



US007898513B2

(12) **United States Patent  
Kong**

(10) **Patent No.: US 7,898,513 B2**  
(45) **Date of Patent: Mar. 1, 2011**

(54) **APPARATUS AND METHOD FOR DRIVING  
LIQUID CRYSTAL DISPLAY DEVICE**

2005/0080975 A1 4/2005 Elledge et al.  
2005/0140626 A1 6/2005 Doyen et al.  
2006/0072664 A1 4/2006 Kwon et al.  
2007/0085794 A1 4/2007 Kawabe et al.

(75) Inventor: **Nam Yong Kong**, Seongnam-si (KR)

**FOREIGN PATENT DOCUMENTS**

(73) Assignee: **LG Display Co., Ltd.**, Seoul (KR)

JP 04-213973 8/1992  
JP 2002-006818 1/2002  
JP 2002-082657 3/2002  
JP 2002-191055 7/2002  
JP 2004-177576 6/2004  
JP 2004-233949 8/2004  
JP 2005-148521 6/2005  
JP 2005-196108 7/2005

(\* ) Notice: Subject to any disclaimer, the term of this patent is extended or adjusted under 35 U.S.C. 154(b) by 0 days.

(21) Appl. No.: **12/874,883**

(22) Filed: **Sep. 2, 2010**

(65) **Prior Publication Data**

US 2011/0018908 A1 Jan. 27, 2011

**Related U.S. Application Data**

(62) Division of application No. 11/476,976, filed on Jun. 27, 2006, now Pat. No. 7,808,464.

(30) **Foreign Application Priority Data**

Dec. 8, 2005 (KR) ..... 10-2005-0119558

(51) **Int. Cl.**  
**G09G 3/36** (2006.01)

(52) **U.S. Cl.** ..... 345/87; 345/89; 345/204

(58) **Field of Classification Search** ..... 345/87-89,  
345/98-100, 690-693, 204

See application file for complete search history.

(56) **References Cited**

**U.S. PATENT DOCUMENTS**

6,509,930 B1 1/2003 Hirano et al.  
6,753,929 B1 6/2004 Sheraizin et al.  
2002/0190940 A1 12/2002 Itoh et al.  
2005/0078069 A1 4/2005 Aiba et al.

**OTHER PUBLICATIONS**

Office Action issued in corresponding Chinese Patent Application No. 2006100903028; issued May 9, 2008.

Office Action issued in corresponding Japanese Patent Application No. 2006-176234; issued Jan. 12, 2010.

English Translation of JP 2004-233949.

*Primary Examiner* — Amare Mengistu

*Assistant Examiner* — Hong Zhou

(74) *Attorney, Agent, or Firm* — Brinks Hofer Gilson & Lione

(57) **ABSTRACT**

An apparatus and method for driving an LCD device is provided. The apparatus for driving an LCD device includes an image display unit that displays an image, and a driving circuit that varies the number of frames of the image displayed in the image display unit in response to motion of the image.

**23 Claims, 21 Drawing Sheets**

110

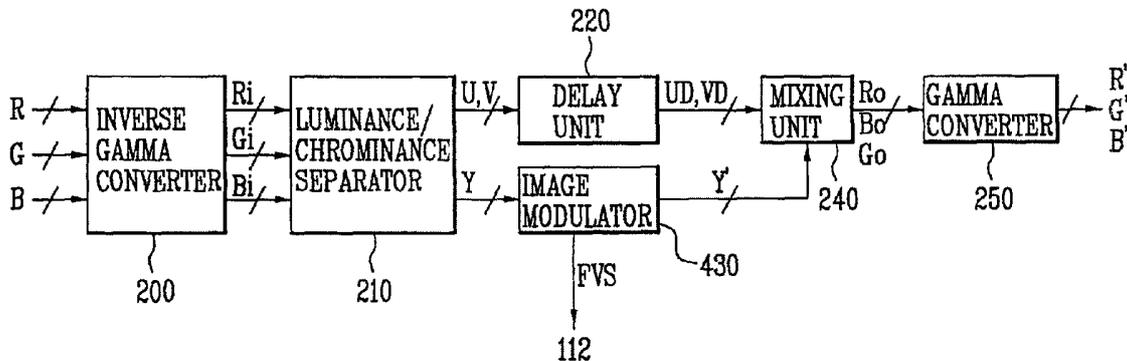
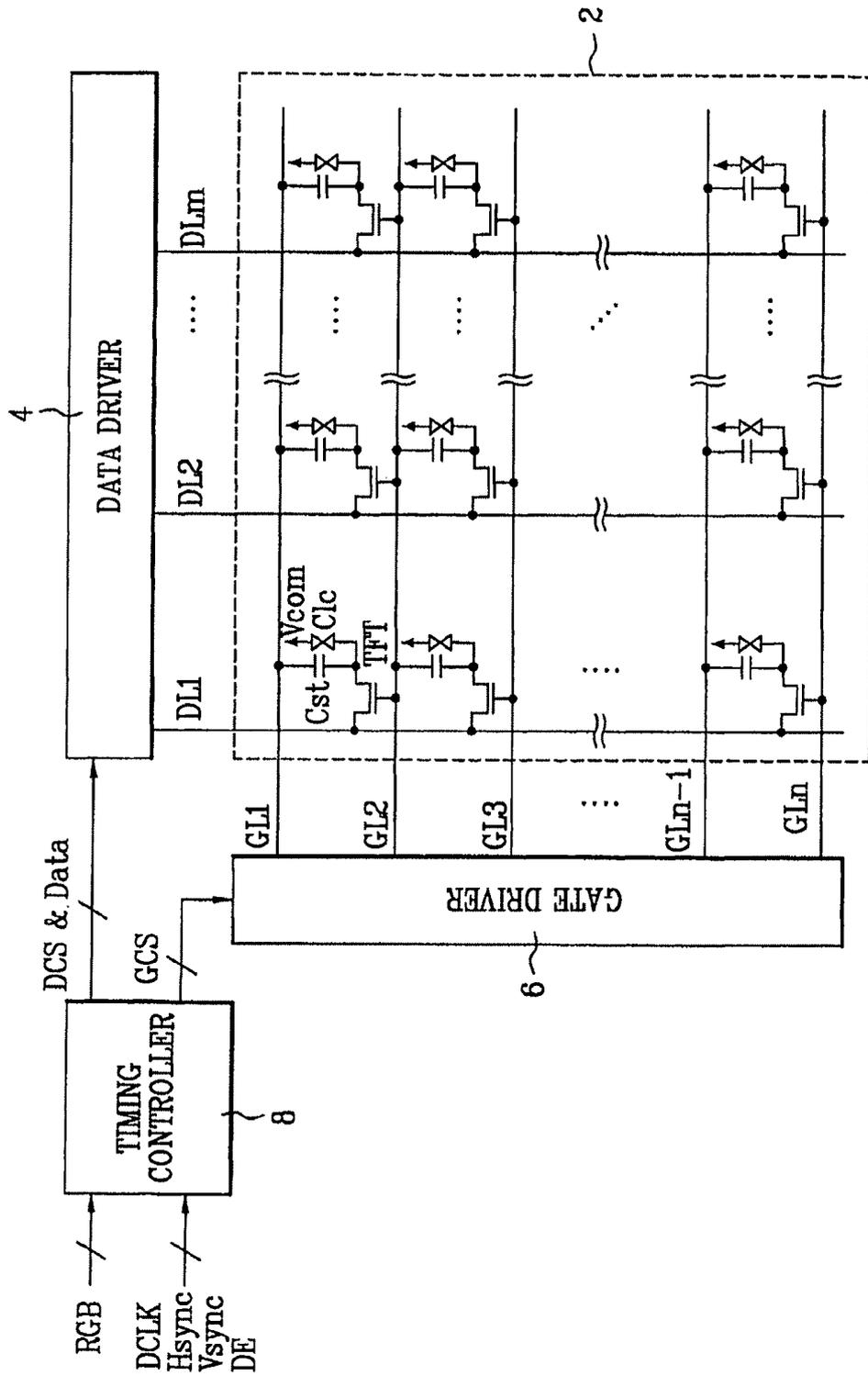
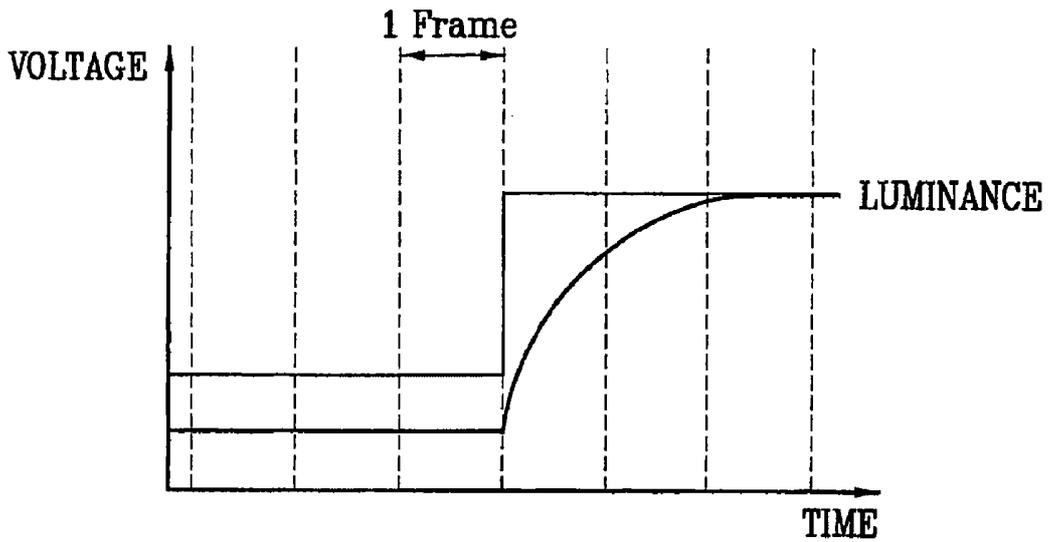


