



US 20030151572A1

(19) United States

(12) Patent Application Publication

Kumada et al.

(10) Pub. No.: US 2003/0151572 A1

(43) Pub. Date: Aug. 14, 2003

(54) DISPLAY DEVICE, DRIVE CIRCUIT FOR THE SAME, AND DRIVING METHOD FOR THE SAME

(52) U.S. Cl. 345/87, 345/92; 345/98

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## ABSTRACT

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(21) Appl. No.: 10/357,480

(22) Filed: Feb. 4, 2003

### Foreign Application Priority Data

Feb. 8, 2002 (JP) P2002-031593  
Dec. 19, 2002 (JP) P2002-367738

In a signal line drive circuit of an active-matrix type liquid-crystal display which is a voltage-controlled type display with a capacitive load,  $n$  selector switches (161 to 16 $n$ ) are provided between buffer circuits (151 to 15 $n$ ) to which voltages responsive to an image to be displayed are inputted from reference voltage selection circuits (131 to 13 $n$ ), and output terminals (T1 to T $n$ ) to which are connected image signal lines. These selector switches (161 to 16 $n$ ), based on a shorting control signal (Csh) that is at a high level when the polarity is reversed to perform AC drive of the liquid-crystal panel, switch the output signals (OUT1 to OUT $n$ ) of the image signal line drive circuit between the output signals of the buffer circuits (151 to 15 $n$ ) and the common electrode signal (Vcom). By doing this, each of the image signal lines is, for a prescribed time only when the polarity is reversed, separated from the buffer circuits (151 to 15 $n$ ) and shorted to the common electrode. This configuration reduces the power consumption of the signal line drive circuit

### Publication Classification

(51) Int. Cl. 7 G09G 3/36

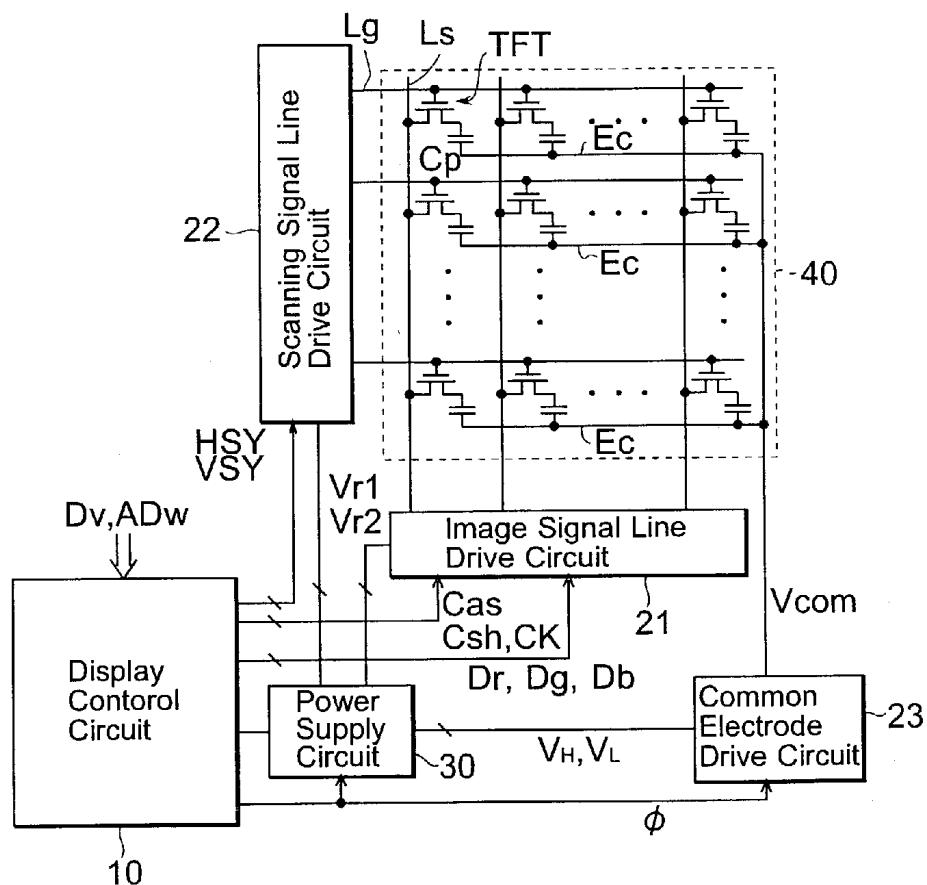


Fig. 1

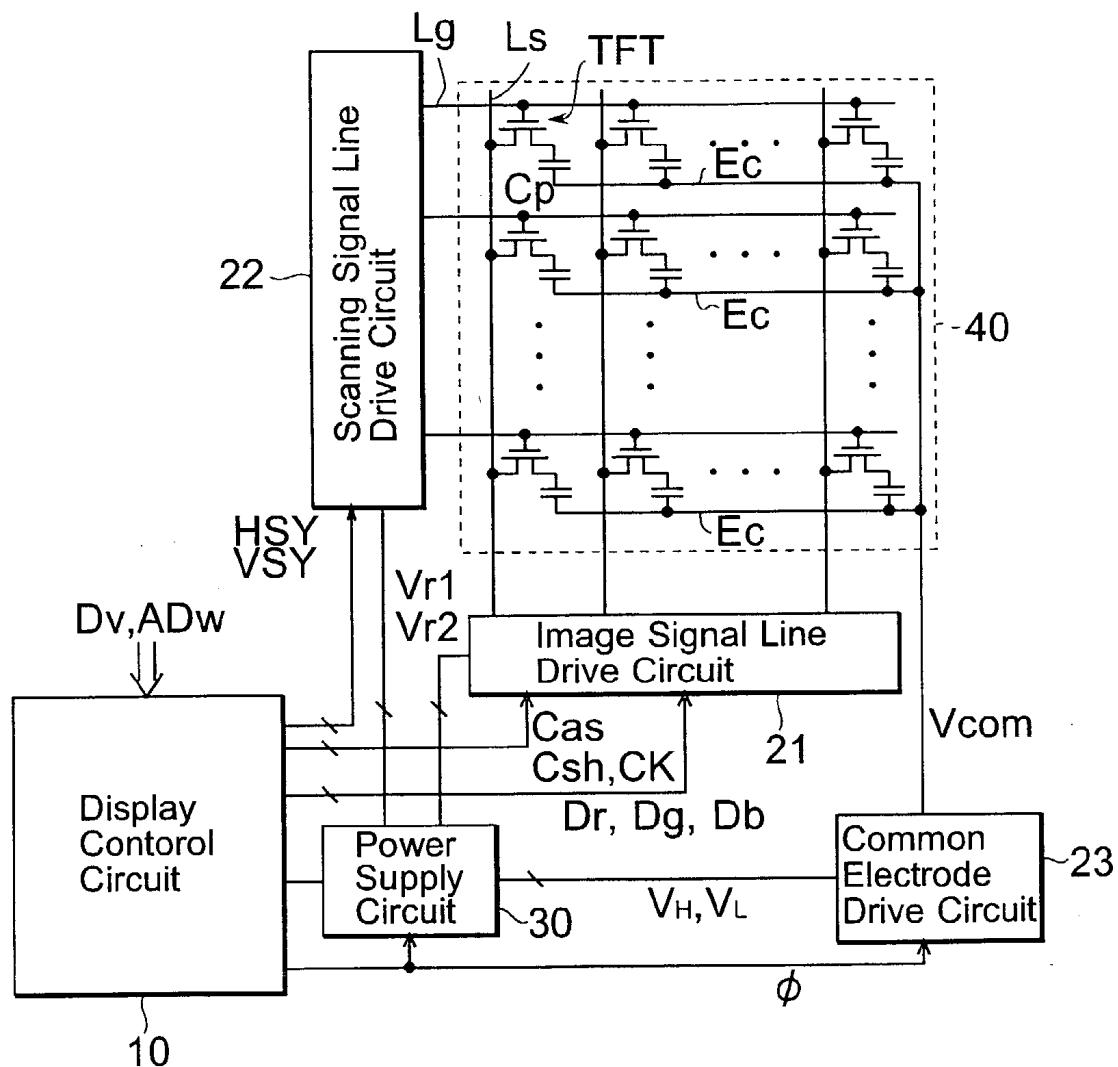


Fig. 2

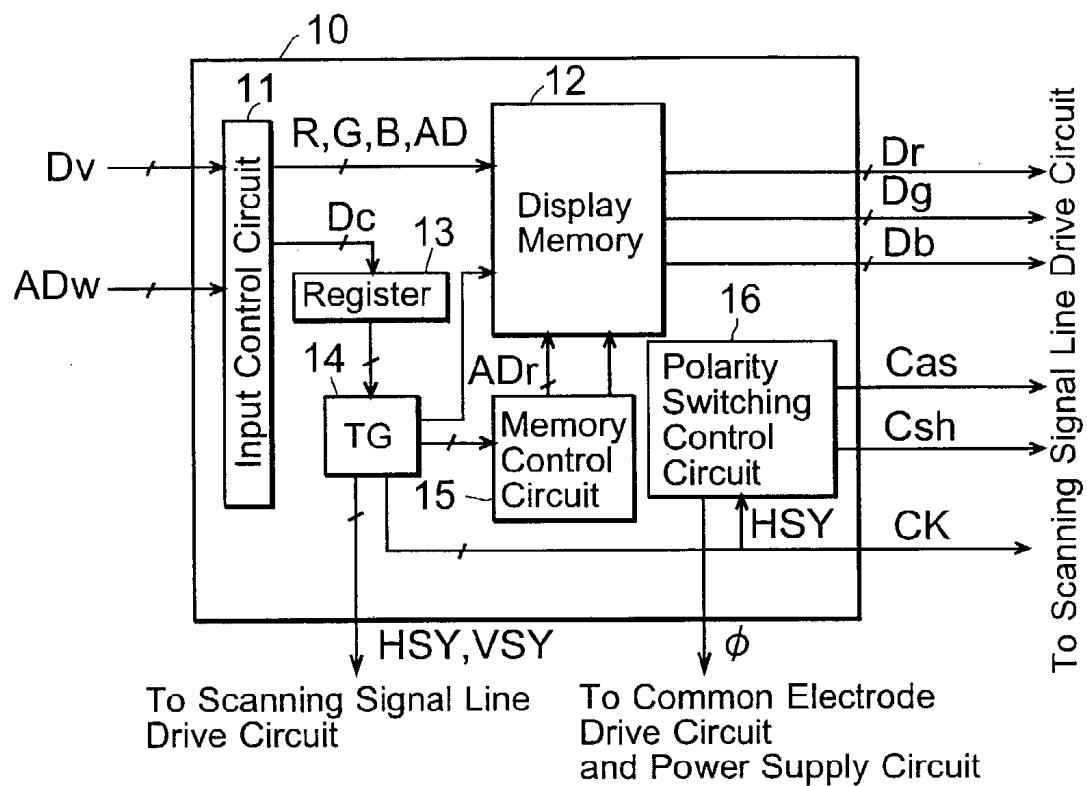
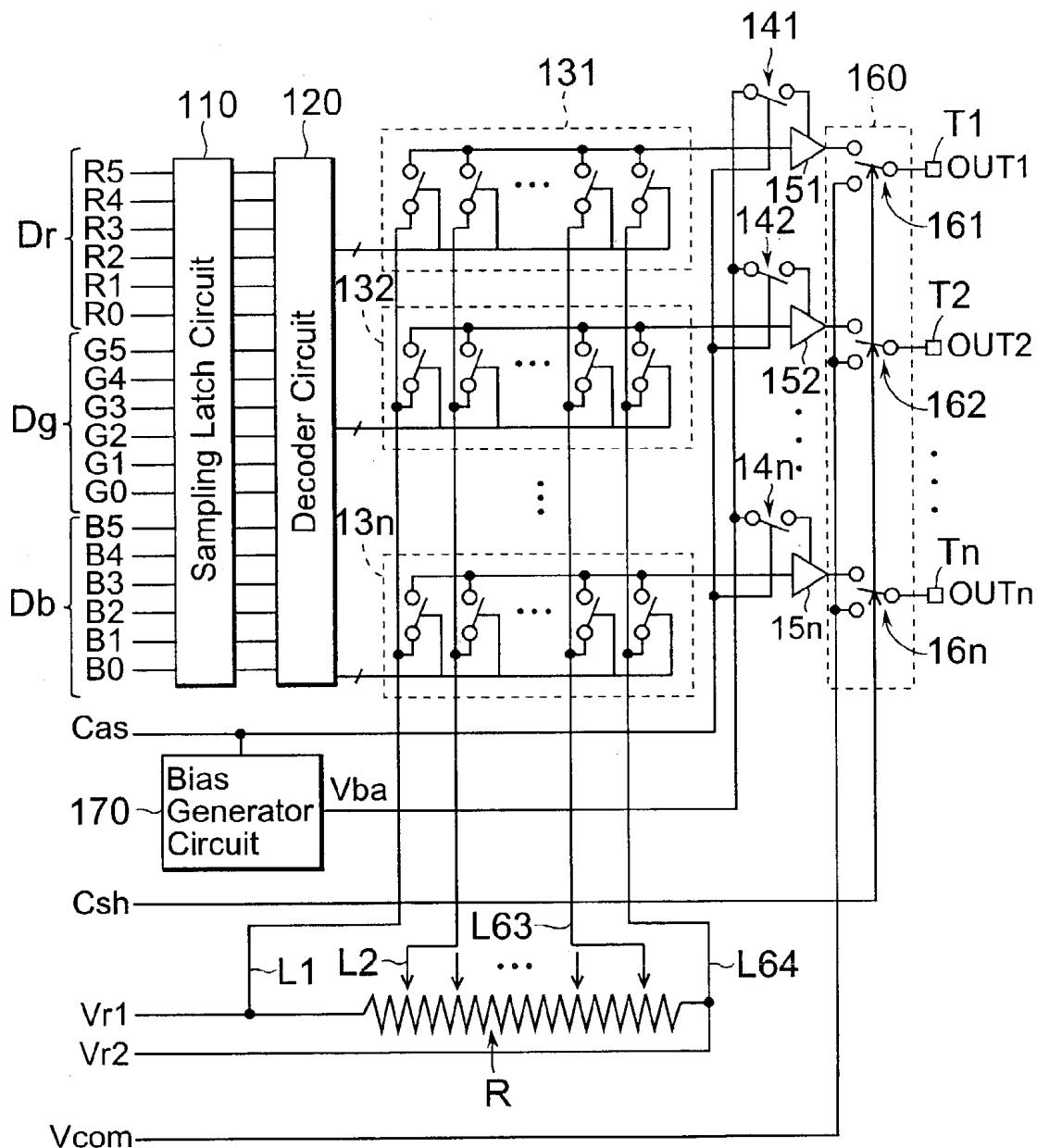
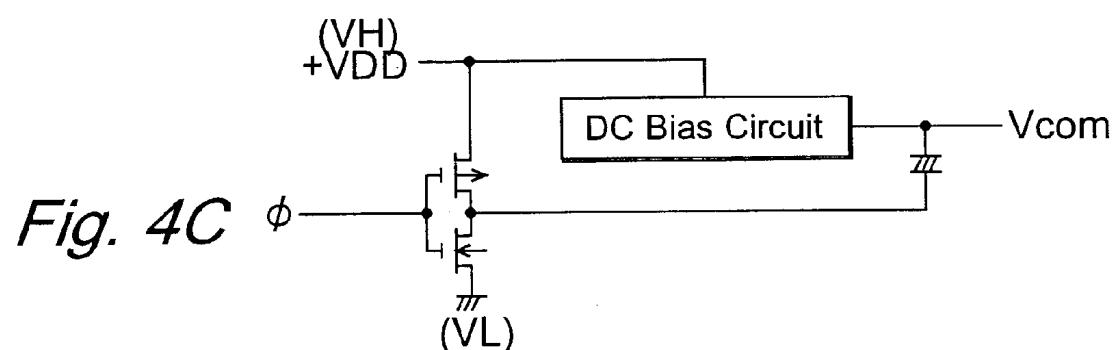
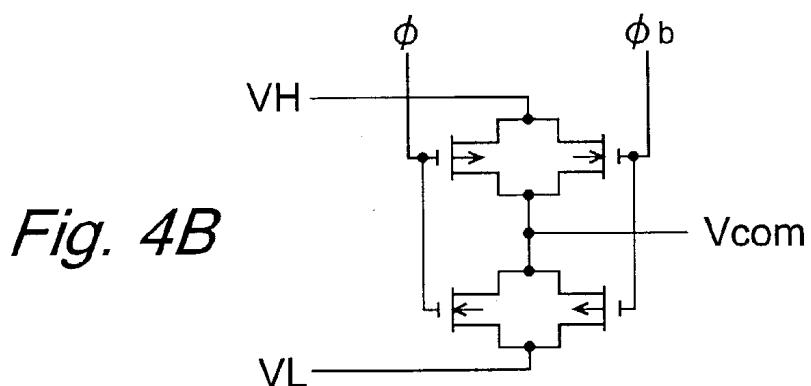
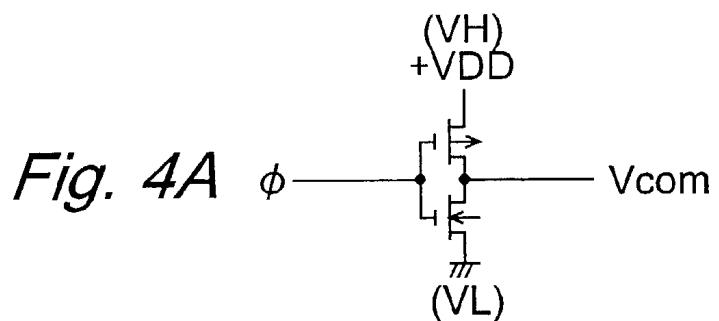
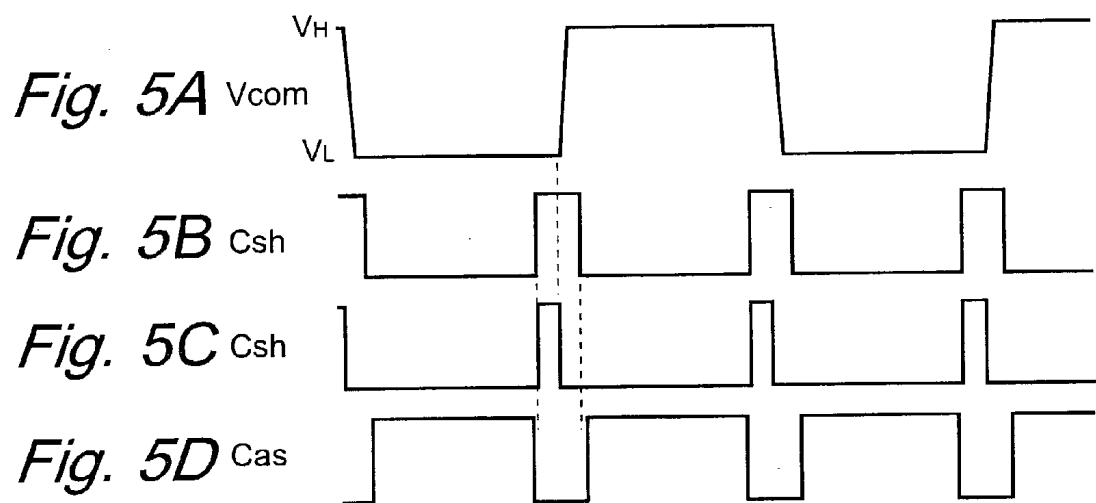


