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(54) **LIQUID CRYSTAL DISPLAY AND
PROCESSING METHOD THEREOF**

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patent is extended or adjusted under 35
U.S.C. 154(b) by 2311 days.

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CPC **G09G 3/3648** (2013.01); **G09G 2360/18**
(2013.01); **G09G 2310/04** (2013.01); **G09G**
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2310/027 (2013.01); **G09G 2320/0233**
(2013.01)
USPC **345/87**; **345/690**

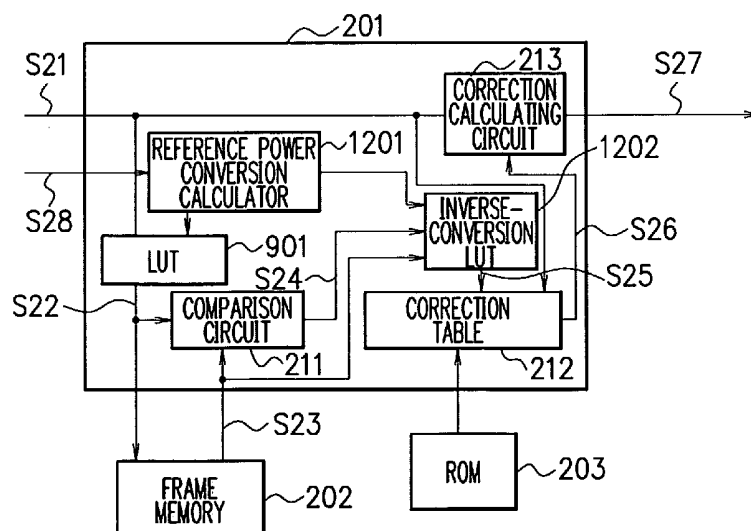
(58) **Field of Classification Search**
USPC 345/87, 90, 94, 98, 99, 100, 101, 698,
345/699; 349/33

See application file for complete search history.

(57) **ABSTRACT**

A liquid crystal display is provided, including a conversion circuit to convert a first image data to a second image data, a frame memory to store the second image data, a difference circuit to output in units of pixel a difference data between the second image data of the present frame to be converted and a third image data of an antecedent frame to be outputted from the frame memory, a correction circuit to correct the difference data based on one of the first to third image data, and an adding circuit to add the corrected difference data and the first image data.

