

FIG. 1

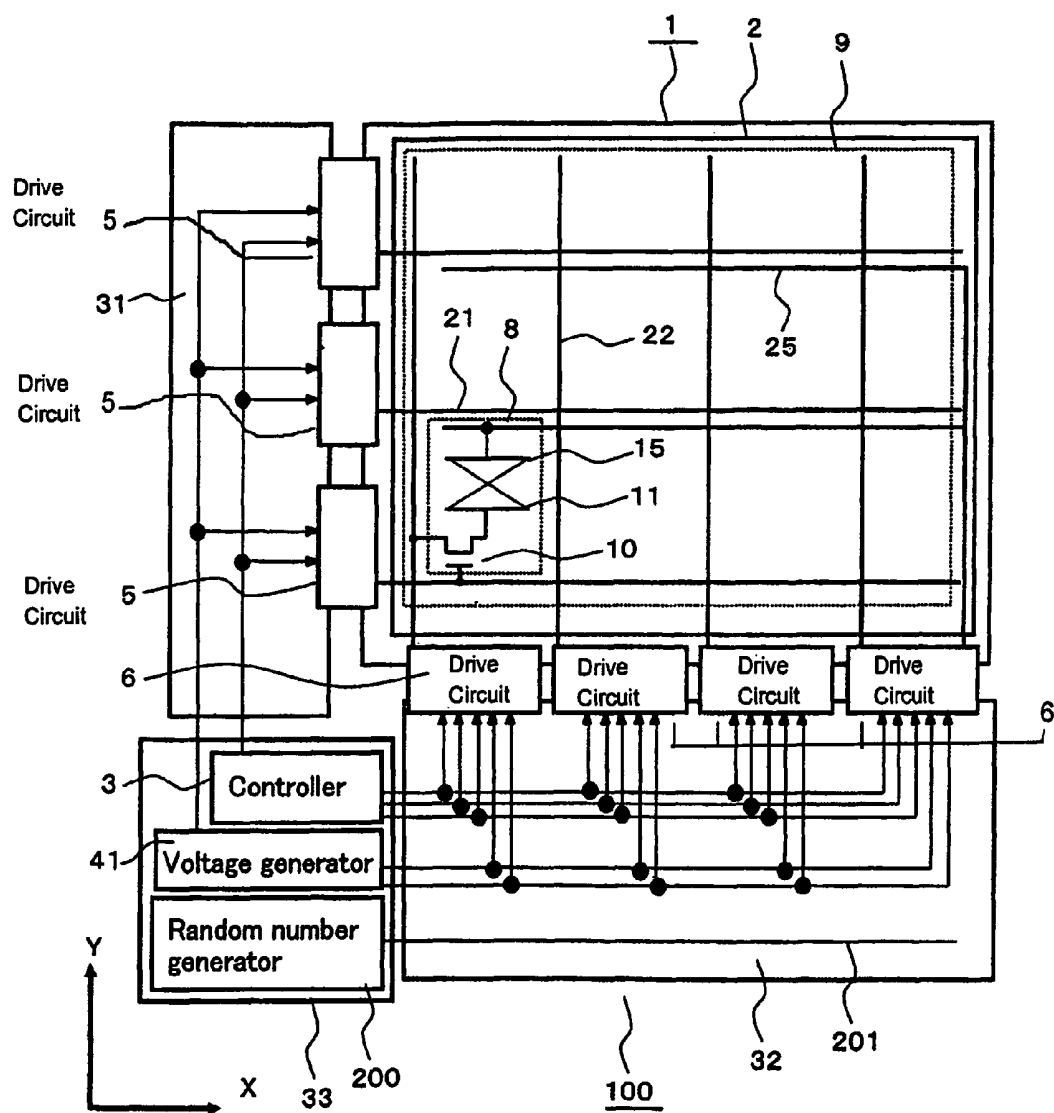


FIG. 2

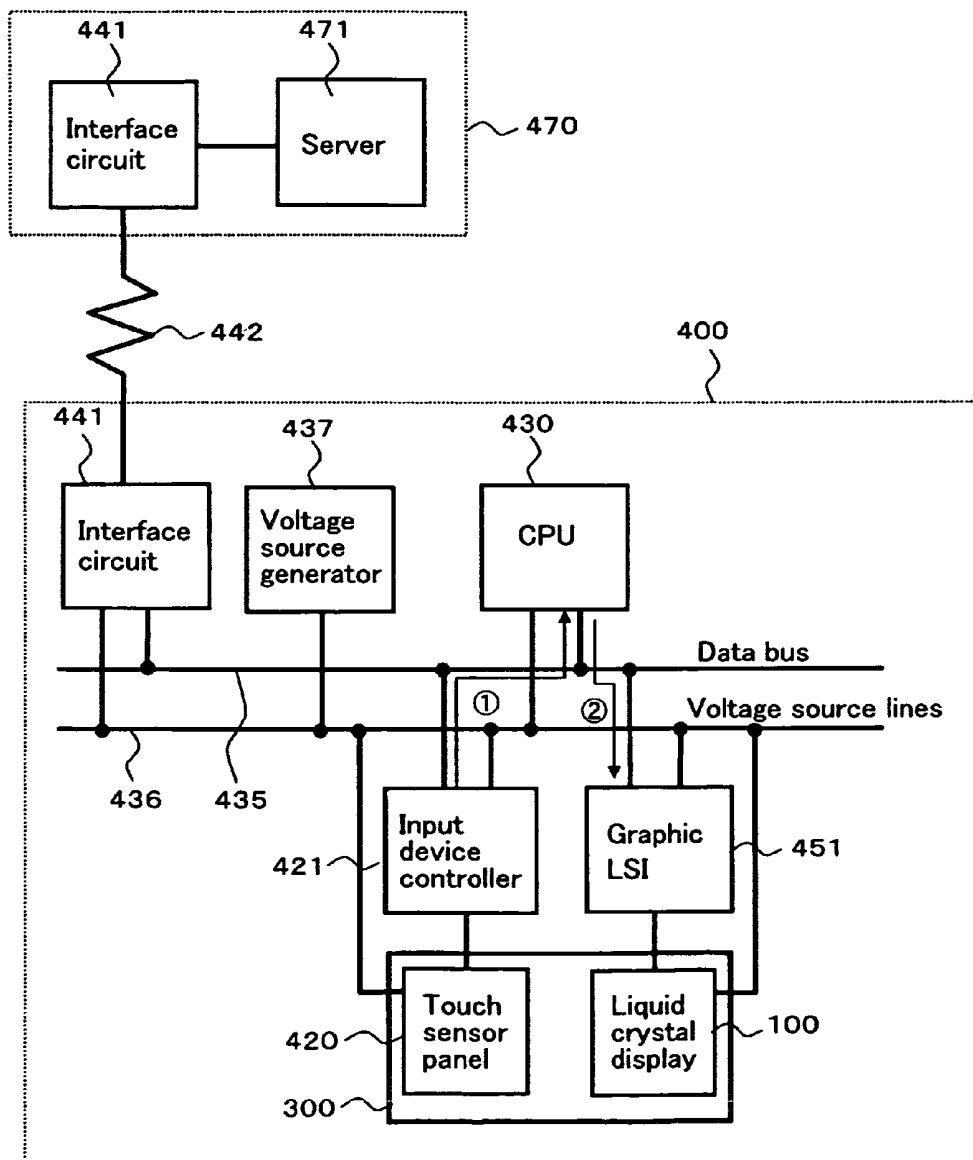


FIG. 3

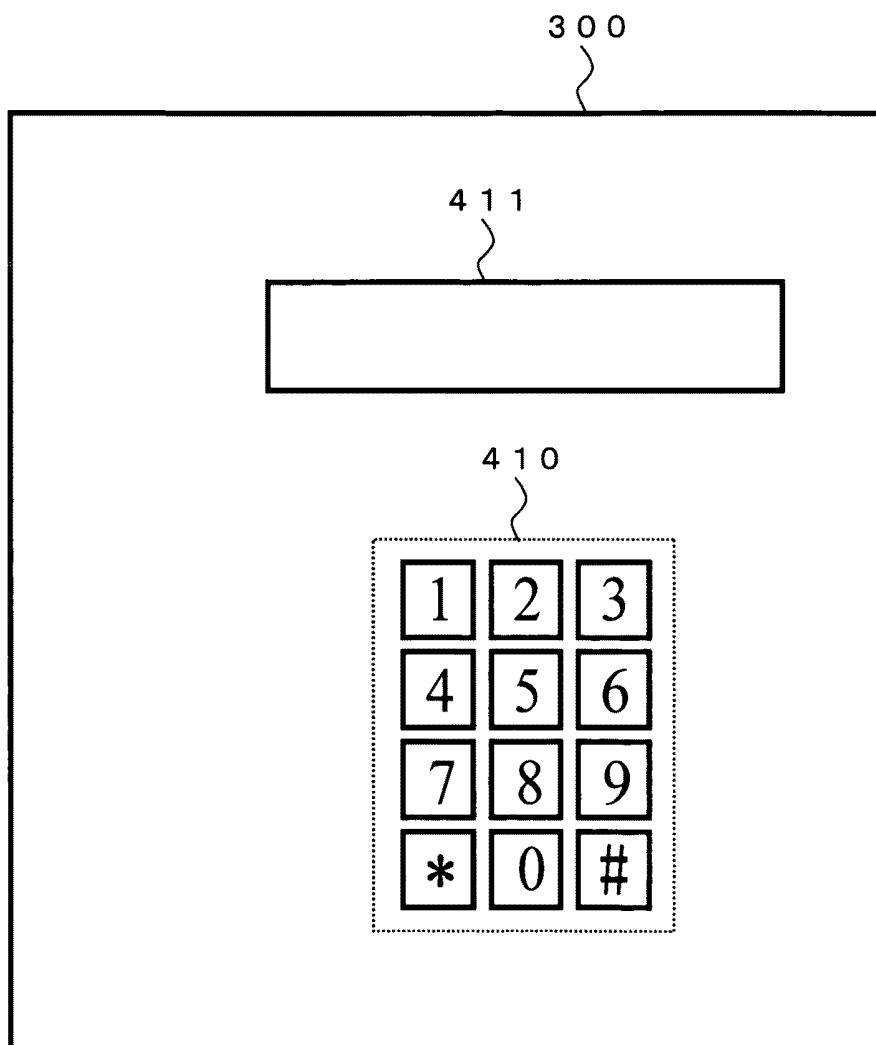


FIG. 4

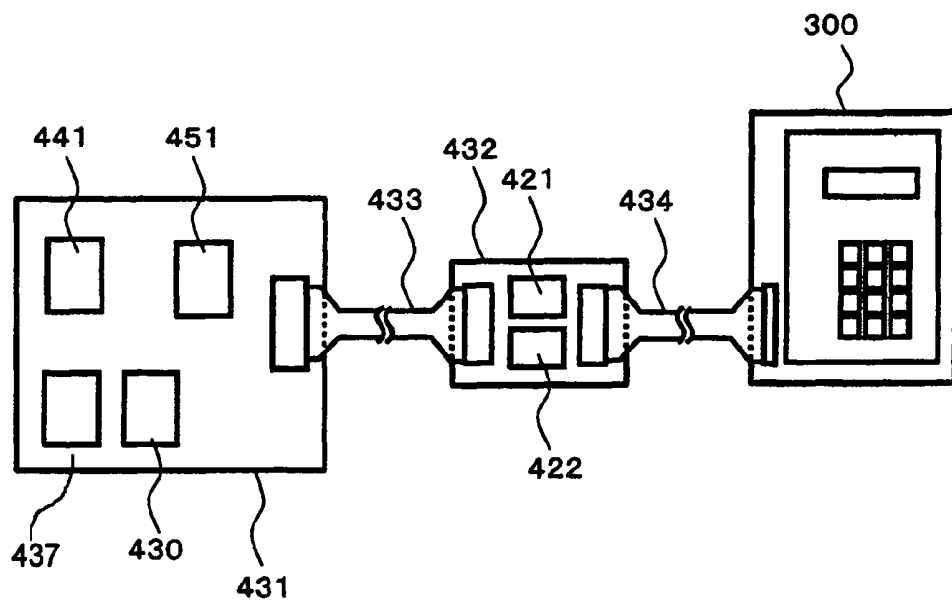


FIG. 5

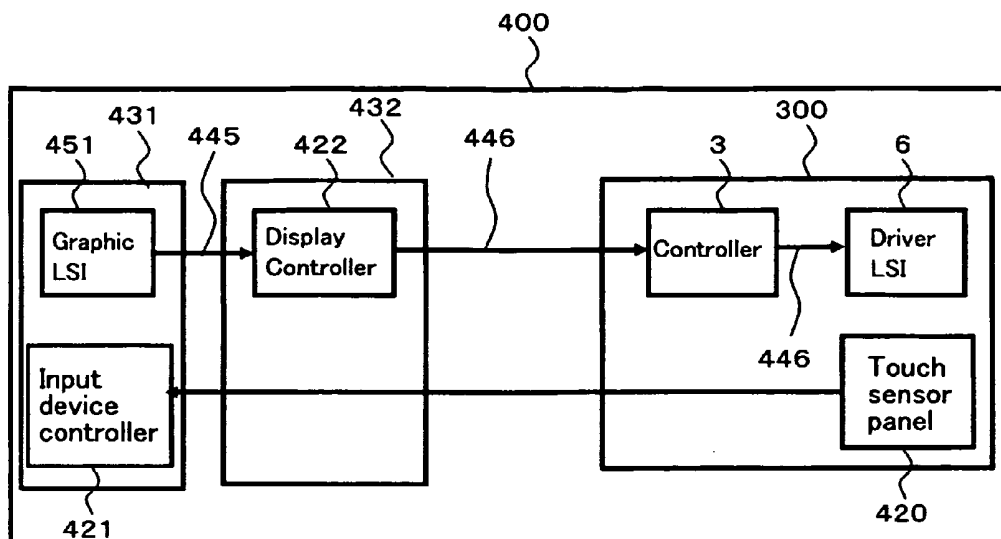


FIG. 6

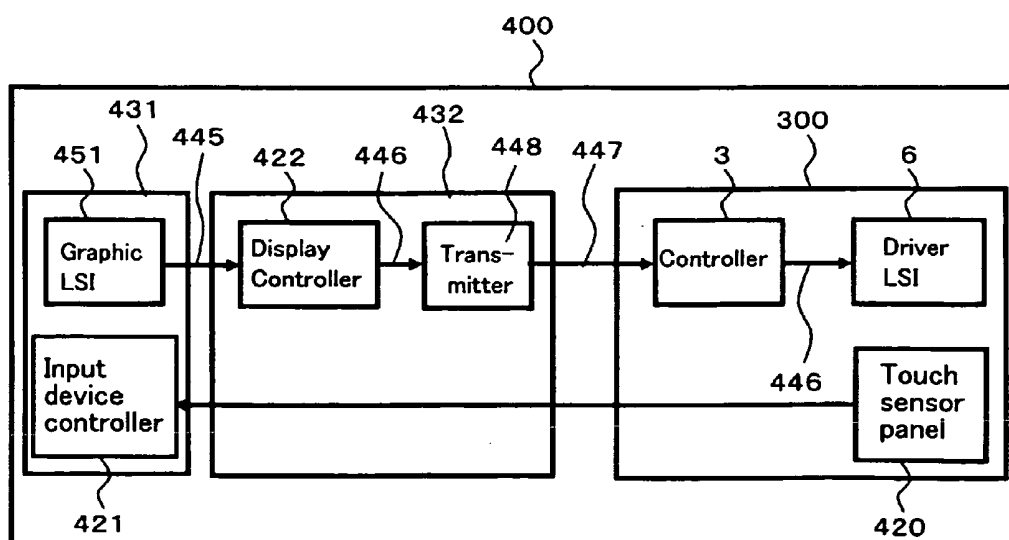


FIG. 7

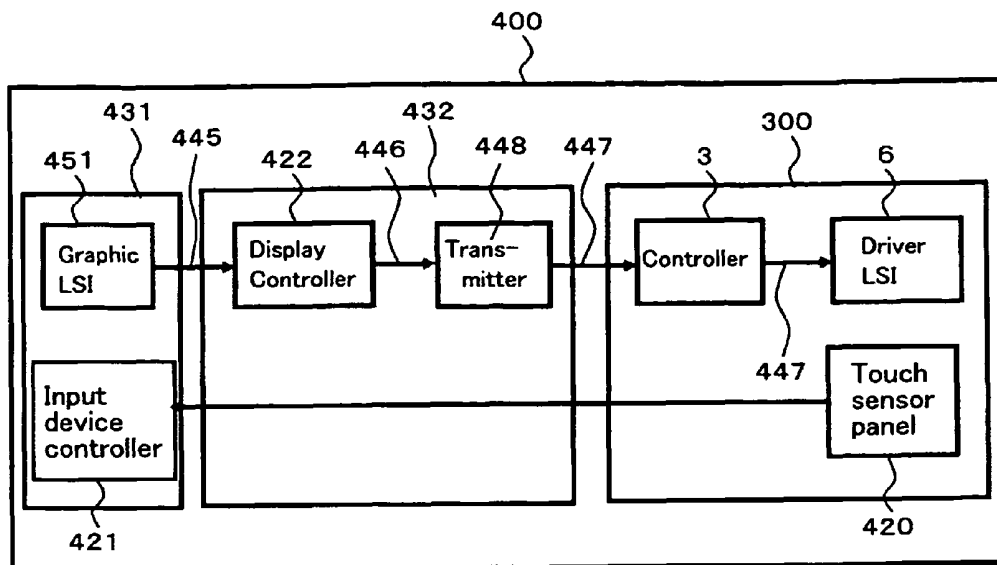


FIG. 8