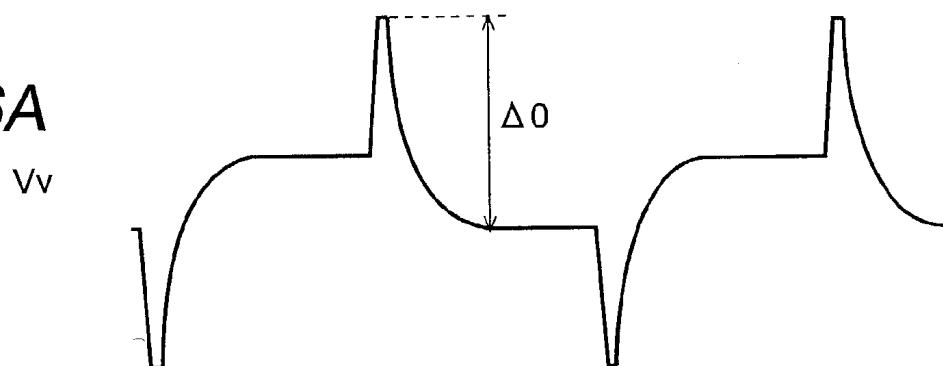
Fig. 3



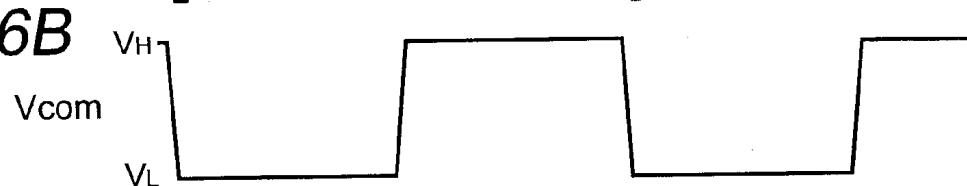


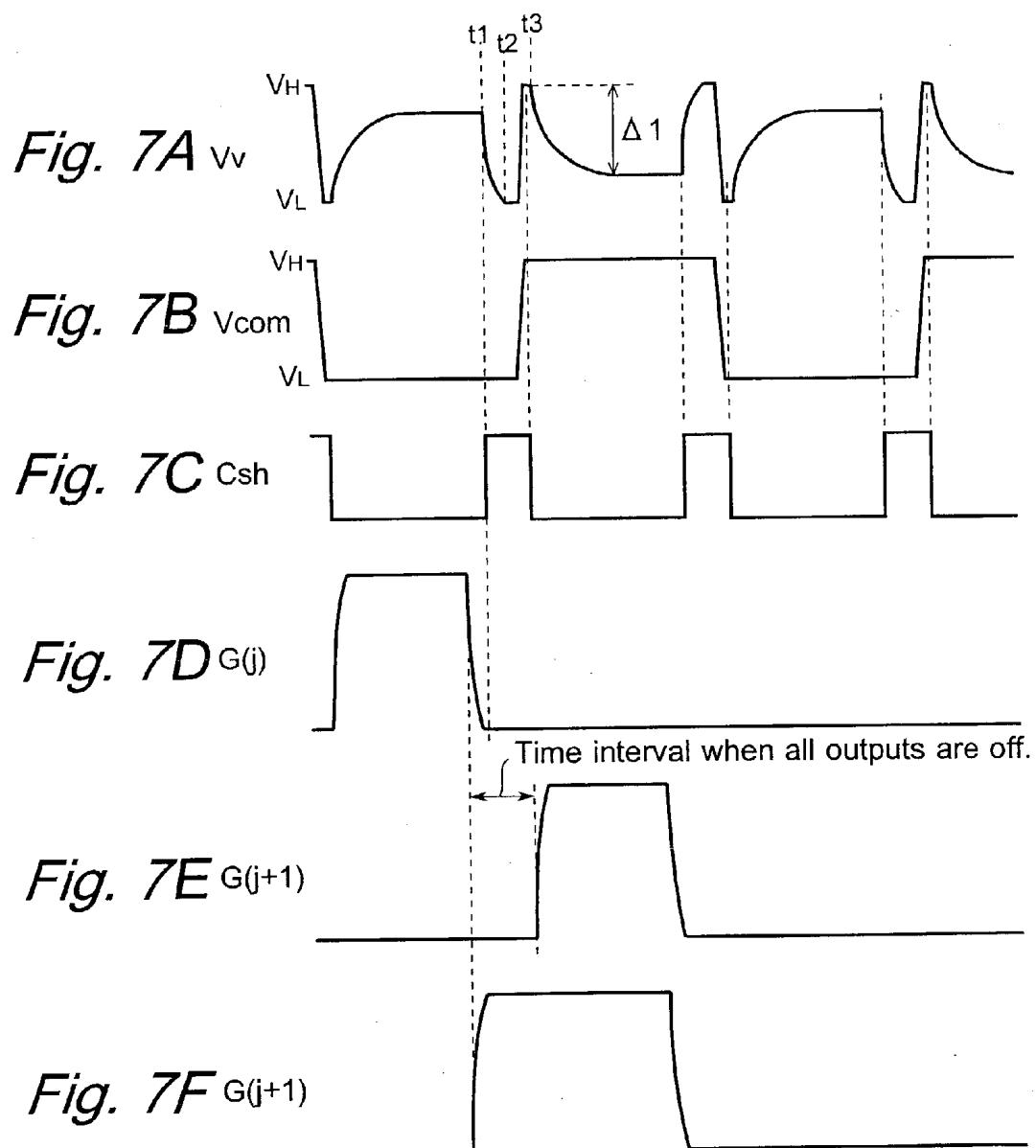


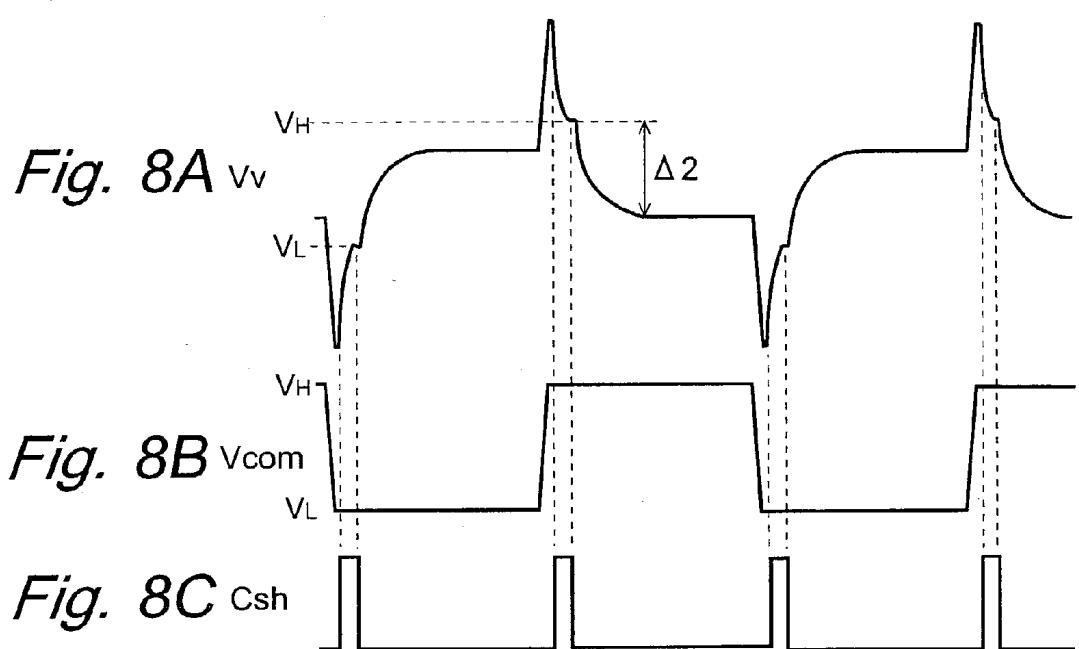
*Fig. 6A*

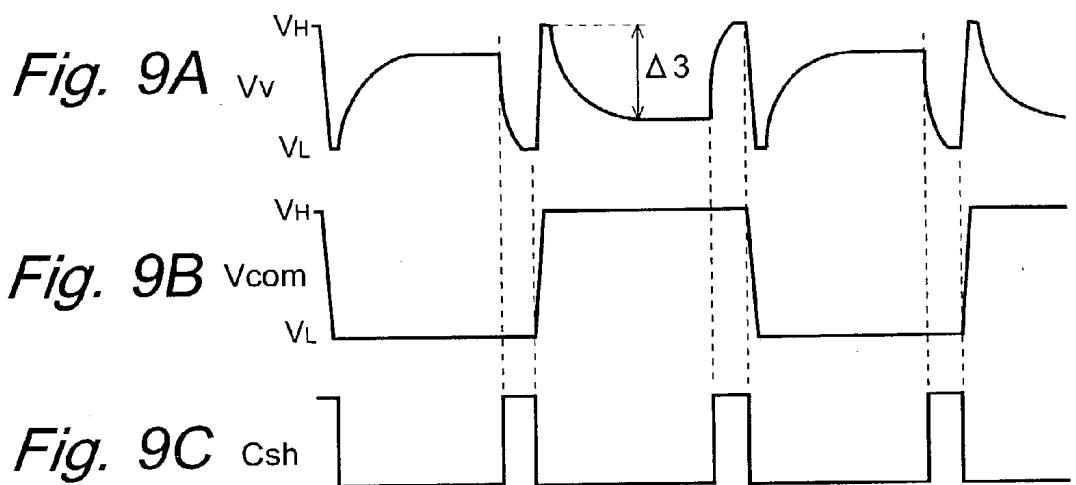


*Fig. 6B*









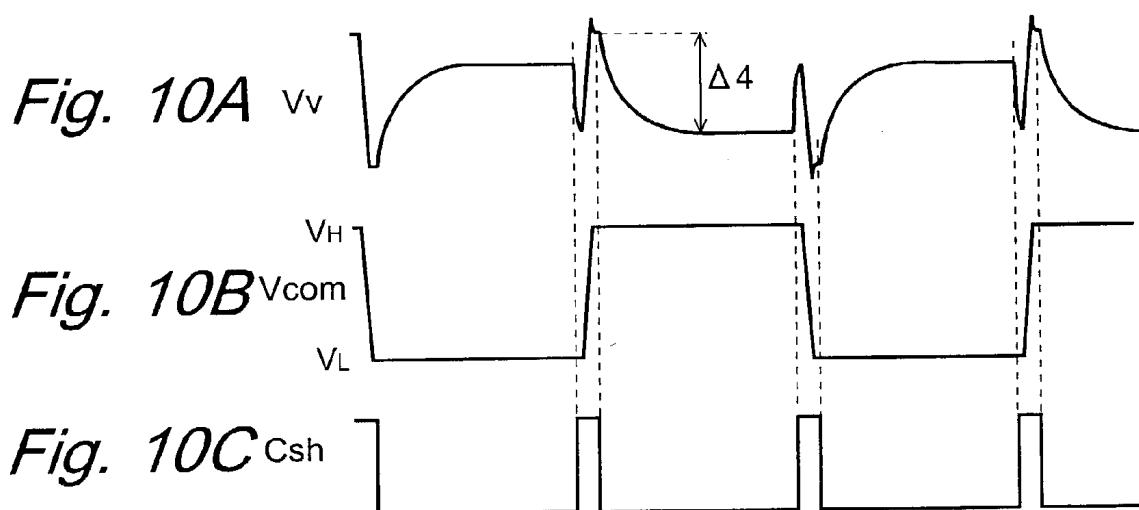


Fig. 11

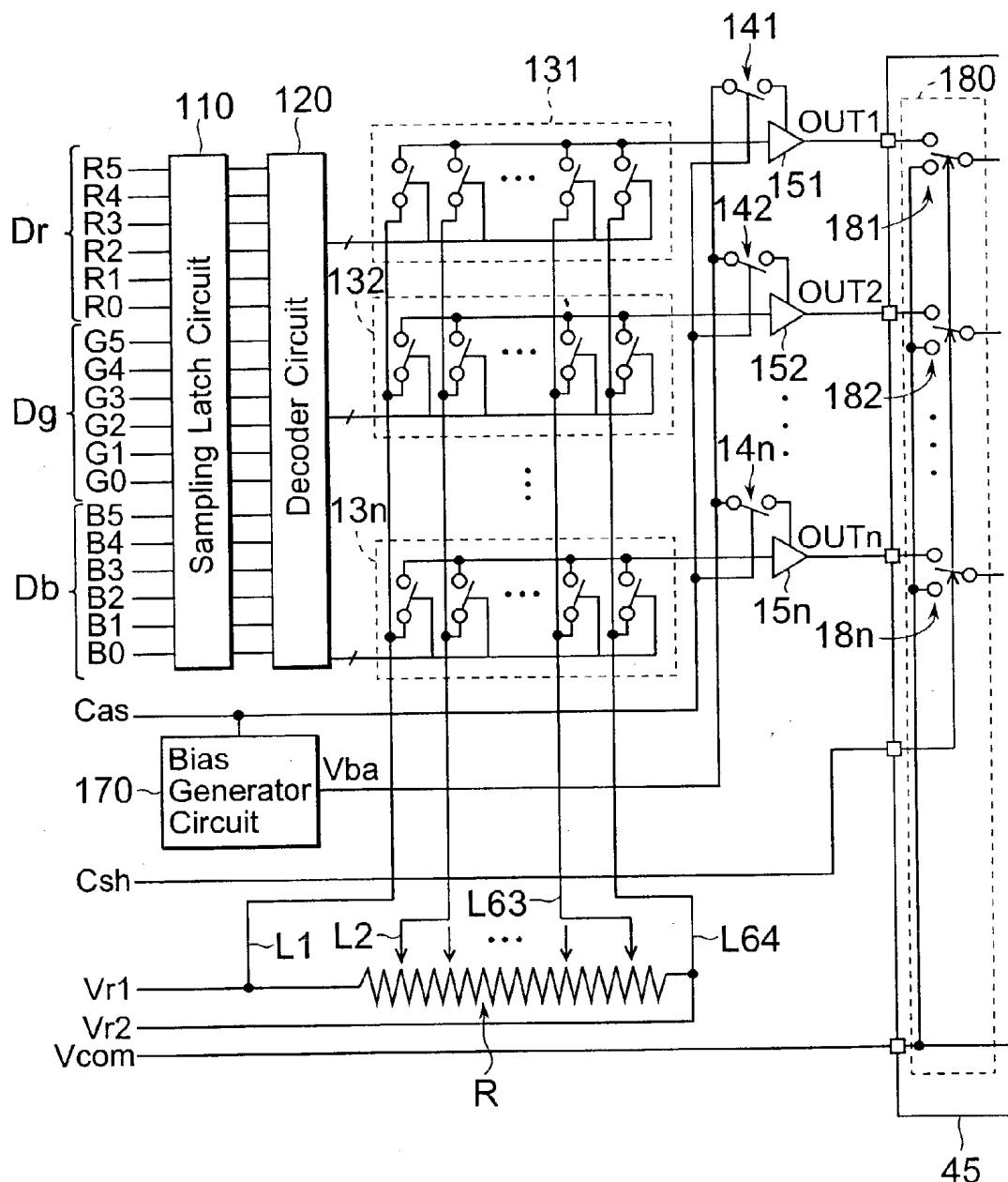


Fig. 12

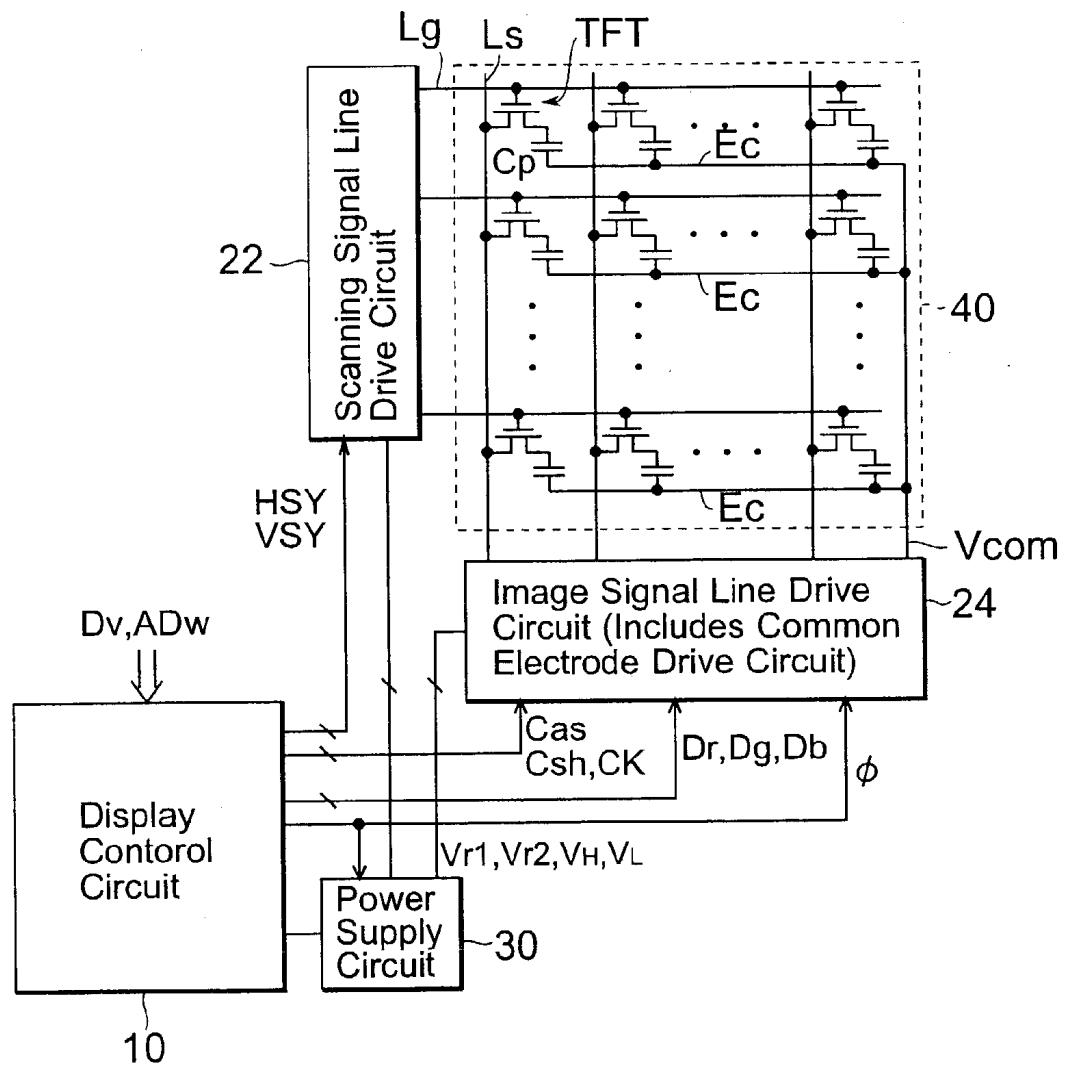


Fig. 13

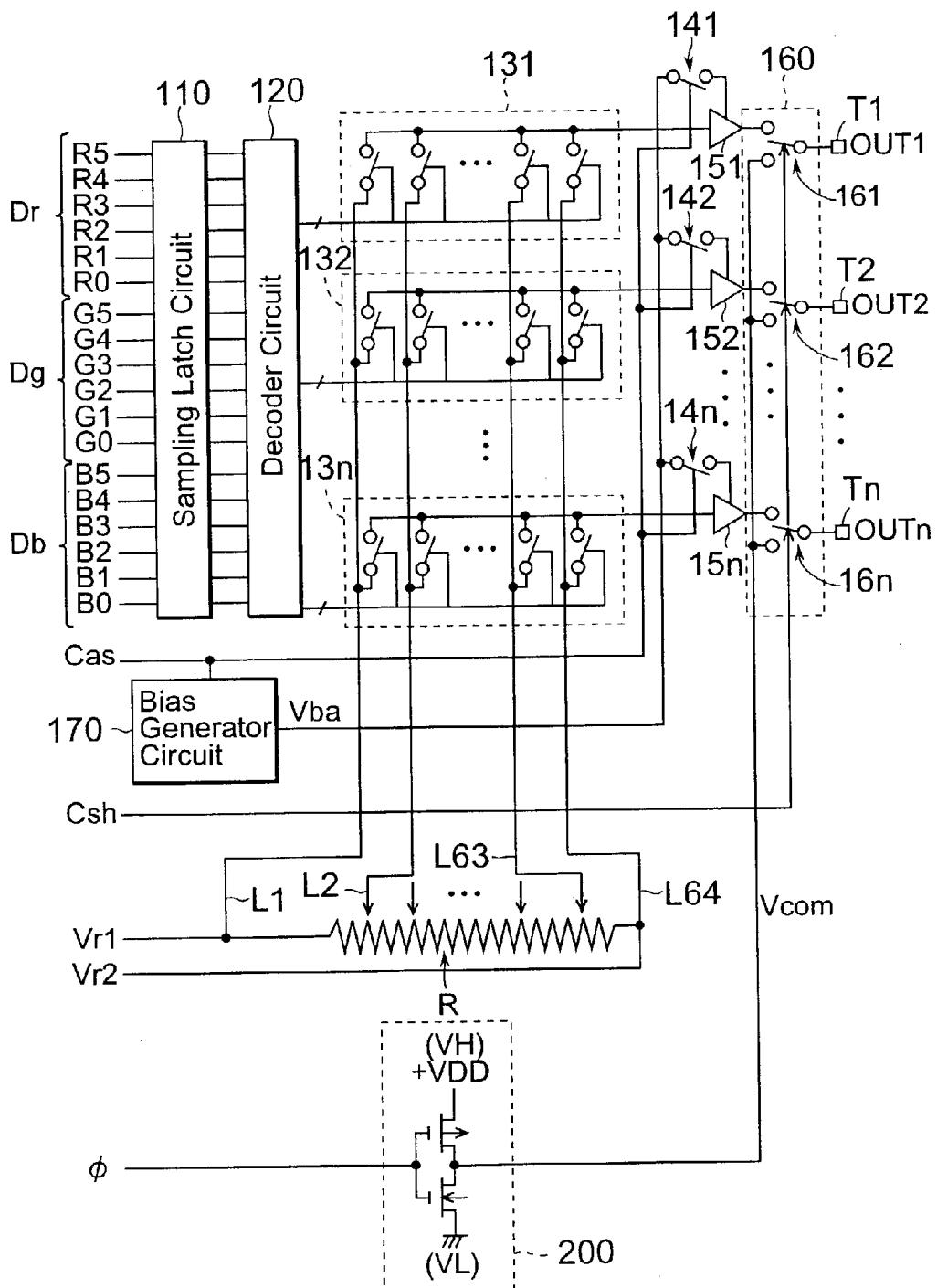


Fig. 14

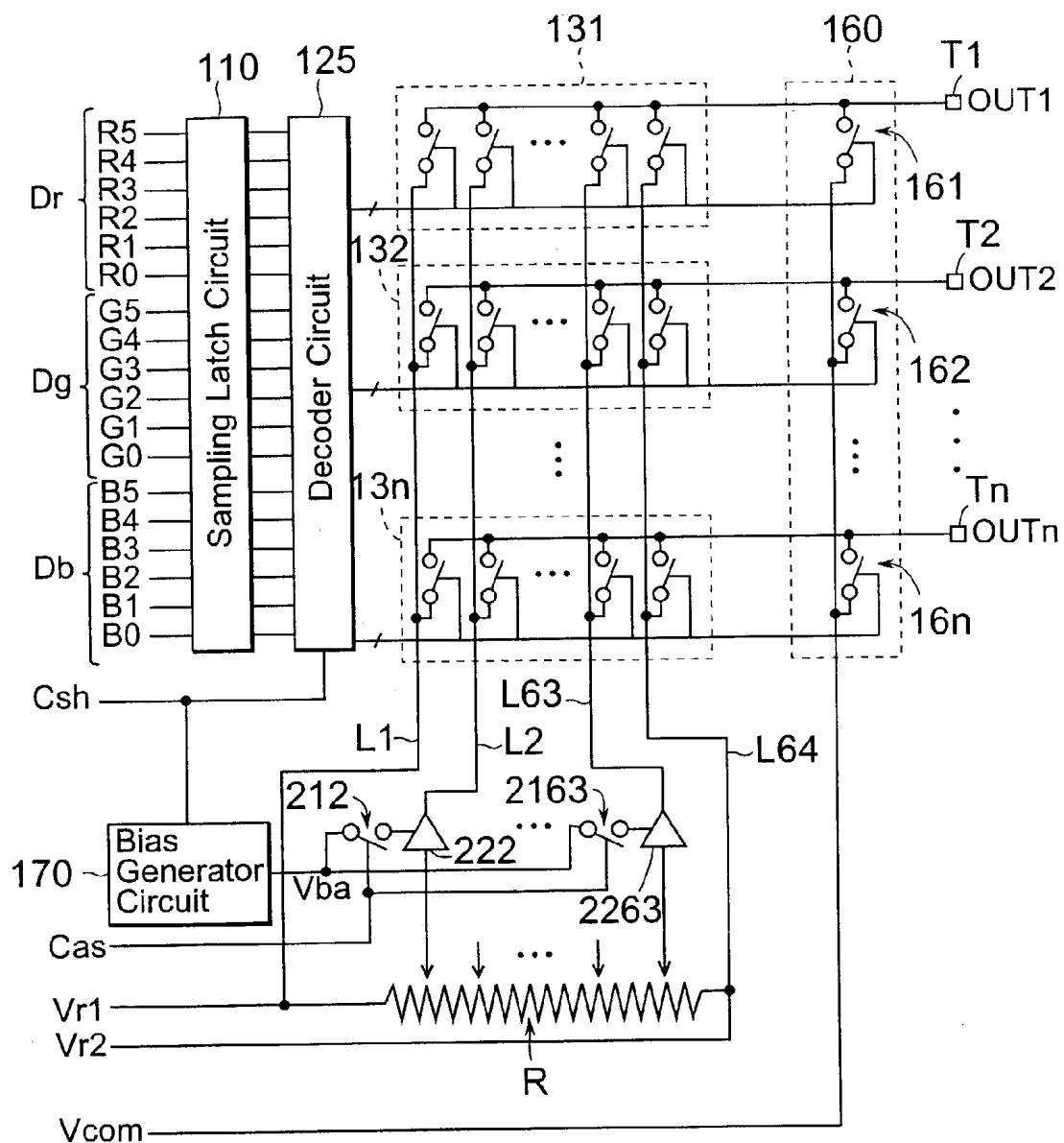
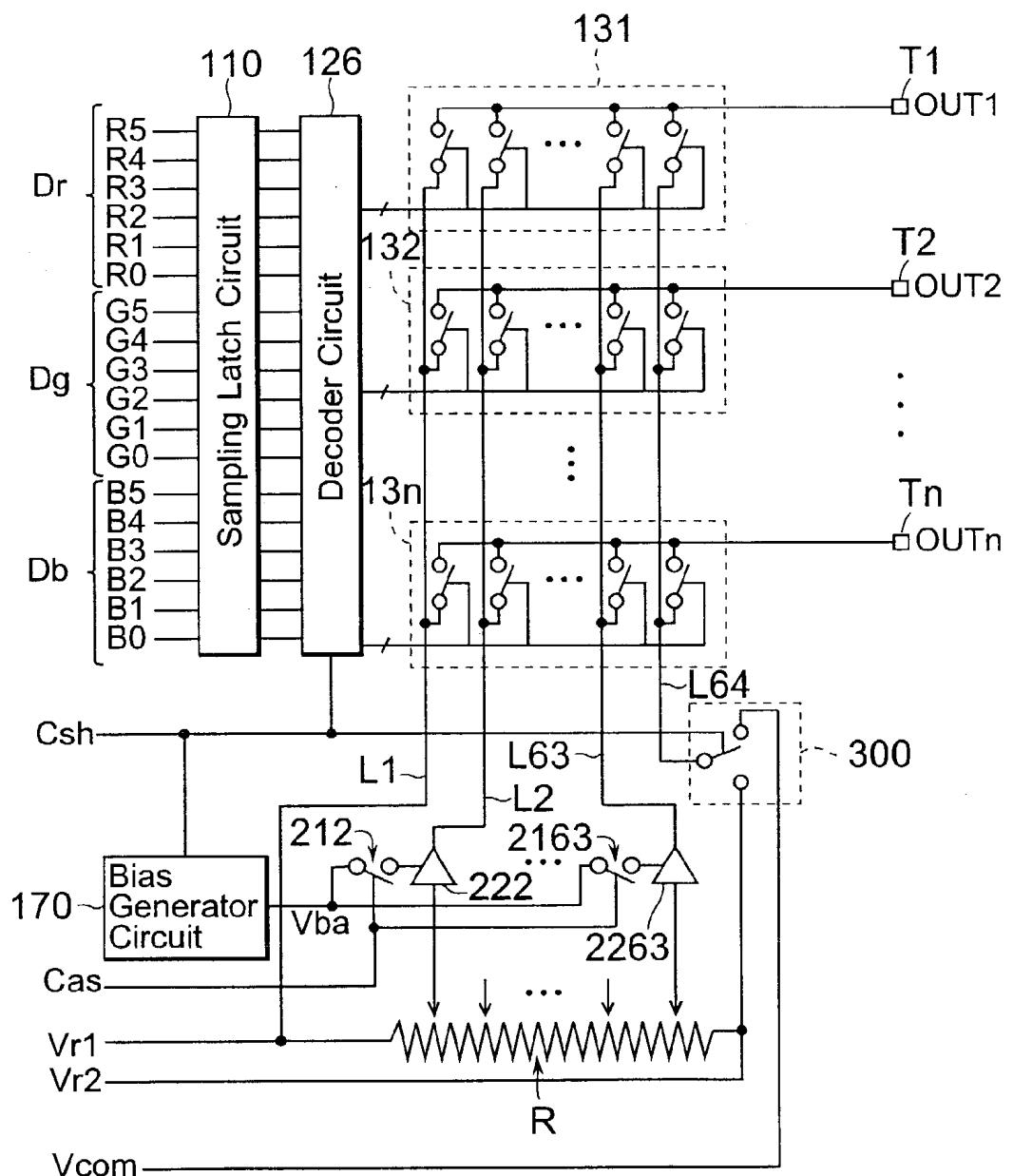
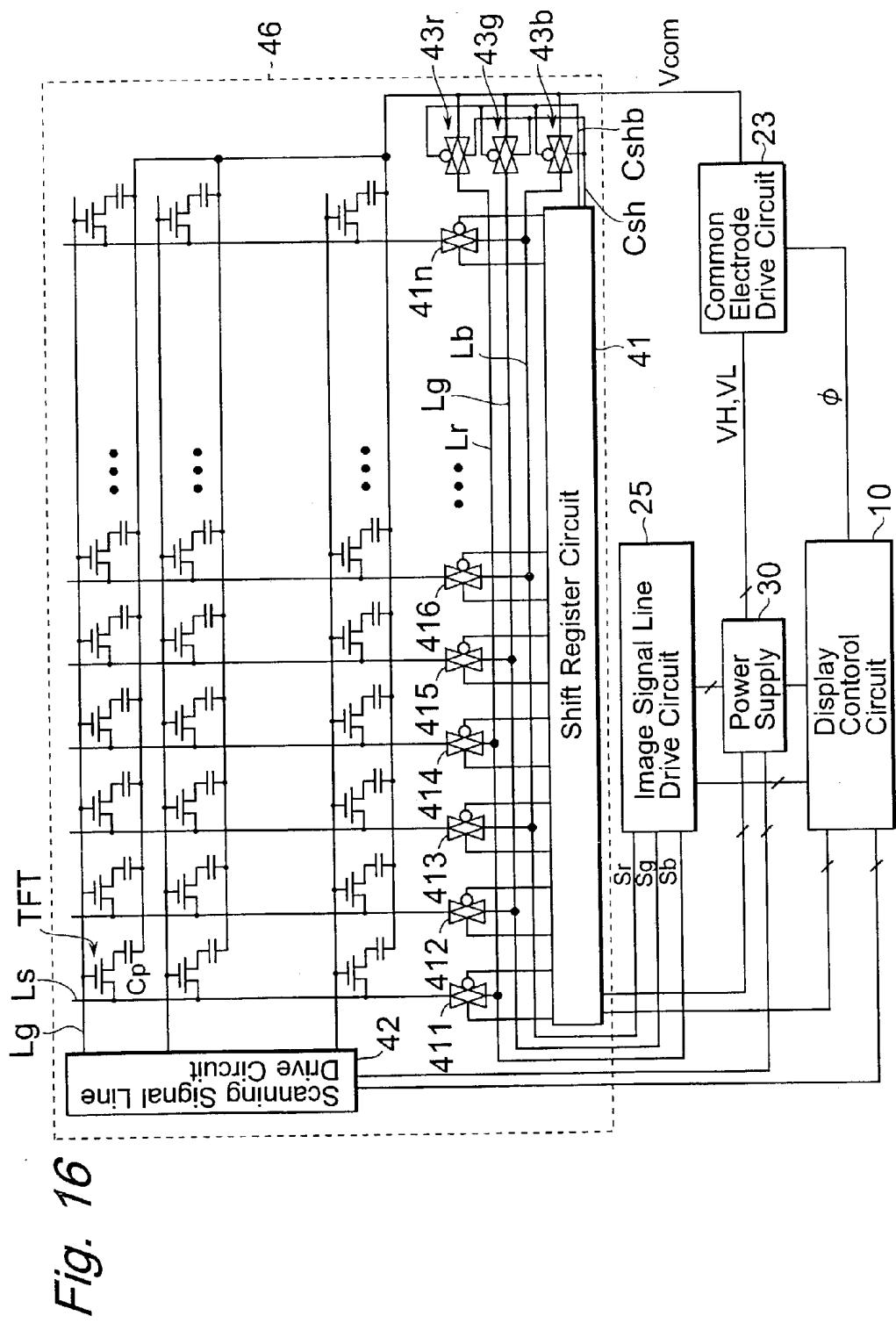


Fig. 15





**DISPLAY DEVICE, DRIVE CIRCUIT FOR THE SAME, AND DRIVING METHOD FOR THE SAME****BACKGROUND OF THE INVENTION****[0001] 1. Field of the Invention**

**[0002]** The present invention relates to a display device with voltage-controlled active-matrix drive, having a capacitive load, such as in an active-matrix type liquid-crystal display device, and more particularly to a drive circuit for such a display device.

**[0003] 2. Background Art**

**[0004]** From the standpoint of extending battery life, there is a serious need to reduce the power consumption in portable information equipment, such as mobile telephones, personal digital assistants (PDAs), and laptop computers. With improvements in processing performance and usable functionality, portable information products, have come to require improved display capabilities, with both high display quality and a larger number of display colors. For this reason, in order to meet the demands for improved display capabilities, active-matrix type liquid-crystal displays (hereinafter referred to as TFT-LCDs) implemented using thin-film transistors (TFTs) are beginning to see use in even such portable information products.

**[0005]** A liquid-crystal panel in a TFT-LCD device (hereinafter referred to as a TFT-LCD panel) has a pair of mutually opposing substrates (a first substrate and a second substrate). These substrates are held fixed with a prescribed distance therebetween (typically several  $\mu\text{m}$ ) with a liquid-crystal material forming a liquid-crystal layer so as to fill the space between the substrates. At least one of these substrates is transparent, and in the case of making a transmissive-type display, both substrates must be transparent. In a TFT-LCD panel, the first substrate is provided with a plurality of mutually parallel scanning signal lines and a plurality of image signal lines, which are perpendicular to the scanning signal lines. Pixel electrodes are provided at the intersection locations between scanning signal lines and image signal lines, as are pixel TFTs, which serve as switching elements for the purpose of making electrical connection between the pixel electrodes and the corresponding image signal line. The gate terminal of a pixel TFT is connected to a scanning signal line the source terminal of the pixel TFT is connected to an image signal line, and the drain terminal of the pixel TFT is connected to the pixel electrode.

