 712 as an output, where

$$R_{fo} = IR - \frac{IR \circ X_{fo}}{|X_{fo}|^2} X_{fo} - \frac{IR \circ Y_{fo}}{|Y_{fo}|^2} Y_{fo}$$
(10)

The remainder R_{f_0} 712 is subtracted 720 from IR 202 to yield

$$Z_{fo} = IR - \left[IR - \frac{IR \circ X_{fo}}{|X_{fo}|^2} X_{fo} - \frac{IR \circ Y_{fo}}{|Y_{fo}|^2} Y_{fo} \right]$$

$$= \frac{IR \circ X_{fo}}{|X_{fo}|^2} X_{fo} + \frac{IR \circ Y_{fo}}{|Y_{fo}|^2} Y_{fo}$$
(11)

where z_{f_O} 722 is the component of IR 202 at the f_O frequency. The magnitude calculation 560 has z_{f_O} 722 as an input and calculates

$$|Z_{fo}| = \left| \frac{|R \circ X_{fo}|}{|X_{fo}|^2} X_{fo} + \frac{|R \circ Y_{fo}|}{|Y_{fo}|^2} Y_{fo} \right|$$

$$= \sqrt{\frac{(|R \circ X_{fo}|^2}{|X_{fo}|^2} + \frac{(|R \circ Y_{fo}|^2)^2}{|Y_{fo}|^2}}$$
(12)

which is equivalent to equation 8a, above.

FIGS. **8**A-B illustrate an iterative embodiment of the frequency calculator **410** (FIG. **4**) described above. An iterative frequency calculator **410** has an initialization **810**, a signal component transform **820**, an extrema detection **850**, a resolution decision **860** and a resolution refinement **880** and provides a component frequency f_O **870**. Initialization **810** defines a window around the pulse rate estimate PR and a frequency resolution within that window.

As shown in FIG. 8A, a signal component transform 820 has an initial frequency selection 822, a frequency cancel-(8a) 50 lation 824 and an magnitude calculation 828. A decision block 830 determines if the magnitude calculation 828 has been performed at each frequency within the window. If not, the loop of frequency cancellation 824 and magnitude calculation 828 is repeated for another selected frequency in (8b) 55 the window. The frequency cancellation 824 removes a frequency component from the IR sensor signal, as described with respect to FIG. 5, above. The magnitude calculation 828 determines the magnitude of the remainder signal, also described with respect to FIG. 5, above. If the decision block 830 determines that the remainder signal magnitudes have been calculated at each of the selected frequencies, then the signal component transform loop 820 is exited to the steps described with respect to FIG. 8B.

As shown in FIG. 8B, the extrema detector 850 finds a minima of a signal component transform 820 and a resolution decision block 860 determines if the final frequency resolution of a signal component transform is achieved. If

not, resolution refinement **880** is performed. If the final resolution is achieved, the component frequency output f_O is equated to the frequency of the minima **870**, i.e. a signal component transform minima determined by the extrema detector **850**.

Further shown in FIG. 8B, the resolution refinement 880 has a set frequency estimate 882, a window decrease 884 and a frequency resolution increase 888. Specifically, the frequency estimate 882 is set to a signal component transform minima, as determined by the extrema detector 850. 10 The window decrease 884 defines a new and narrower window around the frequency estimate, and the frequency resolution increase 888 reduces the spacing of the selected frequencies within that window prior to the next iteration of a signal component transform 820. In this manner, a signal 15 component transform 820 and the resulting frequency estimate are refined to a higher resolution with each iteration of signal component transform 820, extrema detection 850, and resolution refinement 880.

In a particular embodiment, the component calculation 20 requires three iterations. A frequency resolution of 4 beats per minute or 4 BPM is used initially and a window of five or seven selected frequencies, including that of the initial pulse rate estimate PR, is defined. That is, a window of either 16 BPM or 24 BPM centered on PR is defined, and a signal 25 component transform is computed f_O a set of 5 or 7 selected frequencies evenly spaced at 4 BPM. The result is a frequency estimate f₁. Next, the frequency resolution is reduced from 4 BPM to 2 BPM and a 4 BPM window centered on f_1 is defined with three selected frequencies, i.e. 30 f_1 -2 BPM, f_1 , and f_1 +2 BPM. The result is a higher resolution frequency estimate f₂. On the final iteration, the frequency resolution is reduced to 1 BPM and a 2 BPM window centered on f2 is defined with three selected frequencies, i.e. f_2 -1 BPM, f_2 , and f_2 +1 BPM. The final result 35 is the component frequency for determined by a signal component transform to within a 1 BPM resolution. This component frequency f_O is then used to calculate the oxygen saturation, SAT_{f_0} as described above.