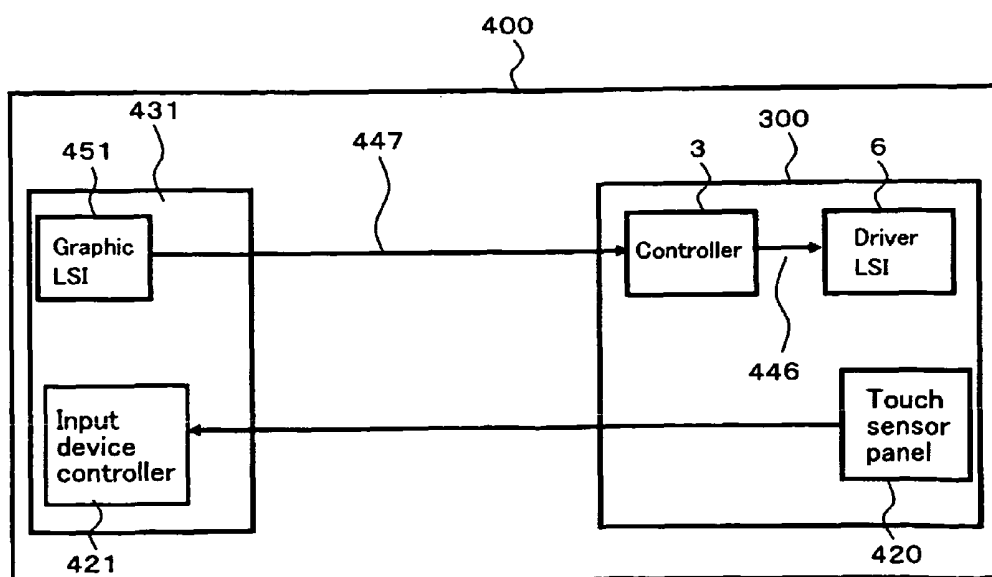


FIG. 9

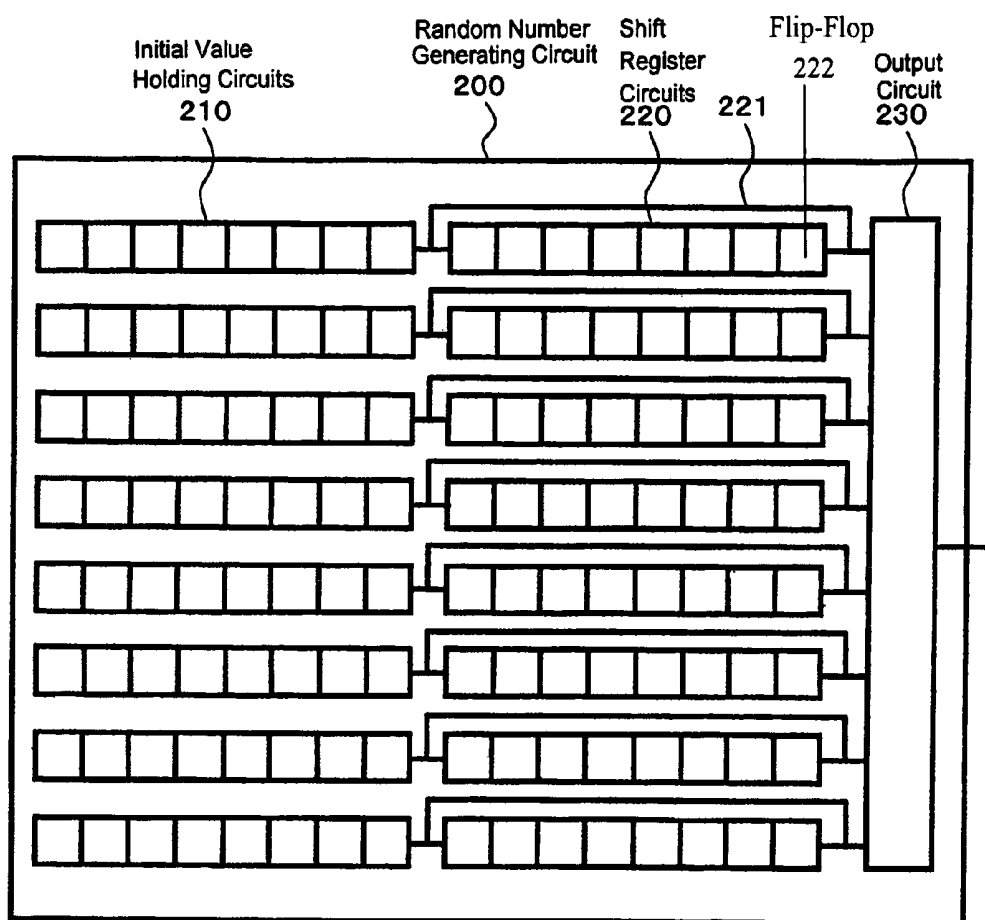


FIG. 10

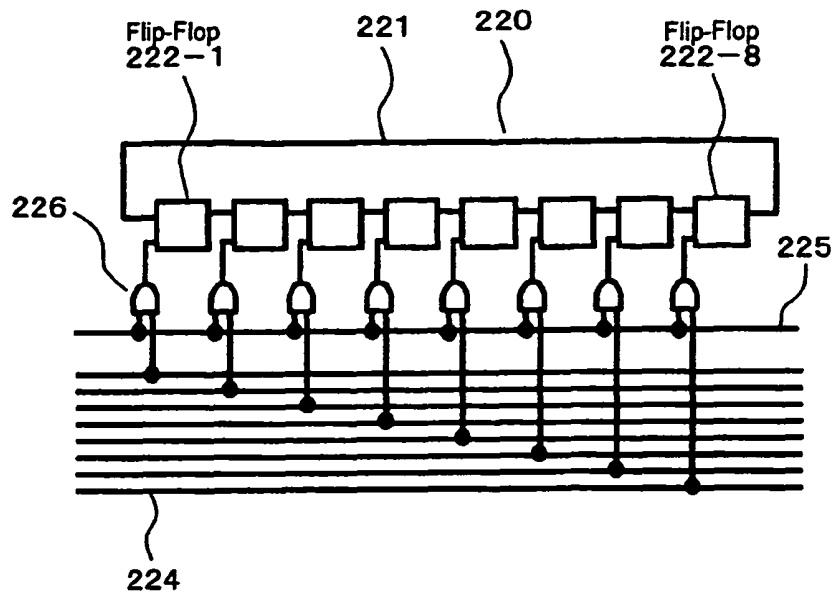


FIG. 11

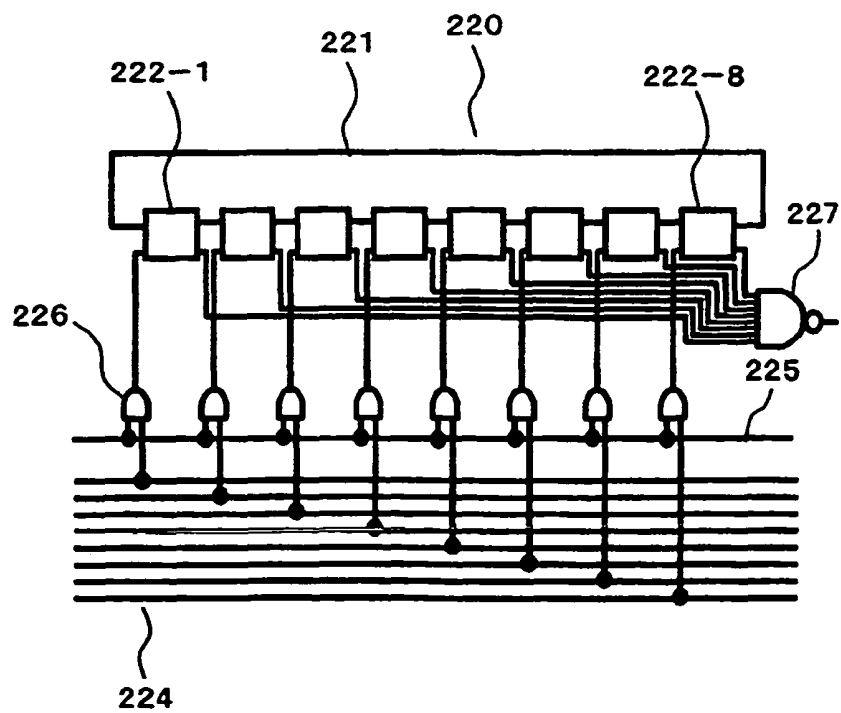


FIG. 12A

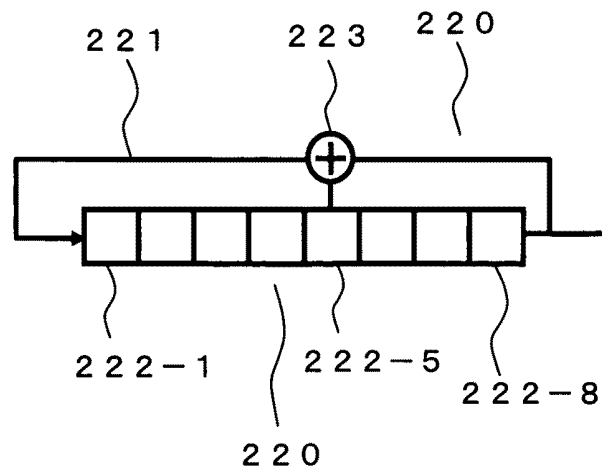
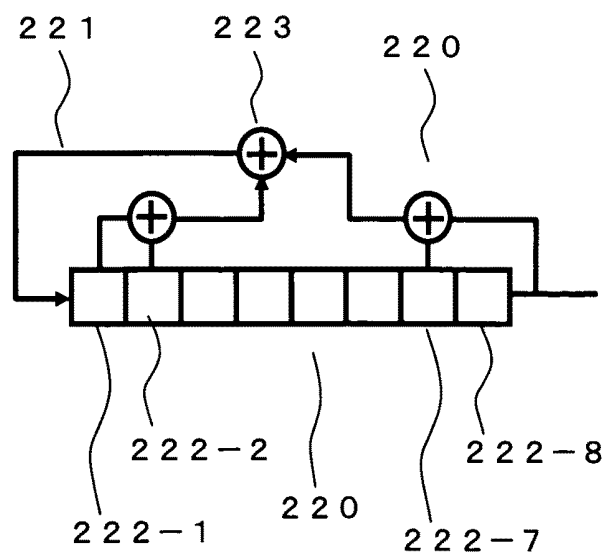


FIG. 12B



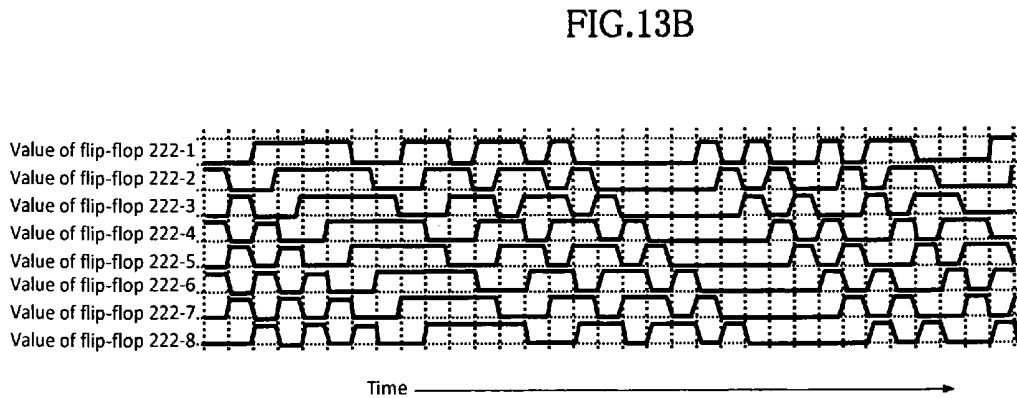
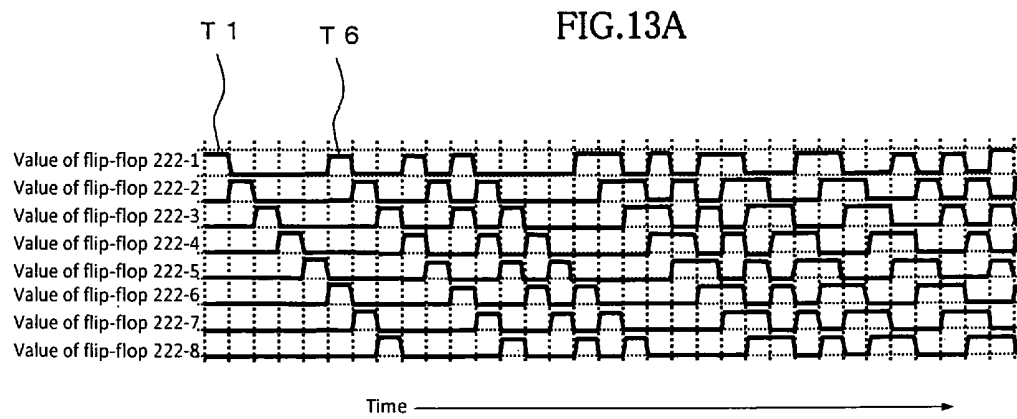


