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(12) United States Patent
Yeon et al.**(10) Patent No.: US 7,129,921 B2****(45) Date of Patent: *Oct. 31, 2006****(54) GRAY VOLTAGE GENERATION CIRCUIT
FOR DRIVING A LIQUID CRYSTAL
DISPLAY RAPIDLY****(75) Inventors:** Yeun-Mo Yeon, Seoul (KR); Kun-Bin
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Suwon-si (KR)**(*) Notice:** Subject to any disclaimer, the term of this
patent is extended or adjusted under 35
U.S.C. 154(b) by 128 days.This patent is subject to a terminal dis-
claimer.**(21) Appl. No.: 10/747,665****(22) Filed: Dec. 30, 2003****(65) Prior Publication Data**

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Related U.S. Application Data**(63)** Continuation of application No. 09/956,146, filed on
Sep. 20, 2001, now Pat. No. 6,670,935.**(30) Foreign Application Priority Data**

Dec. 21, 2000 (KR) 2000-79698

(51) Int. Cl.
G09G 3/36 (2006.01)**(52) U.S. Cl.** 345/89; 345/92; 345/96;
345/99**(58) Field of Classification Search** 345/87-104,
345/690-691, 204, 208-210
See application file for complete search history.**(56) References Cited****U.S. PATENT DOCUMENTS**

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Primary Examiner—Lun-yi Lao**(57) ABSTRACT**

A gray voltage generation circuit for driving a liquid crystal display rapidly outputs an altered gray voltage so that a source driving circuit can charge liquid crystal capacitors constructed in a liquid crystal panel in a short period of time. In response to the gray voltages from the gray voltage generation circuit, while driving a positive polarity, the source driving circuit generates a liquid crystal driving voltage of higher level than the existing liquid crystal driving voltage when applying a gate clock signal of high level, and generates a liquid crystal driving voltage of a level similar to the existing liquid crystal driving voltage when applying a gate clock signal of low level. And, while driving a negative polarity, the source driving circuit generates a liquid crystal driving voltage of lower level than an existing liquid crystal driving voltage when applying a gate clock signal of high level, and generates a liquid crystal driving voltage of a level similar to the existing liquid crystal driving voltage when applying a gate clock signal of low level.

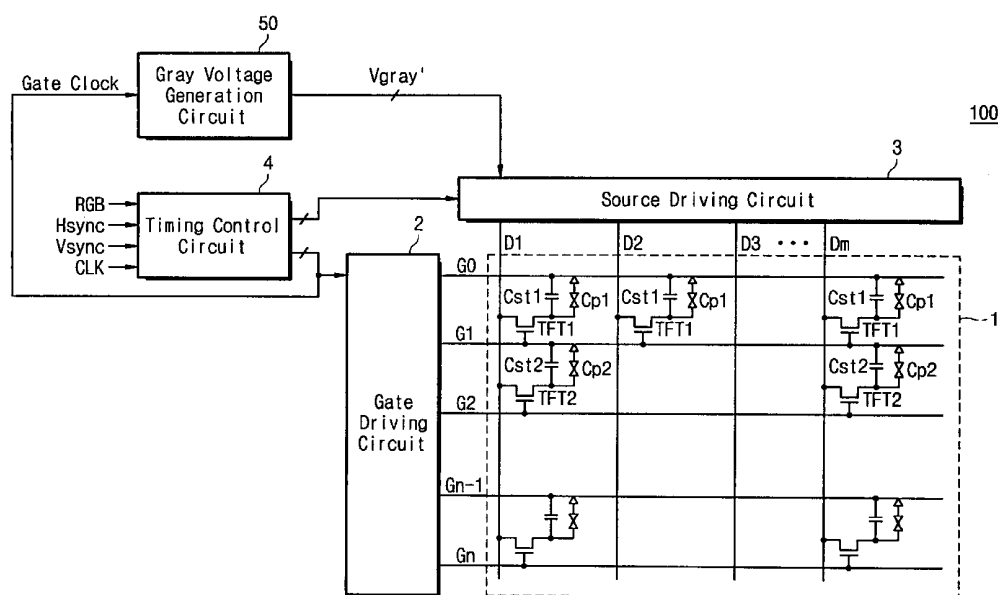
13 Claims, 13 Drawing Sheets

Fig. 1

(Prior Art)