The signal component processor has been described 40 above with respect to pulse oximetry and oxygen saturation measurements based upon a frequency component that optimizes signal energy. The signal component processor, however, is applicable to other physiological measurements, such as blood glucose, carboxy-hemoglobin, respiration rate 45 and blood pressure to name a few. Further, the signal component processor is generally applicable to identifying, selecting and processing any basis function signal components, of which single frequency components are one embodiment, as described in further detail with respect to 50 FIG. 9, below.

FIGS. 9-11 illustrate another embodiment of a signal component processor 900. As shown in FIG. 9, the processor 900 has an index calculator 910 and a measurement calculator 960. The index calculator has a sensor signal input S 5002 and outputs a basis function index κ_O 912, as described with respect to FIG. 10, below. The measurement calculator 960 inputs the basis function index κ_O 912 and outputs a physiological measurement U_{κ_O} 962, as described with respect to FIG. 11, below. The processor 900 also has a basis function indicator 980, which is responsive to the sensor signal input S 902 and provides a parameter ϵ 982 that indicates a set of basis functions to be utilized by the index calculator 910, as described with respect to FIG. 10, below.

As shown in FIG. 10, the functions of the index calculator 65 910 are basis subset generation 1020, component cancellation 1040 and optimization calculation 1060. The basis

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subset generation 1020 outputs a subset φ_κ 1022 of basis function waveforms corresponding to a set of selected basis function indices $\kappa.$ The basis functions can be any complete set of functions such that

$$S = \sum_{k} a_k \Phi_k \tag{13}$$

For simplicity of illustration purposes, these basis functions are assumed to be orthogonal

$$[\overrightarrow{\Phi}_{\nu}, \overrightarrow{\Phi}_{n}] = 0; \gamma \approx \eta$$
 (14)

where [] denotes an inner product. As such

$$\alpha_{\kappa} = [S, \Phi_k]/[\Phi_{\kappa}, \Phi_{\kappa}] \tag{15}$$

$$S_{\kappa} = \alpha_{\kappa} \Phi_{\kappa} \tag{16}$$

In general, the basis functions may be non-orthogonal. The subset of basis functions generated is determined by an input parameter ϵ 982. In the embodiment described with respect to FIG. 5, above, the basis functions are sinusoids, the indices are the sinusoid frequencies and the input parameter ϵ 982 is a pulse rate estimate that determines a frequency window.

As shown in FIG. 10, the component cancellation 1040 generates a remainder output R_{κ} 1042, which is a set of remainder signals corresponding to the subset of basis function waveforms Φ_{κ} 1022. For each basis function waveform generated, component cancellation removes the corresponding basis function component from the sensor signal S 902 to generate a remainder signal. In an alternative embodiment, component cancellation 1040 is replaced with a component selection that generates a corresponding basis function component of the sensor signal S 902 for each basis function generated. The optimization calculation 1060 generates a particular index κ_{O} 912 associated with an optimization of the remainders R_{κ} 1042 or, alternatively, an optimization of the selected basis function signal components.

As shown in FIG. 11, the functions of the measurement calculator 960 are basis function generation 1120, component selection 1140, and physiological measurement calculation 1170. The component selection 1140 inputs the sensor signal S 902 and a particular basis function waveform Φ_{κ_0} 1122 and outputs a sensor signal component S_{κ_0} 1142. The physiological measurement 1170 inputs the sensor signal component S_{κ_0} 1142 and outputs the physiological measurement U_{κ_0} 962, which is responsive to the sensor signal component S_{κ_0} 1142. In the embodiment described with respect to FIG. 6, above, the basis functions Φ_{κ} are sinusoids and the index κ_{Q} is a particular sinusoid frequency. The basis function generation 1120 creates sine and cosine waveforms at this frequency. The component selection 1140 selects corresponding frequency components of the sensor signal portions, RD and IR. Also, the physiological measurement 1170 computes an oxygen saturation based upon a magnitude ratio of these RD and IR frequency components.