FIG. 14

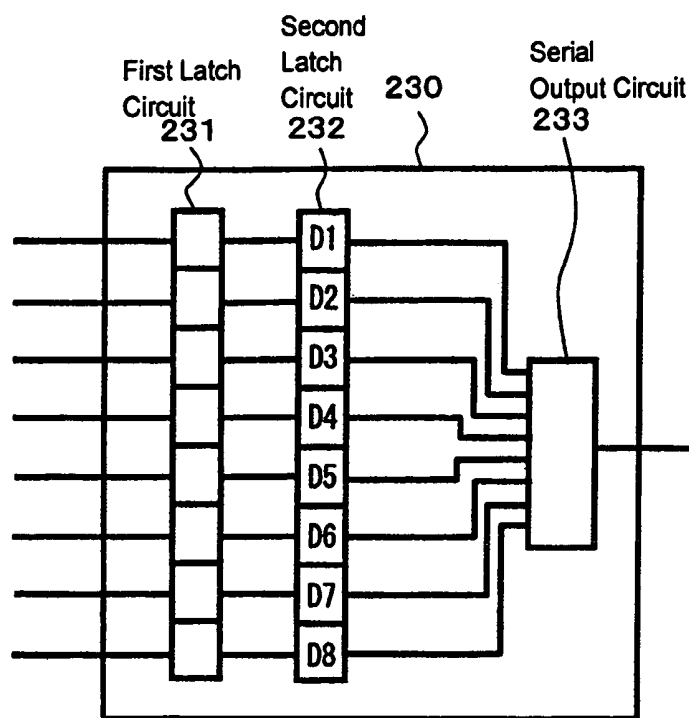


FIG. 15A

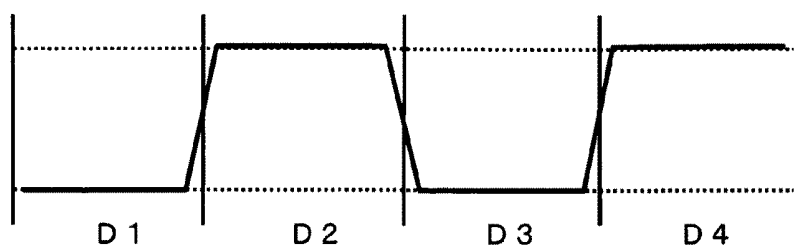


FIG. 15B

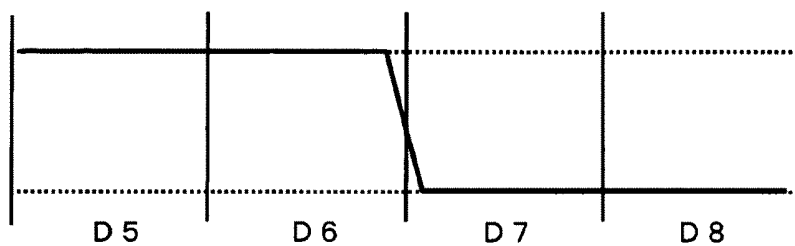


FIG. 15C

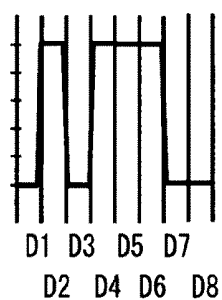
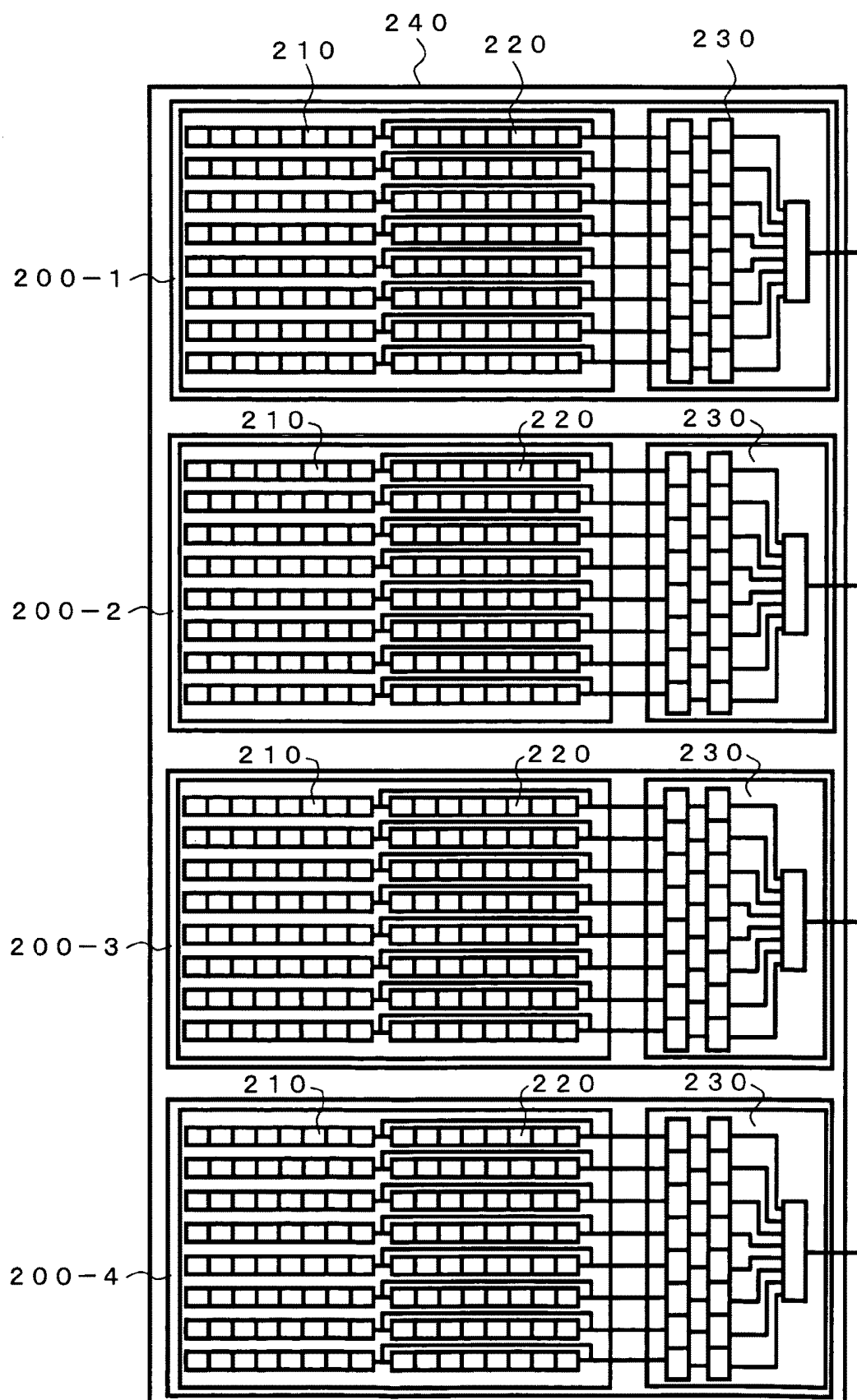


FIG. 16



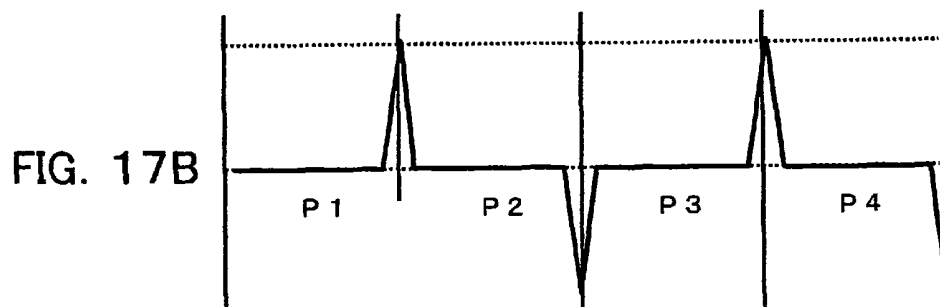
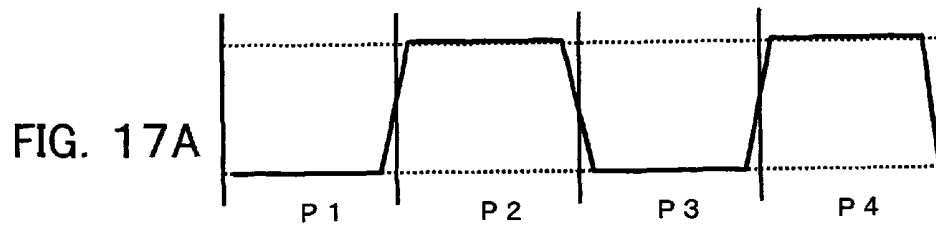