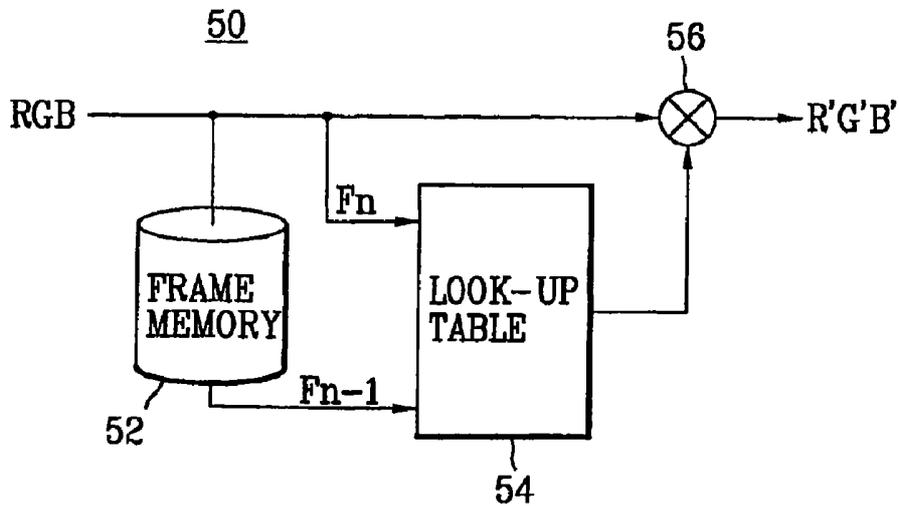
FIG. 1  
RELATED ART



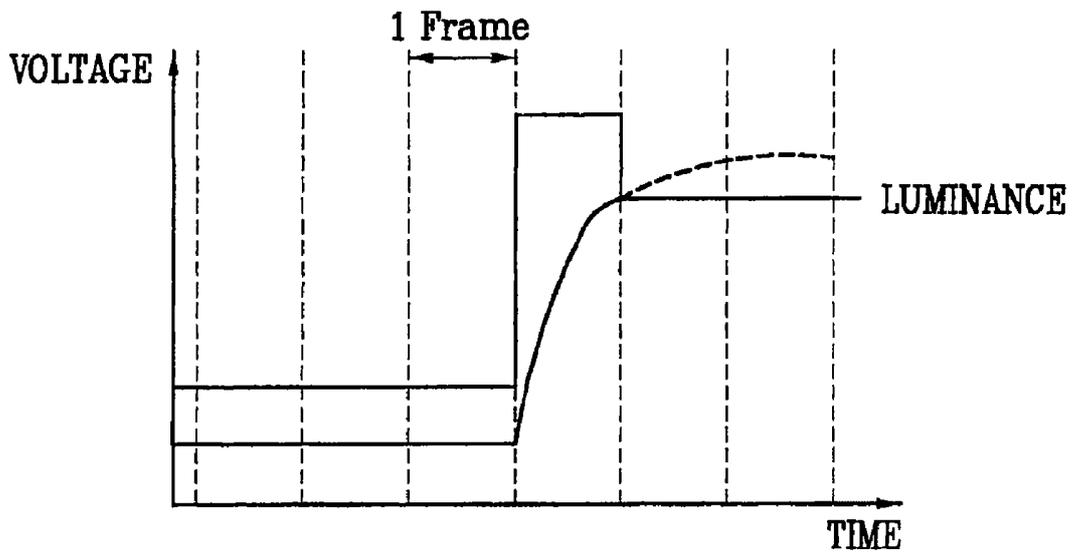
**FIG. 2**  
Related Art



**FIG. 3**  
Related Art



**FIG. 4**  
**Related Art**



**FIG. 5**  
**Related Art**

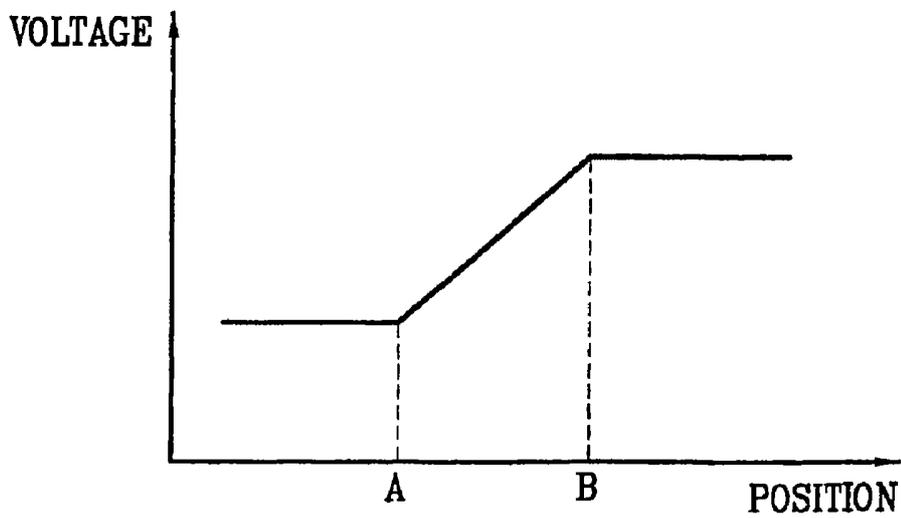


FIG. 6