18 Claims, 8 Drawing Sheets



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FIG. 1

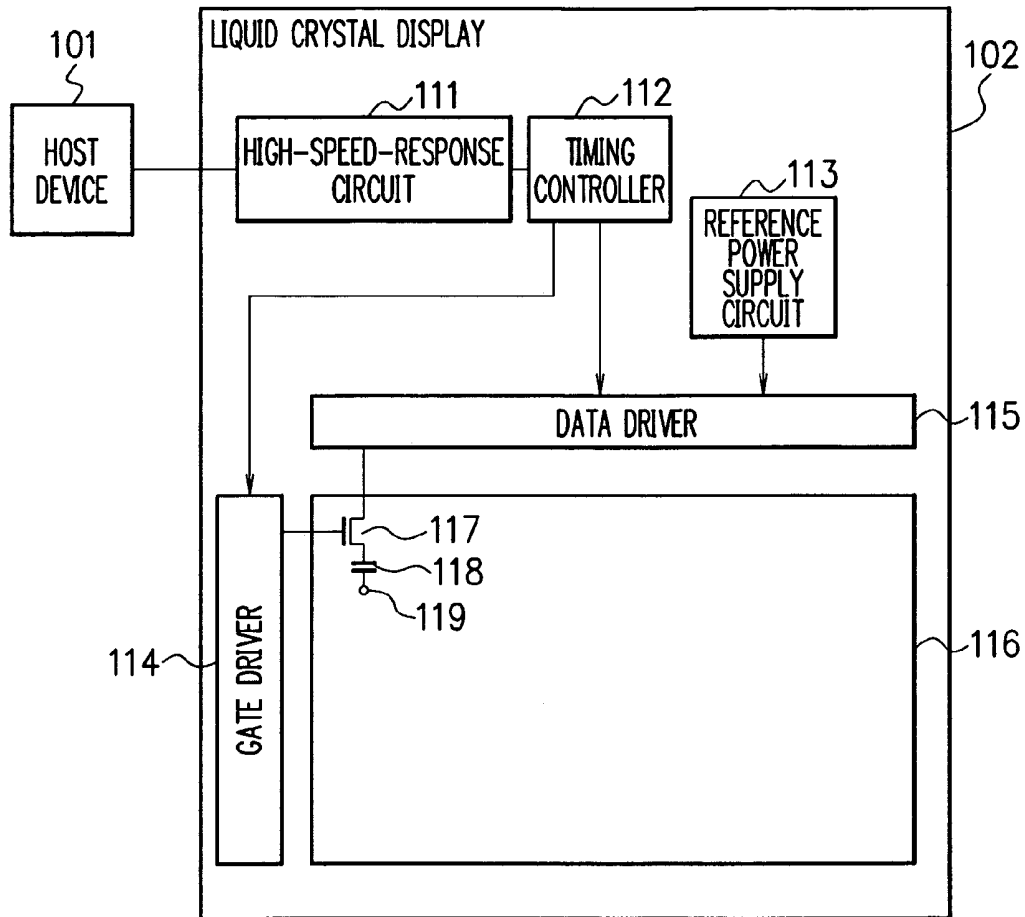


FIG. 2

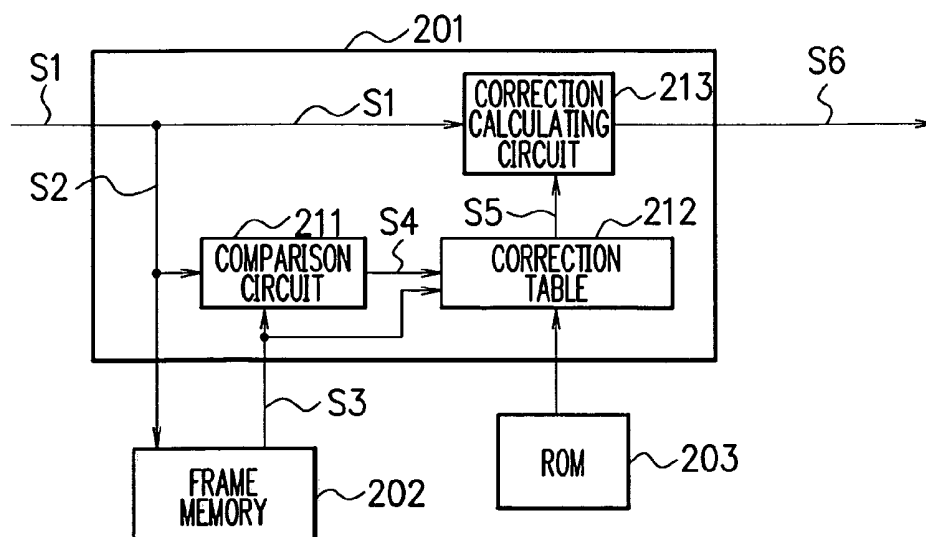


FIG. 3

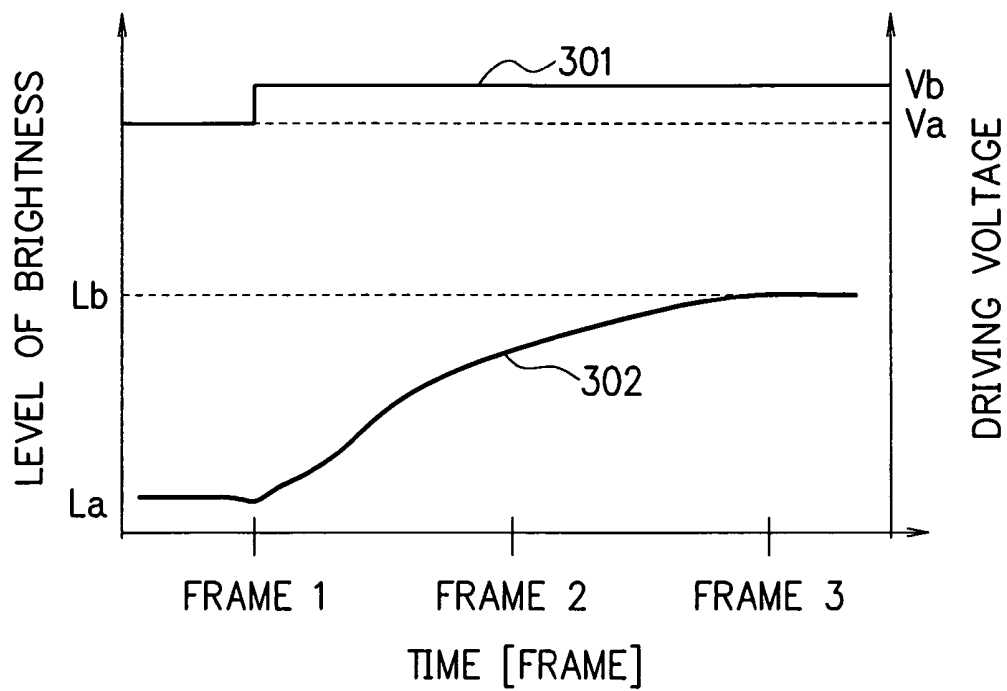


FIG. 4