FIG.18

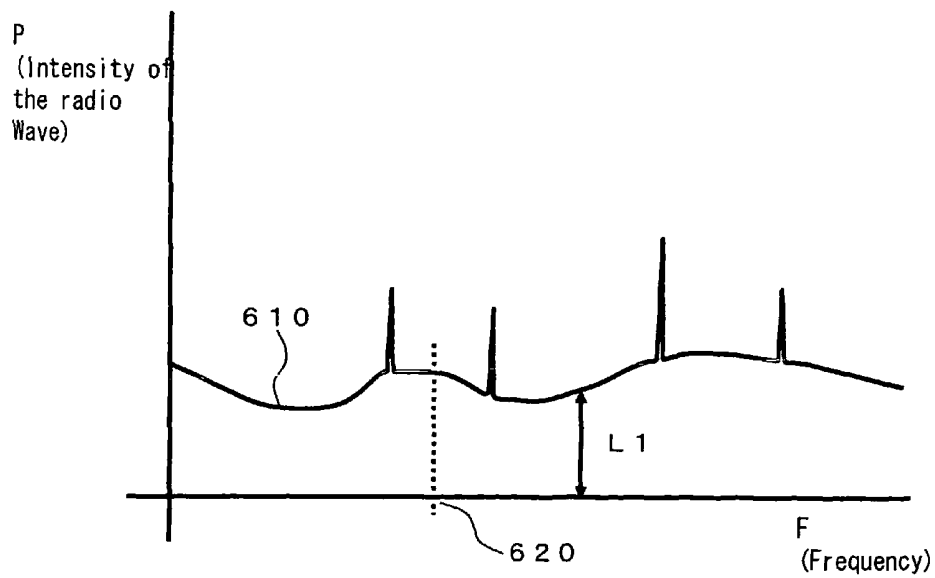


FIG. 19

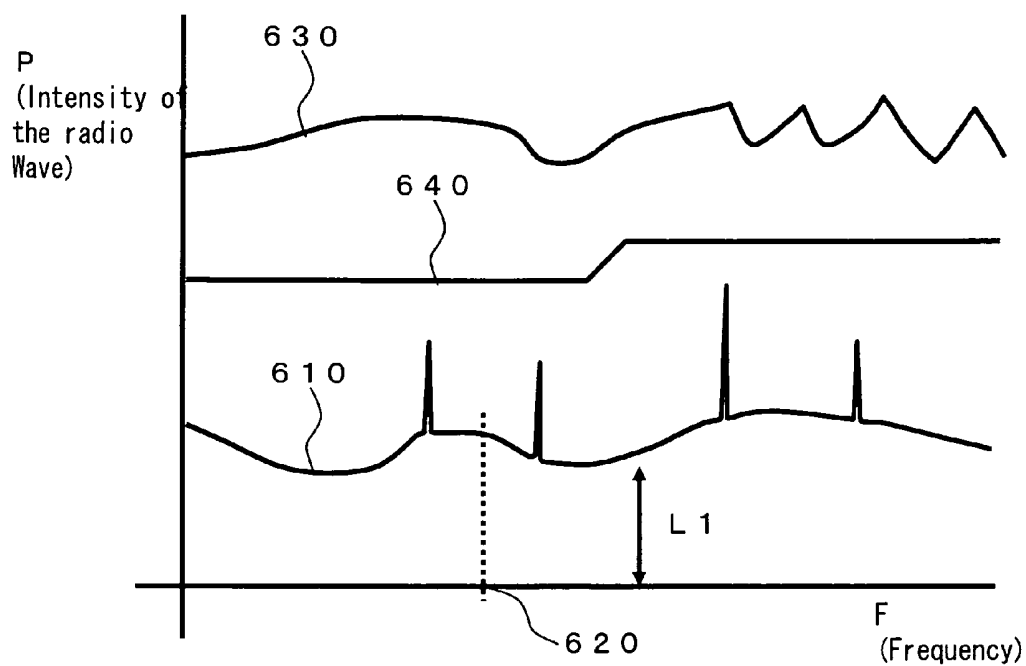


FIG. 20

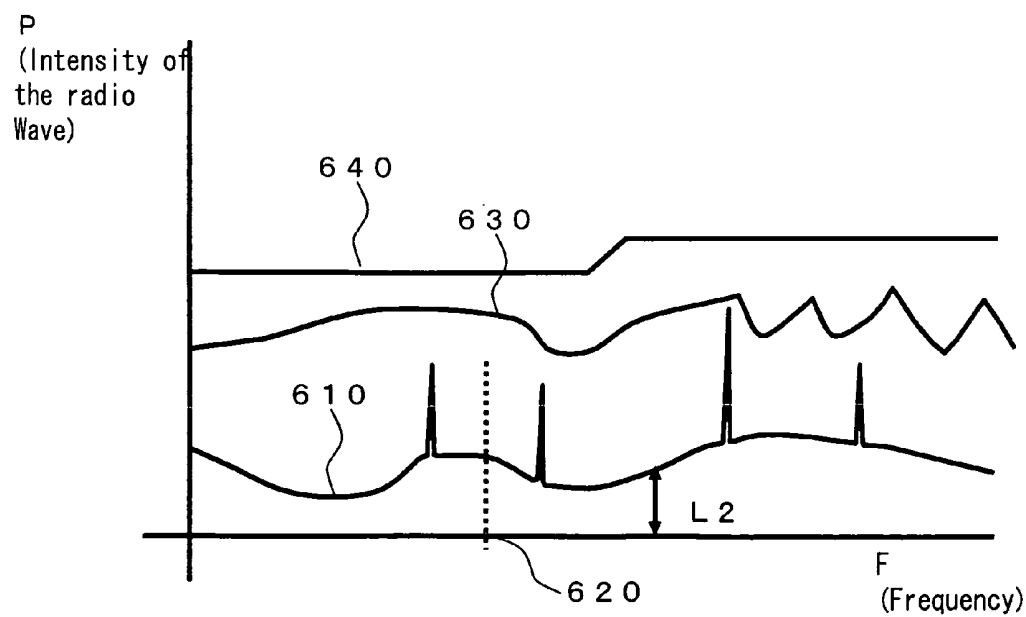


FIG. 21

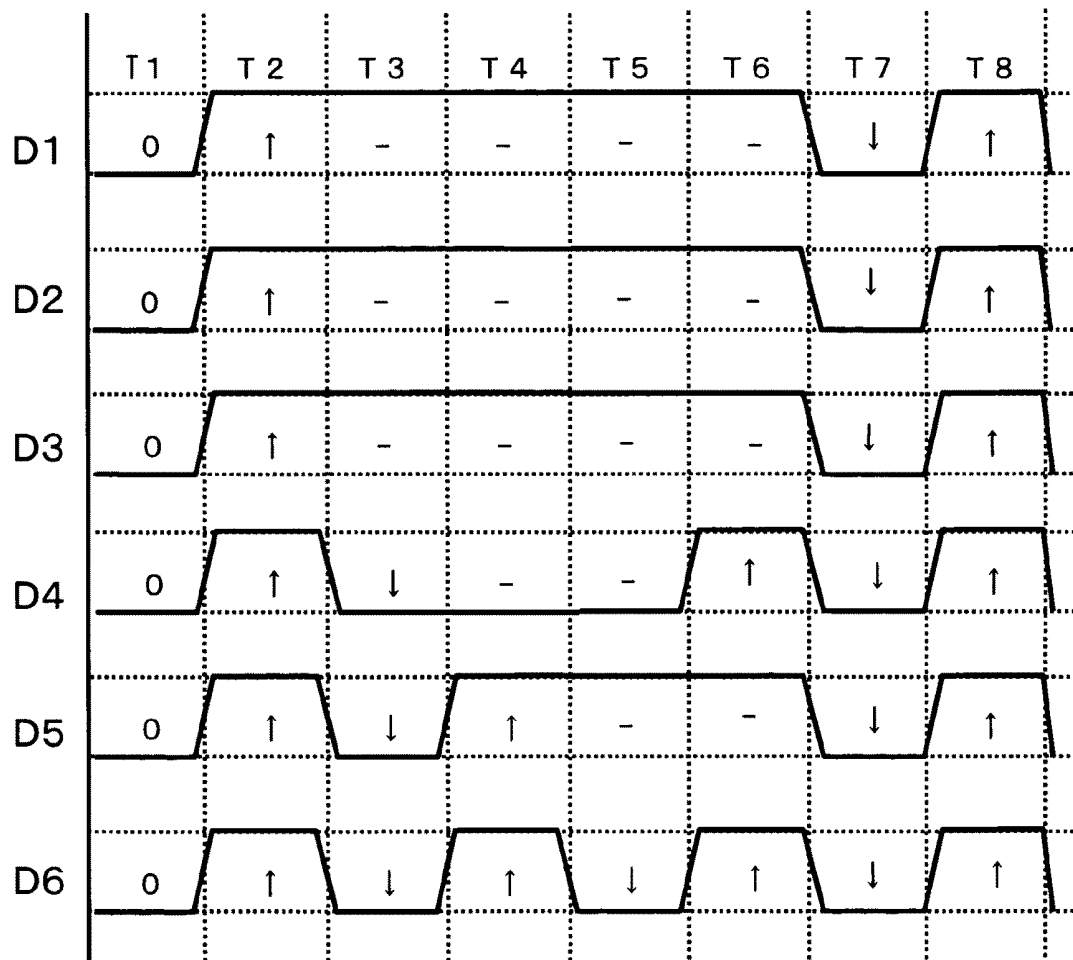
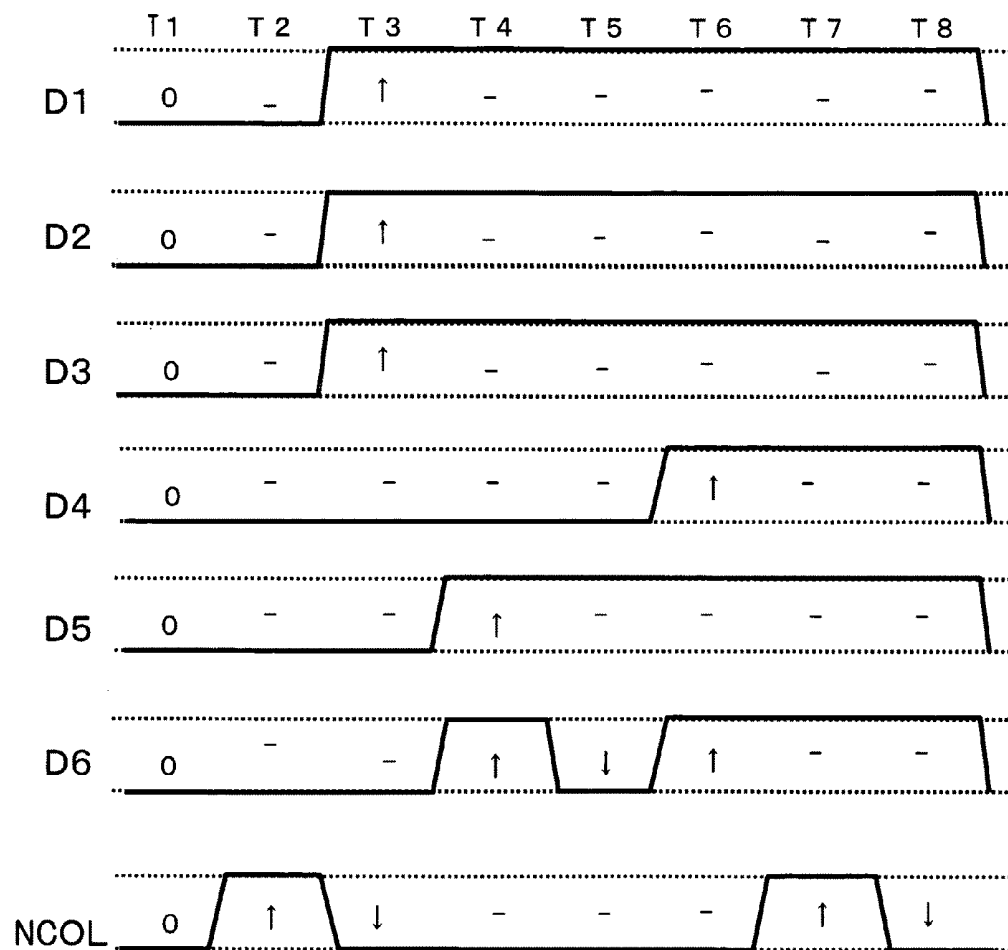