The signal component processor has been disclosed in detail in connection with various embodiments. These embodiments are disclosed by way of examples only and are not to limit the scope of the claims that follow. One of ordinary skill in the art will appreciate many variations and modifications.

What is claimed is:

1. A signal processing system for processing at least one physiological signal obtained from at least one physiological

sensor and computing at least one physiological measurement, the signal processing system comprising:

- at least one input configured to accept at least one physiological signal obtained from at least one physiological sensor;
- at least one basis function index calculator configured to determine at least one basis function index from the at least one physiological signal;
- at least one basis function waveform generator configured to generate at least one basis function waveform 10 according to the at least one basis function index;
- at least one optimization calculator configured to determine one or more optimized basis function waveforms.
- 2. The signal processing system of claim 1, further comprising at least one basis function component waveform 15 calculator configured to derive at least one component waveform from the at least one optimized basis function waveform and the at least one physiological signal.
- 3. The signal processing system of claim 2, further lator configured to calculate at least one physiological measurement responsive to the at least one basis function component waveform.
- 4. The signal processing system of claim 2, wherein the at least one component waveform comprises two or more 25 component waveforms.
- 5. The signal processing system of claim 1, wherein the at least one physiological signal comprises two or more physiological signals.
- 6. The signal processing system of claim 5, wherein the at 30 least one physiological signal comprises three or more physiological signals.
- 7. The signal processing system of claim 1, wherein the at least one basis function waveform comprises two or more basis function waveforms.
- 8. The signal processing system of claim 1, wherein the basis function index comprises a basis function frequency.
- 9. The signal processing system of claim 1, wherein the basis function waveform comprises a sinusoid waveform.
- 10. The signal processing system of claim 1, wherein the 40 at least one basis function index calculator is configured to derive at least one pulse rate estimate from the at least one physiological signal.
- 11. The signal processing system of claim 1, wherein the physiological measurement comprises one or more of a 45 group comprising oxygen saturation, blood glucose, carboxyhemoglobin, and blood pressure.
- 12. A signal processing method for processing, in the time domain, at least one physiological signal obtained from at least one physiological sensor and computing at least one 50 physiological measurement, the signal processing method comprising:

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- obtaining at least one physiological signal from at least one physiological sensor;
- determining at least one basis function index from the at least one physiological signal;
- generating at least one basis function waveform according to the at least one basis function index;
- determining at least one optimized basis function wave-
- 13. The signal processing method of claim 12, further comprising deriving at least one component waveform from the at least one optimized basis function waveform and at least one physiological signal.
- 14. The signal processing method of claim 13, further comprising calculating at least one physiological measurement responsive to the at least one basis function component waveform.
- 15. A signal processing system for processing at least one comprising at least one physiological measurement calcu- 20 physiological signal obtained from at least one physiological sensor and computing at least one physiological measurement, the signal processing system comprising:
 - at least one input configured to accept at least one physiological signal obtained from at least one physiological sensor;
 - means for determining at least one basis function index from the at least one physiological signal;
 - means for generating at least one basis function waveform according to the at least one basis function index;
 - means for determining one or more optimized basis function waveforms.
 - 16. The signal processing system of claim 15, further 35 comprising means for deriving at least one component waveform from the at least one optimized basis function waveform and the at least one physiological signal.
 - 17. The signal processing system of claim 16, further comprising means for calculating at least one physiological measurement responsive to the at least one basis function component waveform.
 - 18. The signal processing system of claim 15, wherein the basis function index comprises a basis function frequency.
 - 19. The signal processing system of claim 15, wherein the basis function waveform comprises a sinusoid waveform.
 - 20. The signal processing system of claim 15, wherein the physiological measurement comprises one or more of a group comprising oxygen saturation, blood glucose, carboxyhemoglobin, and blood pressure.

UNITED STATES PATENT AND TRADEMARK OFFICE CERTIFICATE OF CORRECTION

PATENT NO. : 7,377,899 B2 Page 1 of 1

APPLICATION NO. : 11/417914 DATED : May 27, 2008

INVENTOR(S) : Walter M. Weber et al.