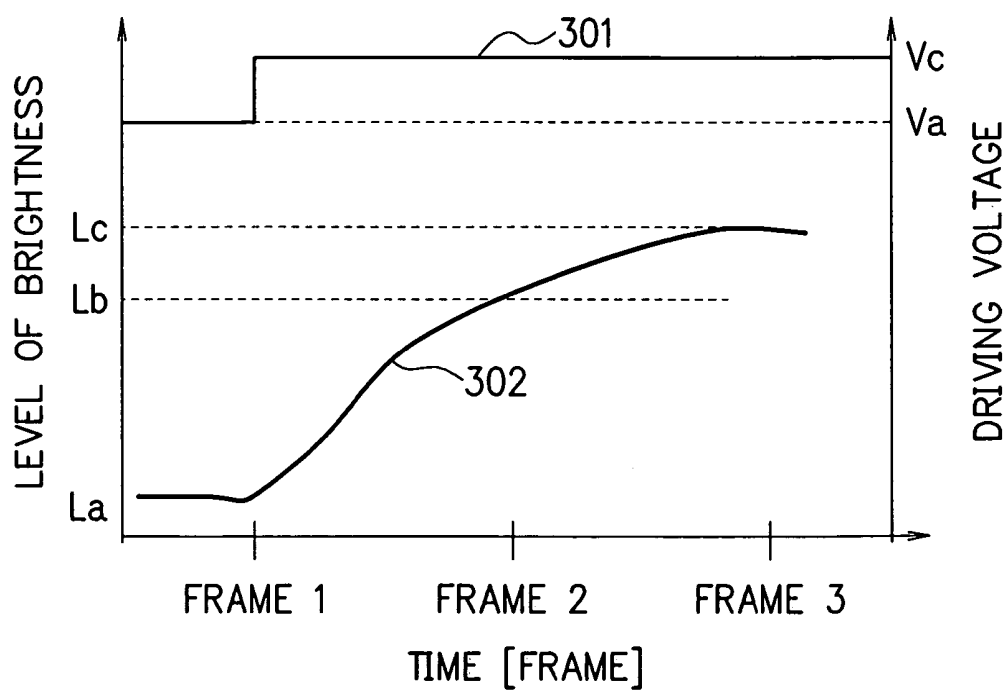


FIG. 5

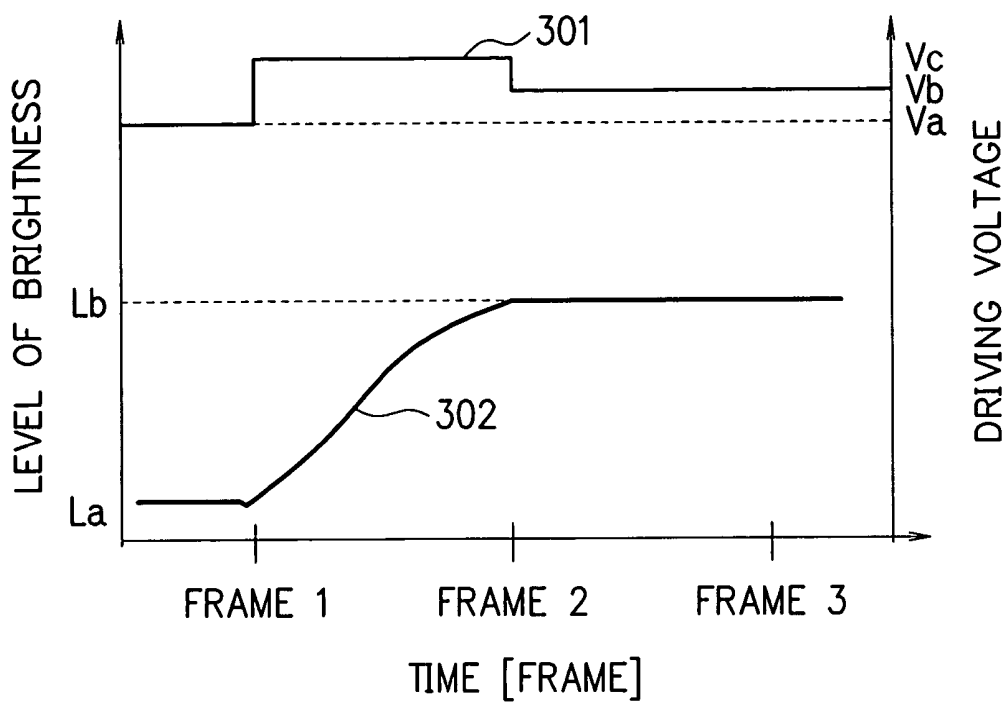


FIG. 6

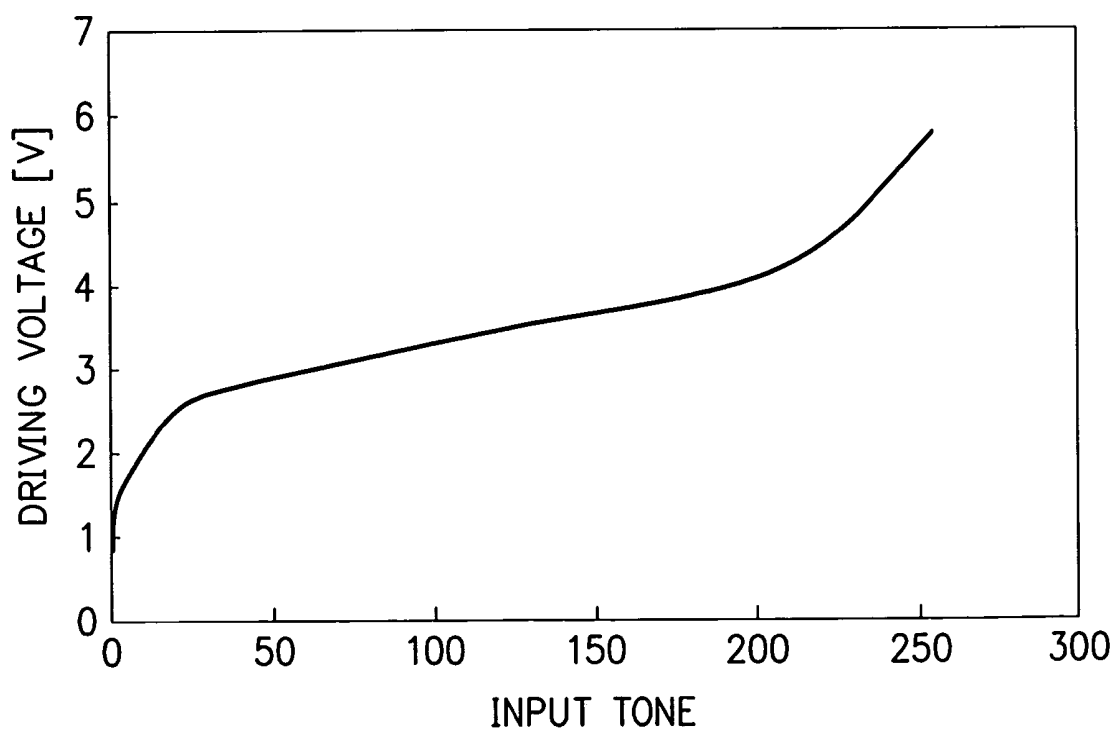


FIG. 7

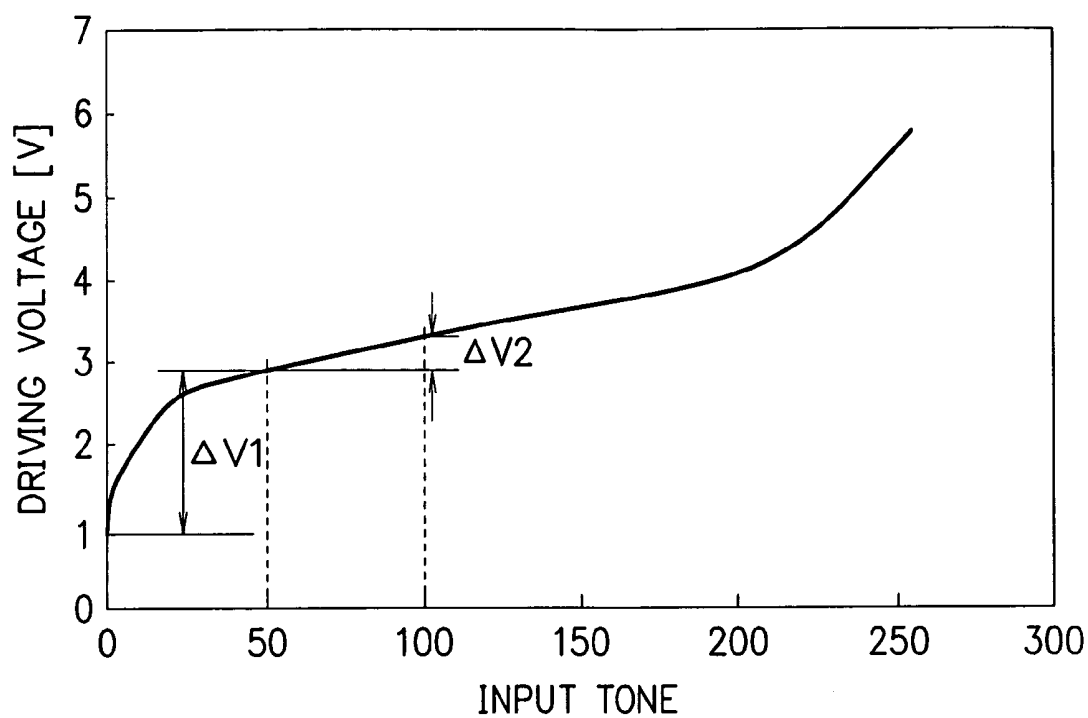


FIG. 8

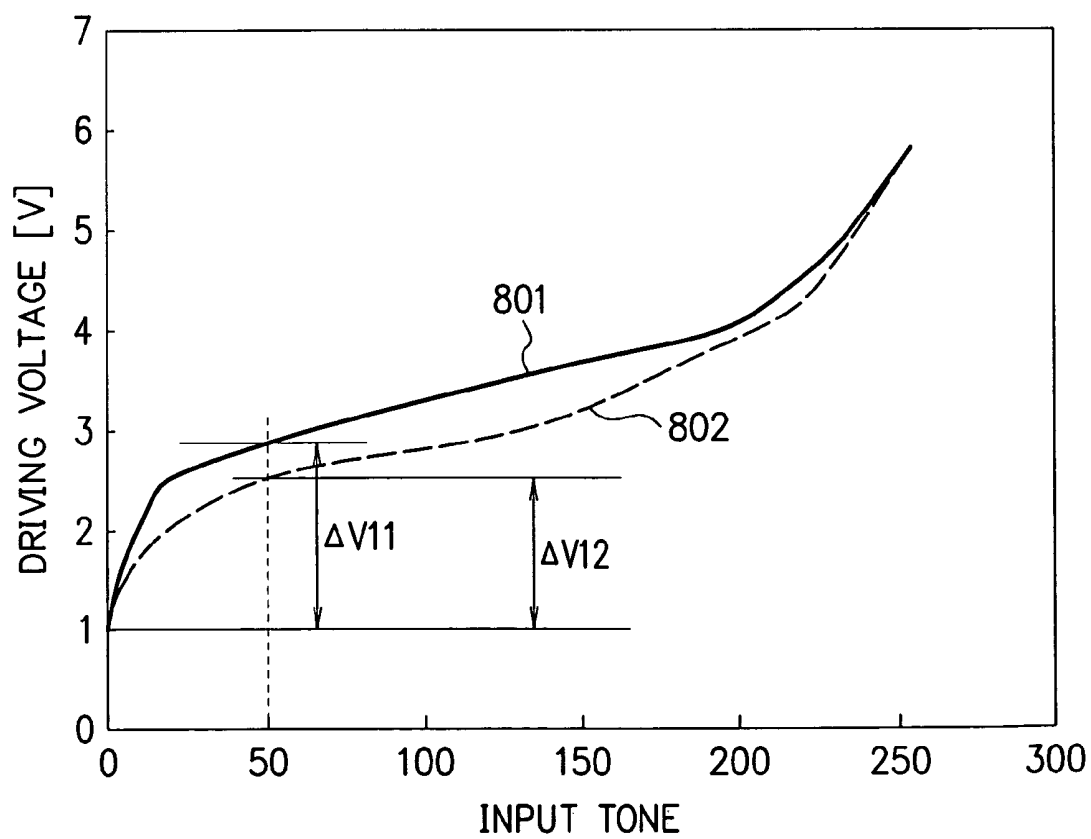