FIG. 22



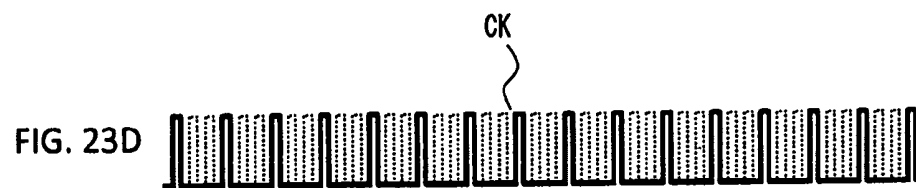
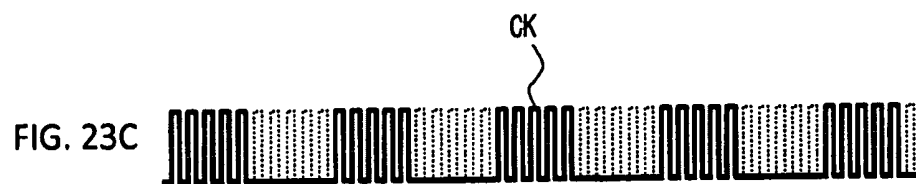
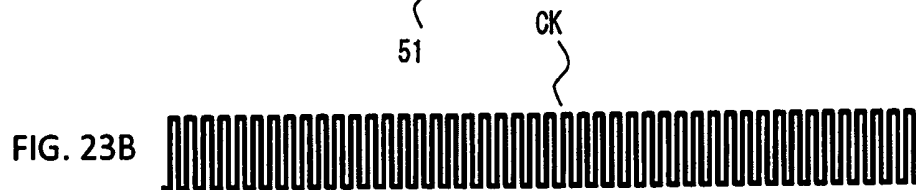
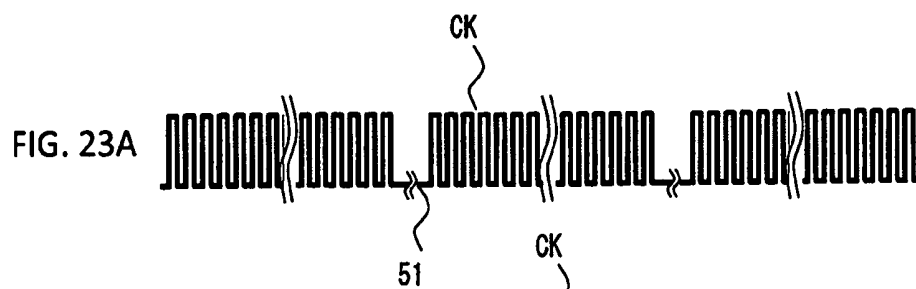
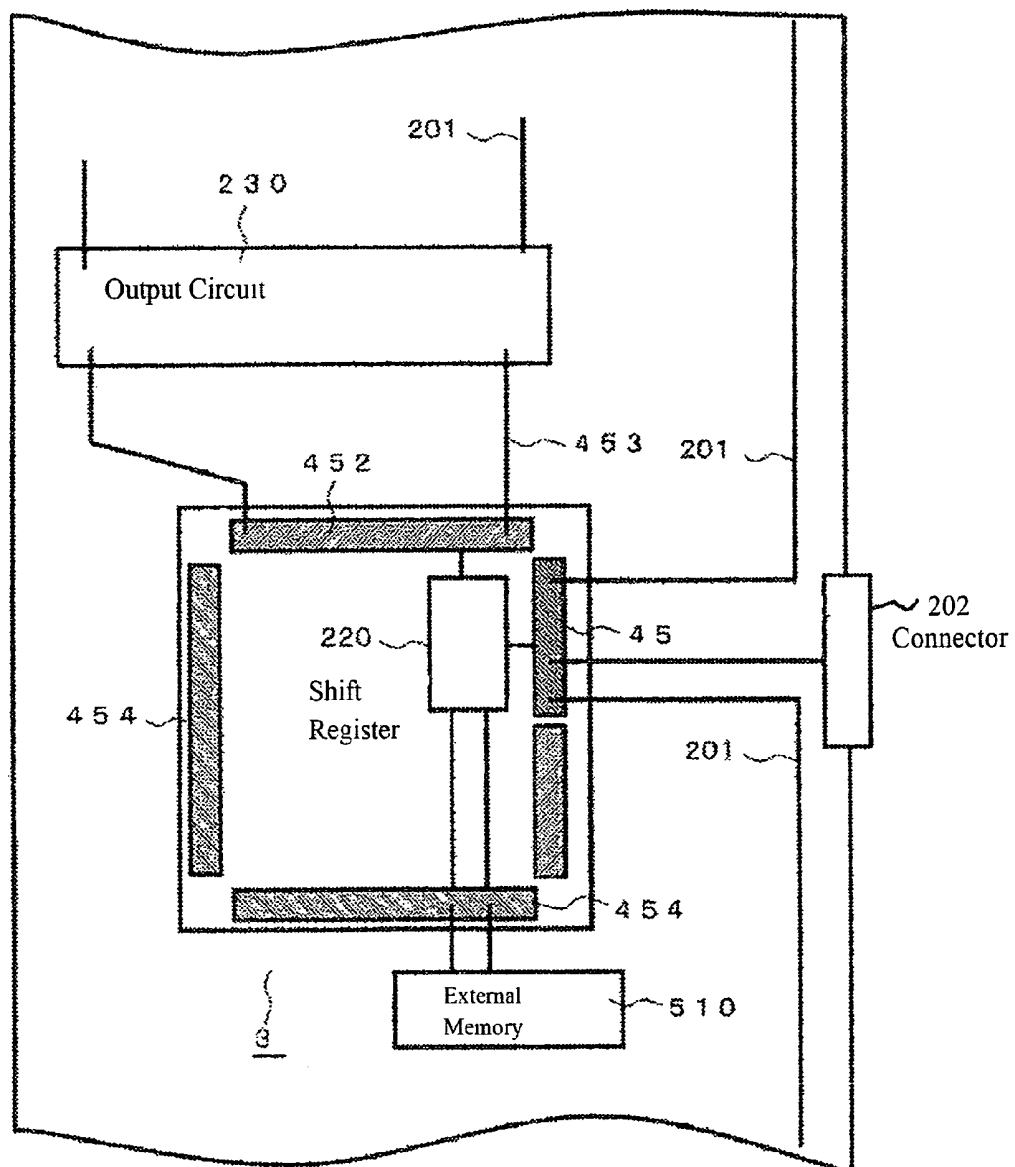


FIG. 24



1 DISPLAY DEVICE

BACKGROUND OF THE INVENTION

1. Field of the Invention

The present invention relates to a display device, and more particularly to a technique which is effectively applicable to a random number generating circuit used in a display device.

2. Description of the Related Art

A TFT (thin film transistor)-type liquid crystal display device having a middle-size liquid crystal display panel has been popularly used as a display part of a personal digital assistant, a notebook-type personal computer or the like.

With respect to this middle-size liquid crystal display device, there has been known a liquid crystal display device which is provided with a touch panel arranged adjacent to the liquid crystal display device as an input device for receiving data inputted from the outside. Further, a keyboard has been also used as an external input device conventionally.

Data inputted from such an input device may be provided with coding such as encryption for securing secrecy. Random numbers are generally used for coding and hence, a display device is also required to include a random number generating circuit.

However, a conventional display device is not provided with a random number generating circuit, and attempts have been made to find out an optimum random number generating device for a general-use display device.

On the other hand, conventionally, there have been proposed various random number generating circuits. JP-A-06-051957 discloses a circuit which generates random numbers referred to as M series by a shift register. However, JP-A-06-051957 neither discloses nor suggests a random number generating circuit used by a display device.

SUMMARY OF THE INVENTION

When a random number generating circuit having high randomness is used in the above-mentioned display device, there arises a drawback that a circuit scale is increased and hence, it is difficult to prevent the elevation of a material cost or to ensure a region where the random number generating circuit is mounted.

Further, a display device is also required to satisfy a demand that a voltage change of a signal outputted from a circuit is suppressed to cope with electromagnetic wave noises (EMI).

The generation of random numbers by the random number generating circuit has a characteristic that randomness is high and hence, there also arises a drawback that a voltage change occurs frequently thus increasing electromagnetic wave noises.

The present invention has been made to overcome the above-mentioned drawbacks, and it is an object of the present invention to provide a technique which can suppress a circuit scale and, at the same time, can cope with electromagnetic wave noises while using an optimum random number generating circuit in a display device.

The above-mentioned and other objects and novel features of the present invention will become apparent from the description of this specification and attached drawings.

To simply explain the summary of typical inventions among inventions disclosed in this specification, they are as follows.

According to one aspect of the present invention, there is provided a display device which includes a display panel, a drive circuit which drives respective pixels of the display

2

panel, and a control circuit which outputs display data to the drive circuit, wherein the control circuit includes a random number generating circuit. The random number generating circuit includes n pieces of shift registers, and a basic clock for the shift registers is outputted from the control circuit as an intermittent drive clock which intermittently repeats stopping thereof. Output signals of n pieces of shift registers are inputted to an output circuit. The output circuit generates an output signal having a frequency n times as large as a frequency of the output signals of n pieces of shift registers from these shift registers. The basic clock of the shift registers may be a clock which is in synchronism with a transmission clock of the display data or may be a clock which is in asynchronism with the transmission clock of the display data. Further, to transmit a digital signal, an inversion control signal which inverts the digital signal based on the number of changes of a signal voltage may be used.

To simply explain advantageous effects acquired by the typical inventions among the inventions disclosed in this specification, they are as follows.

According to the present invention, by providing n pieces of shift registers, the randomness of the random numbers generated by the random number generating circuit is enhanced. Further, with the use of n pieces of shift registers, it is possible to generate the signal having the frequency n times as large as the frequency of the output signal of the shift registers as the random numbers.

By adopting the intermittent drive clock which intermittently repeats stopping thereof as the basic clock of the shift register, the generation of the random number can be stopped intermittently thus suppressing the generation of electromagnetic wave noises attributed to the generation of random numbers.

By allowing the transmission of the digital signal to use the inversion control signal which inverts the digital signal based on the number of changes of the signal voltage, it is possible to suppress the generation of electromagnetic wave noises by suppressing the change of the signal voltage.

BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 is a block diagram showing the schematic constitution of a liquid crystal display device of an embodiment according to the present invention;

FIG. 2 is a block diagram showing the schematic constitution of a terminal device to which the liquid crystal display device of the embodiment according to the present invention is applied;

FIG. 3 is a front view showing a schematic profile of a display/input part of the terminal device to which the liquid crystal display device of the embodiment according to the present invention is applied;

FIG. 4 is a block diagram showing the schematic mounting constitution of the terminal device to which the liquid crystal display device of the embodiment according to the present invention is applied;

FIG. 5 is a block diagram showing a transmission path of a signal of the terminal device to which the liquid crystal display device of the embodiment according to the present invention is applied;

FIG. 6 is a block diagram showing a transmission path of a signal of the terminal device to which the liquid crystal display device of the embodiment according to the present invention is applied;

FIG. 7 is a block diagram showing a transmission path of a signal of the terminal device to which the liquid crystal display device of the embodiment according to the present invention is applied;

FIG. 8 is a block diagram showing a transmission path of a signal of the terminal device to which the liquid crystal display device of the embodiment according to the present invention is applied;

FIG. 9 is a block diagram showing a random number generating circuit used in the liquid crystal display device of the embodiment according to the present invention;

FIG. 10 is a schematic circuit diagram showing the random number generating circuit used in the liquid crystal display device of the embodiment according to the present invention;

FIG. 11 is a schematic circuit diagram showing the random number generating circuit used in the liquid crystal display device of the embodiment according to the present invention;

FIG. 12A and FIG. 12B are schematic circuit diagrams showing the random number generating circuit used in the liquid crystal display device of the embodiment according to the present invention;

FIG. 13A and FIG. 13B are timing charts showing an output waveform of the random number generating circuit used in the liquid crystal display device of the embodiment according to the present invention;

FIG. 14 is a schematic circuit diagram showing an output circuit of the random number generating circuit used in the liquid crystal display device of the embodiment according to the present invention;

FIG. 15A to FIG. 15C are timing charts for explaining an output waveform of the output circuit of the random number generating circuit used in the liquid crystal display device of the embodiment according to the present invention;

FIG. 16 is a schematic block diagram showing the random number generating circuit used in the liquid crystal display device of the embodiment according to the present invention;

FIG. 17A and FIG. 17B are waveforms showing an outputted radio waves of the random number generating circuit used in the liquid crystal display device of the embodiment according to the present invention;

FIG. 18 is a frequency characteristic diagram showing a relationship between frequency and intensity of the radio waves generated from the liquid crystal display device of the embodiment according to the present invention;

FIG. 19 is a frequency characteristic diagram showing the relationship between frequency and intensity of the radio waves generated from the liquid crystal display device of the embodiment according to the present invention before taking a countermeasure;

FIG. 20 is a frequency characteristic diagram showing a relationship between frequency and intensity of the radio waves generated from the liquid crystal display device of the embodiment according to the present invention;

FIG. 21 is a schematic waveform showing a waveform of a digital signal transmitted in the liquid crystal display device of the embodiment according to the present invention;

FIG. 22 is a schematic waveform showing a waveform of a digital signal transmitted in the liquid crystal display device of the embodiment according to the present invention;

FIG. 23A to FIG. 23D are schematic view showing a waveform of a display data transmission clock and a transmission clock of shift register generated in the liquid crystal display device of the embodiment according to the present invention; and

FIG. 24 is a schematic profile showing the arrangement of terminals of a control circuit of the liquid crystal display device of the embodiment according to the present invention.