**[0006]** A common electrode, serving as an opposing electrode, is formed over the entire surface of the second substrate, which opposes the first substrate. An appropriate voltage is applied to the common electrode by a common electrode drive circuit, so that a voltage corresponding to potential difference between the pixel electrode and the common electrode is applied across the liquid-crystal layer. Because this applied voltage can be used to control the light transmissivity of the liquid-crystal layer, it is possible to create a desired pixel display by applying an appropriate voltage from the image signal lines.

**[0007]** In order to prevent deterioration of the liquid crystal and to maintain display quality, the above-described TFT-LCD panel is driven by AC drive. Specifically, the TFT-LCD panel is driven so that the polarity of the voltage applied to the liquid crystal is reversed, for example every

horizontal scan period. Additionally, in order to reduce the amplitude of the voltage on the image signal line, the potential on the common electrode is changed in response to the above-noted AC drive (by applying what is hereinafter referred to as an AC common electrode signal).

**[0008]** However, even if the use of an AC common electrode signal reduces the amplitude of the voltage on the image signal line, there is still a large variation in potential on the image signal lines when the polarity reverses to achieve AC drive. It is therefore necessary for the image signal line drive circuit to have sufficient capacitive load drive capacity to cause this large a potential change with respect to the capacitive load presented by a TFT-LCD panel. For this reason, the image signal line drive circuit consumes a large amount of power, thereby hindering the achievement of low power consumption in a TFT-LCD display.

**SUMMARY OF THE INVENTION**

**[0009]** Accordingly, it is an object of the present invention to provide a display device which reduces the power consumption of the AC drive circuit that applies to a capacitive load a voltage responsive to an image to be displayed while reversing the polarity thereof with a prescribed period.

**[0010]** One aspect of the present invention is a display device in which a voltage serving as an image signal representing an image to be displayed is applied to a capacitive load including a capacitance formed by mutually opposing first and second electrodes, and which has a drive circuit that causes the voltage applied to the capacitive load to reverse polarity periodically, this display device having

**[0011]** an image signal line drive circuit which supplies a voltage signal responsive to an image to be displayed to the first electrode relative to the second electrode as a reference, and

**[0012]** a connection switching circuit which when the polarity of the voltage applied to the capacitive load is reversed, separates the first electrode from the image signal line drive circuit and shorts the first electrode to an electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode.

**[0013]** By adopting the above-noted configuration, when the polarity of the voltage applied to the capacitive load is reversed, the first electrode is electrically separated from the image signal line drive circuit and shorted to an electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode, the charge that had been accumulated in the capacitive load being thereby discharged. By doing this, the amount of change in the potential at the first electrode required after the polarity reversal is reduced. Therefore, even if the drive capacity of the image signal line drive circuit is smaller than in the past, it is possible to reduce the power consumption image signal line drive circuit and also possible to reduce the size of the transistors used to implement the buffer circuit within the image signal line drive circuit. As a result, it is possible to achieve a reduction in both the size and the cost of the display device.

**[0014]** In a display device such as described above, it is preferable that the electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode be the second electrode itself.

[0015] By using the second electrode as the electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode, when reversing the polarity of the voltage applied to the capacitive load, the first electrode is electrically separated from the image signal line drive circuit and shorted to the second electrode, so that the charge accumulated in the capacitive load is directly discharged without going through the power supply. By doing this, similar to the general configuration first described above, it is possible to reduce the power consumption of the image signal line drive circuit and also reduce the size of the buffer circuit in the image signal line drive circuit, thereby enabling a reduction in both the size and the cost of the display device.

[0016] In a display device configured as described above, it is possible for the drive circuit to apply as the image signal to the capacitive load a voltage representing an image to be displayed based on a horizontal scan and a vertical scan, and cause the polarity of the applied voltage to be reversed at the time of switching of a scan line in the horizontal scan.

[0017] By adopting this configuration, in an image display based on horizontal and vertical scanning, that is in the display of an image assembled by repeating a horizontal scan at a prescribed period as the scan starting position is shifted a small amount at a time in the vertical direction, the polarity of the voltage applied to the capacitive load is reversed at the time that the horizontal scan line is switched, and each time the reversal occurs the first electrode is electrically separated from the image signal line drive circuit and shorted to an electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode, the charge that had accumulated in the capacitive load being thereby discharged. By doing this, there is a large effect, for example in reducing the power consumption of the image signal line drive circuit.

[0018] This display device can be further configured so as to have

[0019] a plurality of image signal lines serving as the first electrodes,

[0020] a plurality of scanning signal lines intersecting with the plurality of image signal lines,

[0021] a plurality of pixel formation parts each corresponding to one of points of intersection between the plurality of image signal lines and the plurality of scanning signal lines, and disposed in a matrix arrangement, and

[0022] a scanning signal line drive circuit which selectively drives the plurality of scanning signal lines,

[0023] wherein each pixel formation part includes

[0024] a switching element that is switched on and off by a scanning signal line passing through the corresponding intersection point,

[0025] a pixel electrode connected via the switching element to the image signal line passing through the corresponding intersection point,

[0026] and a common electrode serving as the second electrode, provided in common to the plurality of pixel formation parts, and disposed so

that a prescribed capacitance included in the capacitive load is formed between the common electrode and the pixel electrode,

[0027] the scanning signal line drive circuit applying to a selected scanning signal line a voltage that turns the switching element on, and

[0028] the connection switching circuit, when the polarity of the voltage applied to the capacitive load is reversed, electrically separating the image signal lines from the image signal line drive circuit and shorting the image signal lines to an electrode providing a voltage level that is equivalent to the voltage supplied to the common electrode.

[0029] According to this configuration, when the polarity is reversed to perform AC drive of the pixel formation parts, each image signal line is electrically separated from the image signal line drive circuit and shorted to an electrode supplying a voltage level that is equivalent to the voltage supplied to the common electrode. By doing this, even if the drive capacity of the image signal line drive circuit is smaller than in the past, because it is possible to apply the same voltage as in the past to the capacitive load formed between the pixel electrode and the common electrode and between the image signal line and the common electrode, it is possible to reduce the power consumption of the image signal line drive circuit, and further possible to reduce the size of the transistors that make up the buffer circuit within the image signal line drive circuit. As a result, it is possible to reduce both the size and the cost of the display device.

[0030] In such a display device, it is possible to have the connection switching circuit, after the switching element that had been turned on by the scanning line selected before reversal of the polarity of the voltage applied to the capacitive load is placed in the off state, electrically separate the image signal lines from the image signal line drive circuit and short the image signal lines to an electrode providing a voltage level that is equivalent to the voltage supplied to the common electrode.

[0031] By adopting this configuration, after the switching element that had been turned on by the scanning line selected before reversal of the polarity of the voltage applied to the capacitive load is turned off, the image signal lines are electrically separated from the image signal line drive circuit and shorted to an electrode supplying a voltage level that is equivalent to the voltage supplied to the common electrode, the result being that a pixel value to be written into a pixel formation part by an image signal line is not influenced by this shorting operation.

[0032] In such a display device, it is possible for the connection switching circuit, when the polarity of the voltage applied to the capacitive load is reversed, to short the image signal lines to an electrode providing a voltage level that is equivalent to the voltage supplied to the common electrode for a period of time that is three or more times the delay time constant which is the product of the wiring resistance and wiring capacitance in one image signal line.

[0033] By adopting this configuration, when the polarity of the voltage applied to the capacitive load is reversed, the charge accumulated in the capacitive load (the capacitance formed by each of the image signal lines and the common electrode) is discharged, so that each of the image signal

lines and the common electrode are at substantially the same potential. By doing this, the amount of potential change on an image signal line which is to be made by the image signal line drive circuit after polarity reversal is substantially halved compared to the past.

[0034] In such a display device, it is further possible to provide the image signal line drive circuit with a stopping control circuit which for at least the period of time during which the connection switching circuit is shorting each of the image signal lines to an electrode providing a voltage level equivalent to the voltage supplied by the common circuit, stops at least part of the image signal line drive circuit.

[0035] According to this configuration, it is possible to stop at least part of the image signal line drive circuit without influencing the image display, thereby providing a further reduction in the power consumption of the image signal line drive circuit.

[0036] In such a display device, it is possible to adopt a configuration in which the drive circuit includes a common electrode drive circuit which switches the potential on the common electrode in response to the polarity reversal of the voltage applied to the capacitive load,

[0037] wherein the common electrode drive circuit switches the potential on the common electrode within the period of time during which the connection switching circuit is shorting the image signal lines to an electrode providing a voltage level that is equivalent to the voltage supplied to the common electrode.

[0038] By adopting this configuration, because polarity reversal occurs during the shorting period, the period of time used for pixel writing is lengthened.

[0039] In this display device, it is possible to adopt a configuration in which

[0040] the image signal line drive circuit includes reference voltage selection circuits each of which corresponds to one of the plurality of the image signal lines, selects a voltage responsive to the image signal from a plurality of reference voltages, and supplies the selected voltage to the corresponding image signal line as the voltage signal, and

[0041] in which each of the reference voltage selection circuits includes the connection switching circuit and when the polarity of the voltage applied to the capacitive load is reversed, selects a voltage level equivalent to the common electrode signal which is the voltage supplied to common electrode, instead of a reference voltage from the plurality of reference voltages, and supplies this selected voltage level to a corresponding image signal lines, thereby shorting each of the image signal lines to an electrode supplying a voltage level equivalent to the voltage supplied to the common electrode.

[0042] By adopting this configuration, because the connection switching circuit is included in each of the reference voltage selection circuits, the configuration of the image signal line drive circuit in the above-noted display device is simplified, thereby enabling a reduction in the size of the IC chip used to implement the image signal line drive circuit.

[0043] In this display device, it is further possible to adopt a configuration in which the image signal line drive circuit includes

[0044] a plurality of reference voltage bus lines to each of which is given one of the plurality of reference voltages, and

[0045] a voltage switching circuit which, when the polarity of the voltage applied to the capacitive load is reversed, applies to one reference voltage bus line of the plurality of reference voltage bus lines a voltage level equivalent to the common electrode signal, instead of the reference voltage to be applied to the one reference voltage bus line, wherein

[0046] each of the reference voltage selection circuits, during each horizontal scan period, selects a reference voltage bus line of the plurality of reference voltage bus lines to which is applied a reference voltage responsive to the image signal and connects the selected bus line to a corresponding image signal line and, when the polarity of the voltage applied to the capacitive load is reversed, selects and connects the one reference voltage bus line to the corresponding image signal line.

[0047] By adopting this configuration, although there is an increase of a voltage switching circuit as one switching means, one reference bus line and the switching means within each reference voltage selection circuit corresponding thereto are shared between selection of the reference voltage and selection of the voltage level equivalent to the common electrode signal. By doing this, there is a further reduction in the overall scale of the signal line drive circuitry, thereby making it possible to further reduce the size of the IC chip on which the image signal line drive circuit is implemented.

[0048] The above-noted display device can also be configured such that

[0049] the drive circuit further includes a common electrode drive circuit which switches the potential of the common electrode in response to the reversal of polarity of the voltage applied to the capacitive load, and so that

[0050] the image signal line drive circuit and the common electrode drive circuit are formed on either one and the same substrate or one and the same chip.

[0051] By adopting the above configuration, because the image signal line drive circuit is associated with the common electrode drive circuit via a connection switching circuit, by forming the image signal line drive circuit and the common electrode drive circuit on either one and the same substrate or one and the same chip, it is possible to simplify the configuration of the display device.

[0052] Another aspect of the present invention is a drive circuit in a display device of the AC drive type, in which a voltage serving as an image signal to be displayed is applied to a capacitive load including a capacitance formed by a first electrode and a second electrode which are in mutual opposition, and in which the polarity of the voltage applied to the capacitive load is periodically reversed, this drive circuit having

[0053] an image signal line drive circuit which supplies to the first electrode a voltage signal responsive to the image relative to the second electrode as a reference, and

[0054] a connection switching circuit which when the polarity of the voltage applied to the capacitive load is reversed, electrically separates the first electrode from the image signal line drive circuit and shorts the first electrode to an electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode itself.

[0055] In the above-noted drive circuit, it is preferable that the electrode providing a voltage level equivalent to the voltage supplied to the second electrode be the second electrode.

[0056] Yet another aspect of the present invention is a method for driving using a driving circuit in a display device of the AC drive type, in which a voltage serving as an image signal representing an image to be displayed is applied to a capacitive load including a capacitance formed by a first electrode and a second electrode which are in mutual opposition, and in which the polarity of the voltage applied to the capacitive load is periodically reversed, this method including

[0057] a step of supplying a voltage signal responsive to the image to the first electrode relative to the second electrode as a reference, and

[0058] a step of, when the polarity of the voltage applied to the capacitive load is reversed, electrically separating the first electrode from the part of the drive circuit that supplies the voltage signal to the first electrode, and shorting the first electrode to an electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode.

[0059] In the above-noted drive method, it is preferable that the electrode providing a voltage level equivalent to the voltage supplied to the second electrode be the second electrode itself.

[0060] These and other objects, features, aspects, and advantages of the present invention will be more apparent from the following detailed description of the present invention when taken in conjunction with the accompanying drawings.

#### BRIEF DESCRIPTION OF THE DRAWINGS

[0061] FIG. 1 is a block diagram showing the configuration of a liquid-crystal display according to a first embodiment of the present invention.

[0062] FIG. 2 is a block diagram showing the configuration of a display control circuit in the first embodiment.

[0063] FIG. 3 is a circuit diagram showing the configuration of an image signal line drive circuit in the first embodiment.

[0064] FIG. 4A is a circuit diagram showing a first example of the configuration of a common electrode drive circuit in the first embodiment

[0065] FIG. 4B is a circuit diagram showing a second example of the configuration of a common electrode drive circuit in the second embodiment.

[0066] FIG. 4C is a circuit diagram showing a third example of the configuration of a common electrode drive circuit in the first embodiment.

[0067] FIG. 5A to FIG. 5D are signal waveform diagrams illustrating stopping control of the image signal line drive circuit in the first embodiment.

[0068] FIG. 6A and FIG. 6B are voltage and signal waveform diagrams illustrating the liquid-crystal panel drive method in a conventional liquid-crystal display.

[0069] FIG. 7A to FIG. 7F are voltage and signal waveform diagrams illustrating a first liquid-crystal panel drive method in the first embodiment.

[0070] FIG. 8A to FIG. 8C are voltage and signal waveform diagrams illustrating a second liquid-crystal panel drive method in the first embodiment.

[0071] FIG. 9A to FIG. 9C are voltage and signal waveform diagrams illustrating a third liquid-crystal panel drive method in the first embodiment.

[0072] FIG. 10A to FIG. 10C are voltage and signal waveform diagrams illustrating a fourth liquid-crystal panel drive method in the first embodiment.

[0073] FIG. 11 is a circuit diagram showing the configuration of an image signal line drive circuit in a second embodiment of the present invention.

[0074] FIG. 12 is a block diagram showing the configuration of a liquid-crystal display according to a third embodiment of the present invention.

[0075] FIG. 13 is a circuit diagram showing the configuration of an image signal line drive circuit in the third embodiment.

[0076] FIG. 14 is a circuit diagram showing the configuration of an image signal line drive circuit in a fourth embodiment of the present invention.

[0077] FIG. 15 is a circuit diagram showing the configuration of an image signal line drive circuit in a fifth embodiment of the present invention.

[0078] FIG. 16 is a circuit diagram showing the configuration of a liquid-crystal display according to a sixth embodiment of the present invention.

#### DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

[0079] Embodiments of the present invention are described below in detail, with references made to relevant accompanying drawings.

[0080] (1. First Embodiment)

[0081] (1.1 Overall Configuration and Operation)

[0082] FIG. 1 is a block diagram showing the configuration of a liquid-crystal display according to a first embodiment of the present invention, which has a display control circuit 10, an image signal line drive circuit 21, a scanning signal line drive circuit 22, a common electrode drive circuit 23, a power supply circuit 30, and an active-matrix type liquid-crystal panel 40, wherein AC drive is used in order to reduce deterioration or the like of the liquid crystal, in which

the polarity of a voltage applied to the liquid crystal is reversed every one horizontal scanning interval.

[0083] The liquid-crystal panel 40 serving as a display part in this liquid-crystal display includes a plurality of scanning signal lines Lg each corresponding to a horizontal scanning line in the image of image data Dv, which is received from an external CPU or the like,

[0084] a plurality of image signal lines Ls, which intersect with each of the plurality of scanning signal lines Lg, and

[0085] a plurality of pixel formation parts, provided so as to correspond to points of intersection between the plurality of scanning signal lines Lg and the plurality of image signal lines Ls. The plurality of pixel formation parts are disposed in a matrix arrangement, each of the pixel formation parts having the same type of configuration as the configuration in an active-matrix type liquid-crystal panel in the past, and having

[0086] a TFT as a switching element, the source terminal of which is connected to an image signal line Ls passing through a corresponding intersection point,

[0087] a pixel electrode connected to the drain terminal of the TFT,

[0088] a common electrode Ec, which is an opposing electrode provided in common associated with the plurality of pixel formation parts, and

[0089] a liquid-crystal layer, provided in common for the plurality of pixel formation parts, which is sandwiched between the pixel electrodes and the common electrode Ec.

[0090] In the above-noted configuration, a pixel capacitance Cp is formed by a pixel electrode, the common electrode Ec, and the liquid crystal sandwiched therebetween. This liquid-crystal display is described in detail below, and it will be noted that there are examples of this type of liquid-crystal display in which, for example, a common electrode Ec is formed on an opposing electrode different from the TFT substrate forming the pixel electrodes, and a type in which the common electrode Ec is formed not on the opposing electrode, but rather on the TFT substrate. In the example described below, the liquid-crystal panel 40 has n image signal lines Ls, and the liquid-crystal panel 40 has 64 gradations.

[0091] In this embodiment, image data (in the narrow sense of the term) representing pixels to be displayed on the liquid-crystal panel (including not only images, but also characters and graphics and the like) and display control data, which is data that determines timing and the like of the display operation (for example, data indicating the frequency of the display clock) is sent from an external CPU or the like to a display control circuit 10 (this data sent from outside hereinafter being referred to as image data in the broad sense of the term, and denoted by the reference symbol Dv). Specifically, an external CPU or the like supplies the (narrow-sense) image data and display control data making up the (broad-sense) image as well as an address signal ADw to the display control circuit 10, and thereby writes the (narrow-sense) image data and display

control data respectively into a display memory and a register within the display control circuit 10, which are described below.