FIG. 9

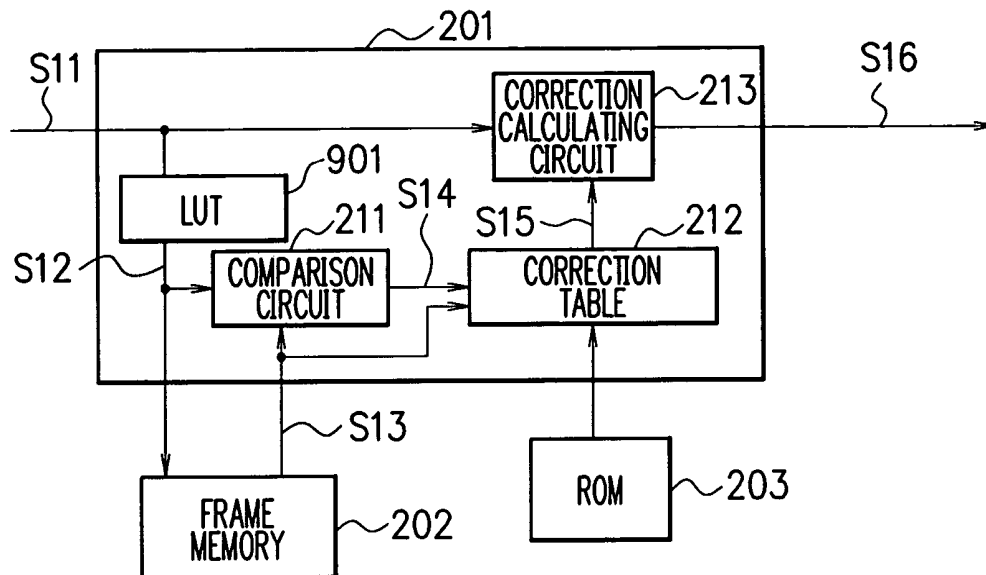


FIG. 10

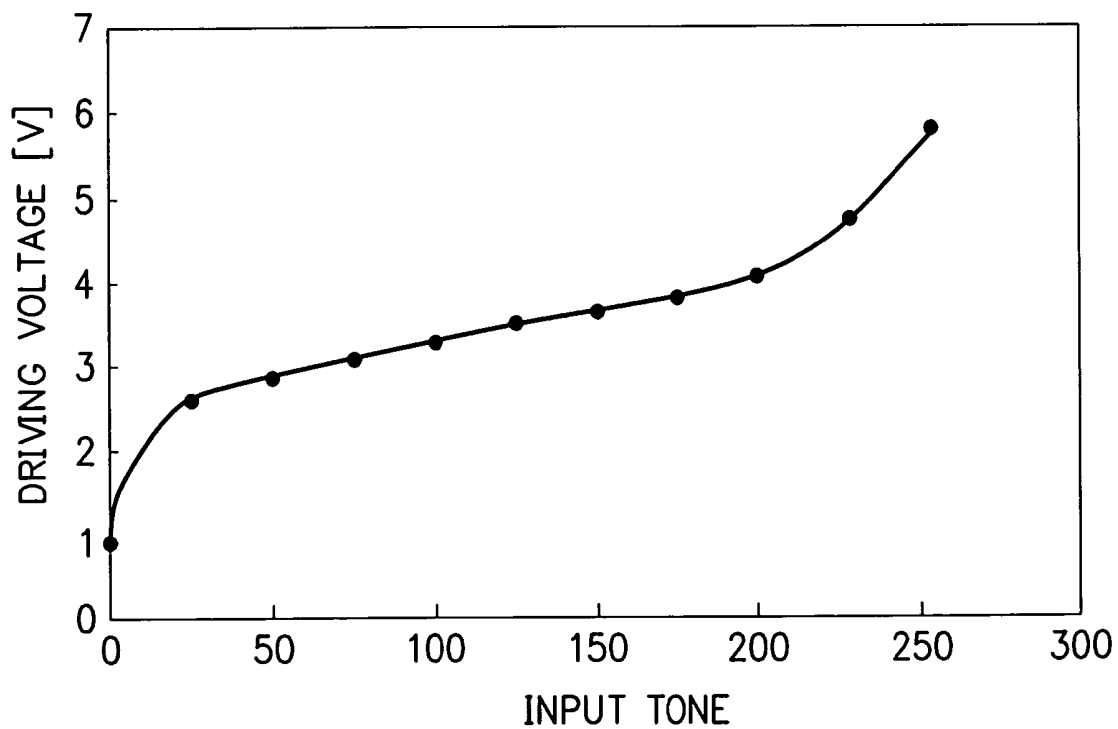


FIG. 11

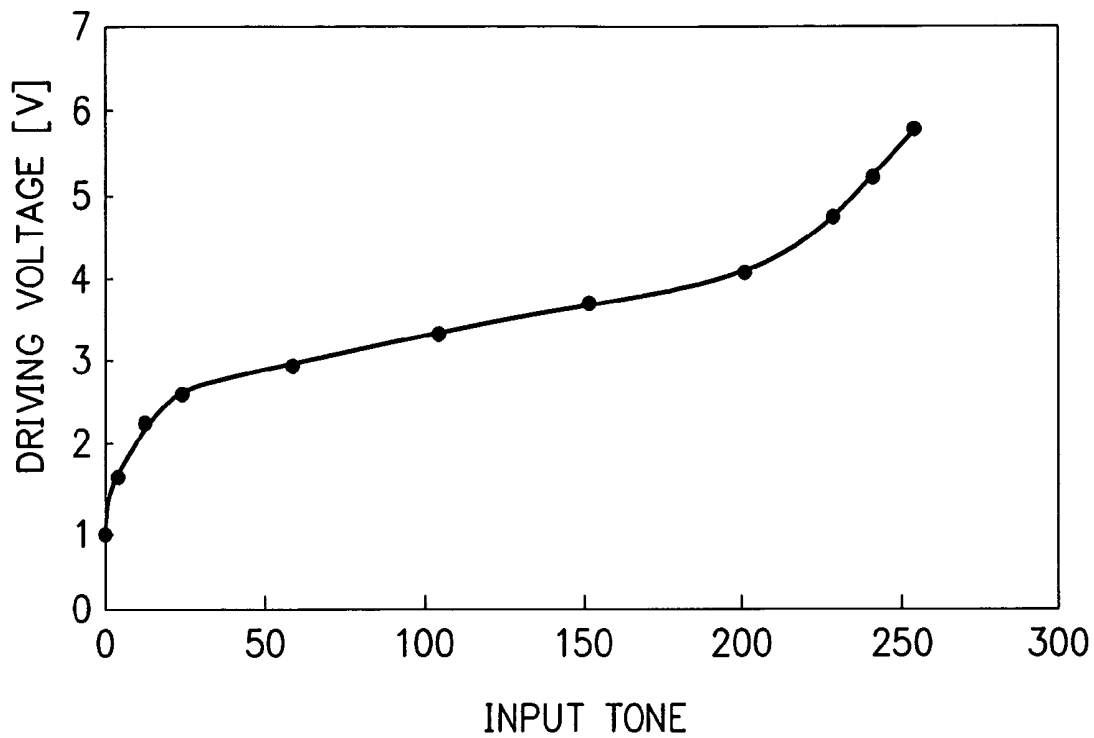
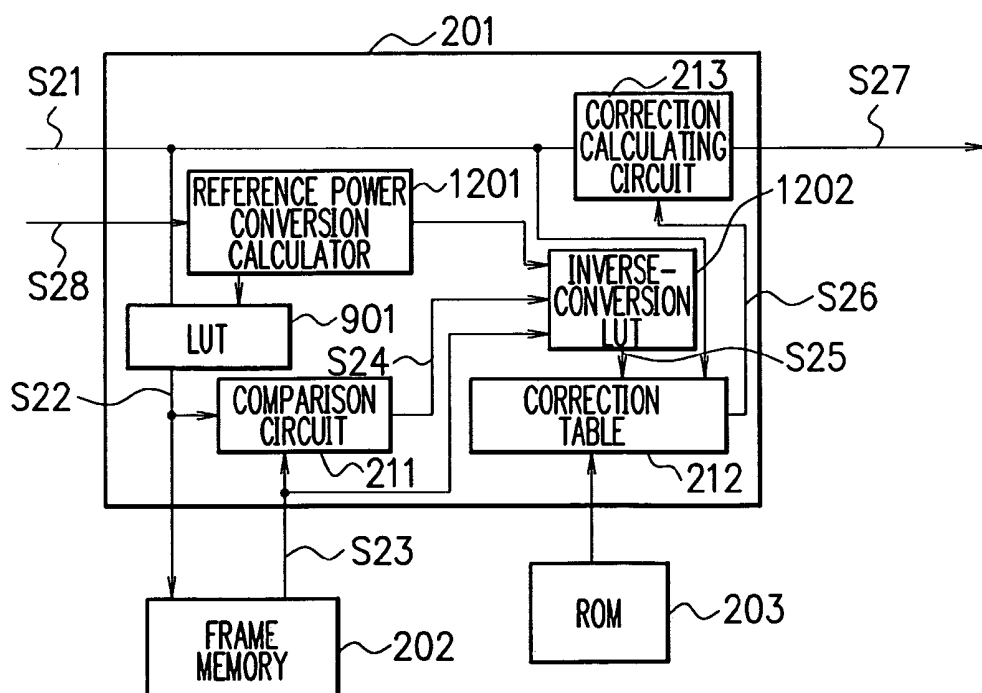
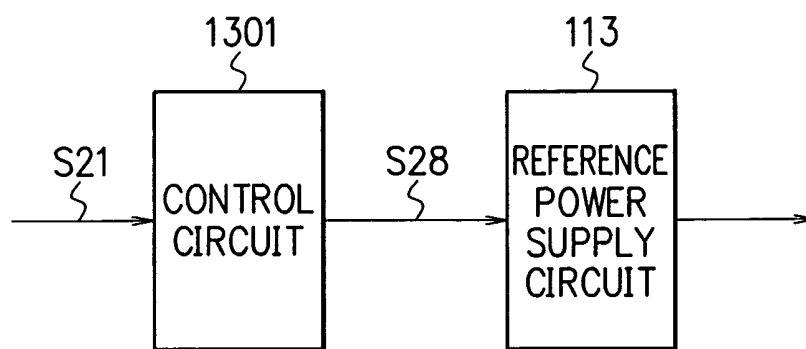