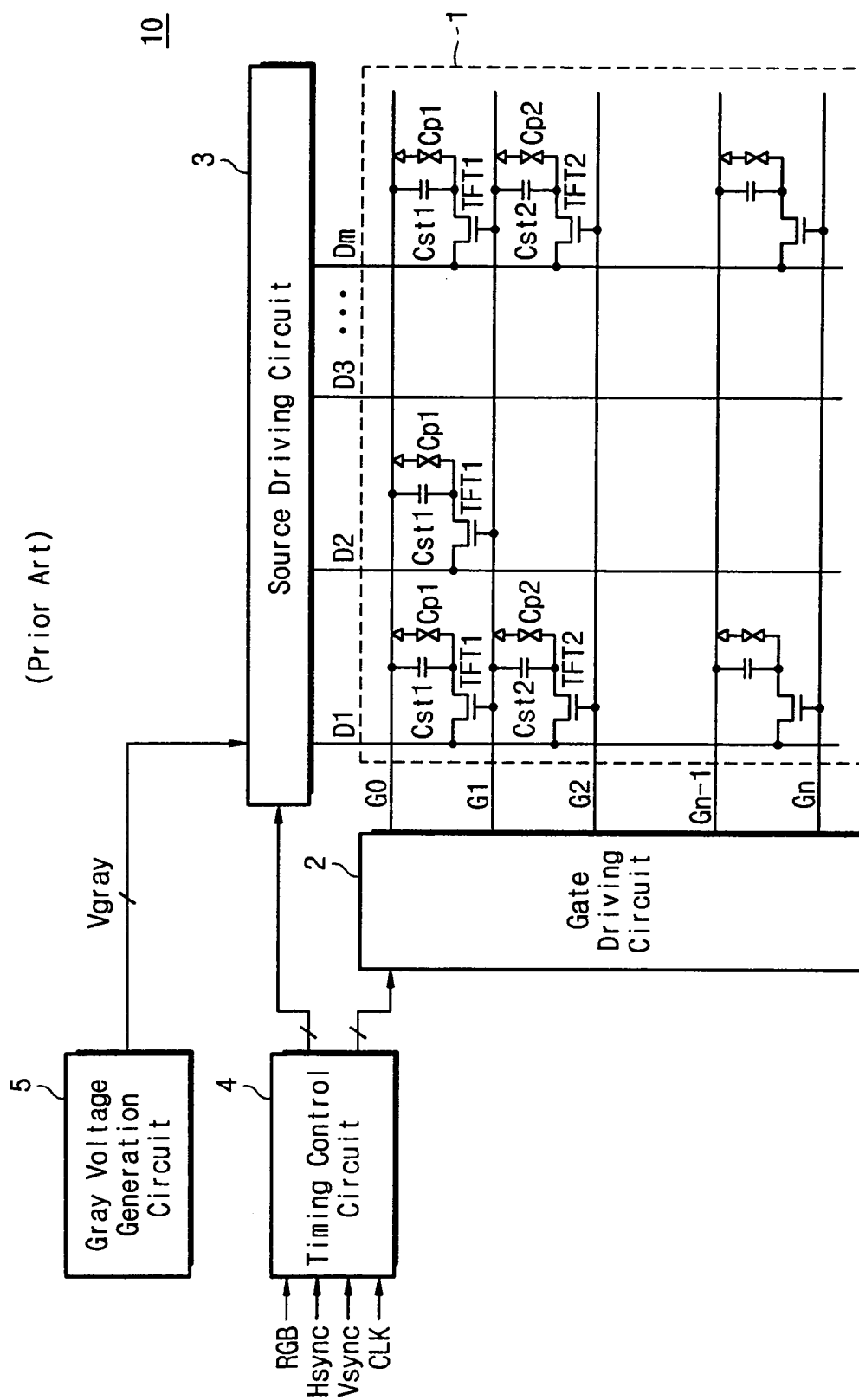


Fig. 2

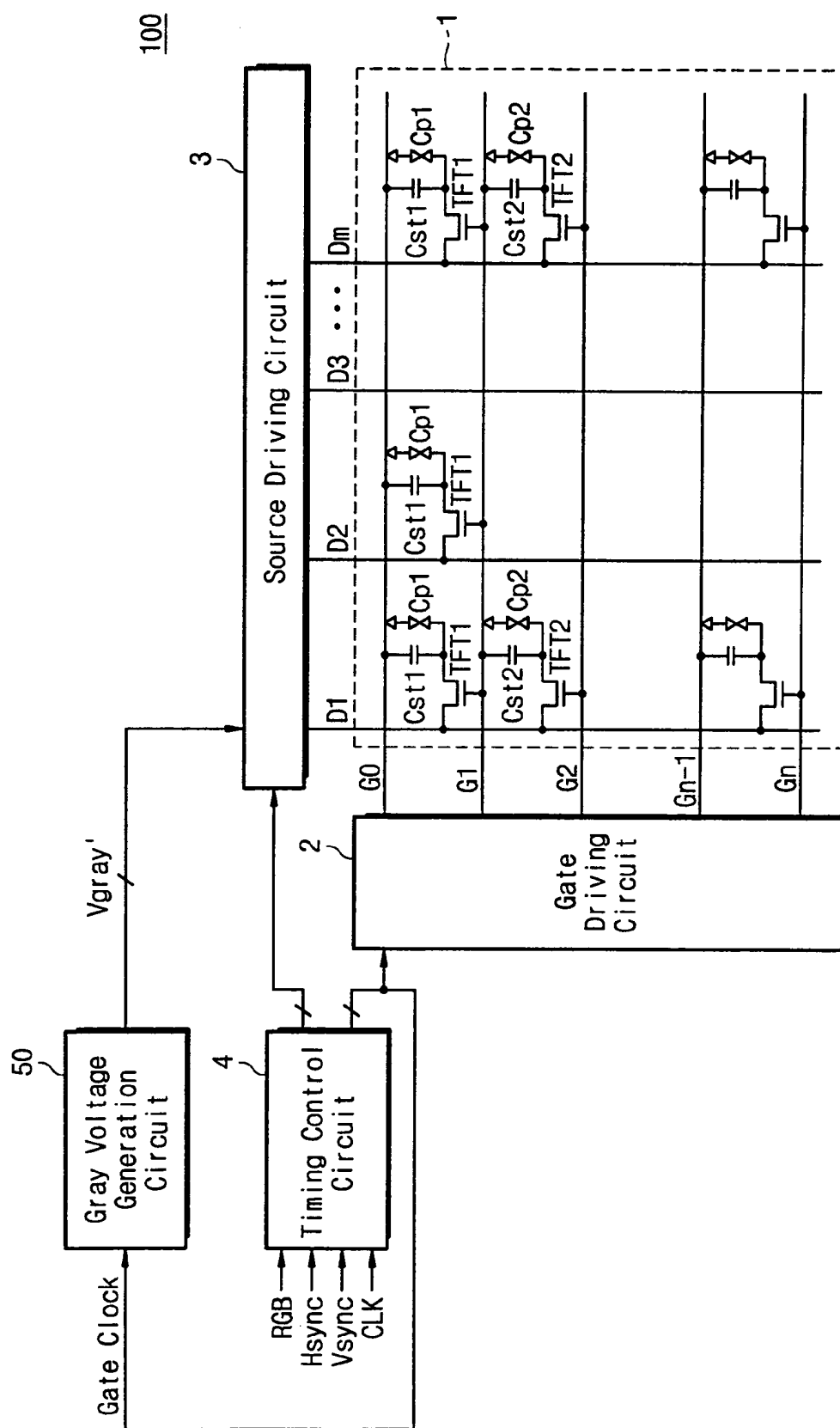


Fig. 3

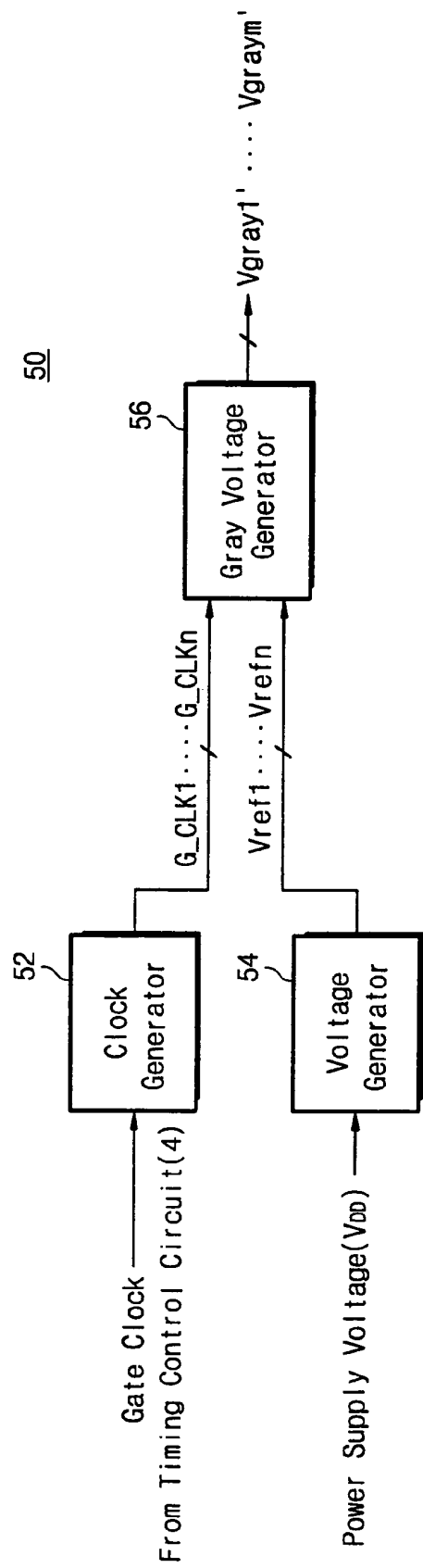


Fig. 4

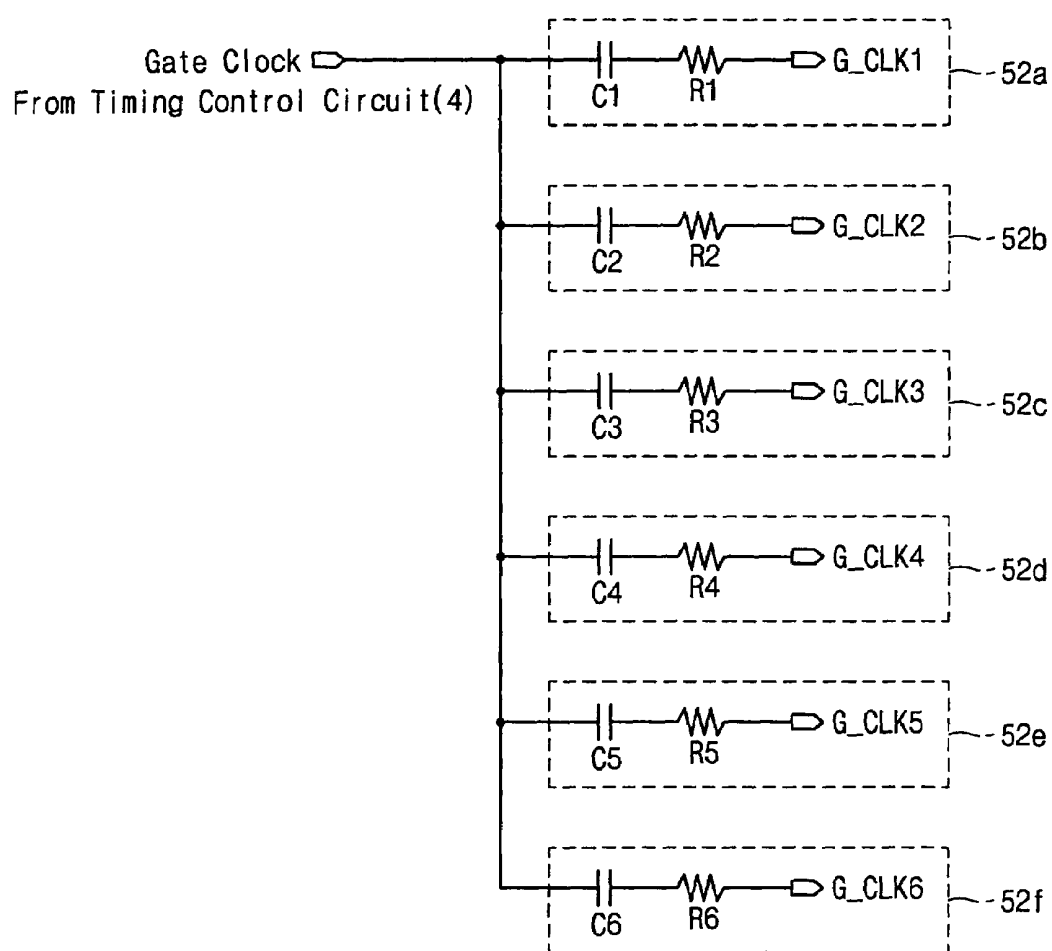
52

Fig. 5

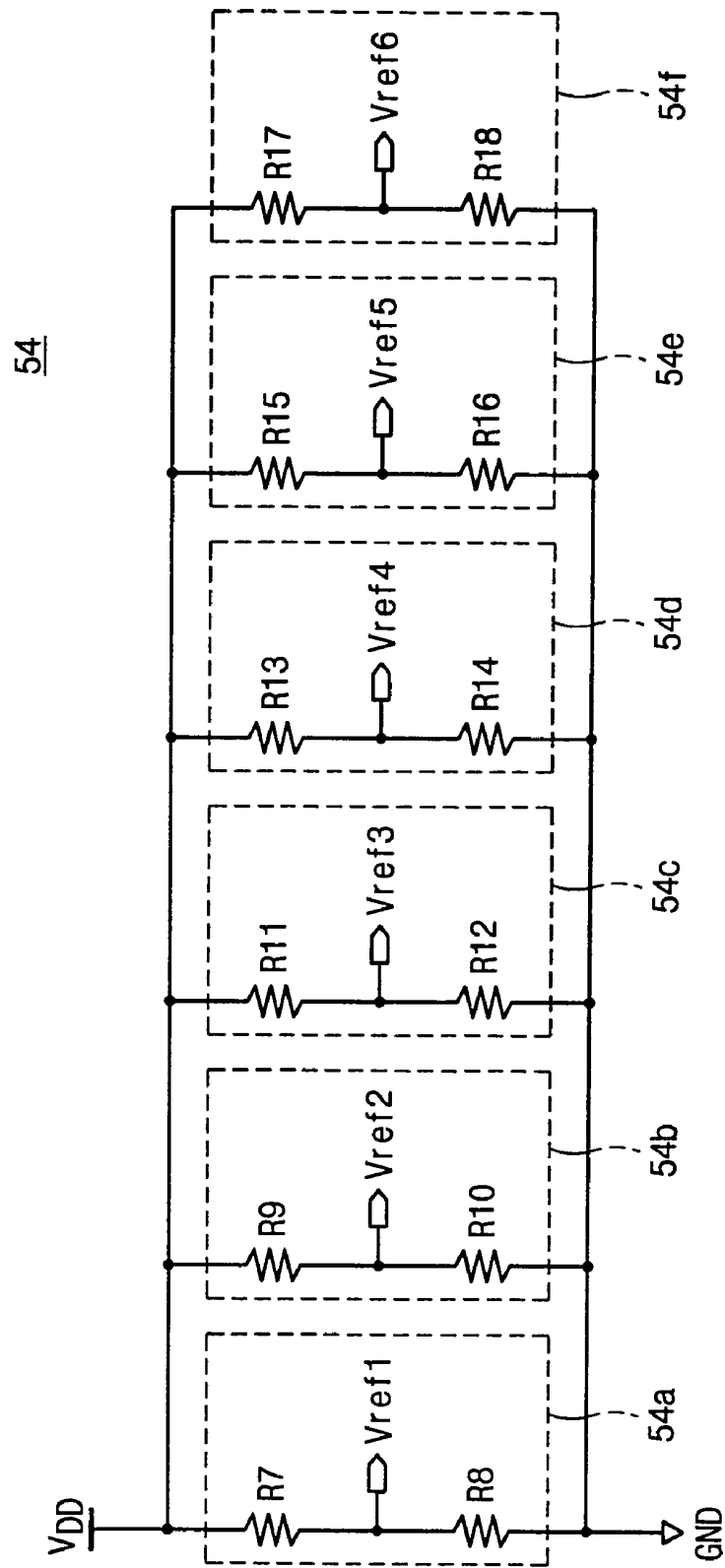


Fig. 6

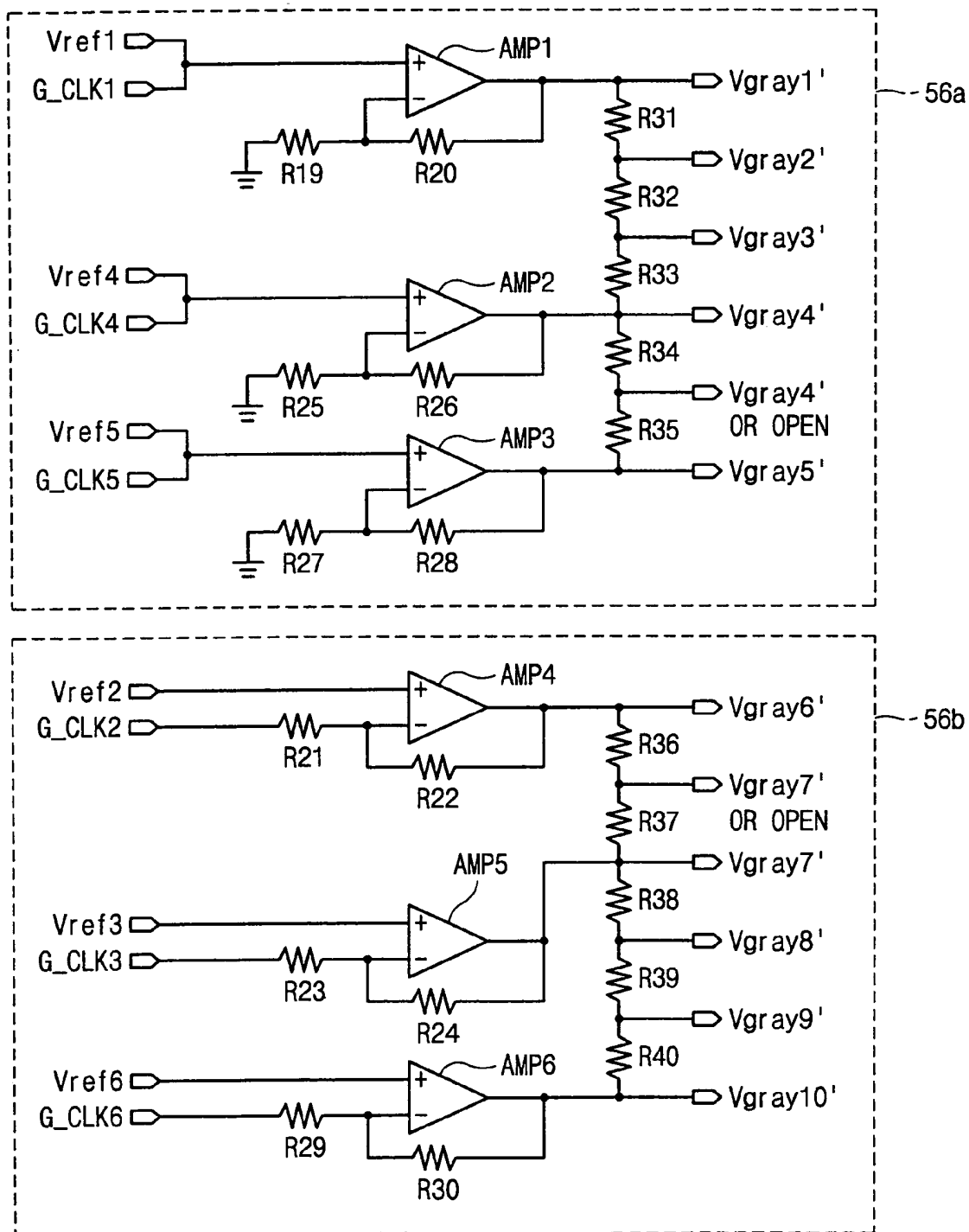
56

Fig. 7A

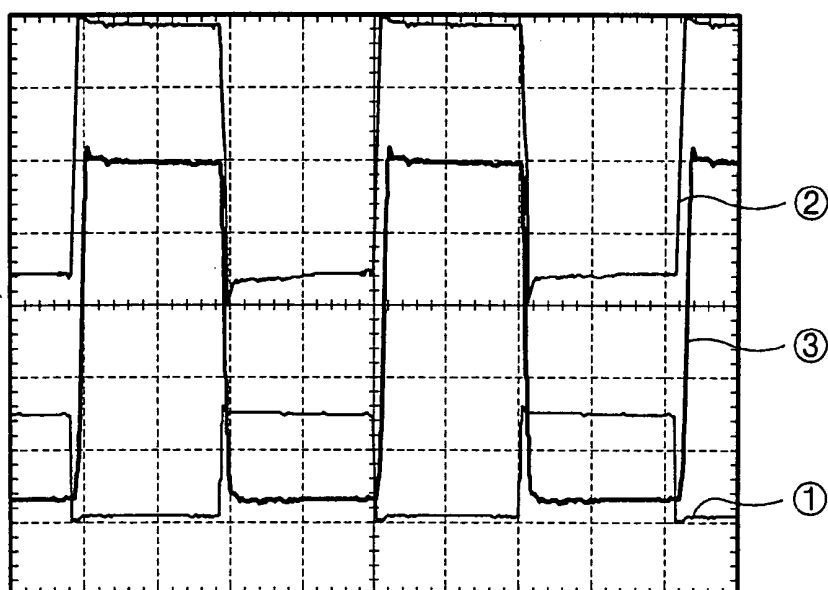


Fig. 7B



Fig. 8

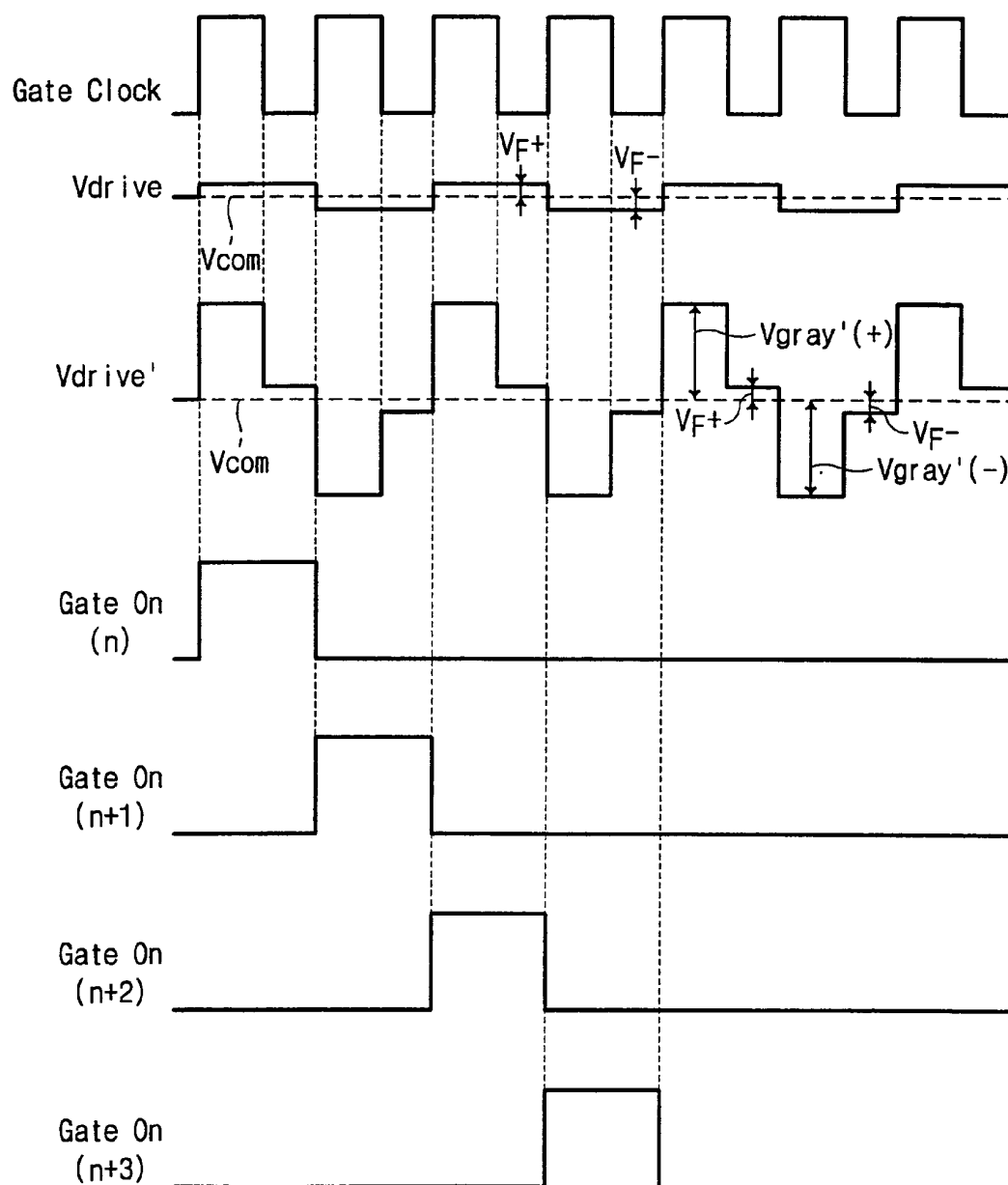


Fig. 9

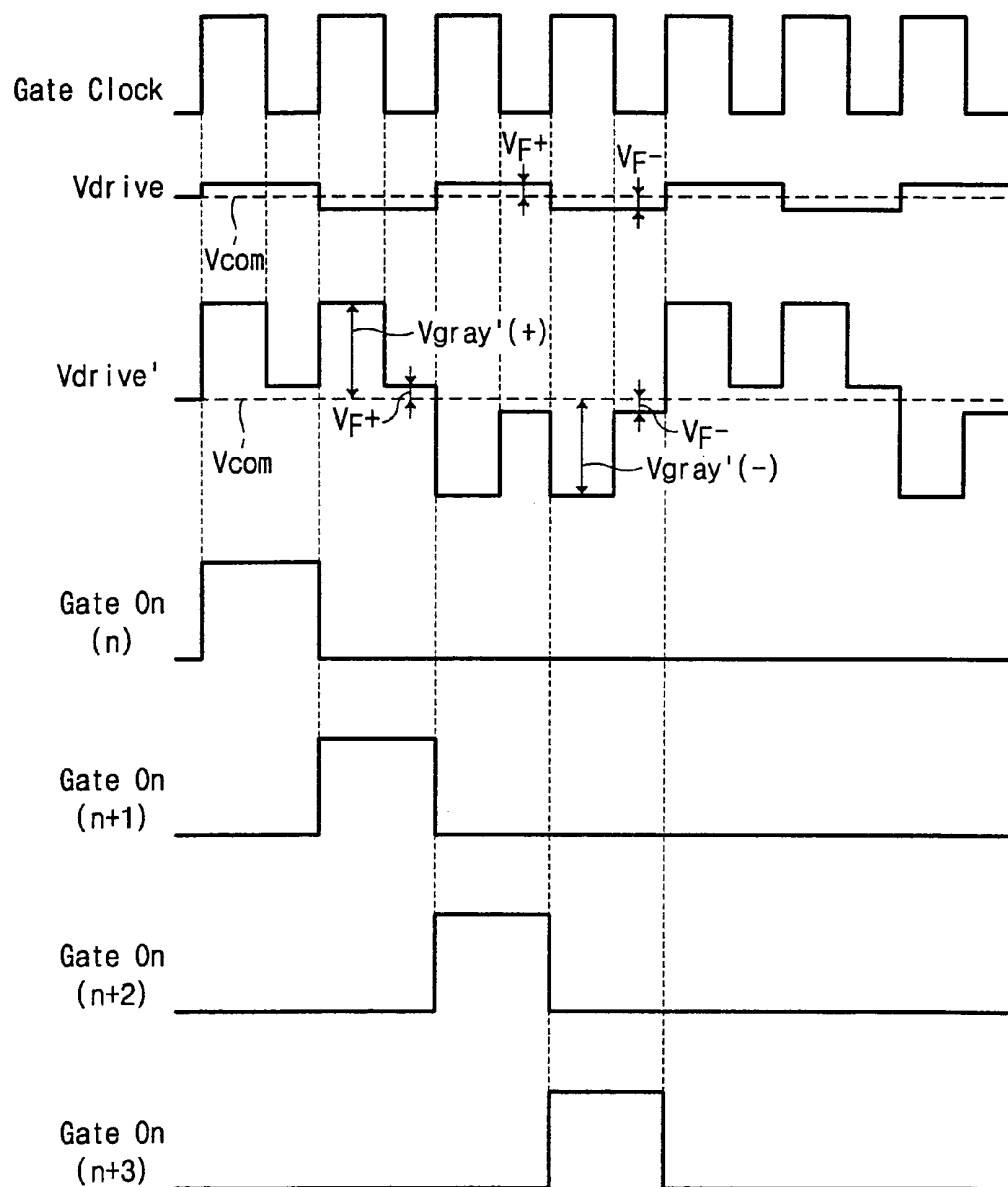


Fig. 10A

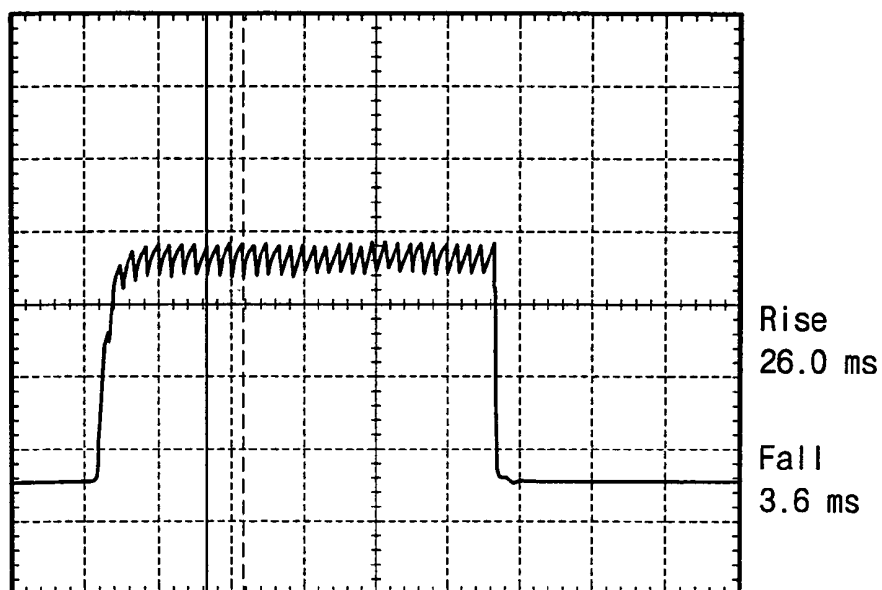


Fig. 10B

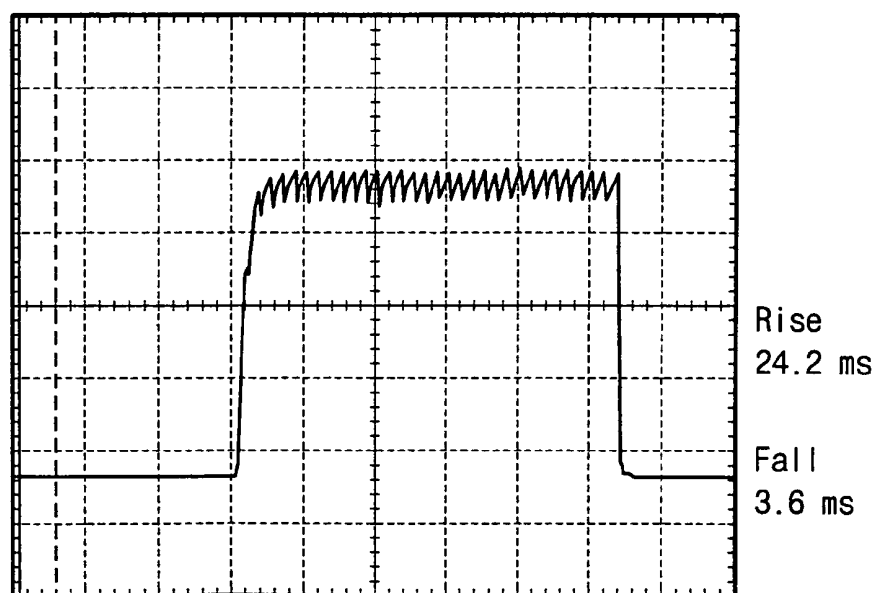


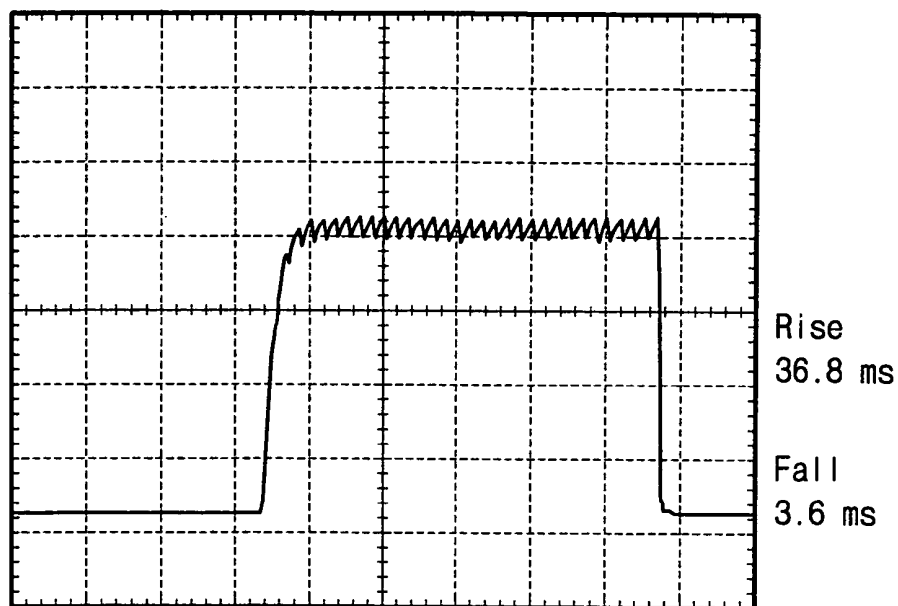
Fig. 11A**Fig. 11B**

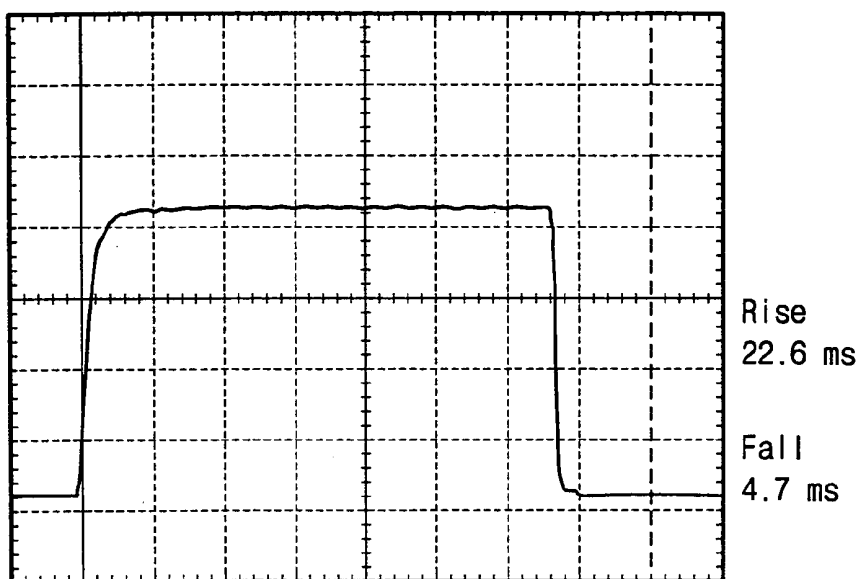
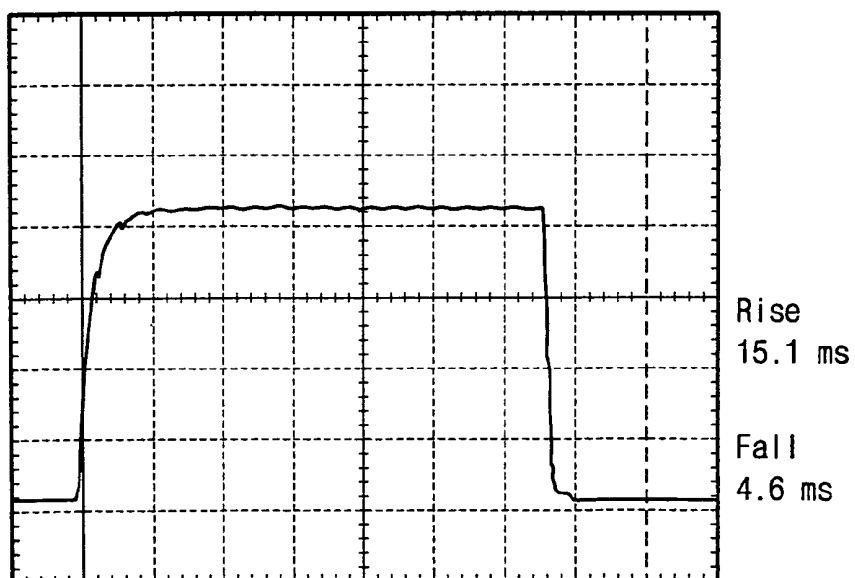
Fig. 12A**Fig. 12B**

Fig. 13A

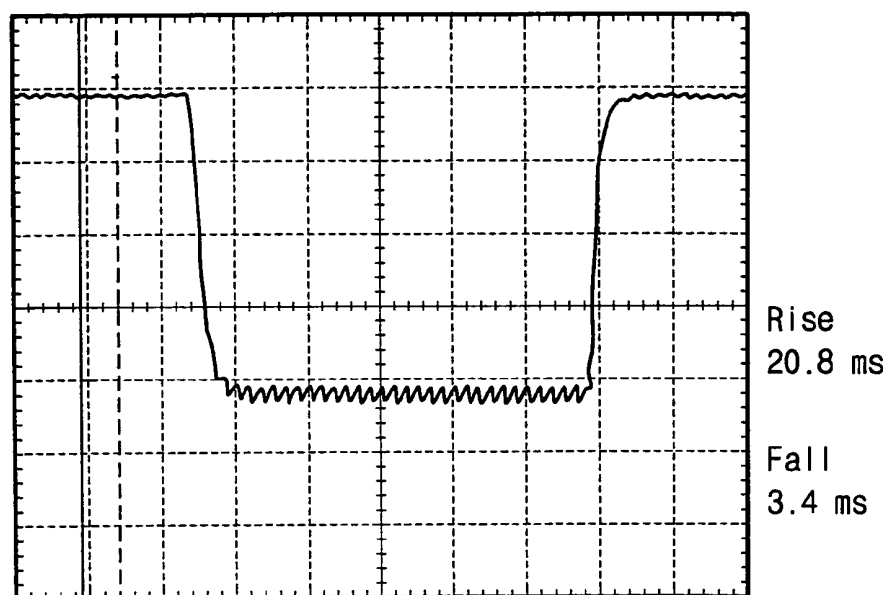
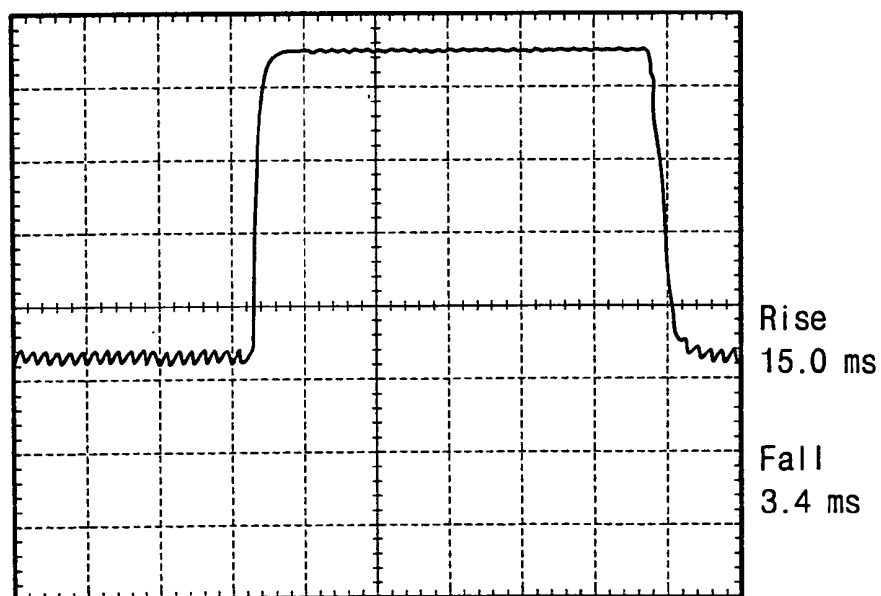


Fig. 13B



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GRAY VOLTAGE GENERATION CIRCUIT FOR DRIVING A LIQUID CRYSTAL DISPLAY RAPIDLY

This application is continuation of Ser. No. 09/956,146
filed Sep. 20, 2001 now U.S. Pat. No. 6,670,935.

FIELD OF THE INVENTION

The present invention relates to a liquid crystal display
and, more particularly, to a gray voltage generation circuit
for driving a liquid crystal display and such a liquid crystal
display.

BACKGROUND OF THE INVENTION

Generally, a liquid crystal is an organic compound having