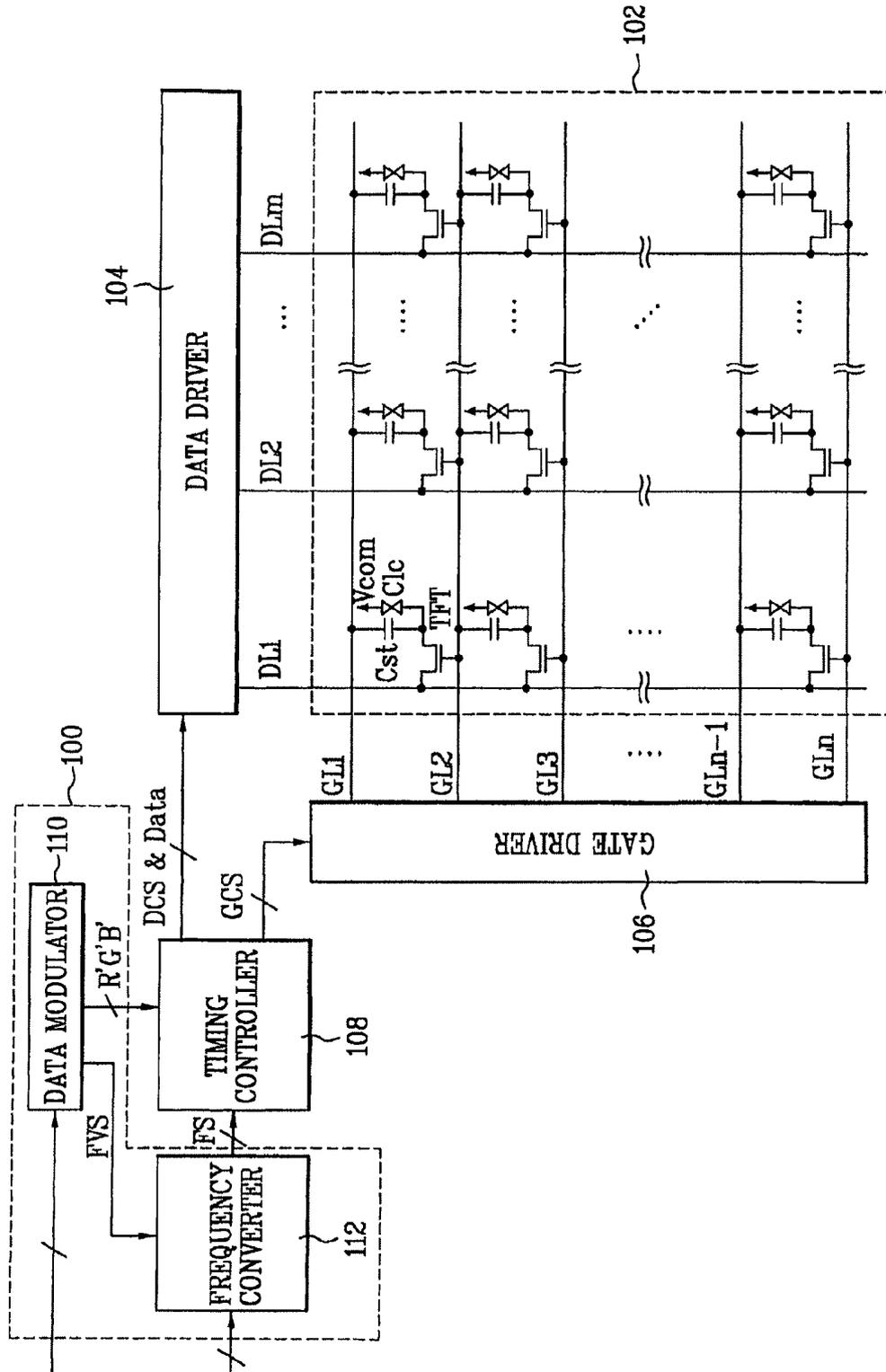


FIG. 7

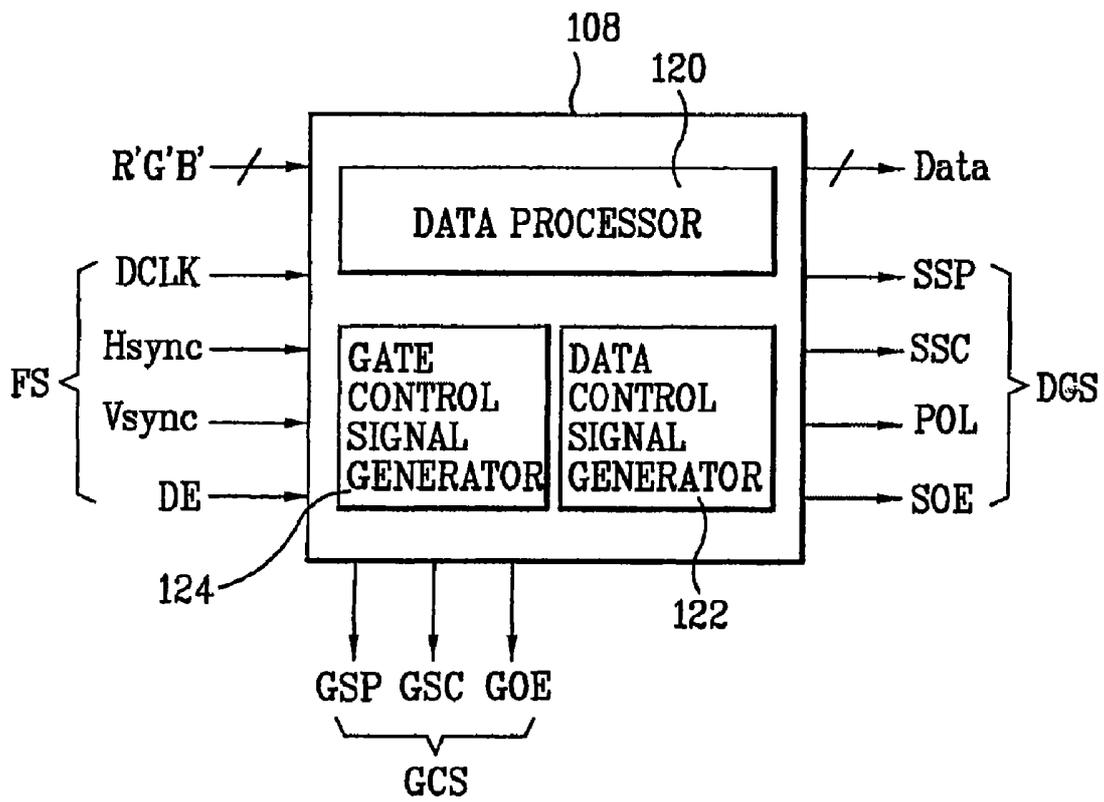


FIG. 8

110

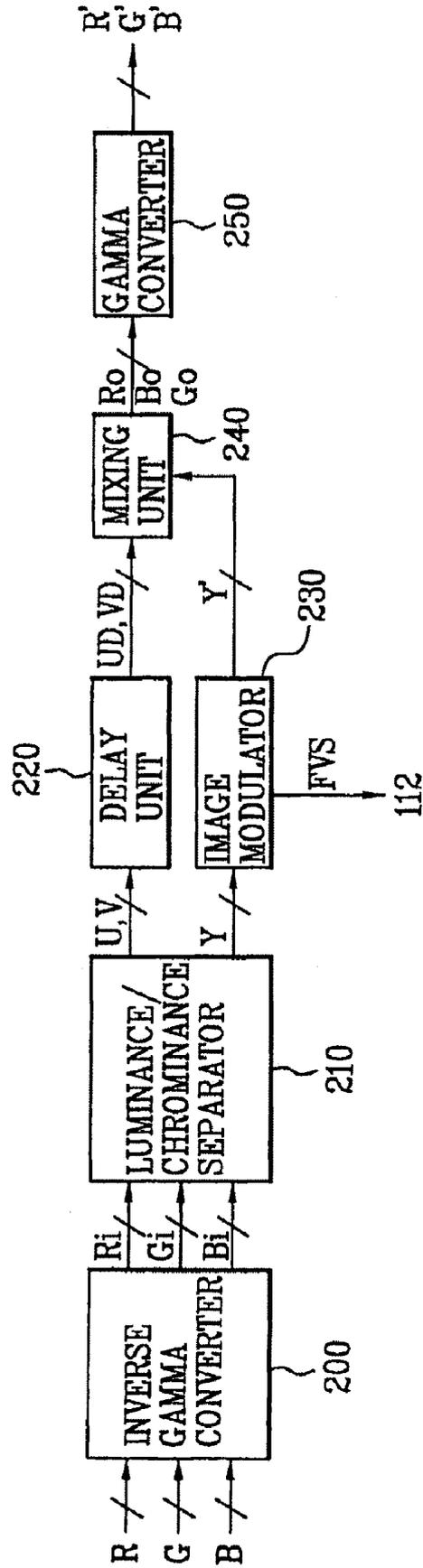


FIG. 9

230 or 630

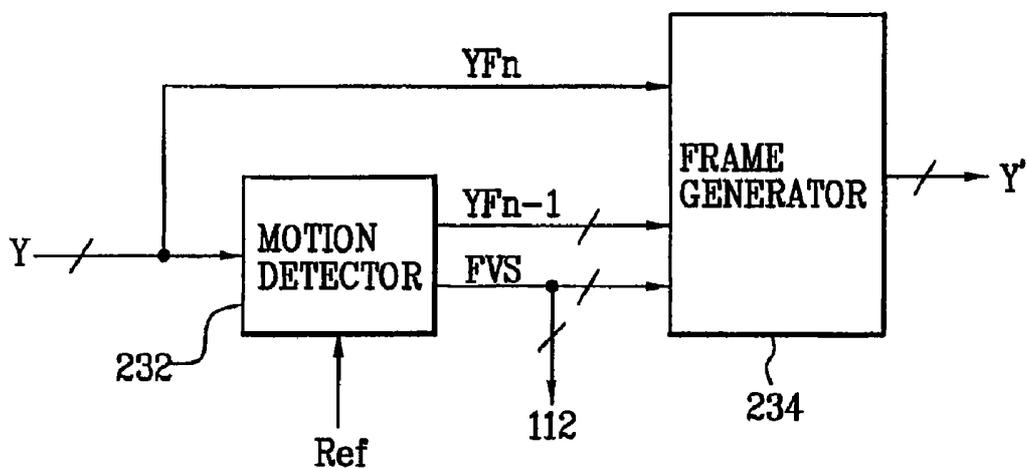


FIG. 10

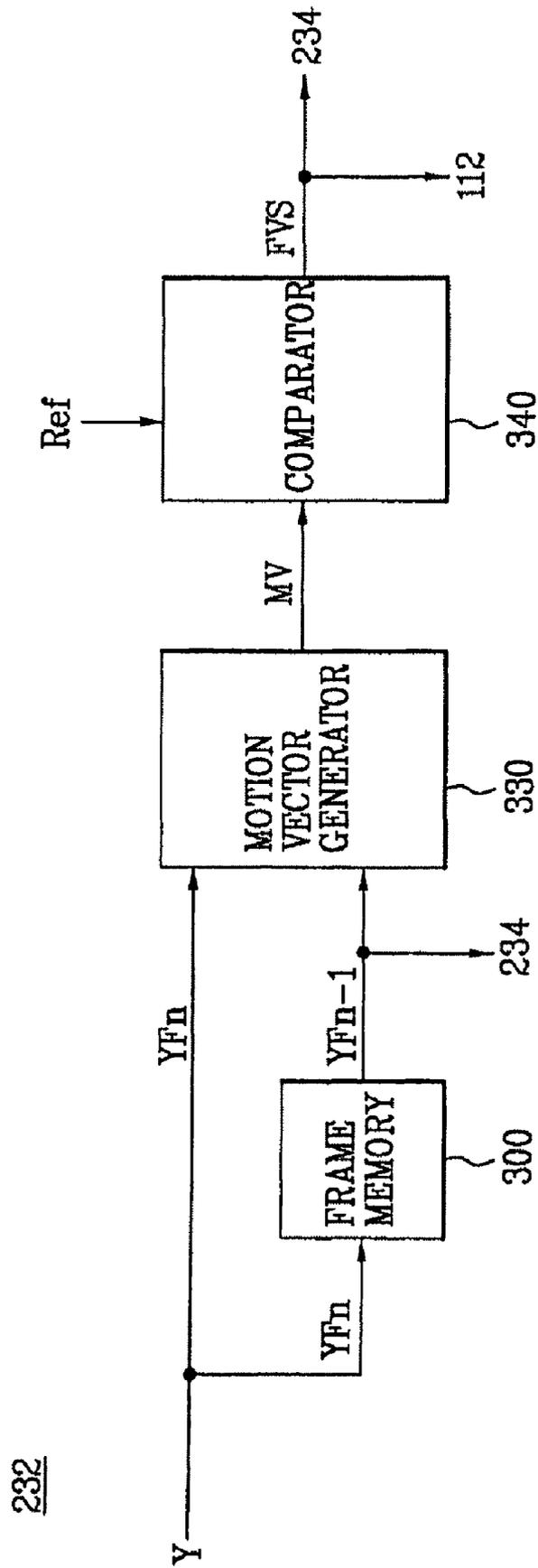


FIG. 11