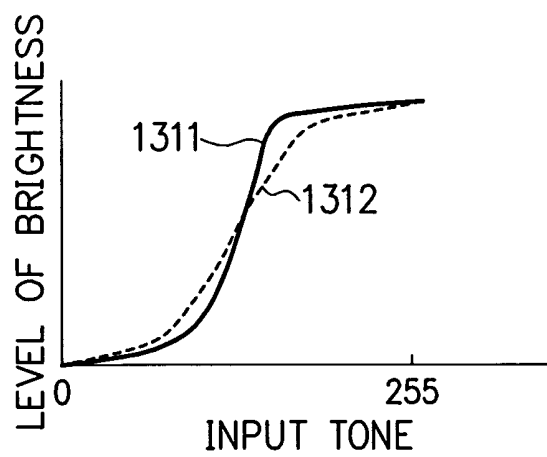
FIG. 12



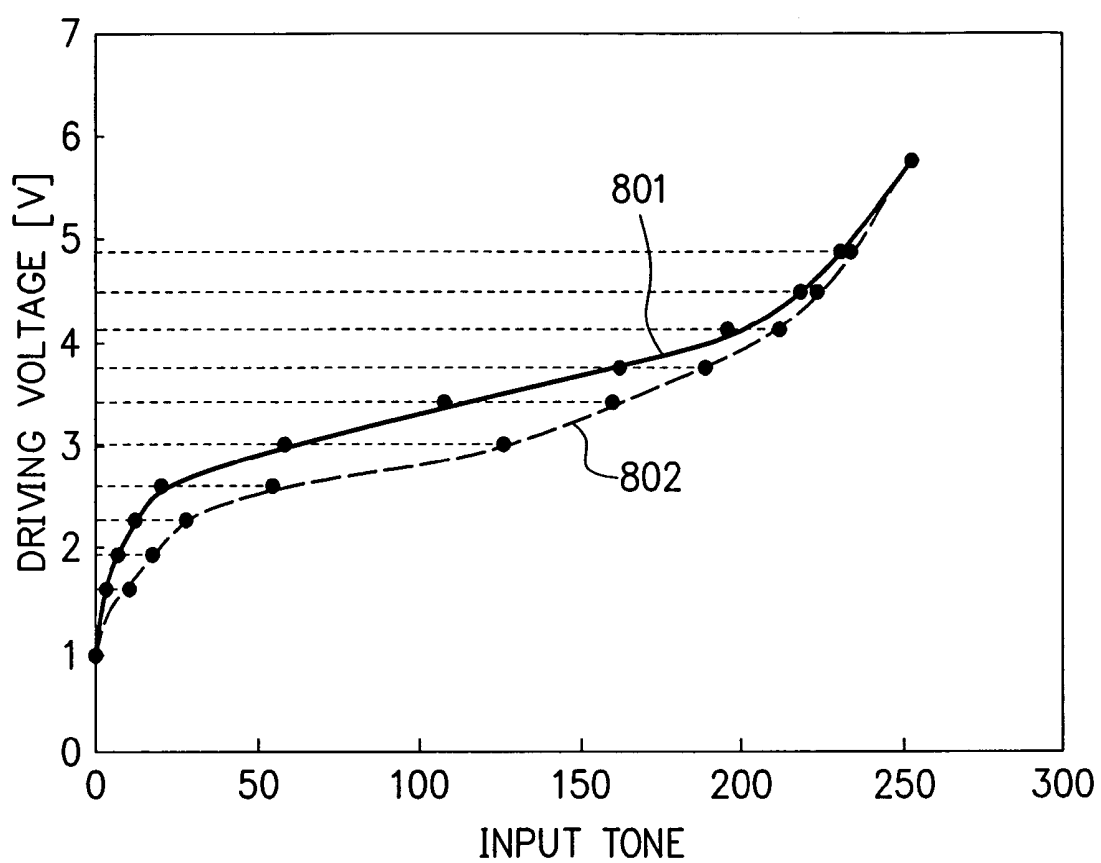
F I G. 13A



F I G. 13B



F I G. 14



LIQUID CRYSTAL DISPLAY AND PROCESSING METHOD THEREOF

CROSS-REFERENCE TO RELATED APPLICATIONS

This application is based upon and claims the benefit of priority from the prior Japanese Patent Application No. 2004-134204, filed on Apr. 28, 2004, the entire contents of which are incorporated herein by reference.

BACKGROUND OF THE INVENTION

1. Field of the Invention

This invention relates to a liquid crystal display, and more particularly, to correction of an image data.

2. Description of the Related Art

In recent years, with demands for energy saving and space saving, notebook PCs (personal computers) or desktop PCs carrying a liquid display monitor are spreading their market. In such a trend, even higher-speed responses are demanded for a liquid crystal display so as to improve the characteristics for displaying moving images and so forth. Accordingly, the improvement of the response of the liquid crystal display is intended through the material characteristics of crystal display, display element structure, and development of a driving method.

In Patent Document 1 described below, a liquid crystal display is disclosed which, in correcting an image data signal and generating a correction data signal, generates a present correction data by a present image data signal and an antecedent correction data signal.

Also in Patent Document 2 described below, a liquid crystal display is disclosed which carries a conversion table to refer to a display-driven data of a present frame through an image data of the present frame and a post-driven-state data of an antecedent frame.

[Patent Document 1] U.S. Patent Application Publication No. U.S. 2002/033813 (Japanese Patent Application Laid-open No. 2002-99249)

[Patent Document 2] U.S. Patent Application Publication No. U.S. 2002/0140652 (Japanese Patent Application Laid-open No. 2002-297104)

SUMMARY OF THE INVENTION

It is an object of the present invention to provide a liquid crystal display which performs high-speed-response driving, has small amounts of memory, and allows the high-definition image display, as well as to provide a processing method thereof.

According to one aspect of the present invention, a liquid crystal display is provided which includes: a conversion circuit to convert a first image data to a second image data having a fewer number of bits; a frame memory to store the second image data; a difference circuit to output, in units of pixel, a difference data between the second image data of the present frame to be converted and a third image data of an antecedent frame to be outputted from the frame memory; a correction circuit to correct the difference data according to one of the first to third image data; and an adding circuit to add the corrected difference data and the first image data.

BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 is a block diagram showing an example of a structure of a host device and a liquid crystal display according to an embodiment of the present invention;

FIG. 2 is a block diagram showing an example of a structure of a high-speed-response circuit;

FIG. 3 is a diagram showing a relationship between time (frame) and liquid crystal driving voltage, and a relationship between time (frame) and level of brightness;

FIG. 4 is a diagram showing a relationship between time (frame) and liquid crystal driving voltage, and a relationship between time (frame) and level of brightness;

FIG. 5 is a diagram showing a relationship between time (frame) and liquid crystal driving voltage and a relationship between time (frame) and level of brightness;

FIG. 6 is a graph showing an example of a relationship between tone value and liquid crystal driving voltage;

FIG. 7 is a graph showing an example of a relationship between tone value and liquid crystal driving voltage;

FIG. 8 is a graph showing an example of a relationship between tone value of an inputted image and liquid crystal driving voltage when a gamma characteristic is switched over;

FIG. 9 is a block diagram showing an example of a structure of a high-speed-response circuit according to a first embodiment of the present invention;

FIG. 10 is a graph showing an example of a relationship between tone value of an inputted data and liquid crystal driving voltage;

FIG. 11 is a graph showing an example of a relationship between tone value of an inputted data and liquid crystal driving voltage;

FIG. 12 is a block diagram of showing an example of a structure of a high-speed-response circuit according to a second embodiment of the present invention;

FIG. 13A is a block diagram showing an example of a structure of a reference power supply circuit and a control circuit thereof, and FIG. 13B is a graph showing an example of a gamma characteristic; and

FIG. 14 is a graph showing an example of a relationship between tone value of an inputted image and liquid crystal driving voltage when the gamma characteristic is switched over.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

FIG. 1 is a block diagram showing an example of a structure of a host device **101** and a liquid crystal display **102**, according to a preferred embodiment of the present invention. The host device **101** is, for example, a personal computer, TV receiver or the like, which outputs image data to the liquid crystal display **102**. The liquid crystal display includes a high-speed-response circuit **111**, a timing controller **112**, a reference power supply circuit **113**, a gate driver **114**, a data driver **115**, and a liquid crystal panel **116**.