a neutral property between liquid and crystal, and changes in
its color or transparency by voltage or temperature. A liquid
crystal display (LCD), which expresses information using the
liquid crystal, occupies a smaller volume and has a lower
power consumption than a conventional display device.
Therefore, lots of attentions are paid to the LCD as a novel
display device.

FIG. 1 schematically illustrates a configuration of a con-
ventional liquid crystal display. A liquid crystal display 10
includes a liquid crystal panel 1, a gate driving circuit 2
coupled to the liquid crystal panel 1, a source driving circuit
3, a timing control circuit 4, and a gray voltage generation
circuit (or gamma reference voltage generation circuit) 5.

The liquid crystal panel 1 is made of a plurality of gate
lines G0 through Gn and a plurality of data lines D1 through
Dm that are vertically interconnected with the gate lines,
respectively. The gate driving circuit 2 is connected to each
of the gate lines G0 through Gn, and the source driving
circuit 3 is connected to each of the data lines D1 through
Dm. One pixel is composed in each interconnection of the
gate lines and the data lines. Each pixel is made of one thin
film transistor (TFT), one storing capacitor Cst, and one
liquid crystal capacitor Cp. Each of pixels composing the
liquid crystal panel 1 further includes three sub-pixels
corresponding to red (R), green (G), and blue (B). A pixel