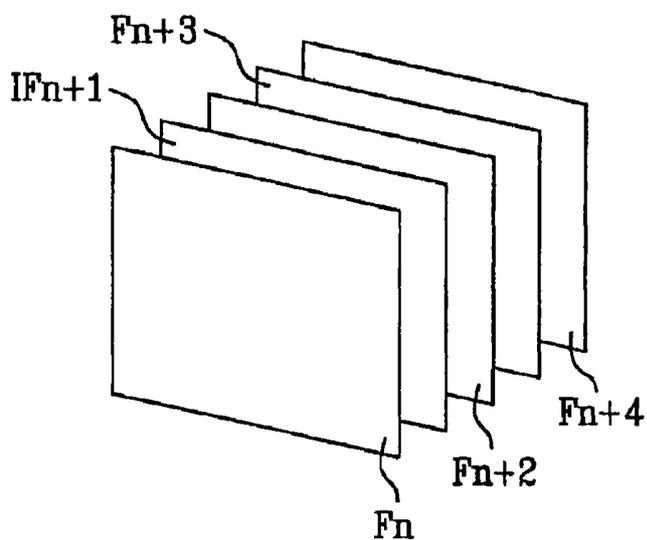


FIG. 12

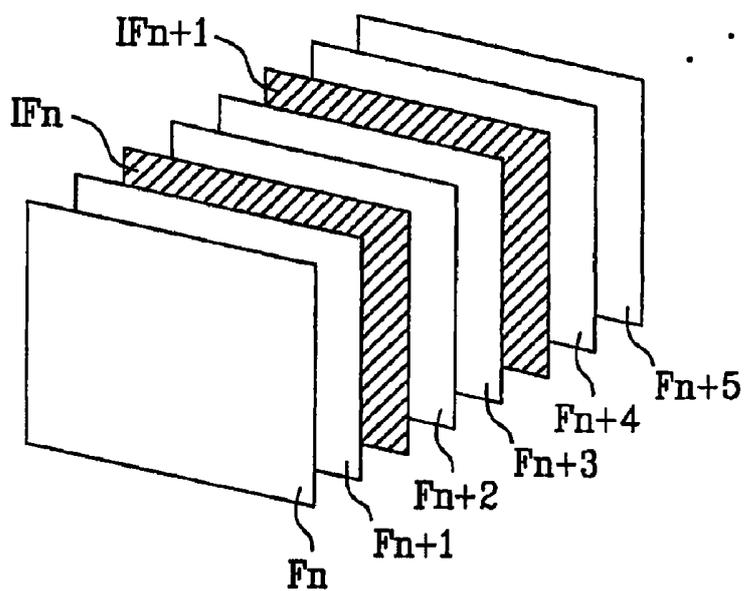


FIG. 13

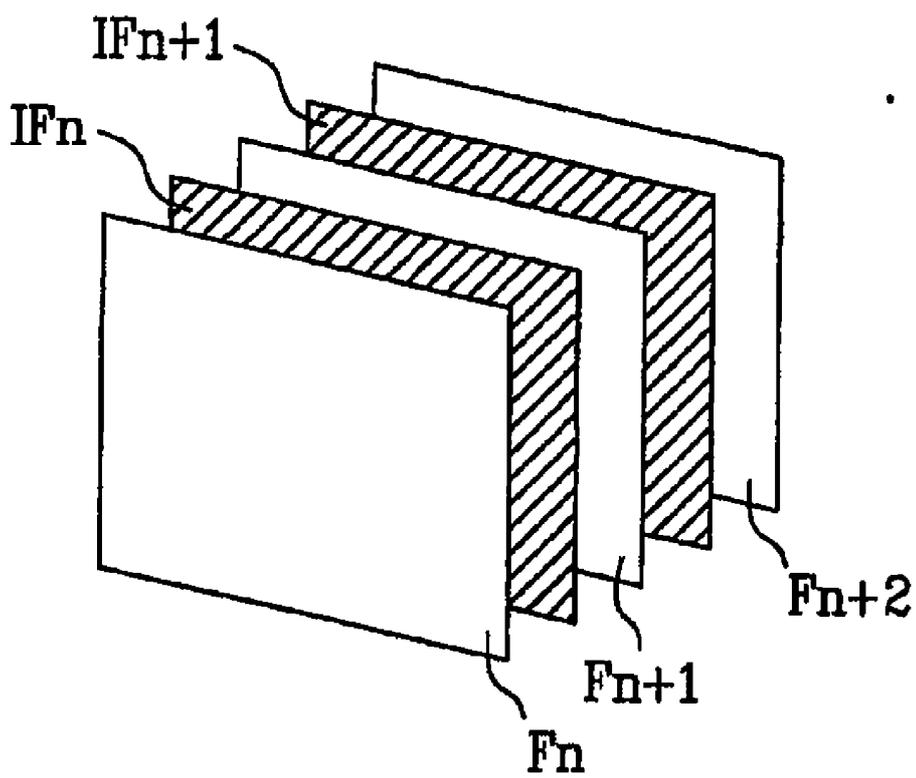


FIG. 14

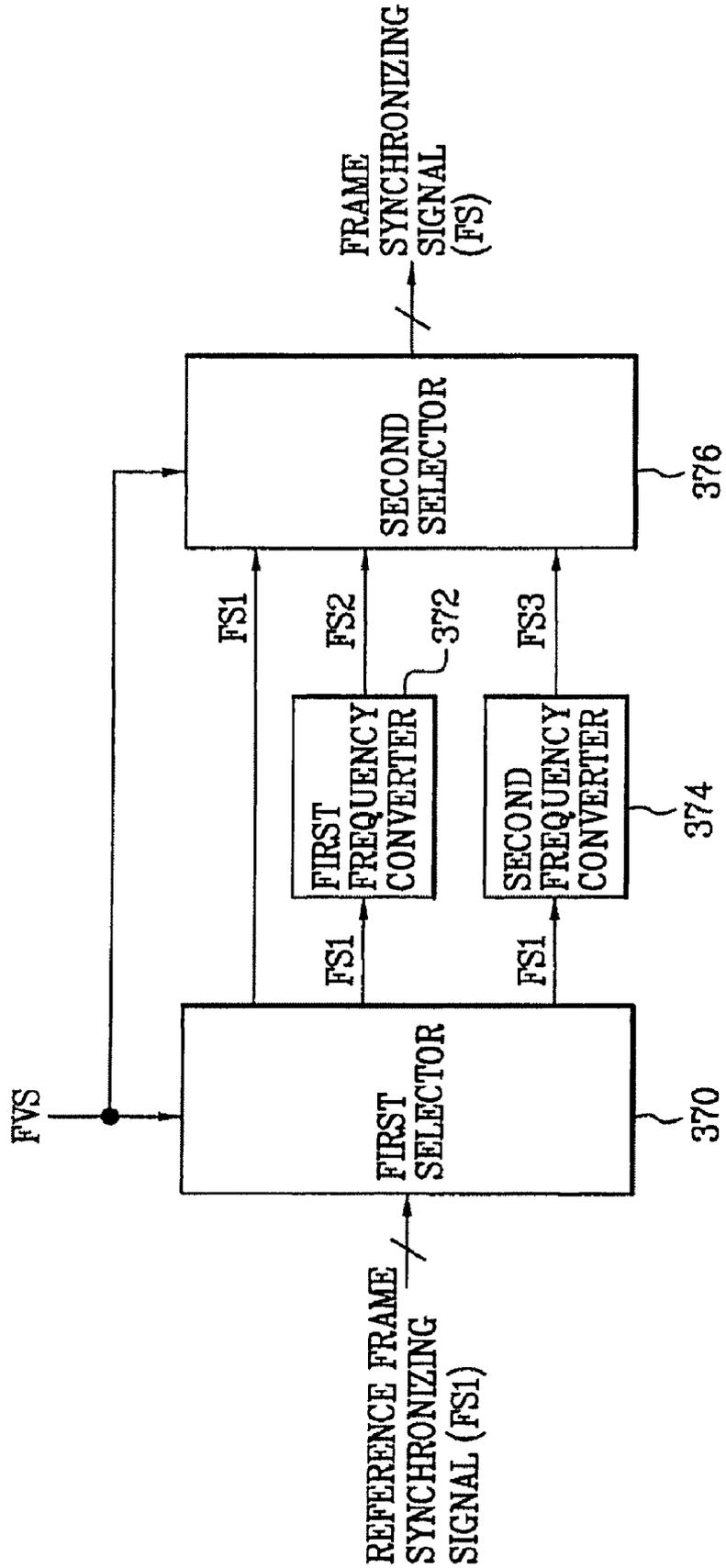
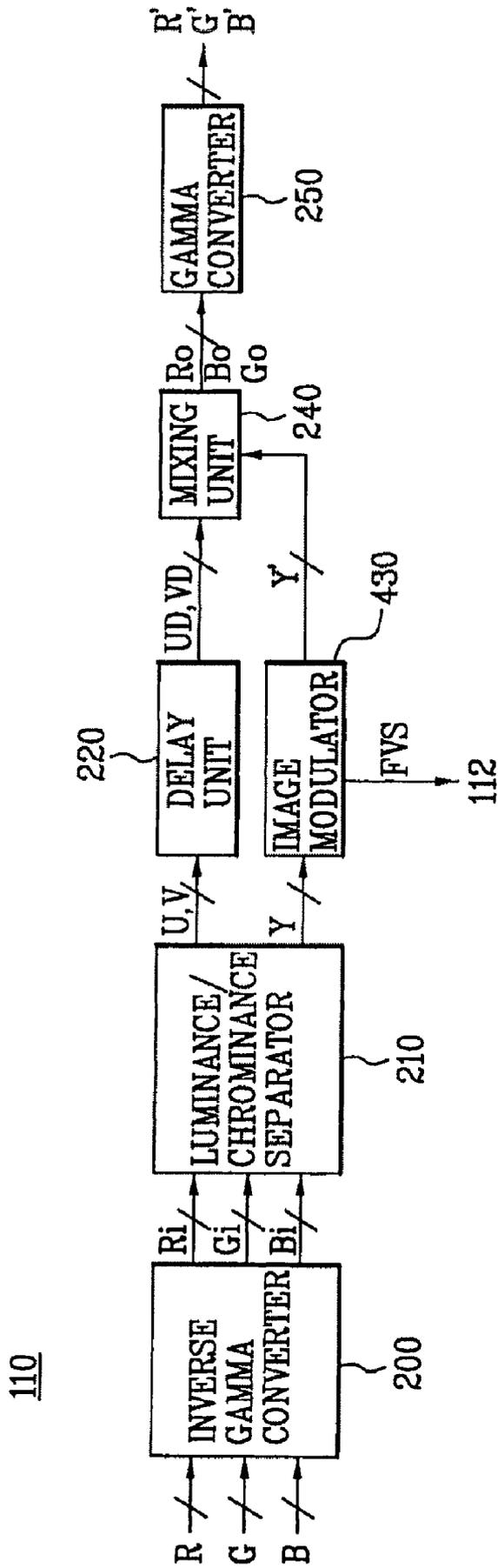


FIG. 15



110

200

210

220

240

250

430

112

R' G' B'