DETAILED DESCRIPTION OF THE EMBODIMENT

Hereinafter, embodiments of the present invention are explained in detail in conjunction with drawings.

Here, in all drawings for explaining the embodiments, parts having identical functions are given same numerals and their repeated explanation is omitted.

FIG. 1 is a block diagram showing the schematic constitution of a liquid crystal display device of the embodiment according to the present invention. As shown in FIG. 1, the liquid crystal display device is constituted of a liquid crystal display panel 1, a scanning-signal-line-side printed circuit board 31, a video-signal-line-side printed circuit board 32 and a control circuit board 33.

The liquid crystal display panel 1 is configured as follows. A TFT substrate 2 on which a plurality of thin film transistors 10, a plurality of pixel electrodes 11, a plurality of counter electrodes 15 and the like are formed and a filter substrate (not shown in the drawing) on which a plurality of color filters and the like are formed so as to overlap with each other with a predetermined gap therebetween. Both substrates are adhered to each other using a frame-shaped sealing material (not shown in the drawing) arranged between both substrates and in the vicinity of peripheral portions of both substrates and, at the same time, liquid crystal composition is filled and sealed in a space defined by both substrates and the sealing material. Further, a polarizer is adhered to outer surfaces of both substrates.

Here, the embodiment of the present invention is applicable to both of a so-called IPS-method type liquid crystal display panel in which the counter electrodes 15 are arranged on the TFT substrate 2 and a so-called vertical-electric-field method type liquid crystal display panel in which the counter electrodes 15 are arranged on the filter substrate in the same manner.

In FIG. 1, scanning signal lines (also referred to as gate signal lines) 21 which extend in the x direction and are arranged parallel to each other in the y direction in the drawing, and video signal lines (also referred to as drain signal lines) 22 which extend in the y direction and are arranged parallel to each other in the x direction in the drawing are arranged on the TFT substrate 2, and a pixel portion 8 is formed in a region surrounded by the scanning signal lines 21 and the drain signal lines 22.

Here, although the liquid crystal display panel 1 includes a large number of pixel portions 8 in a matrix array, for facilitating the understanding of the drawing, only one pixel portion 8 is shown in FIG. 1. The pixel portions 8 arranged in a matrix array form a display region 9, the respective pixel portions 8 play a role of pixels of a display image, and an image is displayed in the display region 9.

The thin film transistor 10 of each pixel portion 8 has a source thereof connected to the pixel electrode 11, has a drain thereof connected to the video signal line 22, and has a gate thereof connected to the scanning signal line 21. The thin film transistor 10 functions as a switch for supplying a display voltage (gray scale voltage) to the pixel electrode 11. Here, although naming of "source" and "drain" may be reversed based on the relationship of biases, in this embodiment, the electrode which is connected to the video signal line 22 is referred to as the drain.

The video signal lines 22 are connected with drive circuits 6, and video signals are supplied from the drive circuits 6. Further, in FIG. 1, numeral 25 indicates a counter electrode (common electrode), and a common voltage is supplied to the counter electrode 25 from the drive circuit 6 in FIG. 1. The

scanning signal lines **21** are connected to the drive circuits **5**, and scanning signals are supplied from the drive circuits **5**.

The drive circuits **5** are connected with a scanning-signal-line-side printed circuit board **31**, control signals are supplied to the drive circuits **5** from the control circuit **3** via the scanning-signal-line-side printed circuit board **31**, and a power source voltage is supplied from a power source circuit **41**. The drive circuits **6** are connected with the video-signal-line-side printed circuit board **32**, and control signals are supplied to the drive circuits **6** from the control circuit **3** via the video-signal-line-side printed circuit board **32**, and a power source voltage is supplied from a power source circuit **41**.

The control circuit **3** and the power source circuit **41** are mounted on the control printed circuit board **33**. A random number generating circuit **200** is mounted on the control printed circuit board **33**, and the random numbers are generated from the random number generating circuit **200**. A transmission line **201** is connected to the random number generating circuit **200** so as to allow the transmission of signals from the random number generating circuit **200** to the video-signal-line-side printed circuit board **32** and the like through the transmission line **201**. The detail of the random number generating circuit **200** is described later.

Further, a display data line **46** and a display data transmission clock line **50** extend to the video-signal-line-side printed circuit board **32** from the control circuit **3**, and the display data line **46** and the display data transmission clock line **50** are connected to the drive circuits **6**, and display data is supplied to the drive circuits **6** in synchronism with a display data transmission clock via the display data lines **46**.

Further, the display data transmission clock or a basic clock generated by the control circuit **3** is supplied to the random number generating circuit **200** from the control circuit **3**. The detail of the basic clock is described later.

An inversion control signal line **203** also extends from the control circuit **3**, and transmits an inversion control signal indicative of whether or not the display data is inverted to the respective drive circuits **6**. The detail of the inversion control signal will be also described later.

Next, a terminal device **400** to which a liquid crystal display device **100** is applied is explained in conjunction with FIG. **2**. The liquid crystal display device **100** is used as a display part of the terminal device **400**. A touch sensor panel **420** is arranged adjacent to the liquid crystal display device **100**, and the liquid crystal display device **100** and the touch sensor panel **420** form a display/input part **300**.

FIG. **3** is a schematic front view of the display/input part **300**. The display/input part **300** shows a touch input part **410** to a user of the terminal device **400**, and displays inputted information on a display part **411**.

The user touches the touch input part **410** with his/her finger or the like so as to input data based on information displayed at a contact position to the terminal device **400**. The inputted data is processed by the terminal device **400** and is also displayed on the display part **411** for confirmation.

In case of data having high secrecy, processing such as encryption is applied to the inputted data. Further, actual data is not displayed on the display part **411** and, for example, asterisk or the like is displayed.

Returning to FIG. **2**, the flow of data processing is explained. Position data inputted from the touch sensor panel **420** is firstly transmitted to an input device control circuit **421** where the position data is subject to data processing. Then, the position data is transmitted to a CPU **430** via a data bus **435**.

In CPU **430**, processing necessary for the input data is selected and executed. With respect to the liquid crystal dis-

play device **100**, a display content is selected based on the input data, and the display content is transmitted to an image integrating circuit **451** via the data bus **435**.

Processing necessary for an image display is performed in the image integrating circuit **451**, and display data is outputted to the liquid crystal display device **100**. Here, when encryption or the like is applied to the display data, a random number generating circuit **200** becomes necessary also on a liquid-crystal-display-device-**100** side.

The terminal device **400** particularly includes an external communication circuit **441**. Due to the transaction of data with a central processing device **470** at a remote site by making use of a communication line **442** such as a LAN, the terminal device **400** requires encryption or the like as inevitable processing. The central processing device **470** is constituted of the external communication circuit **441** and a server **471**, for example. Here, numeral **437** indicates a power source circuit of the terminal device **400**, and numeral **436** indicates a power source line.

Next, the display/input part **300** is explained in conjunction with FIG. **3**. The display/input part **300** is configured such that the liquid crystal display device **100** and the touch sensor panel **420** overlap with each other on a plane. FIG. **3** shows the display/input part **300** as viewed from a user, and a distinct boundary between the liquid crystal display device **100** and the touch sensor panel **420** is not shown.

The display/input part **300** displays the touch input part **410** for allowing the user to input data using the liquid crystal display device **100**. The user can input data by touching the touch input part **410** with his/her finger or the like. The touch sensor panel **420** transmits the position information acquired by touching of the user on the touch sensor panel **420** to the terminal device **400** via the input device control circuit **421**.

The terminal device **400** determines the input data based on the position information from the touch sensor panel **420**. That is, in the display/input part **300**, the position information has the significant meaning and a display image also includes a function of an input device. Here, when the user touches the touch input part **410**, there may be a case that a display in a region which indicates the input data is changed to allow the confirmation of an input operation. For example, when the user touches the inside of a square which surrounds the number **2** of the touch input part **410**, a display of the inside of the square which surrounds the number **2** is inverted or background color is changed.

Next, FIG. **4** shows a mounting mode of the display/input part **300**. The display/input part **300** is mounted apart from a main substrate **431** of the terminal device **400** so as to allow a user to face the display/input part **300**. Further, the display/input part **300** may be also mounted such that an input/output substrate **432** is provided separate from the display/input part **300** and the main substrate **431** and the input/output substrate **432** are connected with each other by a cable **433**, and the input/output substrate **432** and the display/input part **300** are connected with each other by a cable **434**.

The input device control circuit **421** and the display control device **422** are mounted on the input/output substrate **432**. The display control device **422** is provided for imparting the general-use property to the liquid crystal display device **100**. For example, the liquid crystal display device **100** is compatible to not only analog signal inputting but also digital signal inputting due to the display control device **422**.

Further, digital signals differ from each other in accordance with standards and hence, the display control device **422** performs processing for transmitting the display data which the image integrating circuit **451** outputs to the control circuit **3** of the liquid crystal display device **100**. Further, the display

control device **422** may also output a low-voltage differential signal or the like for the liquid crystal display device **100**.

Signals which are transmitted between the main substrate **431** and the display/input part **300** are explained in conjunction with FIG. 5 to FIG. 8. Although the transmission of data is performed via the data bus **435** in the main substrate **431**, a signal **445** which conforms to a display data standard is outputted to the display control device **422** from the image integrating circuit **451**.

The signal **445** outputted from the image integrating circuit **451** is, in general, a digital signal which oscillates at 3V to 5V and also is a parallel signal of 8 to 32 bits. Accordingly, there may arise a drawback that noises are generated by the signal **445** and hence, it is necessary to design wiring of the signal with care.

Next, a digital signal **446** is outputted to the control circuit **3** of the liquid crystal display device **100** from the display control device **422**. The digital signal **446** is also a parallel signal of 8 to 64 bits which oscillates at 3V to 5V. Accordingly, there may arise a drawback that noises are generated by the signal **446** and hence, it is necessary to design wiring of the signal with care.

Further, the digital signal **446** is also outputted to the drive circuit **6** from the control circuit **3**. The digital signal **446** which is outputted to the drive circuit **6** from the control circuit **3** is transmitted via the display data line **46** which is formed on the video-signal-line-side printed circuit board **32** as shown in FIG. 1.

FIG. 6 shows a case in which a low-voltage differential signal **447** is outputted from the input/output substrate **432**. An output from the display control device **422** is converted into the low-voltage differential signal **447** by a transmitter **448**. The low-voltage differential signal **447** is effective to cope with the EMI and hence, noises generated by the low-voltage differential signal **447** are smaller than noises generated by the digital signal **446** or the like. However, depending on a case, it is necessary to take a countermeasure against noises generated by the low-voltage differential signal **447**.

Also in the case shown in FIG. 6, the digital signal **446** is transmitted to the transmitter **448** from the display control device **422** and hence, it is necessary to take a countermeasure against noises between the display control device **422** and the transmitter **448**.

Next, FIG. 7 shows a case in which the low-voltage differential signal **447** is outputted from the input/output substrate **432**, and the low-voltage differential signal **447** is also outputted from the control circuit **3**. By outputting the low-voltage differential signal **447** from the control circuit **3**, it is possible to effectively take a counter measure against the EMI on the video-signal-line-side printed circuit board **32** or the scanning-signal-line-side printed circuit board **31**.

The signal processing is performed using the digital signal in the control circuit **3** and hence, the low-voltage differential signal **447** is once converted into the digital signal in the control circuit **3** and, at the time of being outputted from the control circuit **3**, the digital signal is converted into a low-voltage differential signal **447** for the video-signal-line-side printed circuit board **32** and the scanning-signal-line-side printed circuit board **31**.

FIG. 8 shows a case in which the image integrating circuit **451** has a function of the display control device **422**. In FIG. 8, the low-voltage differential signal **447** is directly outputted to the control circuit **3** from the image integrating circuit **451**. In this case, the generation of noises is suppressed between the image integrating circuit **451** and the control circuit **3**. However, when the digital signal **446** is outputted to the drive circuit **6** or the drive circuit **5** from the control circuit **3**, it is

necessary to take a countermeasure against noises generated in a path which follows the control circuit **3**.