[0092] The display control circuit 10, based on display control data written in the register, generates a clock signal CK, a horizontal synchronization signal HSY, and a vertical synchronization signal VSY, and also generates, based on the horizontal synchronization signal HSY, a polarity reversal control signal  $\phi$  for the purpose of performing AC drive, a shorting control signal Csh, and amplifier stopping control signal Cas. The display control circuit 10 reads image data that is written in the display memory by the external CPU or the like, and outputs three types of digital image signals, Dr, Dg, and Db. Of these, the digital signal Dr is the image signal representing the red component of the image to be displayed (hereinafter referred to as the red image signal), the digital signal Dg is the image signal representing the green component of the image to be displayed (hereinafter referred to as the green image signal), and the digital signal Db is the image signal representing the blue component of the image to be displayed (hereinafter referred to as the blue image signal). In this manner, of the signals generated by the display control circuit 10, the clock signal CK is supplied to the image signal line drive circuit 21, the horizontal synchronization signal HSY and the vertical synchronization signal VSY are supplied to the scanning signal line drive circuit 22, the digital image signals Dr, Dg, and Db, amplifier stopping control signal Cas, and shorting control signal Csh are supplied to the image signal line drive circuit 21, and the polarity reversal control signal  $\phi$  is supplied to the common electrode drive circuit 23 and to the power supply circuit 30. As described above, because the number of gradations is taken to be 64 in this embodiment, each of the digital image signals Dr, Dg, and Db has 6 bits, so that  $6 \times 3 = 18$  lines are required to supply the digital image signals Dr, Dg, and Db from the display control circuit 10 to the image signal line drive circuit 21.

[0093] The power supply circuit 30 supplies a power supply voltage for operation to the display control circuit 10, the image signal line drive circuit 21, the scanning signal line drive circuit 22, and the common electrode drive circuit 23, and also supplies to the image signal line drive circuit 21, the scanning signal line drive circuit 22, and the common electrode drive circuit 23 a reference voltage, which is a voltage serving as a reference for generating a signal to be applied to the liquid-crystal panel 40. The values of reference voltages Vr1 and Vr2 supplied to the image signal line drive circuit 21 are alternatively switched between two pre-established values in response to the polarity reversal control signal  $\phi$ , so that when the polarity reversal control signal  $\phi$  is at a high level the values are such that Vr1 < Vr2, and so that when the polarity reversal control signal  $\phi$  is at a low level the values are such that Vr1 > Vr2. The common electrode drive circuit 23 is supplied with two voltages, VH and VL, serving as reference voltages, such that VH > VL.

[0094] Data representing an image to be displayed by the liquid-crystal panel 40 are supplied as the digital image signals Dr, Dg, and Db in pixel-serial form, as well as a signal indicating timing, the clock signal CK as a control signal, the amplifier stopping control signal Cas, the shorting control signal Csh, and the reference voltages Vr1 and Vr2 are supplied to the image signal line drive circuit 21. The image signal line drive circuit 21, based on these signals and

the reference voltages, generates for each image signal line an image signal (hereinafter referred to as an image drive signal) for driving the liquid-crystal panel **40**, and applies each of the image drive signals to one of the image signal lines  $L_s$  in the liquid-crystal panel **40**.

[0095] The scanning signal line drive circuit **22**, based on the horizontal synchronization signal  $HSY$  and the vertical synchronization signal  $VSY$ , generates scanning signals each to be applied to one of the scanning signal lines  $L_g$  of the liquid-crystal panel **40** for each horizontal scanning period, so as to select the scanning signal lines  $L_g$  alternately and sequentially, repeating the application of an active scanning signal (voltage which turns the TFT on) to each of the scanning signal lines  $L_g$  for sequential selection of all the scanning signal lines  $L_g$ , with one vertical scanning interval as the period.

[0096] The common electrode drive circuit **23** generates a common electrode signal  $V_{com}$  for application of a prescribed potential to the common electrode  $E_c$  of the liquid-crystal panel **40**. In this embodiment, to reduce the amplitude of the voltage of the image signal line  $L_s$ , the potential of the common electrode  $E_c$  is also changed in response to the AC drive. That is, the common electrode drive circuit **23**, in response to the polarity reversal control signal  $\phi$  from the display control circuit **10**, generates a voltage signal that is alternately switched between the two reference voltages  $VH$  and  $VL$  for each one horizontal scanning interval, so that when the polarity reversal control signal  $\phi$  is at a high level the voltage is  $VH$  and when the polarity reversal control signal  $\phi$  is at a low level the voltage is  $VL$ , this voltage being supplied as the common electrode signal  $V_{com}$  to the common electrode  $E_c$  of the liquid-crystal panel **40**. By doing this, the positive and negative polarity of the voltage at the image signal line  $L_s$  relative to the reference potential at the common electrode  $E_c$  can be reversed for each horizontal scanning interval, while reducing the voltage on the image signal line  $L_s$ .

[0097] In a liquid-crystal panel **40** such as described above, image drive signals based on digital image signals  $Dr$ ,  $Dg$ , and  $Db$  from the image signal line drive circuit **21** are supplied to the image signal lines  $L_s$ , a scanning signal is supplied from the scanning signal line drive circuit **22** to the scanning signal line  $L_g$ , and a common electrode signal  $V_{com}$  is supplied from the common electrode drive circuit **23**. By doing this, voltages corresponding to the potential difference between the pixel electrodes and the common electrode  $E_c$ , and responsive to the digital image signals  $Dr$ ,  $Dg$ , and  $Db$ , are applied to the liquid-crystal panel **40**, the polarity of the applied voltages being reversed every one horizontal scanning interval. By making the applied voltages control the transmissivity of the liquid-crystal layer, the liquid-crystal panel **40** displays a color image of the image data received from an external CPU or the like.

[0098] (1.2 Display Control Circuit)

[0099] FIG. 2 is a block diagram showing the configuration of the display control circuit **10** in the above-described liquid-crystal display. This display control circuit **10** has an input control circuit **11**, a display memory **12**, a register **13**, a timing generator circuit **14**, a memory control circuit **15**, and a polarity switching control circuit **16**.

[0100] The display control circuit **10** inputs a signal representing the image data  $Dv$  (which will also hereinafter be

denoted by the reference symbol  $Dv$ ) in the broad-sense of image data received from an external CPU or the like and the address signal  $AD_w$  to the input control circuit **11**. The input control circuit **11**, based on the address signal  $AD_w$ , divides the broad-sense image data  $Dv$  into the three color image data  $R$ ,  $G$ , and  $B$ , and the display control data  $Dc$ . By supplying the signals representing the color image data  $R$ ,  $G$ , and  $B$  (which will also hereinafter be denoted by the reference symbols  $R$ ,  $G$ , and  $B$ ) along with an address signal  $AD$  based on the address signal  $AD_w$  to the display memory **12**, the three image data  $R$ ,  $G$ , and  $B$  are written into the display memory **12** and the display control data  $Dc$  is written into the register **13**. The three image data  $R$ ,  $G$ , and  $B$  are, respectively, data which represent the red component, the green component, and the blue component of the image represented by the image data  $Dv$ . The display control data  $Dc$  includes the frequency of the clock  $CK$ , and the timing information specifying the horizontal scanning interval and the vertical scanning interval for display of the image represented by image data  $Dv$ .

[0101] The timing generator circuit **14**, based on the display control data held in the register **13**, generates the clock signal  $CK$ , the horizontal synchronization signal  $HSY$ , and the vertical synchronization signal  $VSY$ . The timing generator circuit **14** also generates a timing signal for the purpose of synchronizing the operation of the display memory **12** and the memory control circuit **15** to the clock signal  $CK$ .

[0102] The memory control circuit **15** generates an address signal  $AD_r$  for reading out data from the image data  $R$ ,  $G$ , and  $B$  stored in the display memory **12** that represents the image to be displayed on the liquid-crystal panel **40**, and a signal for the purpose of controlling the operation of the display memory **12**. The address signal  $AD_r$  and the control signal are given to the display memory **12**, resulting in data representing the red component, the green component, and the blue component of an image to be displayed on the liquid-crystal panel **40** being read out and outputted to the display control circuit **10** from the display memory **12** as the red image signal  $Dr$ , the green image signal  $Dg$ , and the blue image signal  $Db$ , respectively. These three digital image signals  $Dr$ ,  $Dg$ , and  $Db$  are supplied to the image signal line drive circuit **21**, which is described later.

[0103] The polarity switching control circuit **16** generates the amplifier stopping control signal  $Cas$  and the shorting control signal  $Csh$ , based on the horizontal synchronization signal  $HSY$  generated by the timing generator circuit **14**. The amplifier stopping control signal  $Cas$  is a control signal for the purpose of stopping each of the buffer circuits, to be described later, in the image signal line drive circuit **21**, for a prescribed period of time, when the polarity of the voltage on the image signal line  $L_s$  referenced relative to the common electrode  $E_c$  potential is reversed, and the shorting control signal  $Csh$  is a control signal for the purpose of shorting each of the image signal lines  $L_s$  and the common electrode  $E_c$  at the time of polarity reversal, for just a prescribed amount of time. The amplifier stopping control signal  $Cas$  and the shorting control signal  $Csh$  are supplied to the image signal line drive circuit **21** as described later.

[0104] (1.3 Image Signal Line Drive Circuit)

[0105] FIG. 3 is a circuit diagram showing the configuration of the image signal line drive circuit **21** in the

above-described liquid-crystal display. The image signal line drive circuit **21** is a circuit that generates image drive signals to be supplied to respective image signal lines  $L_s$  in the liquid-crystal display **40**, the image signal line drive circuit **21** supplying  $n$  image drive signals to  $n$  image signal lines  $L_s$  in the liquid-crystal display **40**. The image signal line drive circuit **21** has a sampling latch circuit **110**, a decoder circuit **120**,  $n$  reference voltage selection circuits **131** to **13n**,  $n$  buffer circuits **151** to **15n**,  $n$  stopping control circuit **141** to **14n**, which are on/off switches, a connection switching circuit **160**, which is formed by  $n$  selector switches **161** to **16n**, a bias generator circuit **170**, which generates an amplifier bias  $V_{ba}$  to be supplied to the buffer circuits **151** to **15n**, a voltage divider resistance  $R$ , 64 reference voltage bus lines  $L_1$  to  $L_{64}$  for supplying 64 types of reference voltage to the reference voltage selection circuits **131** to **13n**, this number of reference voltage selection circuits corresponding to the number of gradations in the image display, and  $n$  output terminals  $T_1$  to  $T_n$ , to which  $n$  image signal lines  $L_s$  are respectively connected.

[0106] In the above-described image signal line drive circuit **21**, the sampling latch circuit **110** receives from the display control circuit **10** the red image signal  $D_r$ , formed by 6-bit image signals  $R_5$  to  $R_0$ , the green image signal  $D_g$ , formed by 6-bit image signals  $G_5$  to  $G_0$ , and the blue image signal  $D_b$ , formed by 6-bit image signals  $B_5$  to  $B_0$ , and samples and latches these image signals  $R_5$  to  $R_0$ ,  $G_5$  to  $G_0$ , and  $B_5$  to  $B_0$ , outputting these image signals after latching as internal image signals. These internal image signals are inputted to the decoder circuit **120**.

[0107] The decoder circuit **120**, based on the internal image signals from the sampling latch circuit **110**, generates  $n$  groups of decoded outputs, each corresponding to one of the  $n$  image signal lines  $L_s$ , the  $n$  groups of decoded outputs being input respectively to the  $n$  reference voltage selection circuits **131** to **13n**. Each of the  $n$  groups of decoded output is made up of 64 signals, one signal of each of these groups of 64 signals being made active in response to the above-noted internal image signals, with the other signals being inactive.

[0108] The voltage divider resistance  $R$  has one end which is connected to a first reference voltage  $V_{r1}$ , and another end which is connected to a second reference voltage  $V_{r2}$ , so as to form a voltage divider circuit, this voltage divider circuit generating, in addition to the first and second reference voltages  $V_{r1}$  and  $V_{r2}$ , 62 other types of reference voltages. The 62 reference voltages generated in this manner and the first and second reference voltages  $V_{r1}$  and  $V_{r2}$  are applied, respectively, to the 64 reference voltage bus lines  $L_1$  to  $L_{64}$ , and are, by means of the reference voltage bus lines  $L_1$  to  $L_{64}$ , supplied to each of the reference voltage selection circuits **131** to **13n**. The 64 types of reference voltages are used to apply a voltage to between the pixel electrodes and the common electrode  $E_c$ , this voltage being responsive to each of the gradations of the image display.

[0109] The  $n$  reference voltage selection circuits **131** to **13n** correspond, respectively, to the  $n$  image signal lines  $L_s$ , and include 64 switches, this number being equal to the number of gradations. The 64 switches in each of the reference voltage selection circuits **131** to **13n** input as control signals the 64 signals that make up the decoded output inputted to reference voltage selection circuit to

which the switches belong. Each of the switches is on if the signal inputted thereto is active and off if the signal inputted thereto is inactive. By using this type of switch, each of the reference voltage selection circuits **131** to **13n**, in response to the decoded outputs inputted thereto, selects one of the 64 types of reference voltages supplied thereto by the 64 reference voltage bus lines, and outputs the selected reference voltage (hereinafter referred to as the selected reference voltage). In this manner,  $n$  selected reference voltages outputted from the  $n$  reference voltage selection circuits are inputted respectively to the  $n$  buffer circuits **151** to **15n**.

[0110] During the time when the amplifier bias  $V_{ba}$  is being supplied, each of the buffer circuits **151** to **15n** functions as a voltage follower, meaning that it has an extremely high input impedance and also has a extremely low output impedance, with a gain of substantially 1, but when the amplifier bias  $V_{ba}$  supply is stopped, each of the buffer circuits goes into the stopped condition, in which the power consumption thereof is small enough to neglect, and in which the output impedance is extremely high.

[0111] The buffer circuits **151** to **15n** are provided with stopping control circuits **141** to **14n**, respectively, the stopping control circuits **141** to **14n** acting to control the supply of the amplifier bias  $V_{ba}$  to the respective buffer circuits **151** to **15n**. That is, amplifier stopping control signal  $C_{as}$  such as shown in FIG. 5D is supplied from the display control circuit **10** to the image signal line drive circuit **21**, and when the amplifier stopping control signal  $C_{as}$  is at the high level the stopping control circuits **141** to **14n** allow the supply of the amplifier bias  $V_{ba}$  to the respective buffer circuits **151** to **15n**, but when the amplifier stopping control signal  $C_{as}$  is at the low level, block the supply of the amplifier bias  $V_{ba}$  to the respective buffer circuits **151** to **15n**. The time interval in which the shorting control signal  $C_{sh}$  is at the high level (corresponding to the time interval during which each image signal line  $L_s$  is shorted to the common electrode  $E_c$ ) is either the same as the time interval during which the amplifier stopping control signal  $C_{as}$  is at the low level (this hereinafter being referred to as the amplifier stopped interval, refer to FIG. 5B) or a prescribed time interval that includes the amplifier stopped interval (refer to FIG. 5C). When each of the image signal lines  $L_s$  is shorted to the common electrode  $E_c$ , therefore, the outputs of the buffer circuits **151** to **15n** are always in the high-impedance state.

[0112] The output signals from the  $n$  buffer circuits **151** to **15n** are inputted to  $n$  selector switches **161** to **16n**, respectively, which make up the connection switching circuit **160**. Each of the selector switches **161** to **16n** has a first terminal, a second terminal, and a third terminal, the above-noted output signals that are inputted to the selector switches **161** to **16n** being applied to the respective first terminals thereof. The common electrode signal  $V_{com}$  from the common electrode drive circuit **23** is inputted to the selector switches **161** to **16n**, and is applied to the second terminal of each of the selector switches. The third terminals of the selector switches **161** to **16n** are connected to the output terminals  $T_1$  to  $T_n$ , respectively, of the image signal line drive circuit **21**, these  $n$  output terminals  $T_1$  to  $T_n$  being connected to the  $n$  image signal lines  $L_s$  of the liquid-crystal panel **40**. Each of the selector switches **161** to **16n** makes the third terminal connected to the first terminal when the shorting control signal  $C_{sh}$  is at the low level, and makes the third terminal

connected to the second terminal when the shorting control signal  $C_{sh}$  is at the high level. By doing this, when the shorting control signal  $C_{sh}$  is at the low level, the output signals from the buffer circuits **151** to **15n** are supplied to the respective image signal lines  $L_s$ , and when the shorting control signal  $C_{sh}$  is at the high level, the common electrode signal  $V_{com}$  is supplied to the image signal lines  $L_s$ . When the shorting control signal  $C_{sh}$  is at the high level, therefore, there is a shorting between the signal line which leads to the common electrode signal  $V_{com}$  and each of the image signal lines  $L_s$ , meaning that there is a short between the common electrode  $E_c$  and each of the image signal lines  $L_s$ .

[0113] (1.4 Common Electrode Drive Circuit)

[0114] **FIG. 4A** to **FIG. 4C** are circuit diagrams showing examples of the configuration of the common electrode drive circuit **23** in the liquid-crystal display configured as described above. Because the common electrode drive circuit must in general have a large driving capacity, rather than using an analog buffer, the power consumption of which itself is large, it is common to use a switching circuit type of common electrode drive circuit. Given this, the configurations shown in each of **FIG. 4A** to **FIG. 4C** are not analog buffer type drive circuits, but rather switching circuit type drive circuits using MOS transistors as switching elements.

[0115] In the first configuration example, shown in **FIG. 4A**, the common electrode drive circuit is formed by a p-channel MOS transistor (hereinafter abbreviated as pMOS transistor) and an n-channel MOS transistor (hereinafter abbreviated as nMOS transistor), the drain terminals of both these MOS transistors being mutually connected, the source terminal of the pMOS transistor being connected to the power supply line  $V_{DD}$ , which provides the reference voltage  $V_H$ , and the source of the nMOS transistor being connected to the ground line, which provides the reference voltage  $V_L$ . The polarity reversal control signal  $\phi$  is inputted to the gates of both MOS transistors, and the voltage at the mutually connected drain terminals of the MOS transistors is outputted as the common electrode signal  $V_{com}$ . Therefore, the common electrode signal  $V_{com}$  is  $V_L$  (ground level) when the polarity reversal control signal  $\phi$  is at the high level, and is  $V_H$  (a prescribed positive power supply voltage) when the polarity reversal control signal  $\phi$  is at the low level.