Ri Gi Bi

U V Y

UD VD Y

Ro Bo Go

GAMMA CONVERTER

MIXING UNIT

DELAY UNIT

LUMINANCE/CHROMINANCE SEPARATOR

INVERSE GAMMA CONVERTER

R G B

FIG. 16

430 or 630

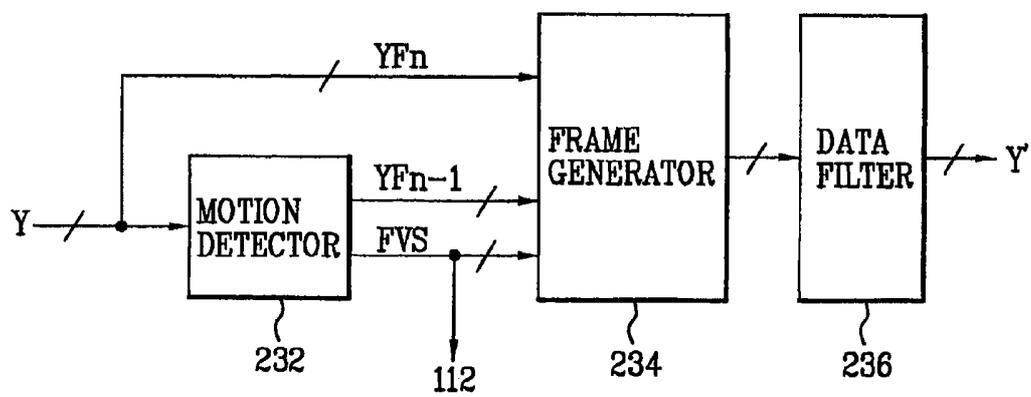


FIG. 17

236

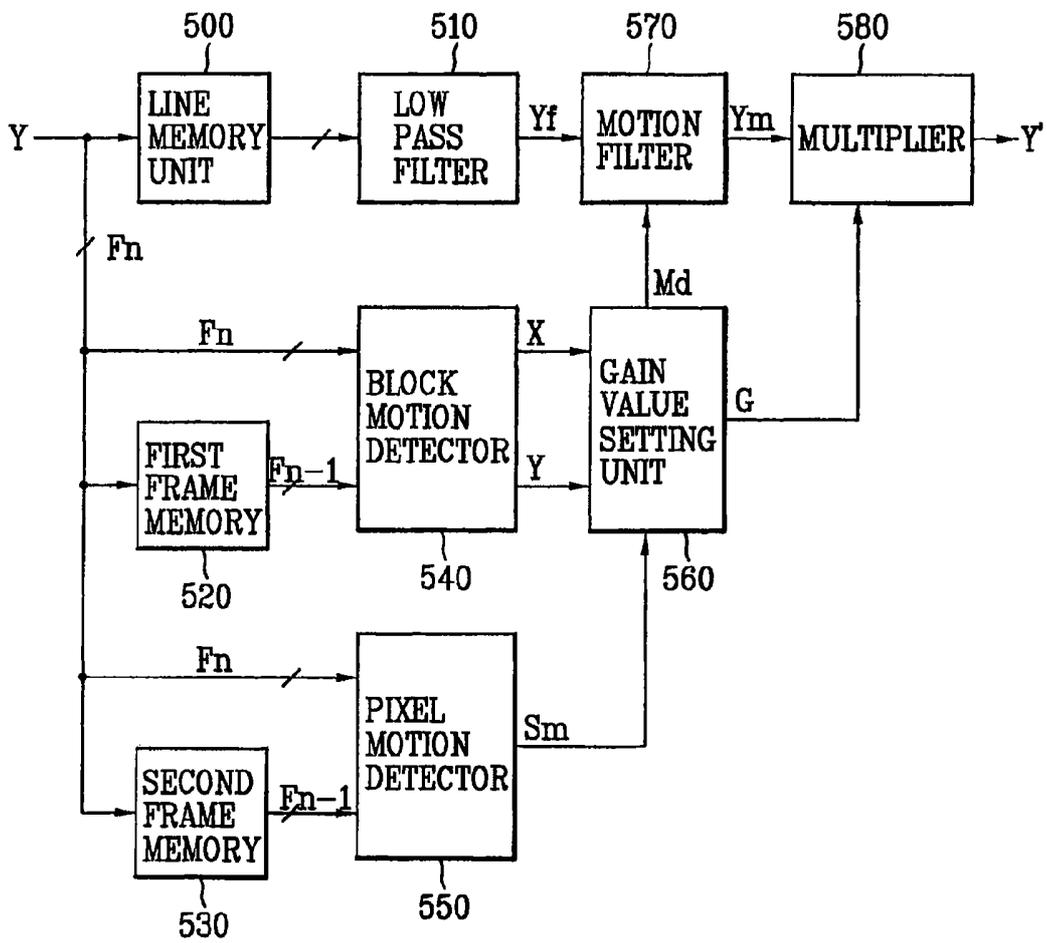


FIG. 18

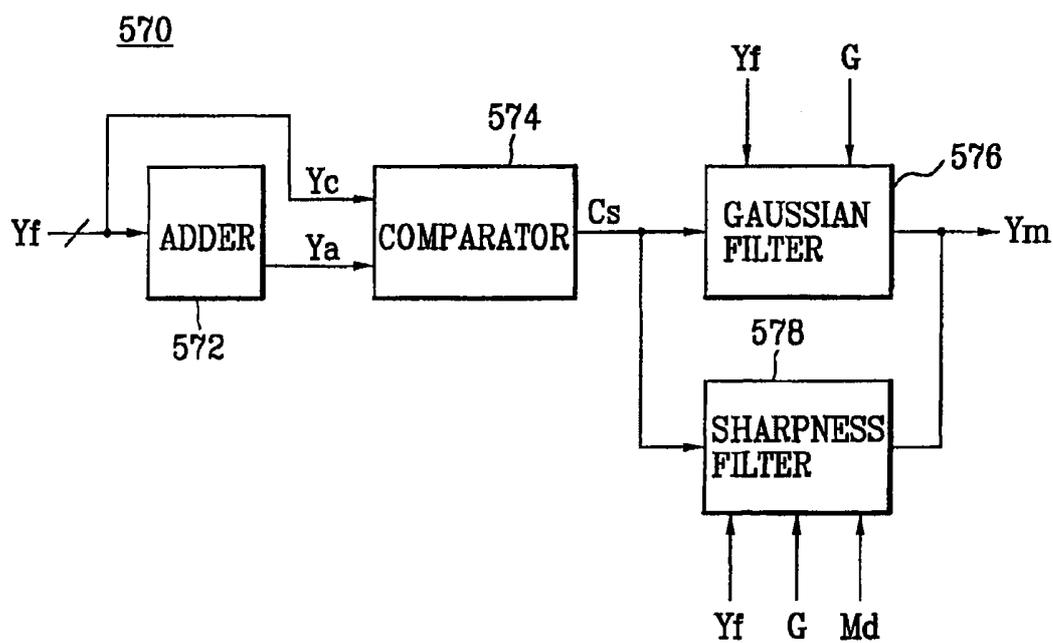


FIG. 19A

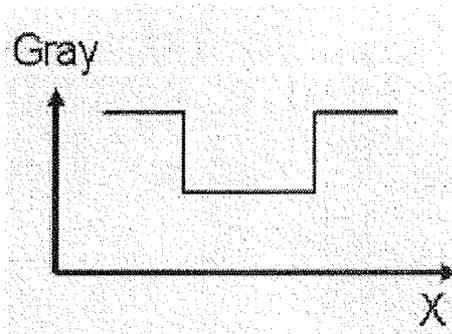
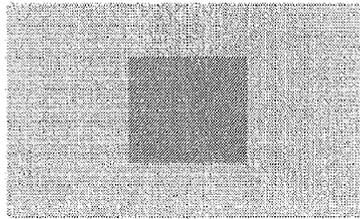


FIG. 19B

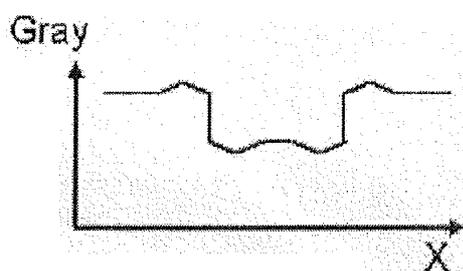
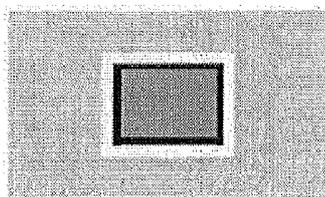