Next, the random number generating circuit **200** is explained in conjunction with FIG. 9. Numeral **210** indicates a register which holds an initial value, numeral **220** indicates a shift register, and numeral **230** indicates an output circuit. The shift register **220** is constituted of plural stages of flip-flops **222**.

Further, the shift register **220** generates random numbers by returning a value of the final-stage flip-flop **222** to a value of the initial-stage flip-flop **222** by a feedback path **221**. The shift register **220**, as described later, takes out a value from the flip-flop **222** of a middle stage, and takes out an exclusive OR of the value of the middle-stage flip-flop **222** and a value of the feedback path **221**, and returns a calculation result to the initial-stage flip-flop **222** thus generating the random number.

Although the shift register **220** may generate the random number in a single form, in FIG. 9, eight shift registers **220-1** to **220-8** are provided in parallel to each other, outputs of the respective shift registers **220** are inputted to the output circuit **230**, and the random number is outputted from the output circuit **230**.

Further, for setting different initial values to eight respective shift registers **220-1** to **220-8**, eight registers **210** are prepared, and the different initial values are stored in the respective registers **210**. By allowing the respective shift registers **220** to read the different initial values from the registers **210** at the time of starting operations of the respective shift registers **220**, it is possible to generate random numbers of different patterns.

Further, to enhance the randomness of the random numbers to be generated, the frequency of the basic clock may differ among the respective shift registers **220-1** to **220-8**.

Further, in the respective shift registers **220-1** to **220-8**, an output stage of the flip-flop **222** which inputs a value thereof to an exclusive-OR circuit **223** may be changed thus forming a different feedback path **221** whereby the randomness of the random number which the random number generating circuit **200** generate may be enhanced.

Further, the randomness of the random number may be also enhanced by changing the number of stages of the flip-flop **222** in the respective shift registers **220-1** to **220-8**. Further, as described later, to reduce electromagnetic wave noises attributed to the generation of the random number, as the basic clocks to be inputted to the respective shift registers **220-1** to **220-8**, intermittent drive clocks which repeat intermittent stopping thereof may be also used.

Next, FIG. 10 shows a transmission method of the initial value. The initial value transmitted by an initial value transmission line **224** is stored in each flip-flop **222** via an AND circuit **226** when an initial value transmission control signal line **225** assumes a high level. In FIG. 10, the flip-flop **222-1** constitutes an initial-stage flip-flop, and the flip-flop **222-8** constitutes a final-stage flip-flop.

FIG. 11 shows a transmission method of the initial value in which a determination circuit **227** is added. The shift register **220** cannot generate the random number when the initial values of all flip-flops **222** are set to 0 and hence, it is determined whether or not the initial values of all flip-flops **222** are set to 0 using the determination circuit **227**.

Next, the feedback path **221** of the shift register **220** is explained in conjunction with FIG. 12A and FIG. 12B. In FIG. 12A, an output of the final-stage flip-flop **222-8** and an output of the fifth-stage flip-flop **222-5** are calculated by an exclusive-OR circuit **223**, and a result of the calculation is inputted to the initial-stage flip-flop circuit **222-1** thus generating the random number.

FIG. 13A shows a pattern which the random numbers generate when the initial values (1, 0, 0, 0, 0, 0, 0, 0) are set. The value "1" is held in the initial-stage flip-flop 222-1 during a period T1, and the value "1" held in the flip-flop 222-1 is transmitted to the next-stage flip-flop in synchronism with the basic clock.

When the value "1" is held in the fifth-stage flip-flop 222-5, an exclusive OR is calculated between the value "1" held by the fifth-stage flip-flop 222-5 and the value "0" held by the final-stage flip-flop 222-8, a result of the calculation "1" is held in the initial-stage flip-flop 222-1, and values (1, 0, 0, 0, 0, 1, 0, 0) are held in the shift register 220 in a period T6. Hereinafter, other remaining values are transmitted between the flip-flops 222 in synchronism with the basic clock in the same manner as the above.

By forming the random number generating circuit 200 using n-stages of shift registers 220 as described above, it is possible to generate random numbers referred to as M series at a cycle of (n powers of 2-1).

FIG. 12B shows the shift register 220 which acquires a calculation result of exclusive OR of the initial-stage flip-flop 222-1 and the second-stage flip-flop 222-2 and a calculation result of exclusive OR of the final-stage flip-flop 222-8 and the flip-flop 222-7 which comes one stage before the final-stage flip-flop 222-8 and, then, inputs a result acquired by calculating an exclusive OR of these calculation results to the initial stage. In this manner, by forming a feedback path 221 which differs from the random number generating circuit 200 which uses the n-stage shift registers 220, it is possible to provide the random number generating circuit 200 which generates a different pattern.

Next, the initial values are explained again in conjunction with FIG. 13A and FIG. 13B. FIG. 13A and FIG. 13B show patterns of random numbers which the same shift register 220 generates. FIG. 13A shows the pattern of random numbers with the initial values (1, 0, 0, 0, 0, 0, 0, 0), and FIG. 13B shows the pattern of random numbers with the initial values (0, 1, 0, 1, 0, 1, 0, 0).

As can be clearly understood from FIG. 13A and FIG. 13B, even when the same shift register 220 is adopted, the patterns of the random number differ from each other when the initial values are different. In the random number generating circuit 200 shown in FIG. 9, the random numbers are generated using the plurality of shift registers 220 in which the different initial values are set thus enhancing the randomness of the random numbers generated by the random number generating circuit 200.

Next, drawbacks which arise when the random number generating circuit 200 is formed using the plurality of shift registers 220 and a method of solving such drawbacks are explained in conjunction with FIG. 14 and FIG. 15.

In FIG. 14, an output of the shift register 220 is temporarily held by a first latch circuit 231. Thereafter, data is held in a second latch circuit 232, and the data is read from the second latch circuit 232 at a frequency 8 times as large as a frequency of the basic clock of the shift register 220, and the data is outputted from a serial output circuit 233 as serial signals.

As shown in FIG. 15A and FIG. 15B, data D1 to D8 held in the second latch circuit 232 have frequency thereof increased 8 times as shown in FIG. 15C and the frequency-increased data is outputted from the serial output circuit 233.

Although the randomness is enhanced when the random numbers are generated using n pieces of shift registers 220, such an operation generates n pieces of different random numbers. In generating a plurality of random numbers, when the number of random numbers which exceeds the number of

random numbers which are actually used, the random numbers which exceeds the number of actually used random numbers become wasteful.

Accordingly, by providing the output circuit 230 so as to output serial signals having n-times frequency, it is possible to make use of the outputted random numbers without a waste. Here, the output of the output circuit 230 is not limited to digital signals, and low-voltage differential signals or the like may be outputted. When the low-voltage differential signals are outputted from the output circuit 230, the transmission line 201 becomes a pair line. Further, the transmission line 201 may be formed of a loop-shaped line in place of the open-ended line shown in FIG. 1.

Next, FIG. 16 shows a set of random number generating circuits 240 which is constituted by arranging 4 pieces of random number generating circuits 200 in parallel, wherein 8 pieces of shift registers 220 are inputted to one piece of output circuit 230. The set of random number generating circuits 240 shown in FIG. 16 can enhance randomness of random numbers thus enabling the further increase of number of the plurality of acquirable random numbers.

The respective random number generating circuits 200-1, 200-2, 200-3, 200-4 differ from each other in the frequency of the basic clock of the shift register 220 thus further enhancing the randomness of random numbers generated by the respective random number generating circuits 200-1, 200-2, 200-3, 200-4. In this case, the respective random number generating circuits 200-1, 200-2, 200-3, 200-4 differ from each other also in the frequency of a serial signal outputted from the output circuit 230.

Further, the respective random number generating circuits 200-1, 200-2, 200-3, 200-4 differ from each other in an output stage of the flip-flop 222 from which a value of the flip-flop 222 is inputted to an exclusive-OR circuit 223 thus forming a different feedback path 221 whereby the randomness of the random numbers generated by the random number generating circuit 200 can be further enhanced.

Further, the randomness of the random numbers can be enhanced also by changing the number of stages of the flip-flop 222 among the respective random number generating circuits 200-1, 200-2, 200-3, 200-4. In this case, the frequency of the serial signal outputted from the output circuit 230 also differs among the respective random number generating circuits 200-1, 200-2, 200-3, 200-4.

Here, to reduce electromagnetic wave noises attributed to the generation of the random numbers, the basic clocks inputted to the respective shift registers 200-1 to 200-4 may be formed of an intermittent drive clock which intermittently repeats stopping thereof.

Next, FIG. 17A shows an output waveform of the output circuit 230. For example, studied is a case in which a voltage of a signal outputted from the output circuit 230 is continuously changed such that the output circuit 230 outputs the signal of low level during a period P1, outputs the signal of high level during a period P2, and outputs the signal of low level during a period P3.

At a boundary between the period P1 and the period P2, the voltage is sharply elevated and hence, an radio waves having waveforms shown in FIG. 17B is generated as electromagnetic wave noises. On the other hand, also at a boundary between the period P2 and the period P3, the voltage is sharply decreased thus generating such an radio waves.

Further, in the same manner as the output circuit 230, an radio waves is generated also by the digital signal 446 which is transmitted between the image integrating circuit 451 and the control circuit 3 described previously. FIG. 18 shows one example of a measurement result of a frequency F and inten-

11

sity P of an radio waves **610** generated from the liquid crystal display device **100** due to the digital signal **446** or the like. In the drawing, symbol L1 indicates average intensity of the radio waves **610** generated from the liquid crystal display device **100**.

It is understood from the drawing that noises are generated not only at a basic frequency **620** of the digital signal **446** but also over the whole frequencies.

FIG. **19** shows a mode of an radio waves **630** generated when the transmission line **201** is provided to the output circuit **230**. The intensity of the radio waves **630** generated by the random number generating circuit **200** is larger than limit radio waves intensity **640** of EMI and hence, the radio waves **630** reaches a level which causes a drawback as noises.

The present inventors have made, first of all, as a countermeasure to cope with electromagnetic wave noises, an attempt in which the radio waves **610** generated from the liquid crystal display device **100** due to the digital signal **446** or the like shown in FIG. **18** is reduced so as to reduce the radio waves **630** generated by the random number generating circuit **200** shown in FIG. **19**.

The radiowaves **610** shown in FIG. **20** indicate an output waveform of the digital signal **446** to which the countermeasure against noises is applied. In FIG. **20**, a countermeasure such as a shield is provided to the liquid crystal display device **100** and, further, an inversion control signal (noise control signal) NCOL which controls the inversion number of the digital signal **446** is added thus reducing the radio waves **610** generated from the liquid crystal display device **100** whereby average intensity of the radio waves **610** can be reduced as indicated by symbol L2 shown in the drawing. Further, the radio waves **630** generated by the random number generating circuit **200** is also reduced by using the clock which intermittently repeats stopping thereof as the basic clock. Further, an attenuator or the like is provided to an output of the output circuit **230** so as to set the intensity of the radio waves **630** to the limit radio waves intensity **640** of the EMI or below.

As shown in FIG. **20**, the intensity of the radio waves **630** generated by the random number generating circuit **200** is the limit radio waves intensity **640** of the EMI or below, and the radio waves **630** can cover the whole frequency band of the radio waves **610** generated from the liquid crystal display device **100**. Accordingly, the radiowaves **630** generated by the random number generating circuit **200** can prevent the radio-waves **610** generated from the liquid crystal display device **100** from being outputted to the outside.

The reason that the radio waves **630** generated by the random number generating circuit **200** can cover the whole frequency band is that the randomness of the random number generating circuit **200** is high, and even when the digital signal **446** is transmitted through the shift register **220** with the basic frequency **620** of the digital signal **446**, by providing n pieces of shift registers **220** thus outputting n-times frequency by the output circuit **230**, the random numbers which fall within a frequency band which is n times as large as the basic frequency **620** can be generated.

Next, the noise control signal NCOL is explained in conjunction with FIG. **21** and FIG. **22**. FIG. **21** shows the change of 6-bit digital signal **446** from a period T1 to a period T8. In the period T1, the values of the digital signal **446** are (0, 0, 0, 0, 0, 0) and hence, a voltage of each signal line is a low voltage.

Subsequently, in the period T2, the values of the digital signal **446** are (1, 1, 1, 1, 1, 1) and hence, voltages of all signal lines is changed from a low voltage to a high voltage. As described previously, the electromagnetic wave noises are generated due to the change of the voltage and hence, a

12

change from the period T1 to the period T2 which brings about the change of voltage in all signal lines provides a worst pattern with respect to the electromagnetic wave noises.

Next, in a change from the period T2 to the period T3, 3 bits which are a half of 6 bits are changed such that the voltage is changed from the high voltage to the low voltage in bits D4 to D6. In a change from the period T3 to the period T4, 2 bits which are less than a half of 6 bits are changed such that the voltage is changed from the low voltage to the high voltage in bits D5 and D6.