The high-speed-response circuit **111** inputs therein an image data from the host device **101**, and corrects the image data for the high-response driving of the liquid crystal panel **116**. The timing controller **112** inputs therein the corrected image data, and controls the timing of the gate driver **114** and data driver **115**. The corrected image data is supplied to the data driver **115** through the timing controller **112**. The image data includes, for example, red, green, and blue image data having 8 bits respectively. The data driver **115** supplies the liquid crystal driving voltage to the liquid crystal panel **116** according to the image data (tone value). The reference power supply circuit **113** generates plural reference power supply voltages corresponding to the tone values of the image data in predetermined intervals, and outputs to the data driver **115**. According to the plural reference power supply voltages, the

data driver **115** generates the liquid crystal driving voltages for all the tone values, selects the liquid crystal driving voltage for each image data, and outputs them to the liquid crystal panel **116**.

The liquid crystal panel **116** includes plural thin-film transistors (TFT) **117** corresponding to the plural pixels arranged two-dimensionally. The transistor **117** has its gate connected to the gate driver **114**, its drain connected to the data driver **115**, and its source connected through a liquid crystal (capacitor) **118** to a common electrode **119**.

The gate driver **114** outputs a gate pulse for sequentially selecting the transistors **117** arranged two-dimensionally to the gate of the transistor **117**. Upon reception of the gate pulse, the transistor **117** is turned on and the liquid crystal driving voltage is provided to the liquid crystal **118** through the drain. According to the liquid crystal driving voltage, the transmittance of the liquid crystal **118** changes, and thereby the level of brightness changes.

FIG. 6 is a graph showing an example of relationship between the tone value of an inputted data and the liquid crystal driving voltage. In accordance with the relationship, the data driver **115** performs conversion from an image data to a liquid crystal driving voltage. The inputted image data is for example 8 bits, and has tone values of 0 (zero) to 255.

FIGS. 3 to 5 illustrates a characteristic **301** showing a relationship between time (frame) and liquid crystal voltage, and a characteristic **302** between time (frame) and level of brightness.

In FIG. 3, when the image data transforms from Da to Db in the first frame, the liquid crystal driving voltage changes from Va to Vb. At that time, the level of brightness changes from La to Lb, but since the response by the liquid crystal is slow, reaching the targeted brightness Lb costs a few frames. For example, the level of brightness reaches Lb at the start point of the third frame.

On the other hand, as shown in FIG. 4, when the voltage which changes from the Va to Vc within the first frame is impressed to the liquid crystal panel, the brightness reaches Lb in the second frame, and Lc in the third frame. Here, the voltage Vc is a liquid crystal driving voltage for the image data Dc, and is higher than the voltage Vb.

As shown in FIG. 5, in the case of the inputted image data transforming from Da to Db, the image data is corrected such that it transforms in an order of Da, Dc, and Db. At the start point of the first frame, the voltage is changed from Va to Vc, and at the start point of the second frame, the voltage is changed from Vc to Vb. As a result, the level of brightness at the start point of the first frame becomes La, while the level of brightness at and after the start point of the second frame becomes Lb. This allows the liquid crystal to respond at a high speed.

FIG. 2 is a block diagram showing an example of a structure of the high-speed-response circuit **111** (FIG. 1) which enables the operation shown in FIG. 5. The high-speed-response circuit **111** contains a processing circuit **201**, a frame memory (SDRAM) **202** and a ROM **203**. An image data S1 is inputted such that red, green, and blue image data respectively having m bits are inputted to the high-speed-response circuit **111** in a parallel manner. An image data S2 is an image data consisting of the upper u bits ($n < m$) in the image data S1 having m bits. The relationship between the image data S1 and S2 will be explained hereinafter, with reference to FIG. 10.

FIG. 10 is a graph showing an example of the relationship between the tone value of the inputted image data and the liquid crystal driving voltage. The solid line represents the image data S1 having m bits. the dots on the solid line repre-

sent the image data S2 having n bits. The image data S2 is mapped to the image data S1 in regular intervals and quantized.

In FIG. 2, the image data S2 is written in the frame memory **202**. The frame memory **202** stores the image data S2 amounting to one frame. Since the image data S2 has fewer bits than the image data S1, the amount of the frame memory **202** can be reduced.

The frame memory **202** delays the image data S2 for one frame, and outputs the image data S3. The comparison circuit **211** compares the image data S2 of the present frame and the image data S3 of the antecedent frame, and outputs a difference data S4. For example, in FIG. 5, the present frame data S2 of the first frame is Db, while the antecedent frame data S3 is Da. The difference data S4 is $Db - (\text{minus}) Da$.

The correction table **212** corrects the difference data S4 according to the image data S3, and outputs a difference data S5. For example, as shown in FIG. 5, at the start point of the first frame, the image data is transformed from Da to Dc, and at the start point of the second frame, the image data is transformed from Dc to Db. Hence, when " $Db - (\text{minus}) Da$ " is inputted as a difference data S4, " $Dc - (\text{minus}) Db$ " is outputted as the difference data S5. In the following frame, 0 (zero) is outputted as the difference data S5. The correction table **212** reads therein the correction data from the ROM **203** in advance.

The correction calculating circuit **213** is an adding circuit, wherein the image data S1 and the difference data S5 are added and the image data S6 is outputted. For example, as shown in FIG. 5, the image data S1 is Db, the difference data S5 is $Dc - (\text{minus}) Db$, and the image data S6 is Dc. Thus, the high-speed response driving shown in FIG. 5 can be realized.

FIG. 7 is a graph showing an example of a relationship between the tone value of the inputted image data and the liquid crystal driving voltage, similarly to FIG. 6. The voltage variation of the liquid crystal voltage when the tone value of the inputted image data is changed from 0 (zero) to 50 is $\Delta V1$, while the voltage variation of the liquid crystal driving voltage when the tone value of the inputted image data is changed from 50 to 100 is $\Delta V2$. The tone variances of the both are identically 50, but $\Delta V1$ is extremely than $\Delta V2$. That is to say, although their tone variances are identical, the variance of their liquid crystal driving voltage varies according to the absolute tone value.

Because the high-speed-response driving is a method to impress the liquid crystal driving voltage suitable for the changed image data, in a region of a large voltage variance, the image data S2 needs to be kept in a fine manner in order to perform a precise high-speed-response driving. That is to say, in the neighborhood of $\Delta V1$, the image data S2 needs to be kept in a fine manner.

One method to enhance the data precision would be to increase the number of bits of the image data S2. However, this method leads to an expanded size of circuits such as of the frame memory **202**, comparison circuit **211**, correction table **212**, and so forth. Further, since the frame memory **202** has a standardized number of bits in general, a frame memory with its number of bits being one rank higher has to be used, leading to a cost increase. In the following, embodiments to solve the above-described problem will be explained.

First Embodiment

FIG. 9 is a block diagram showing an example of a structure of a high-speed-response circuit **111** (FIG. 1), according to a first embodiment of the present invention. FIG. 9 is the structure in FIG. 2 whereto a lookup **901** table is added.

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Explained below are points of difference of the high-speed-response circuit in FIG. 9 compared to that of FIG. 2.

The lookup table **901** converts an image data **S11** having m bits into an image data **S12** having n bits. The image data **S11** consists of red, green and blue image data respectively having m bits. Here, n bits are fewer than m bits. The relationship between the image data **S11** and the image data **S12** are explained below with reference to FIG. 11.