FIG. 19C

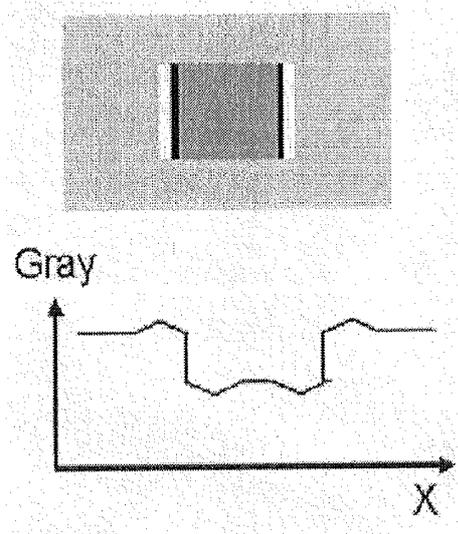


FIG. 19D

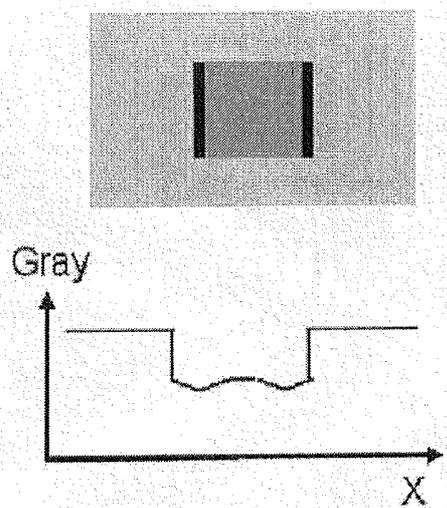


FIG. 20A

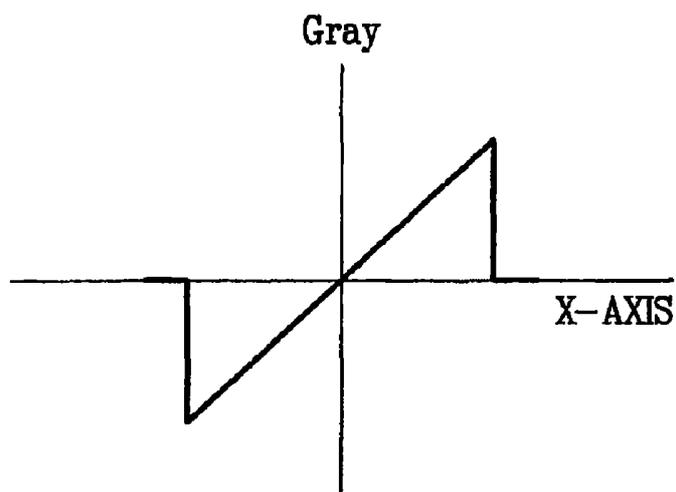


FIG. 20B

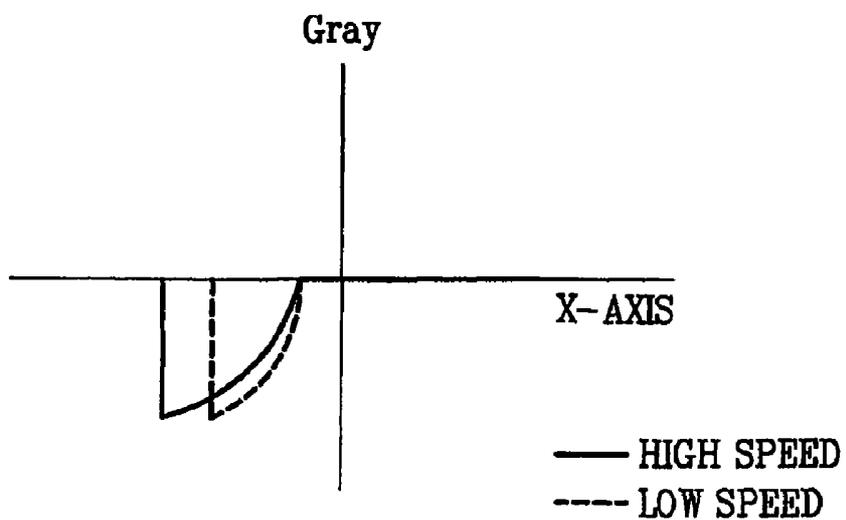
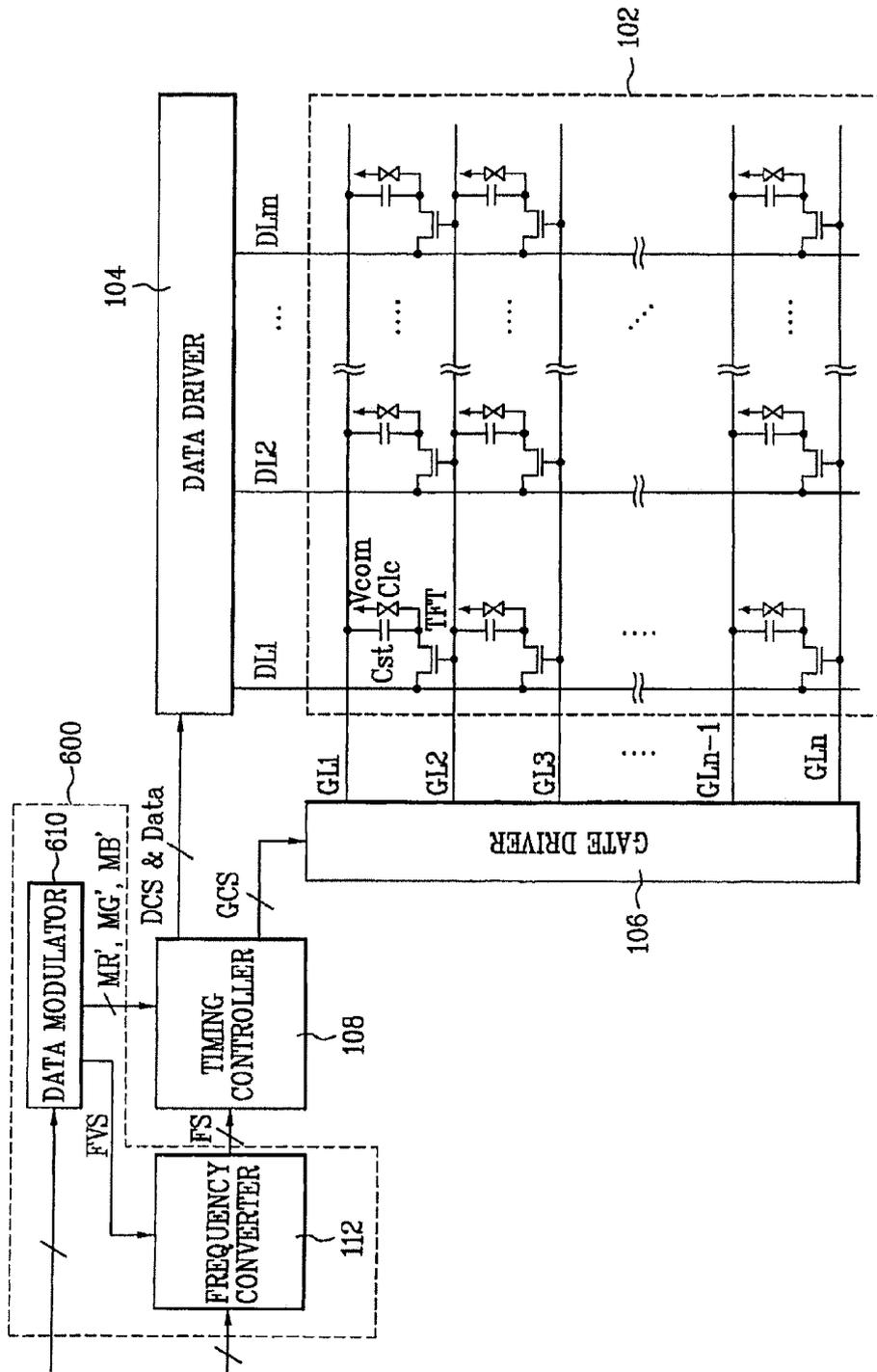


FIG. 21



610

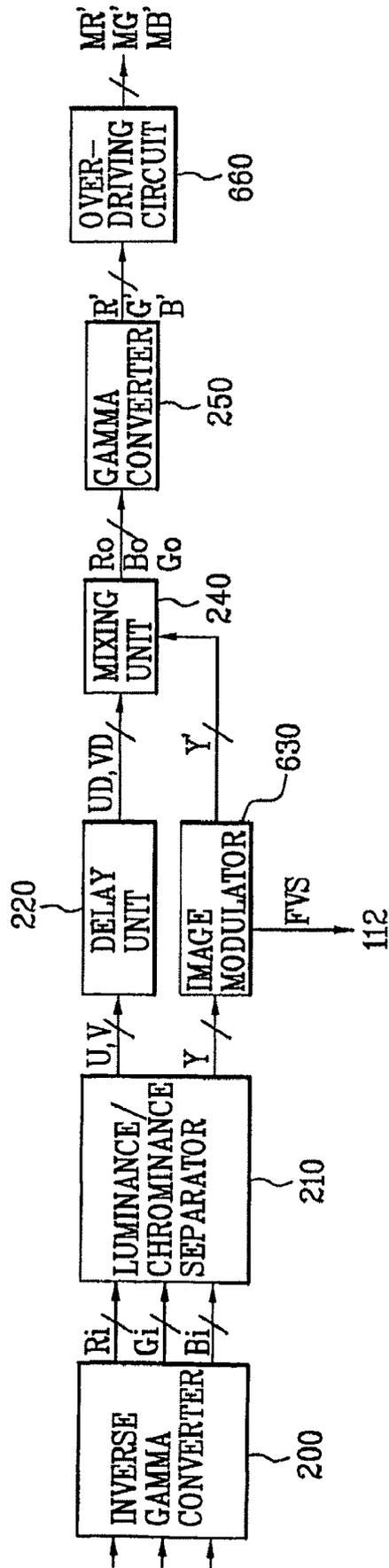
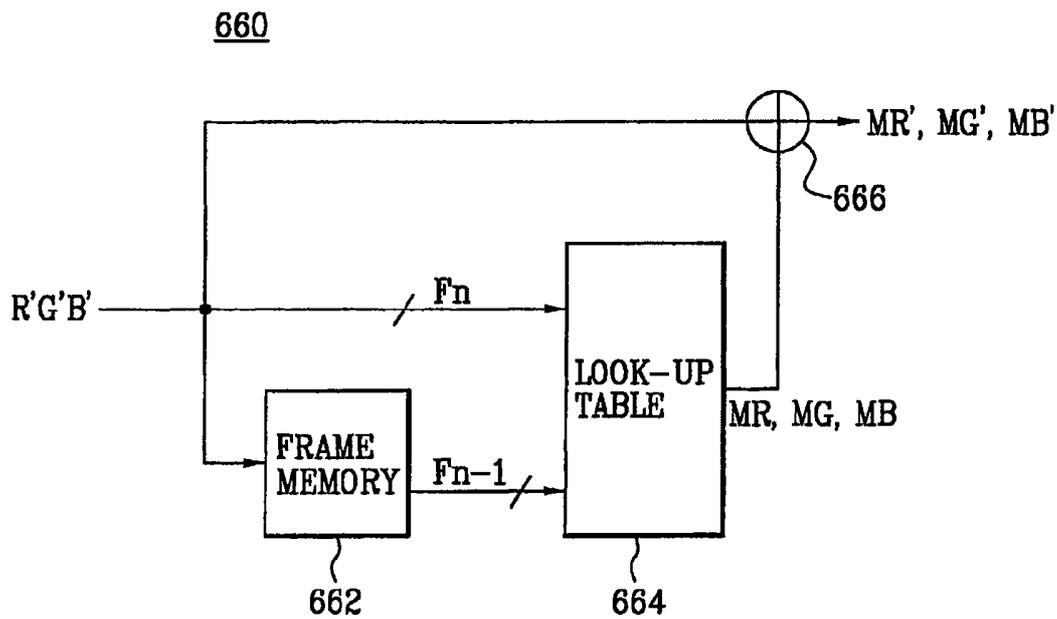


FIG. 22

FIG. 23



## APPARATUS AND METHOD FOR DRIVING LIQUID CRYSTAL DISPLAY DEVICE

The present patent document is a divisional of U.S. patent application Ser. No. 11/476,976, filed Jun. 27, 2006, which claims priority to Korean Patent Application No. P2005-119558 filed in Korea on Dec. 8, 2005, which are hereby incorporated by reference.

### BACKGROUND

#### 1. Field of the Invention

An apparatus and method for driving an LCD device is provided.

#### 2. Discussion of the Related Art

Generally, LCDs adjust light transmittance of liquid crystal cells according to a video signal so as to display an image. An active matrix type LCD device has a switching element formed with every liquid crystal cell and is suitable for the display of a moving image. A thin film transistor (TFT) is mainly used as the switching element of the active matrix type LCD device.

FIG. 1 illustrates a related art apparatus for driving an LCD device.

Referring to FIG. 1, the related art apparatus for driving an LCD device includes an image display unit 2 including liquid crystal cells formed in each region defined by first to n-th gate lines GL1 to GLn and first to m-th data lines DL1 to DLm. A data driver 4 supplies analog video signals to the data lines DL1 to DLm. A gate driver 6 supplies scan signals to the gate lines GL1 to GLn. A timing controller 8 aligns externally input data RGB and supplies them to the data driver 4, generates data control signals DCS to control the data driver 4, and generates gate control signals GCS to control the gate driver 6.