It is certified that error appears in the above-identified patent and that said Letters Patent is hereby corrected as shown below:

In Title Page, Item 75, please delete "Dana Point, CA" and insert -- Laguna Hills, CA --, therefore.

In Column 4, Line 56, please delete "signals(t)" and insert -- signal s(t) --, therefore.

In Column 4, Line 60, please delete "signals(t)" and insert -- signal s(t) --, therefore.

In Column 6, Line 3, please delete "SAT $_{\it f}$ 462" and insert -- SAT $_{\it fo}$ 462 --, therefore.

In Column 6, Line 62, please delete " $r_{n,r}$ " and insert -- $r_{n,f}$ ---, therefore.

In Column 7, Lines 57-59, Equation 8b, please delete " = $\sqrt{\frac{(RD \cdot Y_p)^2}{|X_p|^2}}$ " and insert -- $\sqrt{\frac{(RD \cdot X_p)^2}{|X_p|^2}}$ --, therefore.

Signed and Sealed this

Sixteenth Day of December, 2008

JON W. DUDAS

Director of the United States Patent and Trademark Office



US007377899C1

(12) EX PARTE REEXAMINATION CERTIFICATE (9596th)

United States Patent

Weber et al.

(10) **Number:** US 7,377,899 C1

(45) Certificate Issued: Apr. 12, 2013

(54) SINE SATURATION TRANSFORM

(76) Inventors: **Walter M. Weber**, Laguna Hills, CA (US); **Ammar Al-Ali**, Tustin, CA (US);

Lorenzo Cazzoll, Barcelona (ES)

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Related U.S. Application Data

- (63) Continuation of application No. 11/048,232, filed on Feb. 1, 2005, now Pat. No. 7,373,194, which is a continuation of application No. 10/184,032, filed on Jun. 26, 2002, now Pat. No. 6,850,787.
- (60) Provisional application No. 60/302,438, filed on Jun. 29, 2001.

(51) **Int. Cl. A61B 5/00** (2006.01)

(52) U.S. Cl.

USPC 600/300; 600/316; 600/322; 600/323

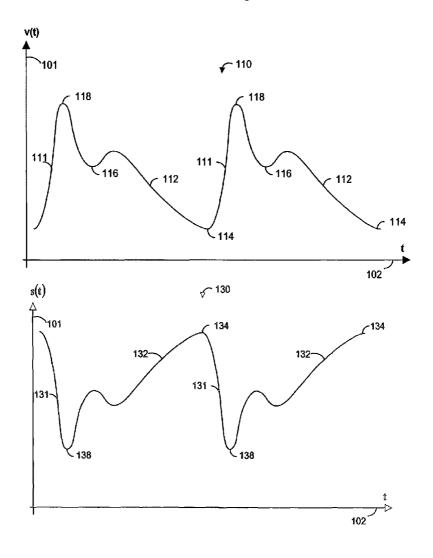
(56) References Cited

To view the complete listing of prior art documents cited during the proceeding for Reexamination Control Number 90/012,538, please refer to the USPTO's public Patent Application Information Retrieval (PAIR) system under the Display References tab.

Primary Examiner — Robert Nasser

(57) ABSTRACT

A transform for determining a physiological measurement is disclosed. The transform determines a basis function index from a physiological signal obtained through a physiological sensor. A basis function waveform is generated based on basis function index. The basis function waveform is then used to determine an optimized basis function waveform. The optimized basis function waveform is used to calculate a physiological measurement.



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EX PARTE REEXAMINATION CERTIFICATE ISSUED UNDER 35 U.S.C. 307

THE PATENT IS HEREBY AMENDED AS INDICATED BELOW.

Matter enclosed in heavy brackets [] appeared in the patent, but has been deleted and is no longer a part of the 10 patent; matter printed in italics indicates additions made to the patent.

AS A RESULT OF REEXAMINATION, IT HAS BEEN DETERMINED THAT:

Claims 2, 13 and 16 are cancelled.

Claims 1, 3, 4, 9, 12, 14, 15, 17 and 19 are determined to be patentable as amended.

Claims 5-8, 10-11, 18 and 20, dependent on an amended 20 claim, are determined to be patentable.