[0116] In the second configuration example, as shown in **FIG. 4B**, the common electrode drive circuit is implemented by two analog switches each formed by a pMOS transistor and an nMOS transistor mutually connected in parallel, the reference voltage  $V_H$  being provided at one end of the first analog switch, and the reference voltage  $V_L$  being provided at one end of the second analog switch, and the other ends of both analog switches being mutually connected. The polarity reversal control signal  $\phi$  is inputted to the gate terminals of the pMOS transistor of the first analog switch and the nMOS transistor of the second analog switch, and signal  $\phi_b$ , which is the polarity reversal control signal  $\phi$  inverted, is inputted at the gate terminals of the nMOS transistor of the first analog switch and the pMOS transistor of the second analog switch. The voltage at the point of connection between the mutually connected analog switches is outputted as the common electrode signal  $V_{com}$ . Therefore, the common electrode signal  $V_{com}$  is  $V_L$  when the

polarity reversal control signal  $\phi$  is at the high level, and is  $V_H$  when the polarity reversal control signal  $\phi$  is at the low level.

[0117] In the third configuration example, as shown in **FIG. 4C**, the common electrode drive circuit, in addition to the circuit of the first configuration example, includes a DC bias circuit and a DC blocking capacitor, the drain terminals of the pMOS transistor and the nMOS transistor being connected to the output terminal of the DC bias circuit via the DC blocking capacitor, and the voltage at the point of connection thereof being output as the common electrode signal  $V_{com}$ . By doing this, the common electrode signal  $V_{com}$  is maintained at an amplitude of  $(V_H - V_L)$ , the same as in the first configuration example, and the DC bias circuit functions to adjust that level.

[0118] (1.5 Liquid-Crystal Panel Drive Method)

[0119] A method for driving a liquid-crystal display configured as described above is described below.

[0120] In a liquid-crystal display of the past, in addition to performing AC drive, whenever the polarity of the voltage applied across the liquid-crystal layer of the liquid-crystal panel being reversed every one horizontal scanning interval, the drive method also includes use of an AC common electrode signal so as to reduce the amplitude of the voltage on the image signal lines, the image signal line potential  $V_v$  in the liquid-crystal panel varies as shown in **FIG. 6A**, and the common electrode signal  $V_{com}$ , which is the potential of the common electrode  $E_c$ , varies as shown in **FIG. 6B**. In this case, however, the image signal line potential  $V_v$  is taken as being the potential at a position that is sufficiently distant from the point of connection between the image signal line drive circuit and the image signal line (this applying below as well). As shown in **FIG. 6A**, in a liquid-crystal display of the past, the image signal line drive circuit, in the case of the normally white mode, must be able to cause a change in the image signal line that is at maximum twice the voltage to be applied across the liquid-crystal layer in order to display black.

[0121] In contrast to the above, in the case of this embodiment, based on the horizontal synchronization signal  $H_{SY}$  a shorting control signal  $C_{sh}$  as shown in **FIG. 7C** is generated, and when the polarity of the voltage of the image signal line  $L_s$  referenced to the potential of the common electrode  $E_c$  is reversed, each of the image signal lines  $L_s$  in the liquid-crystal panel, in accordance with the shorting control signal  $C_{sh}$ , is electrically separated from the image signal line drive circuit **21** and also shorted to the common electrode  $E_c$ . That is, in the reversal of polarity that occurs every one horizontal synchronization interval, the scanning signal  $G(j)$  applied to the scanning signal line  $L_g$  that was immediately previously selected becomes inactive (low level), and after all of the TFTs connected to this scanning signal line  $L_g$  are turned off, for example at some time  $t_1$  (refer to **FIG. 7A**), the shorting control signal  $C_{sh}$  changes to the high level, and each of the image signal lines  $L_s$  in the liquid-crystal panel **40**, by the connection switching circuit **160**, is electrically separated from the image signal line drive circuit **21** and shorted to a signal line that leads to the common electrode signal  $V_{com}$ . During the period of time in which each image signal line  $L_s$  and the common electrode  $E_c$  are shorted (this being referred to as the shorting interval, which can be treated as the same as the time period

during which the shorting control signal  $C_{sh}$  is at the high level), the charge accumulated in the capacitance formed between the image signal lines and the common electrode  $E_c$  is discharged, so that for example at some time  $t_2$  the image signal lines  $L_s$  and the common electrode  $E_c$  are at substantially the same potential. As shown in **FIG. 7B and 7C**, during the shorting interval the positive/negative polarity of the voltage on the image signal line  $L_s$  referenced to the potential of the common electrode  $E_c$  reverses based on the polarity reversal control signal  $\phi$  (this positive/negative polarity reversal being referred to hereinafter as “polarity reversal”). Therefore, the value of the common electrode signal  $V_{com}$  switches between the two reference voltages  $V_L$  and  $V_H$  during the shorting interval. By the switching of the value of the common electrode signal  $V_{com}$ , the potential  $V_v$  of the image signal line  $L_s$  also changes by the amount of the change in the common electrode signal  $V_{com}$ . Thereafter, when the shorting control signal  $C_{sh}$  changes from the high level to the low level, the buffer circuits **151** to **15n** within the image signal line drive circuit **21** are connected to the respective image signal lines  $L_s$ . After the elapse of the polarity reversal interval, for example at some time  $t_3$ , the supply of the inverted-polarity image drive signals to the image signal lines  $L_s$  is started, and when the TFTs connected to the scanning signal line  $L_g$  selected next (refer to **FIG. 7E**), the image drive signals are applied to the pixel electrodes connected to these TFTs.

**[0122]** According to a method such as described above, the waveform (voltage waveform) of the potential  $V_v$  on the image signal line  $L_s$  in the liquid-crystal panel **40** is as shown in **FIG. 7A**. The part of the voltage waveform during the period in which the shorting control signal  $C_{sh}$  is at the low level, is the waveform in accordance with the output signals of the output buffer circuits **151** to **15n** within the image signal line drive circuit **21**. In this embodiment, as can be seen by comparing **FIG. 7A** with **FIG. 6A**, without substantially changing the voltage applied across the liquid-crystal layer, it is possible to achieve a significant reduction in the voltage amplitude on the image signal lines  $L_s$  to be changed in comparison with the past. That is, by the operation of shorting the image signal lines  $L_s$  and the common electrode  $E_c$  by means of the connection switching circuit **160** (hereinafter simply referred to as the shorting operation), the image signal lines  $L_s$  and the common electrode  $E_c$  are at substantially the same potential, so that the amount of change  $\Delta 1$  of the potential  $V_v$  of the image signal lines  $L_s$  in accordance with the buffer circuits **151** to **15n** within the image signal line drive circuit **21** is substantially one-half of the amount of change  $\Delta 0$  of the potential  $V_v$  of the image signal lines  $L_s$  in accordance with the buffer circuits within the image signal line drive circuit of the past (**FIG. 6A**).

**[0123]** In the above, it was assumed that within the shorting interval each image signal line  $L_s$  and the common electrode  $E_c$  become at substantially the same potential, and in order for this assumption to hold it is necessary to set the shorting interval (pulse width of the shorting control signal  $C_{sh}$ ) in accordance with the value of the capacitance formed in the liquid-crystal panel **40** between the image signal line  $L_s$  and the common electrode  $E_c$  and the resistance value of the image signal line  $L_s$ . In a lumped-constant circuit formed by a resistance and a capacitance (integrator circuit) in which the charge accumulated in the capacitor is caused to discharge, when an amount of time that is three times the time constant, which is the product of the resistance value

and the capacitance value, elapses, the approximately 95% of the charge that had been accumulated originally in the capacitor is discharged. In this embodiment, the shorting interval is set so that it is a period of time that is at least three times the delay time constant that is the product of the wiring resistance value and the wiring capacitance value of one image signal line  $L_s$ . In reality, because it is necessary to consider the on-state resistance of the switch in the connection switching circuit **160** and the impedance of the common electrode drive circuit **23** and the like in establishing the shorting interval, the length of the shorting interval is preferably made at least three times the above-noted delay time constant.

**[0124]** The amount of time that can be used for writing a pixel value into a pixel formation part in the liquid-crystal panel **40** (that is, for charging the pixel capacitance  $C_p$  by a voltage corresponding to a pixel value) is the time of the horizontal scanning interval after subtracting the shorting interval and the polarity reversal interval. According to the above-described method, therefore, because the polarity reversal is performed within the shorting interval, if one horizontal scanning interval is held fixed, there is the advantage that the time usable for writing the pixel value is made long.

**[0125]** As shown in **FIG. 5B** to **FIG. 5D**, in at least the shorting interval, the amplifier stopping control signal  $C_{as}$  is at the low level, and all the buffer circuits **151** to **15n** and the bias generator circuit **170** are in the stopped condition.

**[0126]** In the above-described drive method, the scanning line  $G(j+1)$  selected immediately after the above-noted switching interval is at the high level (active), as shown in **FIG. 7C** to **FIG. 7E**, after the shorting control signal  $C_{sh}$  changes to the low level. Therefore, when all the TFTs in the liquid-crystal panel **40** are off, each of the image signal lines  $L_s$  are shorted to the common electrode  $E_c$ . However, during the time period in which the shorting control signal  $C_{sh}$  is at the high level (shorting interval), even if the scanning signal  $G(j+1)$  becomes active and the TFTs connected to scanning lines  $L_g$  leading to the scanning signal  $G(j+1)$  switch on, and pixel electrodes connected to these TFTs are shorted to the common electrode  $E_c$ , because the charging time constant of the pixel capacitance is between several tens of times and several hundred times the charging time constant of the image signal line  $L_s$ , during the shorting operation, there is substantially no change in the potential on these pixel electrodes. Even if the potential on these pixel electrodes should change, the potential change is in the direction that approaches a potential corresponding to the pixel value to be written next. Therefore, as shown in **FIG. 7F**, the scanning signal  $G(j+1)$  of the scanning signal line  $L_g$  selected immediately after the above-noted polarity reversal can change to the high level before the shorting control signal  $C_{sh}$  changes to the high level.

**[0127]** Although in the above-described drive method the polarity reversal is performed during the shorting interval, it is alternatively possible to perform the polarity reversal outside of the shorting interval. For example, in a case in which the polarity reversal is performed before the shorting interval, the waveforms of the potential  $V_v$  on the image signal line  $L_s$ , the common electrode signal  $V_{com}$ , and the shorting control signal  $C_{sh}$  are as shown in **FIG. 8A** to **FIG. 8C**. In this case as well, the amount of change  $\Delta 2$  of the

potential  $V_v$  on the image signal line  $L_s$  according to the buffer circuits **151** to **15n** within the image signal line drive circuit **21** is, by virtue of the shorting operation, substantially one-half of the amount of potential change  $\Delta 0$  (**FIG. 6A**) of the image signal line  $L_s$  in the past, this representing a significant reduction.

[0128] Additionally, in a case in which the polarity reversal is performed after the shorting interval, the waveforms of the potential  $V_v$  on the image signal line  $L_s$ , the common electrode signal  $V_{com}$ , and the shorting control signal  $C_{sh}$  are as shown in **FIG. 9A** to **FIG. 9C**. In this case as well, the amount of change  $\Delta 3$  of the potential  $V_v$  on the image signal line  $L_s$  according to the buffer circuits **151** to **15n** within the image signal line drive circuit **21** is, by virtue of the shorting operation, substantially one-half of the amount of potential change  $\Delta 0$  (**FIG. 6A**) of the image signal line  $L_s$  in the past, this representing a significant reduction.

[0129] Additionally, in a case in which the drive of the image signal lines  $L_s$  by the buffer circuits **151** to **15n** is started before the completion of the discharging by the shorting operation, because the shorting interval is short although the polarity reversal is performed within the shorting interval, the waveforms of the potential  $V_v$  on the image signal line  $L_s$ , the common electrode signal  $V_{com}$ , and the shorting control signal  $C_{sh}$  are as shown in **FIG. 10A** to **FIG. 10C**. In this case, the amount of change  $\Delta 4$  of the potential  $V_v$  on the image signal line  $L_s$  according to the buffer circuits **151** to **15n** within the image signal line drive circuit **21** is larger than one-half of the amount of potential change  $\Delta 0$  (**FIG. 6A**) of the image signal line  $L_s$  in the past, and is however significantly smaller than the amount of potential change  $\Delta 0$  in the past by virtue of the shorting operation.

[0130] As described above, even if the timing of causing the shorting of the image signal line  $L_s$  and the common electrode  $E_c$  (shorting interval) does not completely coincide with the reversal of the polarity of the voltage applied to the image signal line referenced to the common electrode  $E_c$  (the polarity reversal interval), which is the time of the polarity reversal of the voltage applied to the capacitive load, it is sufficient that it be possible to treat this as being synchronized relative to one horizontal scanning interval, and if it is within this range of synchronization, there is no particular problem with regard to relative before/after timing before/after between the shorting and polarity reversal intervals. It will be understood that, although in the above-noted embodiment it is indicated that the image signal line  $L_s$  and the common electrode  $E_c$  are shorted, that is that the common electrode signal  $V_{com}$  is supplied to the image signal line  $L_s$ , at the time of the polarity reversal, the present invention is not restricted to the common electrode signal  $V_{com}$ , it being alternatively possible to supply to the image signal line  $L_s$  a voltage level that is equivalent to the common electrode signal  $V_{com}$ . For example, it is possible to provide a circuit similar to the common electrode drive circuit **23**, and to supply to the image signal line  $L_s$  a voltage level equivalent to that of the common electrode signal  $V_{com}$ . It will be understood that, in an embodiment such as described above, if a common electrode signal  $V_{com}$  itself is used as the voltage level equivalent to that of the common electrode signal  $V_{com}$ , it is not necessary to provide a circuit for generating a voltage level equivalent to the common electrode signal  $V_{com}$  other than the common electrode

drive circuit **23**. This will apply in the description to follow, which is based on the assumption that at the time of polarity reversal the image signal line  $L_s$  or the like is shorted with the common electrode  $E_c$ , that is, that the image signal line  $L_s$  or the like is supplied with the common electrode signal  $V_{com}$ , although it will be understood that there is no restriction to the common electrode signal  $V_{com}$ , it being possible also to supply to the image signal line  $L_s$  or the like a voltage level equivalent to the common electrode signal  $V_{com}$ . That is, the electrode that is to be shorted with the image signal line  $L_s$  or the like at the time of polarity reversal is not restricted to being the common electrode  $E_c$ , but can alternatively be an electrode that supplies a voltage level equivalent to that of the common electrode signal  $V_{com}$ .

[0131] (1.6 Advantageous Effect)

[0132] In the above-described embodiment, polarity reversal is performed every one horizontal synchronization interval in order to implement AC drive of the liquid-crystal panel **40**, and when the polarity reversal occurs each image signal line  $L_s$  is electrically separated from the buffer circuits **151** to **15n** within the image signal line drive circuit **21** and shorted to the common electrode  $E_c$ . By doing this, the charge accumulated in the capacitance formed by image signal lines  $L_s$  and the common electrode  $E_c$  is discharged, after which the image signal lines  $L_s$  are reconnected to the buffer circuits **151** to **15n** within the image signal line drive circuit **21**. Therefore, the amount of change  $\Delta 1$ ,  $\Delta 2$ ,  $\Delta 3$ , or  $\Delta 4$  (amount of change when the shorting control signal  $C_{sh}$  is at the low level) of the potential  $V_v$  on the image signal line  $L_s$  in accordance with the buffer circuits **151** to **15n** is significantly smaller than the amount of potential change  $\Delta 0$  of the image signal line  $L_s$  in the past, so that in the case in which the image signal lines  $L_s$  and the common electrode  $E_c$  become at the same potential within the shorting interval, there is substantially a halving of the amount of change  $\Delta 0$  of the past. That is, whereas the amount of change (voltage change) to be made in the image signal line in the next horizontal synchronization interval in the case of the normal white mode was two times the maximum voltage required for a black display, in the above-described embodiment it is no more than the amount of voltage required for a black display. As a result, in this embodiment even if the buffer circuits used have a smaller drive capacity than the buffer circuits used in the past, it is possible to apply across the liquid-crystal layer in the liquid-crystal panel a voltage equivalent to the voltage applied in the past. For this reason, by using buffer circuits having a lower drive capacity than those used in the past, it is possible to reduce the power consumption of the image signal line drive circuit **21** and reduce the size of the transistors used in the buffer circuits used in the buffer circuits **151** to **15n**, thereby enabling a reduction in the cost of the IC chip used to implement the image signal line drive circuit **21**. By doing this, it is possible to achieve a compact, low-cost liquid-crystal display. A liquid-crystal display according to this embodiment of the present invention, therefore, is suitable for used in portable equipment.

[0133] Because the wiring capacitance in each of the image signal lines  $L_s$  in the liquid-crystal panel **40** is sufficiently larger than the capacitance  $C_p$  for one pixel connected to each of the image signal lines  $L_s$ , the above-described method for enabling use of buffer circuits having

a low drive capacity using a shorting operation is effective regardless of whether the scanning signal line  $L_g$  is active or inactive at the time of the shorting operation. The fact that the amount of potential change at the image signal line  $L_s$  in the next horizontal synchronization interval after the above-noted polarity reversal is small means more generally that the current to be supplied from the power supply to the image signal line  $L_s$  (current consumption) is made small. That is, by the shorting operation at the time of polarity reversal, because the charge accumulated in the capacitance formed between the image signal line  $L_s$  and the common electrode  $E_c$  is directly discharged, without going through the power supply, there is reduction in the current supplied to the image signal line from the power supply (capacitance between the image signal line and the common electrode) commensurate with this direct discharging, resulting in a reduction in the power consumption of the image signal line drive circuit 21. As described above, it is alternatively possible, rather than to cause shorting of the image signal line  $L_s$  to the common electrode  $E_c$  at the time of polarity reversal, to cause shorting thereof to a different electrode supplying a voltage that is equivalent to that of the common electrode signal  $V_{com}$ , in which case, the current for the purpose of causing a discharging of the charge accumulated in the capacitance between the image signal line and the common electrode  $E_c$  is sometimes supplied from a power supply via a prescribed circuit. However, the voltage level that is equivalent to that of the common electrode signal  $V_{com}$  can be supplied to the image signal line  $L_s$  not from buffer circuits 151 to 15n functioning as analog buffers, but rather from a circuit formed by MOS transistors or the like as switching elements, in the same manner as the common electrode drive circuit 23. In this case as well, therefore, compared to the conventional configuration, it is possible to achieve a great reduction in the power consumption.