displayed via the liquid crystal panel 1 is obtained by
combination of R, G, and B color filters. The liquid crystal
display 10 can display not only color pictures but also pure
red, green, blue, and gray scales by combining those pixels.

The timing control circuit 4 issues control signals (e.g.,
gate clock and gate on signals) required in the gate driving
circuit 2 and the source driving circuit 3 in response to color
signals R, G, and B, horizontal and vertical synch signals
HSync and Vsync, and a clock signal CLK. The gray voltage
generation circuit 5 is connected to the source driving circuit
3, generating a gray voltage Vgray or a gamma reference
voltage that is a reference to generate a liquid crystal driving
voltage Vdrive. One example of the gray voltage generation
circuit 5 is disclosed in U.S. Pat. No. 6,067,063 entitled
"LIQUID CRYSTAL DISPLAY HAVING A WIDE VIEW
ANGLE AND METHOD FOR DRIVING THE SAME",
issued to Kim et al., issued on May 23, 2000. A gray voltage
generation circuit 5 disclosed therein includes a plurality of
resistors R1 through Rn+1 that are directly coupled between
a power supply voltage (Vcc) and a ground (GND). Each of
the resistors R1 through Rn+1 distributes the power supply
voltage (Vcc) with a predetermined ratio, generating n-bit
gray voltages VG1 through VGn.

Now, operations of the liquid crystal display 10 having
such a configuration will be described in detail. If the gate

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driving circuit 2 sequentially scans pixels of the panel row
by row, the source driving circuit 3 generates a liquid crystal
driving voltage Vdrive based upon the color signals R, G,
and B inputted through the timing control circuit 4, in
response to the reference voltage Vgray outputted from the
gray voltage generation circuit 5. And then, the source drive
3 applies the generated voltage Vdrive to the panel 1 each
time of scanning.

In such an operation, the TFT acts as a switch. For
example, when the TFT is turned on, the liquid crystal
capacitor Cp is charged by the liquid crystal driving voltage
Vdrive generated from the source driving circuit 3. When
the TFT is turned off, the capacitor Cp prevents the charged
voltage from leaking. This shows that the liquid crystal
driving voltage Vdrive applied from the source driving
circuit 3 has a great influence upon driving each TFT
composing the panel 1.

As the liquid crystal display tends to implement high
speed response, it is required to enhance a response speed of
such a liquid crystal display Cp in order to speed up the
device. This is because if the voltage Vdrive applied from
the source driving circuit 3 has a high value, the capacitor Cp
would quickly be charged to enhance a total driving speed
of a liquid crystal display.