- 1. A signal processing system for processing at least one physiological signal obtained from at least one physiological sensor and computing at least one physiological measurement, the signal processing system comprising:
 - at least one input configured to accept at least one physiological signal obtained from at least one physiological sensor:
 - at least one basis function index calculator configured to determine at least one basis function index from the at 30 least one physiological signal;
 - at least one basis function waveform generator configured to generate at least one basis function waveform according to the at least one basis function index;
 - at least one optimization calculator configured to deter- 35 mine one or more optimized basis function waveforms; and
 - at least one basis function component waveform calculator configured to derive at least one component waveform from the at least one optimized basis function waveform 40 and the at least one physiological signal.
- 3. The signal processing system of claim [2] 1, further comprising at least one physiological measurement calculator configured to calculate at least one physiological measurement responsive to the at least one basis function component 45 waveform.
- 4. The signal processing system of claim [2] 1, wherein the at least one component waveform comprises two or more component waveforms.
- processing system for processing at least one physiological signal obtained from at least one physiological sensor and computing at least one physiological measurement, the signal processing system comprising:
 - at least one input configured to accept at least one physi- 55 ological signal obtained from at least one physiological
 - at least one basis function index calculator configured to determine at least one basis function index from the at least one physiological signal;
 - at least one basis function waveform generator configured to generate at least one basis function waveform according to the at least one basis function index; and

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- at least one optimization calculator configured to determine one or more optimized basis function waveforms, wherein the basis function waveform comprises a sinusoid
- 12. A signal processing method for processing, in the time domain, at least one physiological signal obtained from at least one physiological sensor and computing at least one physiological measurement, the signal processing method comprising:
 - obtaining at least one physiological signal from at least one physiological sensor;
 - determining at least one basis function index from the at least one physiological signal;
 - generating at least one basis function waveform according to the at least one basis function index;
 - determining at least one optimized basis function waveform; and
 - deriving at least one component waveform from the at least one optimized basis function waveform and at least one physiological signal.
- 14. The signal processing method of claim [13] 12, further comprising calculating at least one physiological measurement responsive to the at least one basis function component
- 15. A signal processing system for processing at least one physiological signal obtained from at least one physiological sensor and computing at least one physiological measurement, the signal processing system comprising:
 - at least one input configured to accept at least one physiological signal obtained from at least one physiological sensor;
 - means for determining at least one basis function index from the at least one physiological signal;
 - means for generating at least one basis function waveform according to the at least one basis function index;
 - means for determining one or more optimized basis function waveforms:
 - means for deriving at least one component waveform from the at least one optimized basis function waveform and the at least one physiological signal.
- 17. The signal processing system of claim [16] 15, further comprising means for calculating at least one physiological measurement responsive to the at least one basis function component waveform.
- 19. [The signal processing system of claim 15] A signal processing system for processing at least one physiological signal obtained from at least one physiological sensor and 9. [The signal processing system of claim 1] A signal 50 computing at least one physiological measurement, the signal processing system comprising:
 - at least one input configured to accept at least one physiological signal obtained from at least one physiological sensor;
 - means for determining at least one basis function index from the at least one physiological signal, wherein the basis function waveform comprises a sinusoid waveform:
 - means for generating at least one basis function waveform according to the at least one basis function index; and means for determining one or more optimized basis function waveforms.



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公开(公告)号	US7377899	公开(公告)日	2008-05-27		
申请号	US11/417914	申请日	2006-05-03		
[标]申请(专利权)人(译)	梅西莫股份有限公司				
申请(专利权)人(译)	Masimo公司				
当前申请(专利权)人(译)	Masimo公司				
[标]发明人	WEBER WALTER M AL ALI AMMAR CAZZOLL LORENZO				
发明人	WEBER, WALTER M. AL-ALI, AMMAR CAZZOLL, LORENZO				
IPC分类号	A61B5/00 A61B5/02 A61B5/145 A61B5/1455				
CPC分类号	A61B5/024 A61B5/14551 A61B5/1455 A61B5/14532 A61B5/02433 A61B5/14552 A61B5/7282				
优先权	60/302438 2001-06-29 US				
其他公开文献	US20060270921A1				
外部链接	Espacenet USPTO				

摘要(译)

公开了一种用于确定生理测量的变换。变换根据通过生理传感器获得的 生理信号确定基函数指数。基于基函数索引生成基函数波形。然后使用 基函数波形来确定优化的基函数波形。优化的基函数波形用于计算生理 测量。