FIG. 11 is a graph showing the relationship between the tone value of an inputted image data and the liquid crystal driving voltage. The solid line represents the image data **S11** having m bits. The dots on the solid line represent the image data **S12** having n bits. The image data **S12** is mapped from the image data **S11** in irregular intervals.

The lookup table **901** is a conversion table to store the correspondence between the image data **S11** and the image data **S12**, and maps the image data **S11** to the image data **S12** in irregular intervals. Further, the lookup table **901** maps the image data **S11** to the image data **S12** such that the levels of the liquid crystal driving voltage corresponding to the image data **S2** (vertical axis of FIG. 11) are in regular intervals. With the variance of the liquid crystal driving voltage being constant, if the liquid crystal driving response speed is the same between the two data, this mapping is appropriate. If the liquid crystal driving response speed is not the same, then the lookup table **901** maps the image data **S11** to the image data **S12** in a manner that the response speeds to the liquid crystal driving voltage for the image data **S12** are in regular intervals. Hence, in the relation curve between the image data and the liquid crystal driving voltage, the conversion to the image data **S12** can be carried out such that the sharp curve portion is fine, and the moderate curve portion is rough. This means that the resolution can be enhanced in a critical portion, allowing a high-quality image display.

In FIG. 9, the image data **S12** is written in the frame memory **202**. The frame memory **202** stores the image data **S12** in the amount of one frame. For example, the image data **S11** consists of red, green, and blue image data respectively having 8 bits. The image data **S12** consists of a 5-bit red, 6-bit green, and 5-bit blue image data having 16 bits in total, so that it can be efficiently stored in a memory of a standard size. The number of bits for green is greater than that of red and blue, since green is an important color data having greater influence on the level of brightness. The frame memory **202** delays the image data **S12** for one frame, and outputs an image data **S13**. The comparison circuit **211** compares the image data **S12** of the present frame to the image data **S13** of the antecedent frame, and outputs a difference data **S14** thereof in units of pixel.

The correction table **212** corrects the difference data **S14** according to the image data **S13**, and outputs a difference data **S15**. The correction table **212** reads therein the correction data from the ROM **203** in advance. The correction table **212** may perform correction according to the image data **S11** or **S12** instead of the image data **S13**. The correction calculating circuit **213** is an adding circuit, which adds the image data **S11** and the difference data **S15**, and outputs the image data **S16**. As a result, the high-speed-response driving shown in FIG. 5 can be realized.

The high-speed-response circuit of FIG. 2 stores the image data **S2** in the frame memory **202** in a manner that the intervals on the axis for tone value (the horizontal axis in FIG. 10) of the inputted image data are constant, as shown in FIG. 10. For the portion in which the image data **S2** should be kept in a fine manner, the data is in the regular interval. When the number of bits of the image data **S2** is increased, the size of the circuit has to be larger, so that a frame memory **202** of one

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rank higher must be used. In addition, the portions in the image data **S2** which do not need to be kept in a fine manner are also fragmented, causing inefficiency.

On the other hand, the high-speed-response circuit of FIG. 9 stores in the frame memory **202** the image data **S12** which is shown in intervals on the axis for the liquid crystal driving voltage (the vertical axis in FIG. 11), as shown in FIG. 11. This allows the image data **S12** to keep larger amount of data for the portion requiring finer data, and to keep rough data for the portion not requiring fine data. By using the lookup table **901**, the image data **S12** can be kept optimally in the frame memory **202** without increasing the number of bits thereof.

Since the response of the liquid crystal is evaluated on a basis of brightness, a lookup table **901** having an identical output bits for red, green, and blue may be used. However, considering the size of the frame memory **202**, a lookup table **901** having more bits just for green of which the brightness is high can be used, as it leads to a higher precision. For example, the number of bits for the general frame memory **202** is fixed such as into 16 bits or 32 bits. When the 16-bit frame memory **202** is used in which the lookup table **901** has the same number of bits for red, blue, and green, the respective colors have 5 bits, leaving one extra bit. In such a case, by allocating five bits for red and blue respectively and six bits for green, the frame memory **202** can be used without loss, and at the same time a high-speed-response driving with high precision can be realized.

Second Embodiment

A second embodiment of the present invention is herein-after explained. The reference power supply circuit **113** in FIG. 1 may consist of an amplifier of digital-analog-converter (DAC) type. The DAC-type amplifier **113** can generate plural types of reference power supply voltages (liquid crystal reference driving voltages), and change the reference power supply voltages to be generated according to a control signal. The DAC-type amplifier **113** can change the reference power supply voltage and switch the gamma characteristic depending on the image to be displayed. More details are described below with reference to FIGS. 13A and 13B.

The FIG. 13A shows a structural example of the reference power supply circuit (DAC-type amplifier) **113** and a control circuit **1301** thereof, while FIG. 13B shows a gamma characteristic. The gamma characteristic shows the relationship between the tone value and the level of brightness of the inputted image data.

The control circuit **1301** analyzes the tone distribution of one-frame data of the image data **S12**, and outputs a gamma characteristic signal **S28**. For example, when medium values makes up majority of the range of tones from 0 (zero) to 255, a gamma characteristic **1312** is selected so that the portion is finely quantized. On the other hand, if small and large values make up the majority of the range of tones from 0 (zero) to 255 (for example, where there are only black and white pixels), a gamma **1311** is selected to enhance the contrast of the image. The reference power supply circuit **113** generates reference power supply voltages for realizing the gamma characteristic **1311** or **1312**, depending on a gamma characteristic signal **S28** that is selected.

FIG. 8 is a graph showing a relationship between the tone value of an inputted image data and the liquid crystal driving voltage. Two characteristics **801** and **802** correspond to the two types of gamma characteristics (see FIG. 13B). In actual cases, there exists a combination of characteristics based on the precision of the DAC of the reference power supply circuit

113, but herein, for convenience sake, the two types of characteristics **801** and **802** are presented.

In the characteristic **801**, the variance of the liquid crystal driving voltage when the tone of the inputted image data changes from 0 (zero) to 50 is $\Delta V11$. In the characteristic **802**, similarly, the variance of the liquid crystal driving voltage when the tone of the inputted image data changes from 0 (zero) to 50 is $\Delta V12$. $\Delta V11$ and $\Delta V12$ are clearly different. Here, an issue is the responding characteristic of the liquid crystal. The correction value for $\Delta V11$ and $\Delta V12$ is known not to be in simple proportionality relation. Accordingly, the correction data required for the ROM **203** in FIG. 9 have to be the correction data both for the characteristic **801** and for **802**, suggesting that the amount of data doubles. Further, in the actual liquid crystal driving, not only the two types of characteristics, **801** and **802**, but also additional characteristics may be necessary, suggesting that the method of storing the correction data of each characteristic in the ROM **203** is extremely inefficient and not practical. Shown in FIG. 12 is a high-speed-response circuit in order to solve this problem.

FIG. 12 is a block diagram showing an example of a function of the high-speed-response circuit **111** (FIG. 1) according to the second embodiment of the present invention. It is the circuit of FIG. 9 with a reference power supply conversion calculator **1201** and an inverse-conversion lookup table **1202** added thereto. The difference between the high-speed-response circuits of FIG. 12 and FIG. 9 is explained below.

A reference power supply circuit **113** in FIG. 13A is, for example, a DAC-type amplifier, and changes the reference power supply voltages to be generated according to a control signal **S28**. Subsequently, in FIG. 12, the reference power supply conversion calculator **1201** calculates and rewrites the content of the lookup table **901**. The lookup table **901** converts an image data **S21** having m bits into an image data **S22** having n bits ($n < m$).

FIG. 14 shows, in comparison to FIG. 8, an example of the data of the two-type characteristics **801** and **802** which are written in the lookup table **901**. The solid line and dotted line represent the image data **S21** having m bits. The dots along the solid line and the dotted line represent the image data **S22** having n bits. Similarly to the first embodiment (FIG. 11), the image data **S21** is mapped to the image data **S22** such that the liquid crystal driving voltage levels (the vertical axis of FIG. 14) corresponding to the image data **S22** are in regular intervals. For example, upon conversion from the characteristic **801** to the characteristic **802**, the lookup table **901** is set in a manner that the liquid crystal driving voltage is identical between the characteristic **801** before the conversion and the characteristic **802** after the conversion.