The image display unit 2 includes a transistor array substrate, a color filter array substrate, a spacer, and a liquid crystal. The transistor array substrate and the color filter array substrate face each other and are bonded to each other. The spacer uniformly maintains a cell gap between the two substrates. The liquid crystal is filled in a liquid crystal area prepared by the spacer.

The image display unit 2 includes a TFT formed in the region defined by the gate lines GL1 to GLn and the data lines DL1 to DLm. The liquid crystal cells connect to the TFT. The TFT supplies the analog video signals from the data lines DL1 to DLm to the liquid crystal cells in response to the scan signals from the gate lines GL1 to GLn. The liquid crystal cell is comprised of common electrodes facing each other by interposing the liquid crystal therebetween and pixel electrodes connected to the TFT. Therefore, the liquid crystal cell is equivalent to a liquid crystal capacitor Clc. The liquid crystal cell includes a storage capacitor Cst connected to a previous gate line to maintain the analog video signals filled in the liquid crystal capacitor Clc until the next analog video signals are filled therein.

The timing controller 8 aligns the externally input data RGB to drive the image display unit 2 and supplies the aligned data to the data driver 4. Also, the timing controller 8 generates the data control signals DCS and the gate control signals GCS using a dot clock DCLK, a data enable signal DE, and horizontal and vertical synchronizing signals Hsync and Vsync, which are externally input, so as to control each driving timing of the data driver 4 and the gate driver 6.

The gate driver 6 includes a shift register that sequentially generates scan signals, for example, gate high signals in response to a gate start pulse (GSP) and a gate shift clock

(GSC) are among the gate control signals GCS from the timing controller 8. The gate driver 6 sequentially supplies the gate high signals to the gate lines GL of the image display unit 2 to turn on the TFT connected to the gate lines GL.

The data driver 4 converts the data signals Data aligned from the timing controller 8 into the analog video signals in response to the data control signals DCS supplied from the timing controller 8. The data driver supplies the analog video signals corresponding to one horizontal line per one horizontal period in which the scan signals are supplied into the gate lines GL to the data lines DL. In other words, the data driver 4 selects a gamma voltage having a predetermined level depending on a gray level value of the data signals Data and supplies the selected gamma voltage to the data lines DL1 to DLm. At this time, the data driver 4 inverses polarity of the analog video signals supplied to the data lines DL in response to a polarity control signal POL.

The related art apparatus for driving an LCD device has a relatively slow response speed due to characteristics such as the inherent viscosity and elasticity of the liquid crystal. In other words, although the response speed of the liquid crystal may be different according to the physical properties and cell gap of the liquid crystal, it is common that the rising time is in the range of 20 ms to 80 ms and the falling time is in the range of 20 to 30 ms. Because this response speed is longer than one frame period (16.67 ms in National Television Standards Committee (NTSC)) of a moving image, as shown in FIG. 2, the response of the liquid crystal proceeds to the next frame before the voltage being charged on the liquid crystal cell reaches a desired level.

Since the image of each frame displayed in the image display unit 2 affects the image of the next frame, motion blurring occurs in the moving image due to perception of a viewer.

In the related art apparatus and method for driving an LCD device, motion blurring causes degradation in contrast ratio, and, in turn, degradation in display quality.

In order to prevent motion blurring from occurring, an over-driving apparatus has been suggested that modulates a data signal to obtain the fast response speed of the liquid crystal.

FIG. 3 is a block diagram illustrating a related art over-driving apparatus.

Referring to FIG. 3, the related art over-driving apparatus 50 includes a frame memory 52 that stores RGB data of a current frame Fn. A look-up table 54 generates modulated data that obtains the fast response speed of the liquid crystal by comparing the data RGB of the current frame Fn with data of a previous frame Fn-1 stored in the frame memory 52. A mixing unit 56 mixes the modulated data from the look-up table 54 with the data RGB of the current frame. Fn.

The look-up table 54 lists modulated data R'G'B' that converts a voltage of the data RGB of the current frame Fn into a higher voltage to obtain the fast response speed of the liquid crystal, thereby adapting to a gray level value of an image moving at the fast speed.

In the aforementioned related art over-driving apparatus 50, since a voltage higher than an actual data voltage is applied to the liquid crystal using the look-up table 54 as shown in FIG. 4, the fast response speed of the liquid crystal is adapted to a target gray level voltage until a desired gray level value is actually obtained.

The related art over-driving apparatus 50 can reduce motion blurring of a display image by accelerating the response speed of the liquid crystal using the modulated data R'G'B'.

A problem occurs in that the related art because the LCD device fails to obtain a clear image due to motion blurring occurring in boundaries A and B of each image, as shown in FIG. 5, even though the image is displayed using the over-driving apparatus. In other words, since luminance increases between the boundaries A and B of the image to have a tilt, motion blurring still occurs even though the liquid crystal is driven at a high speed.

If the display image is driven in a frame frequency of 120 Hz, the related art LCD device can reduce motion blurring of the display image. However, there may exist various problems relating to the charge and discharge of the image display unit, a thermal problem of a driver, electromagnetic interference (EMI) caused by high frequency, and difficulty in a circuit design.

### BRIEF SUMMARY

An apparatus and method for driving an LCD device is provided.

An apparatus that drives an LCD device includes an image display unit that displays an image. A driving circuit varies the number of frames of the image displayed in the image display unit in response to motion of the image.

The driving circuit includes a data driver that supplies video signals to the image display unit. A gate driver supplies scan signals to the image display unit. A frame varying unit generates modulated data and a frame variable signal that varies the number of frames of the image displayed in the image display unit by detecting a motion vector from externally input source data. A timing controller aligns the modulated data and supplies the aligned data to the data driver, generates data control signals that drive the data driver, and generates gate control signals to drive the gate driver.

A method for driving an LCD device having an image display unit that displays an image. The method includes detecting a motion vector from externally input source data of the image, and varying the number of frames of the image displayed in the image display unit in response to the motion vector.

The act of varying the number of frames of the image includes generating modulated data and a frame variable signal that varies the number of frames of the image displayed in the image display unit in response to the motion vector, generating the modulated data to obtain the number of frames corresponding to the frame variable signal, generating a frame synchronizing signal by varying an externally input reference frame synchronizing signal in response to the frame variable signal to correspond to the number of frames, generating data and gate control signals using the frame synchronizing signal, supplying scan signals to the image display unit using the gate control signals, and converting the modulated data into analog video signals using the data control signals and supplying the analog video signals to the image display unit to synchronize with the scan signals.

It is to be understood that both the foregoing general description and the following detailed description are exemplary and explanatory and are intended to provide further explanation of the invention as claimed.

### BRIEF DESCRIPTION OF THE DRAWINGS

The accompanying drawings, which are included to provide a further understanding of the embodiments and illustrate the embodiment(s). In the drawings:

FIG. 1 illustrates a related art apparatus for driving an LCD device;

FIG. 2 illustrates the response speed and luminance of a liquid crystal cell shown in FIG. 1;

FIG. 3 is a block diagram illustrating a related art over-driving apparatus;

FIG. 4 illustrates the response speed and luminance of a liquid crystal cell in a related art over-driving apparatus shown in FIG. 3;

FIG. 5 illustrates boundaries of an image according to the related art;

FIG. 6 illustrates an apparatus for driving an LCD device according to a first embodiment;

FIG. 7 is a block diagram that illustrates a timing controller shown in FIG. 6;

FIG. 8 is a block diagram that illustrates a data modulator shown in FIG. 6 in accordance with the first embodiment;

FIG. 9 is a block diagram that illustrates an image modulator shown in FIG. 8 in accordance with the first and third embodiments;

FIG. 10 is a block diagram that illustrates a motion detector shown in FIG. 9;

FIG. 11 illustrates the order of modulated data having a frame frequency of 60 Hz generated by a frame generator shown in FIG. 9;

FIG. 12 illustrates the order of modulated data having a frame frequency of 90 Hz generated by a frame generator shown in FIG. 9;

FIG. 13 illustrates the order of modulated data having a frame frequency of 120 Hz generated by a frame generator shown in FIG. 9;

FIG. 14 is a block diagram that illustrates a frequency converter shown in FIG. 6;

FIG. 15 is a block diagram that illustrates a data modulator shown in FIG. 6 in accordance with the second embodiment;

FIG. 16 is a block diagram that illustrates a data modulator shown in FIG. 15 in accordance with the second embodiment;

FIG. 17 is a block diagram that illustrates a data filter shown in FIG. 16;

FIG. 18 is a block diagram that illustrates a motion filter shown in FIG. 17;

FIG. 19A illustrates luminance components of modulated data supplied to a data filter shown in FIG. 17;

FIG. 19B illustrates overshoot and undershoot occurring if luminance components of modulated data are sharply filtered;

FIG. 19C illustrates overshoot and undershoot that occur if only a moving image is sharply filtered from luminance components of modulated data;

FIG. 19D illustrates undershoot that occurs in a boundary between a still image and a moving image if only the moving image is sharply filtered from luminance components of modulated data;

FIG. 20A is a waveform that illustrates luminance components of a boundary between a still image and a moving image in luminance components of modulated data;

FIG. 20B is a waveform that illustrates the size of undershoot occurring in a boundary between a still image and a moving image in accordance with a gain value obtained by motion speed from luminance components of modulated data;

FIG. 21 illustrates an apparatus for driving an LCD device according to the second embodiment; and

FIG. 22 is a block diagram illustrating a data modulator shown in FIG. 21 in accordance with the third embodiment.

FIG. 23 is a block diagram illustrating an over-driving apparatus shown in FIG. 22.

DETAILED DESCRIPTION OF THE DRAWINGS  
AND THE PRESENTLY PREFERRED  
EMBODIMENTS

Reference will now be made in detail to the preferred embodiments, examples are illustrated in the accompanying drawings. Wherever possible, the same reference numbers will be used throughout the drawings to refer to the same or like parts.

FIG. 6 illustrates an apparatus for driving an LCD device according to the first embodiment.

Referring to FIG. 6, the apparatus that drives an LCD device according to the first embodiment includes an image display unit 102 that includes liquid crystal cells formed in each region defined by first to n-th gate lines GL1 to GLn and first to m-th data lines DL1 to DLm. A driving circuit unit detects a motion vector from externally input source data RGB and generates modulated data R'G'B' and a frame variable signal FVS that varies the number of frames displayed in the image display unit 102 in response to the motion vector.