[0134] In the above-described embodiment, because the buffer circuits 151 to 15n and the bias generator circuit 170 are stopped by the amplifier stopping control signal  $C_{as}$  during the shorting interval or a prescribed period of time that includes the shorting interval, this also contributes to a reduction in the power consumption of the image signal line drive circuit 21.

[0135] In the Japanese unexamined patent application publication H6-337657, there is disclosure of a liquid-crystal display characterized in that during a vertical blanking interval the output potential on an image signal line is made to be the same as the potential on the common electrode of a liquid-crystal pixel. This liquid-crystal display, in so far as it reduces the power consumption by making the potential on the image signal line and the potential on the common electrode the same, is similar to the above-described embodiment. However, whereas in this liquid-crystal display the charging and discharging of the liquid crystal pixel is reduced by eliminating the difference in potential between the image signal line and the common electrode during the vertical blanking interval, which is unrelated to the display, thereby reducing wasteful power consumption, in the above-described embodiment of the present invention, the drive capacity that is required of the buffer circuits 151 to 15n is made small by shorting (equalizing the potential between) the image signal lines  $L_s$  and the common electrode  $E_c$  at the time of polarity reversal for AC drive of the liquid-crystal panel 40, thereby reducing the power consumption of the image signal line drive circuit 21,

the basic ideas for solving the problem of reducing the power consumption being different between the two. Furthermore, in a case such as described above, in which the polarity of the voltage applied to the liquid-crystal panel (liquid-crystal layer) is reversed every one horizontal synchronization interval by AC drive, the two different not only in terms of constitution, but also in the degree of reduction of the power consumption, and it is not possible with the liquid-crystal display noted in the Japanese unexamined patent application publication H6-337657 to achieve a great effect in reducing the power consumption.

[0136] Furthermore, because the common electrode drive circuit 23 in the above-described embodiment is implemented using MOS transistors as switching elements, there is a large drive capacity in spite of a small power consumption, and the shorting of each image signal line  $L_s$  and the common electrode  $E_c$  for each one horizontal synchronization interval such as done in the above-described embodiment does not impose a load on the common electrode drive circuit 23. Also, in an active-matrix type liquid-crystal display such as described regarding the above embodiment, because the voltage level supplied to the image signal line  $L_s$  immediately after the image signal line  $L_s$  is made inactive (that is, the associated TFTs are at the voltage level that turns them off) does not influence the display on the liquid-crystal panel 40, the above-noted shorting operation does not create a display problem.

[0137] (2. Second Embodiment)

[0138] FIG. 11 is a circuit diagram showing the configuration of an image signal line drive circuit in a liquid-crystal display according to a second embodiment of the present invention, and also showing the configuration of the part supplying signals to image signal lines  $L_s$  in the liquid-crystal panel 45. In the liquid-crystal display according to this embodiment, differing from that of the first embodiment, a connection switching circuit 180 that causing shorting of the image signal lines to the common electrode  $E_c$  is built into the liquid-crystal panel 45, and the image signal line drive circuit does not include a connection switching circuit 160. That is, the outputs of the buffer circuits 151 to 15n are outputted as OUT1 to OUTn from the image signal line drive circuit, and inputted to the connection switching circuit 180 within the liquid-crystal panel 45. The shorting control signal  $C_{sh}$  and the common electrode signal  $V_{com}$  are also inputted to the connection switching circuit 180. This connection switching circuit 180, similar to the connection switching circuit 160 in the first embodiment, is formed by n selector switches 181 to 18n, the n output signals OUT1 to OUTn from the image signal line drive circuit being inputted respectively to the selector switches 181 to 18n. Each of the selector switches 181 to 18n has a first, a second, and a third terminal, the output signals OUT1 to OUTn inputted to the selector switches 181 to 18n being applied to the first terminals thereof. The common electrode signal  $V_{com}$  from the common electrode drive circuit 23 is applied to the second terminals of the selector switches 181 to 18n. The image signal lines  $L_s$  in the liquid-crystal panel 45 are connected to the third terminals of the respective selector switches 181 to 18n. The selector switches 181 to 18n make connection to the first terminal when the shorting control signal  $C_{sh}$  is at the low level, and make connection to the shorting control signal  $C_{sh}$  is the high level. By doing this, when the shorting control signal  $C_{sh}$  at the low level,

the output signals from the buffer circuits **151** to **15n** are supplied to the respective image signal lines **Ls**, and when the shorting control signal **Csh** is at the high level, the common electrode signal **Vcom** is supplied to the image signal lines **Ls**. Therefore, when the shorting control signal **Csh** is at the high level, there is a shorting between each of the image signal lines **Ls** and the common electrode **Ec**. Aspects of this embodiment of the present invention other than noted above are the same as described with regard to the first embodiment, and corresponding elements in the second embodiment are assigned the same reference numerals as in the first embodiment and are not explicitly described herein. With regard to the method of driving the liquid-crystal panel as well, description is not provided, as this is the same as described regarding the first embodiment.

**[0139]** In the embodiment described above as well, similar to the case of the first embodiment, when the polarity of the voltages applied to the image signal lines **Ls** in the liquid-crystal panel **45** is reversed for AC drive thereof, the image signal lines **Ls** are electrically separated from the buffer circuits **151** to **15n** within the image signal line drive circuit and shorted to the common electrode **Ec**. By doing this, even if buffer circuits having a drive capacity that is smaller than in the past are used as the buffer circuits **151** to **15n** in the image signal line drive circuit, it is possible to apply a voltage to the liquid-crystal panel that is similar to that applied in the past. Therefore, according to this embodiment, by using a buffer circuit having a drive capacity that is lower than in the past, it is possible to reduce the power consumption of the image signal line drive circuit, and further possible to reduce the size of the transistors used to implement the buffer circuits **151** to **15n**. As a result, it is possible to achieve a reduction in both the size and the cost of the liquid-crystal display.

#### **[0140]** (3. Third Embodiment)

**[0141]** **FIG. 12** is a block diagram showing the configuration of a liquid-crystal display **40** according to a third embodiment of the present invention. In this liquid-crystal display, a circuit that is equivalent to the common electrode drive circuit **23** in the liquid-crystal display according to the first embodiment shown in **FIG. 1** is built into the image signal line drive circuit **24**. Therefore, in this embodiment the reference voltages **VH** and **VL** and the polarity reversal control signal **φ** for generating the common electrode signal **Vcom** are supplied to the image signal line drive circuit **24**, the common electrode signal **Vcom** being applied to common electrode **Ec** of the liquid-crystal panel **40** from the image signal line drive circuit **24**. **FIG. 13** is a circuit diagram showing the configuration of this image signal line drive circuit **24**, which has built therewithin a common electrode drive circuit **200** having the same configuration as the already described circuit shown in **FIG. 4A**, but is otherwise similar in configuration to the first embodiment, so that corresponding elements to elements in the first embodiment are assigned the same reference numerals and are not explicitly described herein. With regard to the method for driving this liquid-crystal panel **40** as well, since the method is the same as for the first embodiment, the method is not described herein. Because the image signal line drive circuit **24** has a built in common electrode drive circuit **200**, the output signal of the common electrode drive circuit **200** is applied as the common electrode signal **Vcom**

to the second terminals of the selector switches **161** to **16n** in the connection switching circuit **160**.

**[0142]** In the above-noted embodiment as well, similar to the case of the first embodiment, when the polarity is reversed as part of AC drive of the liquid-crystal panel **40**, the image signal lines **Ls** in the liquid-crystal panel **40** are electrically separated from the buffer circuits **151** to **15n** within the image signal line drive circuit **24** and shorted to the common electrode **Ec**. By doing this, even if buffer circuits having a drive capacity that is smaller than in the past are used as the buffer circuits **151** to **15n** in the image signal line drive circuit, it is possible to apply a voltage to the liquid-crystal panel that is similar to that applied in the past. Therefore, according to this embodiment, by using a buffer circuit having a drive capacity that is lower than in the past, it is possible to reduce the power consumption of the image signal line drive circuit, and further possible to reduce the size of the transistors used to implement the buffer circuits **151** to **15n**. As a result, it is possible to achieve a reduction in both the size and the cost of the liquid-crystal display.

**[0143]** As is the case in the first and second embodiments, in the case in which the common electrode signal **Vcom** is applied to the image signal lines **Ls** in place of the output signals from the buffer circuit when the polarity is reverse to perform AC drive of the liquid-crystal panel, by packaging the image signal line drive circuit including a common electrode drive circuit on one chip as described above, the configuration of the liquid-crystal display **40** is simplified. There is the same advantage, not only in the case in which the image signal line drive circuit and common electrode drive circuit are implemented as a single chip, but also in the case in which the image signal line drive circuit and the common electrode drive circuit are formed on one and the same substrate making up the liquid-crystal panel.

#### **[0144]** (4. Fourth Embodiment)

**[0145]** **FIG. 14** is a circuit diagram showing the configuration of an image signal line drive circuit in a liquid-crystal display according to a fourth embodiment of the present invention. The liquid-crystal display of this embodiment differs from that of the first embodiment in that, rather than providing the buffer circuits **151** to **15n** between the reference voltage selection circuits **131** to **13n** and the connection switching circuits **160** in the image signal line drive circuit, 62 buffer circuits **222** to **2263** are disposed between a voltage divider circuit formed using a resistance **R** and 62 reference voltage bus lines **L2** to **L63** for passing the 62 types of reference voltages generated by the voltage divider. An on/off switch is provided as stopping control circuits **212** to **2163** for each of the buffer circuits **222** to **2263**, each of the stopping control circuits **2163** controlling the supply of the amplifier bias **Vba** to buffer circuits **222** to **2263**, based on the amplifier stopping control signal **Cas**, as shown in **FIG. 5D**. In this embodiment, the shorting control signal **Csh** is inputted to a decoder circuit **125**, which based on the internal image signals from a sampling latch circuit **110**, generates **n** groups of decoded outputs corresponding to the **n** image signal lines **Ls**, respectively. Because other aspects of the configuration of this embodiment are the same as the first embodiment, corresponding elements are assigned the same reference numerals as in the first embodiment, and are not explicitly described herein. Similarly, since the method

for driving the liquid-crystal panel **40** is the same as in the first embodiment, this method is not described herein.

[0146] During the time when the amplifier bias  $V_{ba}$  is being supplied, the buffer circuits **222** to **2263** in the above-described configuration have an extremely high input impedance and an extremely low output impedance, the gain thereof under this condition being substantially one, so that they function as voltage followers. When the amplifier bias  $V_{ba}$  supply is stopped, however, the buffer circuits go into the stopped condition, in which the power consumption thereof is small enough to neglect, and in which the output impedance is high. Furthermore, because the first and second reference voltages  $V_{r1}$  and  $V_{r2}$  are supplied from the power supply circuit **30**, these reference voltages  $V_{r1}$  and  $V_{r2}$  are applied to the reference voltage bus lines  $L_1$  and  $L_{64}$ , respectively, via a buffer circuit. In the image signal line drive circuit in the above-noted configuration, because there are no buffer circuits between the reference voltage selection circuits **131** to **13n** and the connection switching circuit **160**, the selected reference voltages from the reference voltage selection circuits **131** to **13n** are applied to the first terminals of the selector switches **161** to **16n** in the connection switching circuit **160**.

[0147] In the above-described configuration, the  $n$  groups of decoded outputs from the decoder circuit **125** are each made up of 65 signals (number of gradations plus 1), of which 64 signals are inputted to the reference voltage selection circuits **131** to **13n**. When the shorting control signal  $C_{sh}$  is at the low level, similar to the case of the first embodiment, only one of these 64 signals is made active, responsive to the above-noted internal image signals. One remaining signal of each of the  $n$  groups of decoded signals is inputted to the selector switches **161** to **16n** in the connection switching circuit **160**. The decoded output signal inputted to the selector switches **161** to **16n** are inactive when the shorting control signal  $C_{sh}$  is at the low level, and are active when the shorting control signal  $C_{sh}$  is at the high level. When the shorting control signal  $C_{sh}$  is at the high level, all of the signals of the  $n$  decoded outputs that are inputted to the reference voltage selection circuits **131** to **13n** are inactive. Therefore, when the shorting control signal  $C_{sh}$  is at the low level, the selected reference voltage outputted from the reference voltage selection circuits **131** to **13n** in response to the internal image signals from the sampling latch circuit **110** are output from the image signal line drive circuit as the output signals  $OUT_1$  to  $OUT_n$ , and are supplied to the image signal lines  $L_s$  of the liquid-crystal panel **40**. When the shorting control signal  $C_{sh}$  is at the high level, however, the common electrode signal  $V_{com}$  is supplied to each of the image signal lines  $L_s$  of the liquid-crystal panel **40**. This means that there is a shorting between the common electrode  $E_c$  and the image signal lines  $L_s$  when the shorting control signal  $C_{sh}$  is at the high level. In an image signal line drive circuit such as this, the reference voltage selection circuit includes the connection switching circuit **160** and in the reference voltage selection circuit thereof, it can be thought that one voltage is selected for each image signal line from the 65 types of voltages made up of the 64 types reference voltages corresponding to the number of gradations and the common electrode signal  $V_{com}$ , the selected voltage being output as the output signals  $OUT_1$  to  $OUT_n$ .

[0148] In the above-described embodiment as well, similar to the case of the first embodiment, when the polarity is reversed as part of AC drive of the liquid-crystal panel **40**, the image signal lines  $L_s$  in the liquid-crystal panel **40** are electrically separated from each of the reference voltage selection circuits **131** to **13n** and each of the buffer circuits **222** to **2263** within the image signal line drive circuit and are shorted to the common electrode  $E_c$ . By doing this, even if buffer circuits having a drive capacity that is smaller than in the past are used as the buffer circuits **222** to **2263** in the image signal line drive circuit, it is possible to apply a voltage to the liquid-crystal panel that is similar to that applied in the past. Therefore, according to this embodiment, by using a buffer circuit having a drive capacity that is lower than in the past, it is possible to reduce the power consumption of the image signal line drive circuit, and further possible to reduce the size of the transistors used to implement the buffer circuits **222** to **2263**. As a result, it is possible to achieve a reduction in both the size and the cost of the liquid-crystal display. Additionally, similar to the case of the first embodiment, because the buffer circuits **222** to **2263** and the bias generator circuit **170** are in the stopped condition because of the amplifier stopping control signal for the shorting interval or for a prescribed time interval including the shorting interval (refer to FIG. 5A to FIG. 5D), this also contributes to a reduction of the power consumption of the image signal line drive circuit.

[0149] In addition to the above, in this embodiment, because the configuration is such that the common electrode signal  $V_{com}$  can be treated as one of the reference voltages, with one of the 65 reference voltages selected, that is, the configuration is such that the connection switching circuit **160** is included within the reference voltage selection circuit, compared to a configuration such as shown in FIG. 3, in which the group of switches as the reference voltage selection circuits **131** to **13n**, the group of switches as the buffer circuits **151** to **15n** and the reference voltage selection circuit **160** are disposed in order, it is possible to achieve a compact circuit configuration in the image signal line drive circuit. For this reason, according to this embodiment it is possible to implement an image signal line drive circuit achieving the same type of effect as the first embodiment on a small IC chip, resulting in a more compact and lower-cost liquid-crystal display.

#### [0150] (5. Fifth Embodiment)

[0151] FIG. 15 is a circuit diagram showing the configuration of an image signal line drive circuit in a liquid-crystal display according to a fifth embodiment of the present invention. This liquid-crystal display, rather than the connection switching circuit **160** that is provided in the output section of the image signal line drive circuit of the fourth embodiment, is provided with a voltage switching circuit **300** formed by one selector switch, a reference voltage  $V_{r2}$  supplied from an external power supply circuit being applied to the reference voltage bus line  $L_{64}$  via this voltage switching circuit **300**. The shorting control signal  $C_{sh}$  is inputted to this voltage switching circuit **300** as a control signal to control the switching thereof. Because other aspects of the configuration of this embodiment are basically the same as the fourth embodiment, corresponding elements are assigned the same reference numerals as in the fourth embodiment and are not explicitly described herein. The

method for driving the liquid-crystal panel **40**, being the same as that of the first embodiment, is also not described herein.

[0152] In the image signal line drive circuit of this embodiment, a decoder circuit **126** is substantially the same as the decoder circuit **125** in the fourth embodiment, with the exception that the operation occurs when the shorting control signal *Csh* is at the high level. That is, when the shorting control signal *Csh* is at the high level, whereas in the fourth embodiment all the decoded output signals inputted to the reference voltage selection circuits **131** to **13n** are inactive, in this embodiment of the switches in the reference voltage selection circuits **131** to **13n**, only the decoded output inputted to a switch that is connected to the reference voltage bus line **L64** is active.

[0153] In the image signal line drive circuit of this embodiment, the voltage switching circuit **300** has a first terminal, a second terminal, and a third terminal, the reference voltage bus line **L64** being connected to the first terminal, the reference voltage **Vr2** being connected to the second terminal, and the common electrode signal **Vcom** being connected to the third terminal. The voltage switching circuit **300** makes connection of the first terminal to the second terminal when the shorting control signal *Csh* is at the low level, and makes connection of the first terminal to the third terminal when the shorting control signal *Csh* is at the high level. By doing this, with respect to the reference voltage bus line **L64**, when the shorting control signal *Csh* is at the low level the reference voltage **Vr2** is provided, and when the shorting control signal *Csh* is at the high level the common electrode signal **Vcom** is provided.

[0154] Therefore, when the shorting control signal *Csh* is at the low level, in response to the internal image signals from the sampling latch circuit **110**, the selected reference voltages outputted from the reference voltage selection circuits **131** to **13n** are supplied as the output signals **OUT1** to **OUTn** respectively to the image signal lines **Ls** of the liquid-crystal panel **40** from the image signal line drive circuit. If the shorting control signal *Csh* is at the high level, however, the common electrode signal **Vcom** is supplied to the liquid-crystal panel **40** via the reference voltage selection circuits **131** to **13n**. This means that when the shorting control signal *Csh* is at the high level there is shorting between the common electrode **Ec** and the image signal lines **Ls**.

[0155] In the above-described embodiment as well, it is possible to achieve an effect similar to the fourth embodiment. That is even if buffer circuits having a small drive capacity are used as the buffer circuits **222** to **2263** in the image signal line drive circuit, it is possible to apply a voltage to the liquid-crystal panel that is similar to that applied in the past. Additionally, it is possible to reduce the size of the transistors used to implement the buffer circuits **222** to **2263**. Additionally, because the buffer circuits **222** to **2263** and the bias generator circuit **170** are in the stopped condition because of the amplifier stopping control signal *Cas* for the shorting interval or for a prescribed time interval including the shorting interval (refer to FIG. 5A to FIG. 5D), this also contributes to a reduction of the power consumption of the image signal line drive circuit. Furthermore, there is the advantage, compared with the first embodiment, of a reduction in the number of buffer circuits required.