The DAC in the reference power supply circuit **113** and the reference power supply generating part in the data driver **115** are resistance dividing circuits, so that the reference power supply conversion calculator **1201** can change the content of the lookup table **901** with simple calculations.

The image data **S22** is written in a frame memory **202**. The frame memory **202** stores one-frame amount of the image data **S22**. The frame memory **202** delays the image data **S22** for one frame, and outputs an image data **S23**. A comparison circuit **211** compares the image data **S22** of the present frame and the image data **S23** of the antecedent frame, and outputs a difference data **S24** thereof.

Here, the values for the difference data **S24** differ depending on the characteristics **801** and **802**. In order for a correct table **212** common for the characteristics **801** and **802** to be usable, the inverse-conversion lookup table **1202** is provided.

The inverse-conversion lookup table **1202** inversely converts the difference data **S24** according to the image data **S23**,

and outputs a difference data **S25**. The inverse-lookup table **1202** performs the inverse-conversion with respect to the conversion by the lookup table **901**. The difference data **S24** is inversely converted to the level of the inputted image data **S21** regardless its characteristic is **801** or **802**. The reference power supply conversion calculator **1201** calculates the contents of the lookup tables **901** and **1202**, and rewrites them in the pair form, according to the control signal **S28**. Note that the inverse-conversion lookup table **1202** may perform the inverse-conversion based on the image data **S21** or **S22** instead of the image data **S23**.

The correction table **212** stores one correction data which is common for the characteristics **801** and **802**, corrects the difference data **S25** based on the image data **S21**, and outputs a difference data **S26**. Note that the correction table **212** can perform correction according to the image data **S22** or **S23** instead of the image data **S21**. A correction calculating circuit **213** adds the image data **S21** and the difference data **S26** and outputs the image data **S27**. Consequently, the high-speed-response driving shown in FIG. 5 can be realized.

According to the second embodiment, the gamma character can be switched frame by frame based on a one-frame amount of image data. By converting the image data by the lookup table **901** and thereafter inversely converting it by the inverse-conversion lookup table **1202**, a common correction table **212** can be used. The need to use different correction tables **212** depending on the characteristics **801** and **802** can be eliminated. This effect is significant, in particular where there are a number of switchable characteristics. The ROM **203** no longer needs to store a vast amount of correction data for switching the correction tables **212**.

As has been described, with the first and the second embodiments, the amount of frame memory **202** can be reduced by converting the first image data into a second image data having fewer bits. Further, in the relation curve between the image data and the liquid crystal driving voltage in FIG. 11, the image data is mapped so as to be fine in a sharp curve portion, and rough in a moderate curve portion. In other words, the resolution can be enhanced with respect to an important portion of the image, allowing a high-quality image display. Moreover, by correcting the difference data with the correction table **212**, the high-speed-response driving comes to be possible as shown in FIG. 5.

The conversion of the first image data into the second image data having a small number of bits allows reduction of the amounts of the frame memory. At the same time, the conversion into the second image memory can be carried out such that, in a relation curve between the image data and the liquid crystal driving voltage, a sharp curve portion is converted to a fine image, while a moderate curve portion is converted to a rough image. Further, the correction of the difference data according to any of the first to the third image data allows a high-speed-response driving of the liquid crystal.

The present embodiment is to be considered in all respects as illustrative and no restrictive, and all changes which come within the meaning and range of equivalency of the claims are therefore intended to be embraced therein. The invention may be embodied in other specific forms without departing from the spirit or essential characteristics thereof.

What is claimed is:

1. A liquid crystal display, comprising:

a conversion circuit to convert a first image data to a second image data having a fewer number of bits than said first image data;

a frame memory to store the second image data;

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a difference circuit to output in units of pixel a difference data between the second image data of a present frame and a third image data of an antecedent frame outputted from said frame memory;

a correction circuit to change said difference data based on one of the first to third image data; and

an adding circuit to add the difference data which is changed and the first image data, wherein

said conversion circuit maps the first image data nonlinearly to the second image data in irregular intervals.

2. The liquid crystal display according to claim 1, wherein said conversion circuit is a conversion table to store a correspondence for the first image data and the second image data.

3. The liquid crystal display according to claim 1, wherein said conversion circuit maps the first data to the second data in such a manner that levels of a liquid crystal driving voltage corresponding to the second image data are in regular intervals.

4. The liquid crystal display according to claim 1, wherein said conversion circuit maps the first image data to the second image data in such a manner that response speeds to level changes of the liquid crystal driving voltage corresponding to the second image data are in regular intervals.

5. The liquid crystal display according to claim 1, wherein said conversion circuit changes a mapping method from the first image data to the second image data upon a change in a relationship between image data and a liquid crystal driving voltage.

6. The liquid crystal display according to claim 1, wherein said conversion circuit can perform different mappings depending on each frame.

7. The liquid crystal display according to claim 1, further comprising an inverse-conversion circuit to inversely convert the difference data based on one of the first to third image data, and wherein said correction circuit corrects the difference data which is inversely converted.

8. The liquid crystal display according to claim 7, further comprising a control circuit to change a conversion method of said conversion circuit and an inverse-conversion method of said inverse-conversion circuit, according to a control signal.

9. The liquid crystal display according to claim 8, wherein said control circuit changes the conversion method of said

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conversion circuit and the inverse conversion method of said inverse-conversion circuit in pair upon a change in the relationship between the image data and the liquid crystal driving voltage.

10. The liquid crystal display according to claim 9, wherein said conversion circuit and said inverse-conversion circuit can perform conversion and inverse conversion respectively according to each frame.

11. The liquid crystal display according to claim 10, wherein the conversion method and the inverse-conversion method are determined depending on a tone distribution of one-frame data of the first or the second image data.

12. The liquid crystal display according to claim 11, further comprising a reference voltage generating circuit for generating plural types of liquid crystal reference driving voltages, and wherein said control circuit changes the liquid crystal driving voltages generated according to a control signal.

13. The liquid crystal display according to claim 12, wherein said reference voltage generating circuit is a digital-analog-converter-type amplifier.

14. The liquid crystal display according to claim 13, wherein said conversion circuit maps the first image data to the second image data in irregular intervals.

15. The liquid crystal display according to claim 14, wherein said conversion circuit maps the first image data to the second image data in such a manner that levels of the liquid crystal driving voltage corresponding to the second image data are in regular intervals.

16. The liquid crystal display according to claim 14, wherein said conversion circuit maps the first image data to the second image data in such a manner that response speeds to level changes of the liquid crystal driving voltage corresponding to the second image data are in regular intervals.

17. The liquid crystal display according to claim 1, wherein the first to the third image data include red, green, and blue image data, and

wherein the red, green, and blue image data do not share a common number of bits.

18. The liquid crystal display according to claim 17, wherein the green image data has a greater number of bits than the red and blue image data.

* * * * *

专利名称(译)	液晶显示器及其处理方法		
公开(公告)号	US8803774	公开(公告)日	2014-08-12
申请号	US10/976955	申请日	2004-10-29
[标]申请(专利权)人(译)	富士通显示技术股份有限公司 友达光电股份有限公司		
申请(专利权)人(译)	富士通显示器科技股份有限公司 友达光电股份有限公司		
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IPC分类号	G02F1/133 G09G3/36 G09G3/20 G09G5/10		
CPC分类号	G09G3/3648 G09G2360/18 G09G2310/04 G09G2320/0233 G09G2320/103 G09G2320/0285 G09G2310/027 G09G2340/0407		
优先权	2004134204 2004-04-28 JP		
其他公开文献	US20050243075A1		
外部链接	Espacenet USPTO		

摘要(译)

提供一种液晶显示器，包括：将第一图像数据转换为第二图像数据的转换电路；存储第二图像数据的帧存储器；以像素为单位输出的差分电路，第二图像数据之间的差数据要被转换的当前帧和要从帧存储器输出的先行帧的第三图像数据，用于基于第一到第三图像数据之一校正差数据的校正电路，以及用于添加校正的差异数据和第一图像数据。