There are many methods of boosting a liquid crystal
driving voltage Vdrive applied from the source driving
circuit 3 in order to enhance a driving speed of the liquid
crystal display. For example, it requires a design change of
the gate driving circuit 2 or the source driving circuit to
generate a liquid crystal driving voltage Vdrive of high
level, or a design change of the timing control circuit 4 for
issuing a control signal to the driving circuits 2 and 3.
Unfortunately, changing designs of such high-priced circuits
causes higher costs in a production unit. Furthermore, the
increased liquid crystal driving voltage Vdrive also
increases power consumption of the liquid crystal display in
proportion to the voltage Vdrive rise.

Accordingly, the object of the present invention is to
overcome the foregoing drawbacks, and to provide a gray
voltage generation circuit that can enhance a driving speed
of a liquid crystal display with low cost and power con-
sumption.

SUMMARY OF THE INVENTION

To attain this object, there is provided a liquid crystal
display that includes a liquid crystal panel having a plurality
of pixels, a gray voltage generation circuit for generating a
plurality of gray voltages corresponding to data to be
displayed in the liquid crystal panel, a timing control circuit
for issuing a gate clock signal and a plurality of control
signals, a gate driving circuit for sequentially scanning the
pixels row by row in response to the gate clock signal, and
a source driving circuit for generating a liquid crystal
driving voltage in response to the data and applying the
generated liquid crystal driving voltage to the panel each
time of scanning. In response to the gray voltage, the source
driving circuit generates a liquid crystal driving voltage that
has different values in high and low level intervals.

BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 is a block diagram showing a configuration of a
conventional liquid crystal display.

FIG. 2 is a block diagram showing a configuration of a
liquid crystal display in accordance with the present inven-
tion.

FIG. 3 is a block diagram showing a configuration of a gray voltage generation circuit in accordance with the present invention.

FIG. 4 is a circuit diagram showing a detailed configuration of a clock generator shown in FIG. 3.

FIG. 5 is a circuit diagram showing a detailed configuration of a voltage generator shown in FIG. 3.

FIG. 6 is a circuit diagram showing a detailed configuration of a gray voltage generation circuit shown in FIG. 3.

FIGS. 7A and 7B are waveform diagrams showing one example of waveforms of gray voltages that are generated from a gray voltage generation circuit in accordance with the present invention.

FIGS. 8 and 9 are waveform diagrams showing one example of waveforms of outputs of a source driving circuit, which are generated by applying the gray voltage shown in FIGS. 7A and 7B.

FIGS. 10A, 10B, 11A, 11B, 12A, 12B, 13A and 13B are timing diagrams showing response speed measuring results of 0–32, 0–48, 0–64, and 32–84 grays of the source driving circuits by means of the gray voltage shown in FIGS. 7A and 7B.

DESCRIPTION OF THE PREFERRED EMBODIMENT

A new and improved gray voltage generation circuit of a liquid crystal display is provided to the present invention. The gray voltage generation circuit generates a high-potential liquid crystal driving voltage for a predetermined interval so that liquid crystal capacitors may be charged in a short time, and alters and outputs a gray voltage after the predetermined interval in order to generate a normal liquid crystal driving voltage. As a result, a driving speed of the liquid crystal display can be enhanced.

FIG. 2 schematically illustrates a configuration of a liquid crystal display 100 according to the present invention. The liquid crystal display 100 includes a liquid display panel 1, a plurality of gate driving circuits 2 coupled to the panel 1, a plurality of source driving circuits 3, a timing control circuit 4, and a gray voltage generation circuit 50. Such a configuration is identical to the configuration of the conventional liquid crystal display shown in FIG. 1, except for a gray voltage generation circuit 50 for generating a gray voltage Vgray' in response to a gate clock signal Gate Clock issued from a timing control circuit. Same numerals denote same elements throughout the drawings, and their description will be skipped herein so as to avoid duplicate description.

It is well known that the source driving circuit 3 selects one of a plurality of gray voltages according to color signals (R, G, and B), and applies a liquid crystal driving voltage Vdrive to a liquid crystal panel in response to the selected one gray voltage. A function of the source driving circuit 3 is closely bound up with a charging speed of the liquid crystal display Cp constructed in the liquid crystal panel 1. The liquid crystal driving voltage Vdrive is dependent upon the gray voltage Vgray' generated from the gray voltage generation circuit 50. Therefore, a liquid crystal display 100 of the invention changes a liquid crystal driving voltage Vdrive generated from the source driving circuit 3 so as to enhance a charging speed of the liquid crystal capacitor Cp constructed in the panel 1. Without modifying designs of expensive and complex circuits such as the gate driving circuit 2, the source driving circuit 3, and the timing control circuit 4, a gray voltage generation circuit 50 of much lower

price than the above circuits is made to enhance a driving speed of the liquid crystal display 100.

FIG. 3 schematically illustrates a configuration of a gray voltage generation circuit according to the present invention. A gray voltage generation circuit 50 includes a clock generator 52, a voltage generator 54, and a gray voltage generator 56. The clock generator 52 generates n-bit clock signals G_CLK1, . . . , and G_CLKn that are not overlapped with each other, in response to a gate clock signal GATE CLOCK. The voltage generator 54 generates n-bit reference voltages Vref1, . . . , and Vrefn each having different level, in response to a power supply voltage VDD that is an analog signal and is used as a power supply voltage of a source driving circuit 3.

If the n-bit clock signals G_CLK1, . . . , and G_CLKn and the n-bit reference voltages Vref1, . . . , and Vrefn are inputted to the gray voltage generator 56, the gray voltage generator 56 generates m-bit gray voltages Vgray1', . . . , and Vgraym' that are synchronized with the clock signals G_CLK1, . . . , and G_CLKn to have different potentials based upon levels of the reference voltages Vref1, . . . , and Vrefn. Although described in detail hereinbelow, the gray voltages Vgray1', . . . , and Vgraym' makes the source driving circuit 3 generate a liquid crystal driving voltage Vdrive' that has different values in high and low intervals of the clock signal CLOCK during one period of the gate clock GATE CLCK. The liquid driving voltage Vdrive' of the source driving circuit 3 having such a characteristic can enhance a driving speed of a liquid crystal display 100.

FIGS. 4, 5 and 6 illustrate the clock generator 52, the voltage generator 54, and the gray voltage generator 56 that are shown in FIG. 3, respectively. The clock generator 52 issues six clock signals C_CLK1, . . . , and C_CLK6. The voltage generator 54 generates six reference voltages Vref1, . . . , and Vref6. And, the gray voltage generator 56 generates ten clock signals G_CLK1', . . . , and G_CLK10' in response to the six clock signals C_CLK1, . . . , and C_CLK6 and the six reference voltages Vref1, . . . , and Vref6. According to a circuit configuration, the number of generated signals can be changed. The circuits shown in the drawings are merely one example of the circuit configuration.

Referring now to FIG. 4, the clock generator 52 consists of an input terminal for receiving a gate clock signal GATE CLOCK generated from the timing control circuit 4, first and sixth clock generation units 52a–52f each being coupled to the input terminal in parallel, and first and sixth output terminals each being coupled to the units 52a–52f. Each of the units 52a–52f has a capacitor C1, . . . , or C6 and a resistor R1, . . . , or R6 that are serially connected between the input terminal and the output terminal. And, each of the units 52a–52f outputs first and sixth clock signals G_CLK1, . . . , and G_CLK6 not to be overlapped with each other. A period of the clock signals G_CLK1, . . . , and G_CLK6 is identical to that of the gate clock signal GATE CLOCK generated from the timing control circuit 4.

Referring to FIG. 5, the voltage generator 54 consists of six voltage generation units 54a–54f for generating six reference voltages Vref1, . . . , and Vref6 by dividing a power supply voltage VDD at a predetermined ratio to generate six reference voltages of different levels. The units 54a–54f are connected between the power supply voltage VDD and a ground voltage GND in parallel. Each of the units 54a–54f includes two resistors serially connected between VDD and GND, and an output terminal coupled to a contact point between the resistors.

Referring to FIG. 6, the gray voltage generator **56** consists of first and second gray voltage generation units **56a** and **56b**. The first gray voltage unit **56a** generates first to fifth gray voltages Vgray1', . . . , and Vgray5' that are used to drive a positive polarity of a liquid crystal. The second gray voltage unit **56b** generates sixth to tenth gray voltages Vgray6', . . . , and Vgray10' that are used to drive a negative polarity of a liquid crystal.

The first gray voltage unit **56a** includes first to sixth input terminals for receiving clock signals G_CLK1, G_CLK4, and G_CLK5 generated from a clock generator **52** and reference voltages Vref1, Vref4, and Vref5 generated from a voltage generator **54**. It also includes a first amplifier AMP1, a second amplifier AMP2 and a third amplifier AMP3 for respectively adding and amplifying G_CLK1, G_CLK4, and G_CLK5 to a predetermined ratio to generate gray voltages Vgray1', Vgray4', and Vgray5', and output terminals for outputting Vgray1', Vgray4', and Vgray5'. The first amplifier circuit AMP1 adds G_CLK1 to Vref1, and amplifies it to a predetermined ratio to generate Vgray1'. The second amplifier circuit AMP2 adds G_CLK4 to Vref4, and amplifies it to a predetermined ratio to generate Vgray4'. And, the third amplifier circuit AMP3 adds G_CLK5 to Vref5, and amplifies it to a predetermined ratio to generate Vgray5'.

The gray voltages Vgray1', Vgray4', and Vgray5' are given by the following equations;

$$V_{gray1'} = \frac{R19 + R20}{R19} \left[V_{ref1} + \frac{R1}{R1 + R19} V_{G_CLK1} \right] \quad \text{<Equation 1>}$$

$$V_{gray4'} = \frac{R25 + R26}{R25} \left[V_{ref4} + \frac{R4}{R4 + R25} V_{G_CLK4} \right] \quad \text{<Equation 2>}$$

$$V_{gray5'} = \frac{R27 + R28}{R27} \left[V_{ref5} + \frac{R5}{R5 + R27} V_{G_CLK5} \right] \quad \text{<Equation 3>}$$

wherein V_{G_CLKn} represents an alternative element of a gate clock signal GATE CLOCK.

The first gray voltage generation unit **56a** generates second and third gray voltages Vgray2' and Vgray3', as well as Vgray1', Vgray4', and Vgray5'. These gray voltages Vgray2' and Vgray3' have the level of a voltage that is divided by resistors R31, R32, and R33 that are serially connected between output terminals of the first and second amplifier circuit AMP1 and AMP2.