[0156] In this embodiment, of the 64 reference voltage bus lines one bus line, **L64**, is shared between passing the reference voltage **Vr2** and passing the common electrode signal **Vcom**, and furthermore in place of the connection switching circuit **160** of the earlier described embodiments, one switch in the reference voltage selection circuits **131** to **13n** is used. With this configuration, compared with the configuration in which the shorting operation is performed by a connection switching circuit **160** made up of *n* selector switches **161** to **16n**, although one switch having a low on resistance is required as the voltage switching circuit **300**, the need for the connection switching circuit **160** is eliminated, thereby eliminating the need for a large amount of wiring for control signals. For this reason, this embodiment, in addition to achieving the above-noted effects, has the effect of enabling a further reduction in the size of an IC chip used to implement the image signal line drive circuit.

[0157] (6. Sixth Embodiment)

[0158] FIG. 16 shows the configuration of a liquid-crystal display according to a sixth embodiment of the present invention, as a combination of a block diagram and a circuit diagram. In the description below, constituent elements and signals that are the same as the above-noted embodiment are assigned the same reference numerals and will not be explicitly described herein.

[0159] The liquid-crystal display according to this embodiment is an analog-driver type liquid-crystal display, having a display control circuit **10**, a common electrode drive circuit **23**, and a power supply circuit **30** configured the same as in the first embodiment, but differing therefrom with regard to an image signal line drive circuit **25** and a liquid-crystal panel **46**.

[0160] The image signal line drive circuit **25** generates a red image signal **Sr**, which is an analog signal representing the red component of an image to be displayed, a green image signal **Sg**, which is an analog signal representing the green component of an image to be displayed, and a blue image signal **Sb**, which is an analog signal representing the blue component of an image to be displayed. These analog image signals **Sr**, **Sg**, and **Sb** are polarity-reversed every one horizontal synchronization interval to achieve AC drive.

[0161] The liquid-crystal panel **46** is an active-matrix display panel having as switching elements TFTs using polysilicon, and having a pair of mutually opposing substrates (hereinafter referred to as the first substrate and the second substrate). These substrates are held fixed with a prescribed distance therebetween (typically several  $\mu\text{m}$ ), with a liquid crystal material forming a liquid-crystal layer so as to fill the space between the substrates. At least one of these substrates is transparent. On the first substrate is disposed a plurality of image signal lines **Ls** (the number of image signal lines below taken to be *n*) and a plurality of scanning signal lines **Lg**, in a lattice arrangement, and a plurality of pixel formation parts, disposed in a matrix arrangement, each corresponding to one of points of intersection between the plurality of image signal lines **Ls** and the plurality of scanning signal lines **Lg**. Each of the pixel formation parts has a TFT, the source terminal of which is connected to an image signal line **Ls** and the gate terminal of which is connected to a scanning signal line **Lg**, a pixel electrode connected to the drain terminal of the TFT, a common electrode, provided in common to the plurality of

pixel formation parts and, formed over the entire surface of the second substrate as an opposing electrode, so that a capacitance  $C_p$  is formed between the common electrode and the pixel electrode, and a liquid-crystal layer provided in common to all the pixel electrodes and sandwiched between the pixel electrodes and the common electrode. In addition to the above, on the first substrate of the liquid-crystal panel 46, there is formed a scanning signal line drive circuit 42, which supplies a scanning signal to the plurality of scanning signal lines  $L_g$ , image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  for the purpose of passing analog image signals  $S_r$ ,  $S_g$ , and  $S_b$  from the image signal line drive circuit 25, a sampling circuit formed by  $n$  analog switches 411 to 41n for sampling the analog image signals  $S_r$ ,  $S_g$ , and  $S_b$  passed by these image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  and supplying them to the plurality of image signal lines  $L_s$ , and a connection switching circuit for shorting the image signal line bus lines  $L_r$ ,  $L_g$ , and  $L_b$  to the common electrode at the time of polarity reversal. In the liquid-crystal panel 46, therefore, the plurality of pixel formation part disposed in a matrix arrangement, the image signal lines  $L_s$  and scanning signal lines  $L_g$  formed in a lattice arrangement, and part of the drive circuit are formed as one.

[0162] In a liquid-crystal panel configured as described above, a shift register circuit 41 sequentially sends one pulse from an input terminal to an output terminal during one horizontal synchronization interval, and also generates a shorting control signal  $C_{sh}$  that is at the high level for a prescribed amount of time at the time of polarity reversal, this being a signal at the high level each time the above-noted pulse reaches the output terminal, and the inverted shorting control signal  $C_{shb}$  which is derived by inverting the signal  $C_{sh}$ .

[0163] The  $n$  analog switches 411 to 41n of the sampling circuit are sequentially turned on by a pulse that is transferred by the shift register circuit 41, the result of this on operation being that the analog image signals  $S_r$ ,  $S_g$ , and  $S_b$  on the image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  are supplied to the image signal lines  $L_s$ , and passed to the pixel electrodes via TFTs that have been turned on by the scanning signal line drive circuit 42.

[0164] The connection switching circuit includes three analog switches 43r, 43g, and 43b provided for the respective image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  and inserted between the respective image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  and a signal line that passes the common electrode signal  $V_{com}$ . These analog switches 43r, 43g, and 43b input the above-noted shorting control signal  $C_{sh}$  and inverted shorting control signal  $C_{shb}$  as control signals. By doing this, the image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  are supplied with the common electrode signal  $V_{com}$  only during the time interval in which the shorting control signal  $C_{sh}$  is at the high level. This means that the image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  are shorted to the common electrode only when the shorting control signal  $C_{sh}$  is at the high level (that is, only during a prescribed amount of time when polarity reversal occurs).

[0165] As described above, the image signal line drive circuit 25 supplies the analog image signals  $S_r$ ,  $S_g$ , and  $S_b$  with polarity reversed every one horizontal synchronization interval to the image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  in the liquid-crystal panel 46. During at least the shorting interval,

however, for example by placing the output of a buffer circuit within the image signal line drive circuit 25 in the high-impedance state, the image signal drive circuit 25 is electrically separated from the image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$ . Because the image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  are formed on the first substrate and the common electrode is formed over the entire surface of the second substrate, although not shown in the drawing, a capacitance is formed between the image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  and the common electrode. The image signal line drive circuit 25 of this embodiment therefore is the same as the image signal line drive circuits of the earlier described embodiments in that the image signal line drive circuit drives the capacitive load by supplying the capacitive load with a signal which has its polarity reversed for each fixed period.

[0166] In this embodiment, as described earlier, by a connection switching circuit formed by three analog switches 43r, 43g, and 43b, the image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  are shorted to the common electrode for a prescribed amount of time (the time interval during which the shorting control signal  $C_{sh}$  is at the high level) each time the polarity of the analog image signals  $S_r$ ,  $S_g$ , and  $S_b$  passed by the image signal lines  $L_r$ ,  $L_g$ , and  $L_b$  is reversed, during which time the image signal line drive circuit 25 is electrically separated from the image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$ . Therefore, similar to other previously described embodiments, even if a buffer circuit having a lower drive capacity than in the past is used as the buffer circuit in the image signal line drive circuit 25, it is possible to supply to the image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  signals that are the same as in the past. For this reason, according to this embodiment by using a buffer circuit having a smaller drive capacity than in the past, it is possible to reduce the power consumption of the image signal line drive circuit 25, and further possible to reduce the size of the transistors used to implement the buffer circuits 151 to 15n. As a result, it is possible to achieve a reduction in both the size and the cost of the liquid-crystal display.

[0167] (7. Variations)

[0168] It will be understood that the foregoing descriptions do not restrict the present invention, and that it is possible to make other variations of the present invention within the scope of the present invention. For example, although in the above embodiments the polarity of the signals (voltages) supplied to the image signal lines  $L_s$  and those supplied to the image signal bus lines  $L_r$ ,  $L_g$ , and  $L_b$  relative to the common electrode potential as a reference are reversed every one horizontal synchronization interval, the period for polarity reversal is not restricted to one horizontal synchronization interval, and can for example be made two horizontal synchronization intervals. In this case as well, when the polarity is reversed a shorting operation as described above is performed so as to reduce the required buffer circuit drive capacity, thereby enabling a reduction in the power consumption of the image signal line drive circuit and in the size of the circuit.

[0169] Additionally, whereas in the first to the fifth embodiments the amplifier stopping control signal  $C_{as}$  stops the buffer circuit and bias generator circuit so as to reduce power consumption, it is alternatively possible not to stop these circuits, or to cause the bias generator circuit to operate at all times, and stop only the buffer circuits. In the case of

not stopping the buffer circuits, however, it is preferable to perform output control so that the output of each buffer circuit is in the high-impedance state during the shorting interval.

[0170] It will further be understood that the above embodiments are described in the form of examples of liquid-crystal displays, and that the present invention can be applied as well to other display devices in which a voltage signal that is polarity reversed with a prescribed period is supplied to a capacitive load so as to drive the capacitive load. Also, although the above-described embodiments the potential on the common electrode (common electrode signal Vcom) is AC driven in order to reduce the amplitude of the voltage on the image signal lines Ls, the present invention can also be applied in the case in which the potential on the common electrode is fixed, for example the case of a drive method in which polarity reversal of the voltage applied across the liquid-crystal layer is performed for each one horizontal synchronization interval and for each image signal line while performed for each frame, this being dot reversal drive.

[0171] While the invention has been described in detail, the foregoing description is in all aspects illustrative and not restrictive. It is understood that numerous other modifications and variations can be devised without departing from the scope of the invention.

[0172] The present application is based on a claim for priority deriving from the Japanese patent application 2002-031593, titled "Image display device and drive circuit and drive method for same" filed on Feb. 8, 2002, the content of which is incorporated herein by reference.

What is claimed is:

1. A display device in which a voltage serving as an image signal representing an image to be displayed is applied to a capacitive load including a capacitance formed by mutually opposing first and second electrodes, and which has a drive circuit that causes the voltage applied to the capacitive load to reverse polarity periodically, the display device comprises:

an image signal line drive circuit for supplying a voltage signal responsive to an image to be displayed to the first electrode relative to the second electrode as a reference, and

a connection switching circuit, when the polarity of the voltage applied to the capacitive load is reversed, for electrically separating the first electrode from the image signal line drive circuit and shorting the first electrode to an electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode.

2. A display device according to claim 1, wherein the electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode is the second electrode.

3. A display device according to claim 1, wherein the drive circuit applies as the image signal to the capacitive load a voltage representing an image to be displayed based on a horizontal scan and a vertical scan, and causes the polarity of the applied voltage to be reversed at the time of switching of a scan line in the horizontal scan.

4. A display device according to claim 3, further comprising:

a plurality of image signal lines serving as the first electrode,

a plurality of scanning signal lines intersecting with the plurality of image signal lines,

a plurality of pixel formation parts each corresponding to one of points of intersection between the plurality of image signal lines and the plurality of scanning signal lines, and disposed in a matrix arrangement, and

a scanning signal line drive circuit for selectively driving the plurality of scanning signal lines,

wherein each pixel formation part comprises

a switching element that is switched on and off by a scanning signal line passing through a corresponding intersection point,

a pixel electrode connected via the switching element to an image signal line passing through the corresponding intersection point, and

a common electrode serving as the second electrode, providing in common to the plurality of pixel formation parts, and disposed so that a prescribed capacitance included in the capacitive load is formed between the common electrode and the pixel electrode,

wherein the scanning signal line drive circuit applies to a selected scanning signal line a voltage that turns the switching element on, and

the connection switching circuit, when the polarity of the voltage applied to the capacitive load is reversed, electrically separates the image signal lines from the image signal line drive circuit and shorting the image signal lines to an electrode providing a voltage level that is equivalent to the voltage supplied to the common electrode.

5. A display device according to claim 4, wherein the connection switching circuit, after the switching element that had been turned on by the scanning line selected before reversal of the polarity of the voltage applied to the capacitive load is placed in the off state, electrically separates the image signal lines from the image signal line drive circuit and shorts the image signal lines to an electrode providing a voltage level that is equivalent to the voltage supplied to the common electrode.

6. A display device according to claim 4, wherein the connection switching circuit, when the polarity of the voltage applied to the capacitive load is reversed, shorts the image signal lines to an electrode providing a voltage level that is equivalent to the voltage supplied to the common electrode for a period of time that is three or more times the delay time constant which is the product of the wiring resistance and wiring capacitance in one image signal line.

7. A display device according to claim 4, wherein the image signal line drive circuit includes a stopping control circuit for stopping at least part of the image signal line drive circuit for at least the period of time during which the connection switching circuit is shorting the image signal lines to an electrode providing a voltage level equivalent to the voltage supplied by the common circuit.

8. A display device according to claim 4, wherein the drive circuit includes a common electrode drive circuit for

switching the potential on the common electrode in response to the polarity reversal of the voltage applied to the capacitive load,

wherein the common electrode drive circuit switches the potential on the common electrode within the period of time during which the connection switching circuit is shorting the image signal lines to an electrode providing a voltage level that is equivalent to the voltage supplied to the common electrode.

**9. A display device according to claim 4, wherein**

the image signal line drive circuit includes a reference voltage selection circuit provided correspondingly for each of the image signal lines, for selecting a voltage responsive to the image signal from a plurality of reference voltages and supplying the selected voltage to the corresponding image signal line as the voltage signal, and

wherein each of the reference voltage selection circuits includes the connection switching circuit, and when the polarity of the voltage applied to the capacitive load is reversed, selects a voltage level equivalent to the common electrode signal that is the voltage supplied to common electrode, instead of a reference voltage from the plurality of reference voltages, and supplies the selected voltage level to a corresponding image signal line, thereby shorting each of the image signal lines to an electrode providing a voltage level equivalent to the voltage supplied to the common electrode.

**10. A display device according to claim 9, wherein the image signal line drive circuit further comprises:**

a plurality of reference voltage bus lines to which are given the plurality of reference voltages, respectively; and

a voltage switching circuit, when the polarity of the voltage applied to the capacitive load is reversed, for applying to one reference voltage bus line of the plurality of reference voltage bus lines a voltage level equivalent to the common electrode signal, instead of the reference voltage to be applied to the one reference voltage bus line, and wherein

each of the reference voltage selection circuits, during each horizontal scan period, selects a reference voltage bus line of the plurality of reference voltage bus lines to which is applied a reference voltage responsive to the image signal and connects the selected bus line to a corresponding image signal line and, when the polarity of the voltage applied to the capacitive load is reversed, selects and connects the one reference voltage bus line to the corresponding image signal line.

**11. A display device according to claim 10, wherein the voltage level equivalent to the common electrode signal is the common electrode signal.**

**12. A display device according to claim 4, wherein the electrode providing a voltage level that is equivalent to the voltage supplied to the common electrode is the common electrode signal.**

**13. A display device according to claim 4, wherein the drive circuit further comprises:**

a common electrode drive circuit for switching the potential of the common electrode in response to the reversal of polarity of the voltage applied to the capacitive load,

wherein the image signal line drive circuit and the common electrode drive circuit are formed on either one and the same substrate or one and the same chip.

**14. A drive circuit in a display device of the AC drive type, in which a voltage serving as an image signal representing an image to be displayed is applied to a capacitive load including a capacitance formed by mutually opposing first and second electrodes, and in which the polarity of the voltage applied to the capacitive load is periodically reversed, the drive circuit comprising:**

an image signal line drive circuit for supplying a voltage signal responsive to the image to the first electrode relative to the second electrode as a reference, and

a connection switching circuit, when the polarity of the voltage applied to the capacitive load is reversed, for electrically separating the first electrode from the image signal line drive circuit and shorting the first electrode to an electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode.

**15. A drive circuit according to claim 14, wherein the electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode is the second electrode.**

**16. A method for driving by means of a driving circuit in a display device of the AC drive type, in which a voltage serving as an image signal representing an image to be displayed is applied to a capacitive load including a capacitance formed by mutually opposing first and second electrodes, and in which the polarity of the voltage applied to the capacitive load is periodically reversed, the method comprising:**

a step of supplying a voltage signal responsive to the image to the first electrode relative to the second electrode as a reference; and

a step of, when the polarity of the voltage applied to the capacitive load is reversed, electrically separating the first electrode from the part of the drive circuit that supplies the voltage signal to the first electrode, and shorting the first electrode to an electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode.

**17. A method according to claim 16, wherein the electrode providing a voltage level that is equivalent to the voltage supplied to the second electrode is the second electrode.**

专利名称(译)	显示装置，其驱动电路及其驱动方法		
公开(公告)号	<a href="#">US20030151572A1</a>	公开(公告)日	2003-08-14
申请号	US10/357480	申请日	2003-02-04
[标]申请(专利权)人(译)	熊田浩二 OHTA TAKASHIGE KAGAWA HARUHITO		
申请(专利权)人(译)	熊田浩二 OHTA TAKASHIGE KAGAWA HARUHITO		
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发明人	KUMADA, KOUJI OHTA, TAKASHIGE KAGAWA, HARUHITO		
IPC分类号	G02F1/133 G09G3/20 G09G3/36		
CPC分类号	G09G3/3614 G09G3/3688 G09G2330/021 G09G2310/027 G09G2310/0248		
优先权	2002031593 2002-02-08 JP 2002367738 2002-12-19 JP		
其他公开文献	US7098885		
外部链接	<a href="#">Espacenet</a> <a href="#">USPTO</a>		

## 摘要(译)

在有源矩阵型液晶显示器的信号线驱动电路中，该有源矩阵型液晶显示器是具有电容性负载的电压控制型显示器，n个选择器开关 ( 161至16n ) 设置在电压缓冲电路 ( 151至15n ) 之间响应于要显示的图像，从参考电压选择电路 ( 131到13n ) 和连接图像信号线的输出端子 ( T1到Tn ) 输入。这些选择开关 ( 161至16n ) 基于短路控制信号 ( Csh )，当极性反转以执行液晶面板的交流驱动时，该短路控制信号处于高电平，切换输出信号 ( OUT1至OUTn )。在缓冲电路 ( 151至15n ) 的输出信号和公共电极信号 ( Vcom ) 之间的图像信号线驱动电路。通过这样做，每个图像信号线仅在极性反转时的规定时间内，从缓冲电路 ( 151至15n ) 分离并短接到公共电极。这种配置降低了信号线驱动电路的功耗