FIG. **22** shows a case in which the noise control signal NCOL is added to the digital signals. In FIG. **22**, when a change of voltage occurs in most of bits D1 to D6 as indicated by a change from the period T1 to the period T2, the noise control signal NCOL is set to "1" so that bits D1 to D6 output a voltage of polarity opposite to polarity of the original voltage. For example, although all bits are assumed to exhibit a high voltage during the period T2, all bits are set to a low voltage.

Next, during the period T3, the bits D1 to D3 assume a high voltage, and the bits D4 to D6 assume a low voltage. However, the change of voltage takes place in the half of 6 bits and hence, the original voltage is outputted by setting the noise control signal NCOL to "0" whereby the bits D1 to D3 assume a high voltage and the bits D4 to D6 assume a low voltage.

Next, in a change from the period T3 to the period T4, the number of bits where the voltage is changed is 2 bits which is less than the half of 6 bits and hence, the original voltage is outputted by setting the noise control signal NCOL to "0" whereby the bits D1 to D3 assume a high voltage, the bit D4 assumes a low voltage, and the bits D5 and D6 assume a high voltage.

In the control performed using the noise control signal NCOL shown in FIG. **22**, when the voltage is changed in a large number of signal lines, the signal is controlled such that the voltage is not changed and hence, the change of voltage can be suppressed. In FIG. **22**, the number of arrows which indicate the change of the voltage is 12 and hence, the number of arrows is largely reduced compared to the number of arrows shown in FIG. **21** which is 26.

In this manner, by lowering levels of electromagnetic wave noises generated from the original digital signal **446** using the noise control signal NCOL, it is possible to reduce the radio waves **610** generated from the liquid crystal display device **100** as shown in FIG. **20**.

When the electronic wave **610** generated from the liquid crystal display device **100** can be reduced, it is also possible to reduce the radio waves **630** generated by the random number generating circuit **200** which can conceal the radio waves **610** generated from the liquid crystal display device **100** and hence, the intensity of the radio waves **630** generated by the random number generating circuit **200** can be suppressed to a level lower than the limit radio waves intensity **640** of the EMI.

Although the case in which 6-bit digital signal **446** is used has been explained as an example, the present invention is not limited to the 6-bit digital signal. For example, in inputting a 24-bit digital signal, when 13 bits or more are changed, the noise control signal NCOL is used so as to suppress the voltage change of the digital signal **446**, while in inputting a 18-bit digital signal, when 10 bits or more are changed, the noise control signal NCOL is used so as to suppress the voltage change of the digital signal **446**.

The above-mentioned explanation has been made to explain a typical function of the noise control signal NCOL. However, functions which the noise control signal NCOL

possesses are not limited to such a function. That is, the noise control signal NCOL may be used in combination with the inversion of the display data or the like. In other words, an essential role of the noise control signal NCOL is to control a change quantity of the display data.

That is, the noise control signal NCOL can change the intensity of electromagnetic wave noises by suppressing the change quantity of the display data signal and, at the same time, can reduce the power consumption of the liquid crystal display device **100**.

The utilization of a vacant terminal (empty cable or the like) of a connector of an interface which connects information devices or the like as a signal line for transmitting the noise control signal NCOL is considered as an effective way with no waste. However, a terminal (cable or the like) for the noise control signal NCOL may be provided intentionally. Further, with respect to the arrangement of the interface cable and the signal line for transmitting the noise control signal NCOL on the printed circuit board for driving the liquid crystal display device **100**, these lines may preferably be arranged adjacent to the display data line to acquire a more advantageous effect. However, the arrangement of these lines is not limited to such an arrangement.

Further, the number of signal lines which transmit the noise control signal NCOL is not limited to one. That is, it may be possible to adopt the circuit constitution in which a plurality of signal lines are arranged, and a total change quantity of display data is made uniform for every display data transmitting timing. Here, by making the change quantity of the display data uniform, the intensity of an electromagnetic field leaked from the liquid crystal display device **100** can be also made uniform.

Next, the basic clocks of the random number generating circuit are explained in conjunction with FIG. **23A** to FIG. **23D**. A display data transmission clock generated by the control circuit **3** or a clock generated by the control circuit **3** is used as the basic clock.

FIG. **23A** shows the display data transmission clock. Although the display data transmission clock is a clock which plays a role of supplying the display data to the drive circuit **6** from the control circuit **3**, the supply of the display data is unnecessary during a period in which the clock does not contribute to the display region **9**, that is, during a retracing period **51** and hence, the clock may be stopped for the purpose of low power consumption. By using the display data transmission clock as the basic clock of the random number generating circuit **200**, the generation of the random number can be stopped during the retracing period **51** and hence, the electromagnetic wave noises **630** can be reduced.

FIG. **23A** indicates a waveform of the display data transmission clock when outputting of the clock is stopped during the retracing period **51**. However, the clock may be continuously outputted during the retracing period **51**. That is, provided that the display data transmission clock satisfies the condition that the intensity of the radiowaves **630** generated by the random number generating circuit **200** is equal to the limit radio waves intensity **640** of the EMI or below, the display data transmission clock which is continuously outputted can be used.

FIG. **23B** shows a waveform of the display data transmission clock which is continuously outputted without interrupting the basic clock generated by the control circuit **3**. In this case, the random number generated by the random number generating circuit **200** is continuously outputted and hence, the reduction of the output radio waves **630** generated by the random number generating circuit **200** cannot be expected. However, provided that such a display data transmission

clock satisfies the condition that the intensity of the radio-waves **630** generated by the random number generating circuit **200** is equal to the limit radio waves intensity **640** of the EMI or below in the same manner as the display data transmission clock continuously outputted, the display data transmission clock which is continuously outputted can be also used.

FIG. **23C** shows a waveform of a display data transmission clock when the basic clock generated by the control circuit **3** is intermittently stopped, wherein the waveform is formed by repeating outputting of 5 clocks and stopping of 5 clocks. This waveform is similar to the waveform of the display data transmission clock when the clock is stopped during the retracing period. Since the generation of the random numbers from the random number generating circuit **200** is also stopped during the clock stopping period, the radio waves **630** can be reduced. Outputting of the waveform which intermittently stops the clock can be easily realized, and the waveform which intermittently stops the clock is not limited to such a waveform.

FIG. **23D** is a waveform obtained by repeating outputting of 1 clock and stopping of 2 clocks based on the basic clock. In the same manner as such a waveform, a frequency of the basic clock may be substantially lowered thus realizing a waveform substantially equal to the clock which changes a duty ratio. This clock is also provided with a clock stopping period so that the radio waves **630** can be reduced.

The stopping period of the basic clock is provided for generating the radio waves **630** by the random number generating circuit **200** which realizes covering of the radio waves **610** generated from the liquid crystal display device **100** over the whole frequency band as low as possible. That is, the level of the radio waves **630** generated by the random number generating circuit **200** can be adjusted by adjusting the intermittent period of the basic clock.

Next, FIG. **24** shows the arrangement of terminals when the shift register **220** is provided in the control circuit **3** and the output circuit **230** is provided outside the control circuit **3**. Numerals **452** to **454** indicate the arrangement positions of the terminals connected with the shift register **220**. The terminal which is arranged at the terminal position **452** is connected with the output circuit **230**, and a signal inputted to the output circuit **230** is outputted to the transmission line **201** from the output circuit **230** as the random numbers.

The transmission line **201** plays a role of an antenna and can control the intensity of the radio waves **630** generated by the random number generating circuit **200** by changing a shape or the arrangement of the transmission line **201**. Here, an attenuator or the like may be connected to the transmission line **201** for reducing the intensity of the radio waves **630**.

The terminal arranged at the terminal position **453** directly takes out an output of the shift register **220**, and the output before increasing a frequency by the output circuit **230** is transmitted to the transmission line **201**. By transmitting the output before increasing the frequency to the transmission line **201**, it is possible to broadly output the radio waves on a low frequency side.

Further, numeral **202** indicates a connector for external connection which is used when the transmission line **201** is arranged outside. The terminal arranged at the terminal position **453** is used for connection with the connector **202** for external connection. The terminal at the terminal position indicated by numeral **454** is used for connection with an external memory element **510** which stores the initial value, for example.

Although the embodiment in which the present invention is applied to the liquid crystal display device has been explained

15

heretofore, the present invention is not limited to such an embodiment. For example, the present invention is applicable to display devices in general including an organic EL display device.

Although the invention made by inventors of the present invention has been specifically explained based on the embodiment, it is needless to say that the present invention is not limited to such embodiment, and various modifications can be made without departing from the gist of the present invention.

What is claimed is:

1. A display device comprising:

a display panel;

a drive circuit driving respective pixels of the display panel; and

a control circuit supplying display data to the drive circuit, wherein the control circuit includes a random number generating circuit, the random number generating circuit includes an output circuit which increases a frequency of a pulse signal generated by the random number generating circuit n times, where n is an integer greater than one, and outputs a pulse signal of n -times increased frequency, the random number generating circuit includes n pieces of shift registers, a basic clock for n pieces of shift registers is an intermittent drive clock which intermittently alternates between a clock stopping period during which the random number generating circuit stops generating a random number and a clock outputting period, the basic clock being configured to controllably adjust an intermittent period thereof so as to control a level of radio waves generated by the random number generating circuit, and the control circuit outputs an inversion control signal indicative of inversion of the display data.

2. A display device according to claim 1, wherein the basic clock is a clock which is in synchronism with a transmission clock of the display data.

3. A display device according to claim 1, wherein the basic clock is a clock which is in asynchronism with a transmission clock of the display data.

4. A display device according to claim 1, wherein the random number generating circuit includes a register which stores n pieces of initial values.

5. A display device comprising:

a display panel;

a drive circuit which drives respective pixels of the display panel;

a control circuit which outputs display data to the drive circuit;

a random number generating circuit which is provided to the control circuit;

a memory element which stores initial values of the random number generating circuit; and

an electromagnetic wave noise generating circuit into which signals generated by the random number generating circuit are inputted,

wherein the random number generating circuit includes n pieces of shift registers (where n is an integer greater than one), a basic clock of n pieces of shift registers is an intermittent drive clock which intermittently alternates between a clock stopping period during which the random number generating circuit stops generating a ran-

16

dom number and a clock outputting period, the basic clock being configured to controllably adjust an intermittent period thereof so as to control a level of radio waves generated by the random number generating circuit, and the control circuit outputs an inversion control signal indicative of inversion of the display data.

6. A display device according to claim 5, wherein the basic clock is a clock which is in synchronism with a transmission clock of the display data.

7. A display device according to claim 5, wherein the basic clock is a clock which is in asynchronism with a transmission clock of the display data.

8. A display device according to claim 5, wherein the random number generating circuit includes a register which stores n pieces of initial values.

9. A display device according to claim 5, wherein the random number generating circuit includes an output circuit which is connected to n pieces of shift registers.

10. A display device comprising:

a display panel;

a touch panel which overlaps with the display panel in plane;

a drive circuit which drives respective pixels of the display panel;

a control circuit which outputs display data to the drive circuit;

a random number generating circuit which is provided to the control circuit;

a memory element which stores initial values of the random number generating circuit; and

an electromagnetic wave noise generating circuit into which signals generated by the random number generating circuit are inputted,

wherein the random number generating circuit includes n pieces of shift registers (where n is an integer greater than one), a basic clock of n pieces of shift registers is an intermittent drive clock which intermittently alternates between a clock stopping period during which the random number generating circuit stops generating a random number and a clock outputting period, the an electromagnetic wave noise circuit changes frequency of signals generated by the random number generating circuit, the basic clock being configured to controllably adjust an intermittent period thereof so as to control a level of radio waves generated by the random number generating circuit, and the control circuit outputs a noise control signal and controls the number of inversion of the display data.

11. A display device according to claim 10, wherein the basic clock is a clock which is in synchronism with a transmission clock of the display data.

12. A display device according to claim 10, wherein the basic clock is a clock which is in asynchronism with a transmission clock of the display data.

13. A display device according to claim 10, wherein the random number generating circuit includes a register which stores n pieces of initial values.

14. A display device according to claim 10, wherein the random number generating circuit includes an output circuit which is connected to n pieces of shift registers.

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当前申请(专利权)人(译)	日立显示器有限公司. 松下液晶显示CO. , LTD.		
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摘要(译)

一种显示装置，包括显示面板，控制电路和随机数产生电路。随机数产生电路包括多个移位寄存器，输出电路和保持初始值的寄存器。通过提供多个初始值，可以增强随机数的随机性。此外，可以从多个移位寄存器输出彼此不同的随机数，因此，可以通过输出电路增加频率并输出增加的频率。此外，通过添加抑制数字信号的反转次数的噪声控制信号，可以减少从液晶显示装置产生的电磁波噪声。此外，通过采用间歇地重复停止的间歇驱动时钟作为多个移位寄存器的基本时钟，可以减少从显示装置产生的电磁波噪声。基本时钟可以是与显示数据的传输时钟同步的时钟，或者是与显示数据的传输时钟异步的时钟。