The second gray voltage generation unit **56b** includes seventh to twelfth input terminals for receiving clock signals G_CLK2, G_CLK3, and G_CLK6 generated from the clock generator **52** and reference voltages Vref2, Vref3, and Vref6 generated from the voltage generator **54**. It also has a fourth amplifier AMP4, a fifth amplifier AMP5, and a sixth amplifier AMP6 for subtracting G_CLK2, G_CLK3, and G_CLK6 from Vref2, Vref3, and Vref6 to generate gray voltages Vgray6', Vgray8', and Vgray10', and output terminals for outputting Vgray6', Vgray8', and Vgray10' generated from AMP4, AMP5 and AMP6. The fourth amplifier circuit AMP4 subtracts G_CLK2 from Vref2, and amplifies it to a predetermined ratio to generate Vgray6'. The fifth amplifier circuit AMP5 subtracts G_CLK3 from Vref3, and amplifies it to a predetermined ratio to generate Vgray8'. And, the sixth amplifier circuit AMP6 subtracts G_CLK6 from Vref6, and amplifies it to a predetermined ratio to generate Vgray10'.

The gray voltages Vgray6', Vgray8', and Vgray10' are given by the following equations;

$$V_{gray6'} = \frac{R2 + R21 + R22}{R22} \left[V_{ref2} - \frac{R22}{R2 + R21} V_{G_CLK2} \right] \quad \text{<Equation 4>}$$

$$V_{gray8'} = \frac{R3 + R2 + R24}{R24} \left[V_{ref3} - \frac{R24}{R3 + R23} V_{G_CLK3} \right] \quad \text{<Equation 5>}$$

$$V_{gray10'} = \frac{R6 + R29 + R30}{R30} \left[V_{ref6} - \frac{R30}{R6 + R29} V_{G_CLK6} \right] \quad \text{<Equation 6>}$$

wherein V_{G_CLKn} represents an alternative element of the gate clock signal GATE CLOCK.

The second gray voltage generation unit **56b** generates eighth and ninth gray voltages Vgray8' and Vgray9', as well as Vgray6', Vgray7', and Vgray10'. These gray voltages Vgray8' and Vgray9' have the level of a voltage that is divided by resistors R38, R39, and R40 that are serially connected between output terminals of the fifth and the sixth amplifier circuit AMP5 and AMP6.

In the drawings, the fourth and seventh gray voltages Vgray4' and Vgray7' can be outputted through one or two terminals. For example, the fourth gray voltage Vgray4' generated through a fourth output terminal indicates that it uses an output of the second amplifier circuit AMP2 naturally. And, the fourth gray voltage Vgray4' generated through a fifth output terminal indicates that it divides the output of the second amplifier circuit AMP2 through a resistor to a predetermined ratio for output. Based upon a circuit configuration, the gray voltages Vgray1', . . . , and Vgray10' generated from the gray voltage generator **56** may use an output of an amplifier circuit naturally, or may divide and use the output of the amplifier circuit to a predetermined rate. Although Vgray4' and Vgray7' are illustrated in the drawing, they are simply examples. This can be applied to any other gray voltages.

FIGS. 7A and 7B exemplarily illustrate waveforms of gray voltages generated from a gray voltage generation according to the present invention. In particular, FIG. 7A shows a waveform of a gray voltage of a positive polarity, and FIG. 7B shows a waveform of a gray voltage of a negative polarity. Waveforms ① and ①', ② and ②', and ③ and ③' denote a gate clock signal GATE CLOCK issued from a timing control circuit **4**, a 48-gray voltage, and a 64-gray voltage, respectively.

FIGS. 8 and 9 exemplarily illustrate waveforms of outputs of a source driving circuit, which are generated by applying the gray voltage shown in FIGS. 7A and 7B. In particular, FIG. 8 shows a waveform in driving dot inversion, and FIG. 9 shows a waveform in driving 2-line inversion (i.e., normally white mode that white presents when a power is not applied).

In the drawings, illustrated elements are a gate clock signal GATE CLOCK outputted from a timing control circuit **4**, an output signal Vdrive of a source driving circuit in a conventional liquid crystal display, an output signal of a source driving circuit **3** in a liquid crystal display according to the present invention, and gate on signals GATE ON(n), GATE ON(n+1), GATE ON(n+2) and GATE ON(n+3) that are outputted from the timing control circuit **4** in order to drive (n)th, (n+1)th, (n+2)th and (n+3)th lines.

The source driving circuit in the conventional liquid crystal display generates a liquid crystal driving voltage Vdrive having voltage level of V_{F+} and V_{F-} in each period of the gate clock GATE CLOCK. The voltage Vdrive is symmetric to positive and negative directions on the basis of a common voltage Vcom.

The source driving circuit 3 in the liquid crystal display 100 according to the present invention generates a liquid crystal driving voltage $V_{drive}' = V_{gray}(t)$ that is changed by a gray voltage in each period of the gate clock signal GATE CLOCK. In each period of the gate clock signal GATE CLOCK, the voltage V_{drive}' generates a liquid crystal driving voltage V_{drive}' having different levels in high and low level intervals. That is, the liquid crystal driving voltage $V_{drive}' = V_{gray}(t)$ generates positive and negative high voltage that are enough to rapidly charge liquid crystal capacitors C_p constructed in a liquid crystal panel 1. In this case, the liquid crystal driving voltage $V_{drive}' = V_{gray}(t)$ generates the high voltages only for a predetermined interval, in order to prevent power consumption caused by generating such high voltages.

With reference to FIG. 8, in driving dot inversion, how to drive a positive polarity when applying a gate on signal Gate On(n) for driving an (n)th line, is now explained. If a gate clock signal Gate Clock is laid to high level, a source driving circuit 3 generates a liquid crystal driving voltage V_{drive}' having first voltage level that is still higher than that of an existing liquid crystal driving voltage V_{drive} . If Gate Clock is laid to low level, the source driving circuit 3 generates a liquid crystal driving voltage V_{drive}' having a second voltage level of V_{F+} with the same polarity as V_{drive} . In this case, both the first voltage level and the second voltage level are higher than a common voltage V_{com} . And, the first voltage level is higher than the second voltage level.

When a gate-on signal Gate On(n) for driving an (n+1)th line is applied, driving a negative polarity is explained. If the gate clock signal Gate Clock is laid to high level, the source driving circuit 3 generates a liquid crystal driving voltage V_{drive}' having third voltage level that is still lower than that of the existing liquid crystal driving voltage V_{drive} . If Gate Clock is laid to low level, the source driving circuit 3 generates a liquid crystal driving voltage V_{drive}' having fourth voltage level of V_{F-} with the same polarity as V_{drive} . In this case, both values of the third voltage level and the fourth voltage level are lower than the common voltage V_{com} . And, the third voltage level is lower than the fourth voltage level.

With reference to FIG. 9, in driving 2-line inversion, when a gate on signal Gate On(n) for driving (n)th and (n+1)th lines is applied, driving a positive polarity is explained. If a gate clock signal Gate Clock is laid to high level, a source driving circuit 3 generates a liquid crystal driving voltage V_{drive}' whose level is still higher than that of an existing liquid crystal driving voltage V_{drive} . If Gate Clock is laid to low level, the source driving circuit 3 generates a liquid crystal driving voltage V_{drive}' having voltage level of V_{F+} the same as V_{drive} .

When a gate on signal Gate On(n) for driving (n+2)th and (n+3)th lines is applied, driving a negative polarity is explained. If the gate clock signal Gate Clock is laid to high level, the source driving circuit 3 generates a liquid crystal driving voltage V_{drive}' whose level is still lower than that of the existing liquid crystal driving voltage V_{drive} . If Gate Clock is laid to low level, the source driving circuit 3 generates a liquid crystal driving voltage V_{drive}' of V_{F-} with the same polarity as V_{drive} .

In FIGS. 7 and 8, output waveforms of the source driving circuit 3 can be changed according to a kind of line driving methods, and are applicable to various kinds of line driving methods (e.g., n-line inversion driving method).

FIGS. 10A, 10B, 11A, 11B, 12A, 12B, 13A and 13B show response speed measuring results of 0 through 32, 0 through 48, 0 through 64, and 32 through 84 gray levels of the source

driving circuits by means of the gray voltage shown in FIGS. 7A and 7B. In particular, FIG. 10A, FIG. 10B, FIG. 11A, and FIG. 11B show a response speed of 0 through 32 gray levels of a conventional source driving circuit, a response speed of 0 through 32 gray levels of a source driving circuit according to the invention, a response speed of 0 through 48 gray levels of the conventional source driving circuit, and a response speed of 0 through 48 gray levels of the source driving circuit according to the invention, respectively. FIG. 12A, FIG. 12B, FIG. 13A, and FIG. 13B show a response speed of 0 through 64 gray levels of the conventional source driving circuit, a response speed of 0 through 64 gray levels of the source driving speed according to the invention, a response speed of 32 through 64 gray levels of the conventional source driving circuit, and a response speed of 32 through 64 gray levels of the source driving circuit according to the invention, respectively.

The result can be obtained by measuring the 48-gray voltages ② and ②' and the 64-gray voltages ③ and ③' (see FIGS. 7A and 7B) that were changed and applied with respect to five source driving circuits each having positive and negative polarities. A rising time of each waveform is denoted on the basis of a luminance, and corresponds to a falling time of a liquid crystal based on its movement.

Referring to FIGS. 10A and 10B, in response speeds of a source driving circuit with respect to 0 through 32 gray levels, a conventional rising time (i.e., a falling time of a liquid crystal) is 26.0 ms and a conventional falling time (i.e., a rising time of the liquid crystal) is 3.6 ms. According to the present invention, a rising time (i.e., a falling time of a liquid crystal) is 24.2 ms and a falling time (i.e., a rising time of the liquid crystal) is 3.6 ms. In this case, a luminance-based falling time is not changed, while a luminance-based rising time is reduced from 26 ms to 24.2 ms by 1.8 ms.

Referring to FIGS. 11A and 11B, in response speeds of a source driving circuit with respect to 0 through 48 gray levels, a conventional rising time (i.e., a falling time of a liquid crystal) is 36.8 ms and a conventional falling time (i.e., a rising time of the liquid crystal) is 3.6 ms. According to the invention, a rising time (i.e., a falling time of a liquid crystal) is 26.2 ms and a falling time (i.e., a rising time of the liquid crystal) is 4.4 ms. In this case, a luminance-based falling time increases in 0.8 ms, while a luminance-based rising is reduced from 36.8 ms to 26.2 ms by 10.6 ms.

Referring to FIGS. 12A and 12B, in response speeds of a source driving circuit with respect to 0 through 64 gray levels, a conventional rising time (i.e., a falling time of a liquid crystal) is 22.6 ms, and a conventional falling time (i.e., a rising time of the liquid crystal) is 4.7 ms. According to the invention, a rising time (i.e., a falling time of a liquid crystal) is 15.1 ms, and a falling time (i.e., a rising time of the liquid crystal) is 4.6 ms. In this case, a luminance-based falling time is reduced by 0.1 ms, and a luminance-based rising time is reduced from 22.6 ms to 15.1 ms by 7.5 ms.

Referring to FIGS. 13A and 13B, in response speeds of 32 through 64 gray levels with respect to a source driving circuit, a conventional rising time (i.e., a falling time of a liquid crystal) is 20.8 ms, and a falling time (i.e., a rising time of the liquid crystal) is 3.4 ms. According to the invention, a rising time (i.e., a falling time of a liquid crystal) is 15.0 ms, and a falling time (i.e., a rising time of the liquid crystal) is 3.4 ms. In this case, a luminance-based falling time is not changed, and a luminance-based rising time is reduced from 20.8 ms to 15.0 ms by 5.8 ms.

In FIGS. 10A through 13B, response speeds of a source driving circuit 3 according to the present invention change as follows. In 0 through 32 gray levels, a response speed is reduced from 26 ms to 24.2 ms by 1.8 ms. In 0 through 48 gray levels, a response speed is reduced from 36.8 ms to 26.2 ms by 10.6 ms. In 0 through 64 gray levels, a response speed is reduced from 22.6 ms to 15.1 ms by 7.5 ms. And, in 32 through 64 gray levels, a response speed is reduced from 20.8 ms to 15.0 ms by 5.8 ms. The following table [TABLE 1] represents these response speeds.

TABLE 1

Falling Times of Liquid Crystal		
	Prior Art	Present Invention
0-32 Gray Levels	26.0 ms (1.00)	24.2 ms (0.96)
0-48 Gray Levels	36.8 ms (1.00)	26.2 ms (0.71)
0-64 Gray Levels	22.6 ms (1.00)	15.1 ms (0.67)
32-64 Gray Levels	20.8 ms (1.00)	15.0 ms (0.72)

wherein these falling times are results of simulation that is carried out in the same condition, and numerals in parentheses denote normalized results on the basis of falling times of a conventional liquid crystal, respectively.

Referring to the normalized results in TABLE 1, in 0 through 32 gray levels, the falling time of the liquid crystal is improved by 7%. In 0 through 48 gray levels, the falling time is improved by 29%. In 0 through 64 gray levels, the falling time is improved by 33%. And, in 32 through 64 gray levels, the falling time is improved by 28%. In other words, the speed of the falling time of the liquid crystal is improved in proportion to the gray values.

As described above, a gray voltage generation circuit of this invention outputs an altered gray voltage V_{gray}' so that a source driving circuit can generate a liquid crystal driving voltage V_{drive}' having a voltage level as shown in FIGS. 7 and 8. Thus, the source driving circuit 3 generates a liquid crystal driving voltage $V_{drive}' = V_{gray}'(t)$ that changes according to a gray voltage in each period of a gate clock signal Gate Clock. Liquid crystal capacitors C_p constructed in a liquid crystal panel 1 are rapidly charged by the liquid crystal driving voltage V_{drive}' applied from the source driving circuit 3. As a result, a falling time of the liquid crystal is reduced to improve a driving speed of a liquid crystal display.

While an illustrative embodiment of the present invention has been shown and described, numerous variations and alternate embodiments will occur to those skilled in the art, without departing from the spirit and scope of the invention. Accordingly, it is intended that the present invention not be limited solely to the specifically described illustrative embodiment. Various modifications are contemplated and can be made without departing from the spirit and scope of the invention as defined by the appended claims.

What is claimed is:

1. A liquid crystal display (LCD), comprising:

a liquid crystal panel having a plurality of pixels;

a timing control circuit generating a gate clock signal and a plurality of control signals;

a gray voltage generation circuit generating a first gray voltage in a first interval of the gate clock signal and a second gray voltage in a second interval of the gate clock signal in response to a voltage level of the gate clock signal and a plurality of reference voltages obtained by dividing a power supply voltage to a

predetermined ratio, the first gray voltage having a magnitude greater than that of the second gray voltage; a gate driving circuit sequentially scanning the pixels row by row in response to the gate clock signal; and

a source driving circuit generating a first driving voltage corresponding to the first gray voltage and a second driving voltage corresponding to the second gray voltage,

wherein the first driving voltage is applied to the panel in the first interval of the gate clock signal and the second driving voltage is applied to the panel in the second interval of the gate clock signal.

2. The LCD of claim 1, wherein the source driving circuit, while driving a positive polarity of the panel, generates a driving voltage having a first voltage level in the first interval of the gate clock signal, and generates a driving voltage having a second voltage level in the second interval of the gate clock signal, and

both the first voltage level and the second voltage level are higher than a common voltage level, and the first driving voltage level is higher than of the second driving voltage level.

3. The LCD of claim 2, wherein the source driving circuit, while driving a negative polarity of the panel, generates a driving voltage having a third voltage level in the first interval of the gate clock signal, and generates a driving voltage having a fourth voltage level in the second interval of the gate clock signal, and

both the first voltage level and the second voltage level are lower than the common voltage level, and the third driving voltage level is lower than the fourth driving voltage level.

4. The LCD of claim 1, wherein the gray voltage generation circuit comprises:

a clock generator generating a plurality of clock signals having a same period as the gate clock signal, in response to the gate clock signal;

a voltage generator dividing the power supply voltage to a predetermined ratio to generate a plurality of voltages as reference for generating the first gray voltage and the second gray voltage; and

a gray voltage generator outputting either the first gray voltage or the second gray voltage to the source driving circuit, in response to the gate clock signals issued from the clock generator and the voltages generated by the voltage generator.

5. The LCD of claim 4, wherein the voltage generator includes n-bit voltage generation units for dividing the power supply voltage to a predetermined ratio to generate the n-bit voltages each having different voltage level, and

wherein each of the voltage generation unit includes at least two and more resistors coupled between the power supply voltage and a ground voltage, and an output terminal coupled to one of contact points between the resistors.

6. The LCD of claim 4, wherein the clock generator comprises:

an input terminal for receiving the gate clock signal;

n-bit clock generation units coupled to the input terminal in parallel; and

n-bit output terminals each being coupled to the n-bit clock generation units,

wherein each of the clock generation units has a capacitor and a resistor that are serially connected between the input terminal and the output terminal, and generates a clock signal having a same period as the gate clock signal.

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7. The LCD of claim 4, wherein the gray voltage generator comprises:

- a positive gray voltage generation unit for generating (m/2)-bit positive first and second gray voltages having a same polarity as the gate clock signal and each having different voltage level in the first interval and the second interval of the gate clock signal so as to drive a positive polarity of the panel; and
- a negative gray voltage generation unit for generating (m/2)-bit negative first and second gray voltages having a polarity opposite to the gate clock signal and each having different voltage level in the first interval and the second interval of the gate clock signal so as to drive a negative polarity of the panel.

8. The LCD of claim 7, wherein the positive gray voltage generation unit includes at least one or more amplifier circuits having a first input terminal for receiving one of the n-bit clock signals from the clock generator and one of the n-bit reference voltages from the voltage generator, a second input terminal connected to a ground through a resistor, and an amplifier circuit having a feedback resistor connected between the second input terminal and the output terminal.

9. The LCD of claim 8, wherein the amplifier circuit adds the clock signal to the reference voltage, and amplifies the same to generate the positive first and second gray voltages.

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10. The LCD of claim 8, wherein the amplifier circuit further includes a resistor for dividing the positive first and second gray voltages, and an output terminal connected to the contact point of the resistor, for outputting the divided gray voltage.

11. The LCD of claim 7, wherein the negative gray voltage generation unit includes a first input terminal for receiving one of the n-bit reference voltages from the voltage generator, a second input terminal for receiving one of the n-bit clock signals from the clock generator, and an amplifier circuit having a feedback resistor connected between the second input terminal and the output terminal.

12. The LCD of claim 11, wherein the amplifier circuit subtracts the clock signal from the reference voltage, and amplifies it to a predetermined ratio to generate the negative first and second gray voltages.

13. The LCD of claim 11, wherein the amplifier circuit further includes a resistor for dividing the negative first and second gray voltages, and an output terminal to the contact point of the resistor, for outputting the divided gray voltage.

* * * * *

专利名称(译)	用于快速驱动液晶显示器的灰度电压产生电路		
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申请号	US10/747665	申请日	2003-12-30
[标]申请(专利权)人(译)	YEON允MO 李健熙BIN		
申请(专利权)人(译)	YEON允-MO 李健熙-BIN		
当前申请(专利权)人(译)	三星DISPLAY CO. , LTD.		
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摘要(译)

用于驱动液晶显示器的灰度电压产生电路快速输出改变的灰度电压，使得源极驱动电路可以在短时间内对液晶面板中构造的液晶电容器充电。响应于来自灰度电压产生电路的灰度电压，在驱动正极性的同时，当施加高电平的栅极时钟信号时，源极驱动电路产生比现有液晶驱动电压更高电平的液晶驱动电压，并且当施加低电平的栅极时钟信号时，产生与现有液晶驱动电压相似的电平的液晶驱动电压。并且，在驱动负极性时，源极驱动电路在施加高电平的栅极时钟信号时产生比现有液晶驱动电压低的液晶驱动电压，并产生与以下相似的电平的液晶驱动电压。当施加低电平的栅极时钟信号时，现有的液晶驱动电压。