The image display unit 102 includes a transistor array substrate, a color filter array substrate, a spacer, and a liquid crystal. The transistor array substrate and the color filter array substrate face each other and are bonded to each other. The spacer uniformly maintains a cell gap between the two substrates. The liquid crystal is filled in a liquid crystal area prepared by the spacer.

The image display unit 102 includes a TFT formed in the region defined by the gate lines GL1 to GLn and the data lines DL1 to DLm, and the liquid crystal cells connects to the TFT. The TFT supplies the analog video signals from the data lines DL1 to DLm to the liquid crystal cells in response to the scan pulses from the gate lines GL1 to GLn. The liquid crystal cell is comprised of common electrodes that face each other by interposing the liquid crystal therebetween and pixel electrodes connect to the TFT. Therefore, the liquid crystal cell is equivalent to a liquid crystal capacitor Clc. The liquid crystal cell includes a storage capacitor Cst connected to a previous gate line to maintain the analog video signals filled in the liquid crystal capacitor Clc until the next analog video signals are filled therein.

The driving circuit unit includes a data driver 104 that supplies analog video signals to the data lines DL1 to DLm. A gate driver 106 supplies scan signals to the gate lines GL1 to GLn. A frame varying unit 100 detects a motion vector from source data RGB and generates modulated data R'G'B' and a frame variable signal FVS that varies the number of frames of an image displayed in the image display unit 102. A timing controller 108 aligns the modulated data R'G'B' from the frame varying unit 100 and supplies the aligned data to the data driver 104, generates data control signals DCS that drive the data driver 104, and generates gate control signals GCS that drives the gate driver 106.

The frame varying unit 100 includes a data modulator 110 and a frequency converter 112.

The data modulator 110 detects the motion vector from luminance components of the externally input source data RGB, and generates the frame variable signal FVS in response to the detected motion vector. The data converter 110 generates modulated data R'G'B' by modulating the luminance components of the source data RGB to obtain the number of frames corresponding to the frame variable signal FVS, and supplies the generated modulated data R'G'B' to the timing controller 108.

The frequency converter 112 generates a frame synchronizing signal FS by varying an externally input reference frame synchronizing signal FS1 in response to the frame

variable signal FVS from the data modulator 110, and supplies the generated frame synchronizing signal FS to the timing controller 108.

The frame varying unit 100, which includes the data modulator 110 and the frequency converter 112, may be provided inside the timing controller 108.

The timing controller 108, as shown in FIG. 7, includes a data processor 120, a data control signal generator 122, and a gate control signal generator 124.

The data processor 120 aligns the modulated data R'G'B' supplied from the data modulator 110 to a data signal Data that drives the image display unit 102, and supplies the aligned data signal Data to the data driver 104.

The data control signal generator 122 generates the data control signals DCS, which include a source start pulse SSP, a source shift clock SSC, a polarity signal POL, and a source output enable signal SOE, using the frame synchronizing signal FS input from the frequency converter 112. The frame synchronizing signal FS may be a main clock MCLK, a data enable signal DE, and horizontal and vertical synchronizing signals Hsync and Vsync.

The gate control signal generator 124 generates the gate control signals GCS, which include, for example, a gate start pulse GSP, a gate shift clock GSC, and a gate output enable signal GOE, using the frame synchronizing signal FS, and supplies the generated gate control signals GCS to the gate driver 106.

The gate driver 106 includes a shift register that sequentially generates scan signals, for example, gate high signals in response to the gate control signals GCS from the timing controller 108. The gate driver 106 sequentially supplies the gate high signals to the gate lines GL of the image display unit 102 to turn on the TFT connected to the gate lines GL.

The data driver 104 converts the data signal Data aligned from the timing controller 108 into the analog video signals in response to the data control signals DCS supplied from the timing controller 108, and supplies to the data lines DL the analog video signals corresponding to one horizontal line per one horizontal period in which the scan signals are supplied to the gate lines GL. The data driver 104 generates the analog video signals by selecting a gamma voltage having a predetermined level depending on a gray level value of the data signal Data, and supplies the generated analog video signals to the data lines DL1 to DLm. The data driver 104 inverses polarity of the analog video signals supplied to the data lines DL in response to a polarity control signal POL.

According to the first embodiment, it is possible to remove motion blurring of a moving image by detecting the motion vector from the input data RGB, generating the frame variable signal FVS in response to the detected motion vector, and varying the number of frames of the image displayed in the image display unit 102 in response to the generated frame variable signal FVS.

FIG. 8 is a block diagram that illustrates the data modulator 110 shown in FIG. 6 in accordance with the first embodiment.

Referring to FIG. 8 in connection with FIG. 6, the data modulator 110 includes an inverse gamma converter 200, a luminance/chrominance separator 210, a delay unit 220, an image modulator 230, a mixing unit 240, and a gamma converter 250.

The inverse gamma converter 200 converts the externally input source data RGB into first linear data Ri, Gi and Bi using the following equation (1) because the externally input data RGB has undergone gamma correction considering output characteristics of a cathode ray tube.

$$R_i = R^\lambda$$

$$Gi=B^{\lambda}$$

$$Bi=B^{\lambda} \quad (1)$$

The luminance/chrominance separator **210** separates the first data  $R_i$ ,  $G_i$  and  $B_i$  of a frame unit into a luminance component  $Y$  and chrominance components  $U$  and  $V$ . The luminance component  $Y$  and the chrominance components  $U$  and  $V$  are respectively obtained by the following equations (2) to (4).

$$Y=0.229 \times Ri+0.587 \times Gi+0.114 \times Bi \quad (2)$$

$$U=0.493 \times (Bi-Y) \quad (3)$$

$$V=0.887 \times (Ri-Y) \quad (4)$$

The luminance/chrominance separator **210** supplies the luminance component  $Y$  separated from the first data  $R_i$ ,  $G_i$  and  $B_i$  by the equations (2) to (4) to the image modulator **230** and also supplies the chrominance components  $U$  and  $V$  separated from the first data  $R_i$ ,  $G_i$  and  $B_i$  to the delay unit **220**.

The image modulator **230** according to the first embodiment detects the motion vector using the luminance component  $Y$  from the luminance/chrominance separator **210**, and generates the frame variable signal FVS using the detected motion vector. The image modulator **230** generates a luminance component  $Y'$  to obtain the number of frames corresponding to the frame variable signal FVS and supplies the luminance component  $Y'$  to the mixing unit **240**.

The image modulator **230**, as shown in FIG. 9, includes a motion detector **232** and a frame generator **234**.

The frame detector **232**, as shown in FIG. 10, includes a frame memory **300**, a motion vector generator **330**, and a comparator **340**.

The frame memory **300** stores the luminance component  $Y$  supplied from the luminance/chrominance separator **210** for each unit of frame. The luminance component  $Y$  stored in the frame memory **300** for each unit of frame is supplied to the motion vector generator **330** and the frame generator **234**.

The motion vector generator **330** generates a motion vector  $MV$  using a luminance component  $Y_{Fn}$  of a current frame supplied from the luminance/chrominance separator **210** and a luminance component  $Y_{Fn-1}$  of a previous frame supplied from the frame memory **300**.

Specifically, the motion vector generator **330** detects a point equal to average luminance of a block unit of  $i \times i$  by comparing the luminance component of the current frame  $F_n$  with the luminance component of the previous frame  $F_{n-1}$ , so as to generate the motion vector  $MV$  corresponding to motion speed from the distance between a current pixel and a similar pixel.

The comparator **340** generates a frame variable signal FVS having a logic state of a 2-bit signal by comparing the motion vector  $MV$  supplied from the motion vector generator **330** with a plurality of reference values. Supposing that the size of the maximum motion vector  $MV$  for a block unit of  $i \times i$  is 10 in case of the image moving for a unit of 10 pixel/frame, the reference values are sent as a first reference value Ref1 having a value of '2' and a second reference value Ref2 having a value of '5'. The reference values may be reset as other values by a user.

The comparator **340** generates a frame variable signal FVS having a first logic state if the motion vector  $MV$  is smaller than the first reference value Ref1, and generates a frame variable signal FVS of a second logic state if the motion vector  $MV$  is between the first and second reference values Ref1 and Ref2. The comparator **340** generates a frame variable signal FVS of a third logic state if the motion vector  $MV$

is greater than the second reference value Ref2. The frame variable signal FVS includes any one of the first to third logic states, generated by the comparator **340** that are supplied to the frame generator **234** and the frequency converter **112**, respectively.

If the frame variable signal FVS of the first logic state is supplied from the motion detector **232**, the frame generator **234** shown in FIG. 9 bypasses the luminance component  $Y_{Fn}$  of the current frame that is supplied from the luminance/chrominance separator **210** as shown in FIG. 11 and then supplies it to the mixing unit **240**. For example, the luminance component  $Y'$  supplied from the frame generator **234** to the mixing unit **240** in response to the frame variable signal FVS of the first logic state has the frame frequency of 60 Hz.

If the frame variable signal FVS of the second logic state is supplied from the motion detector **232**, the frame generator **234** generates a luminance component of a reference frame by comparing the luminance component  $Y_{Fn}$  of the current frame supplied from the luminance/chrominance separator **210** with the luminance component  $Y_{Fn-1}$  of the previous frame that is supplied from the frame memory **300**, and generates a luminance component of an insertion frame by comparing the luminance component of the reference frame with the luminance component  $Y_{Fn}$  of the current frame. The frame generator **234** generates the reference frame as an intermediate luminance component by comparing the luminance component of the previous frame with the luminance component of the current frame for each unit of block, and generates the insertion frame as the intermediate luminance component by comparing the luminance component of the reference frame with the luminance component of the current frame for each unit of block.

The frame generator **234**, as shown in FIG. 12, supplies the luminance component  $Y'$  of a frame unit to the mixing unit **240** in the order of the previous frame  $F_{n-1}$ , the current frame  $F_n$  and the insertion frame  $I_{Fn}$  in response to the frame variable signal FVS of the second logic state. In other words, the frame generator **234** supplies the luminance component of frame 3 to the mixing unit **240** using the luminance component of frame 2. For example, the luminance component  $Y'$  supplied from the frame generator **234** to the mixing unit **240** in response to the frame variable signal FVS of the second logic state has a frame frequency of 90 Hz.

If the frame variable signal FVS of the third logic state is supplied from the motion detector **232**, the frame generator **234** generates the luminance component of the insertion frame by comparing the luminance component  $Y_{Fn}$  of the current frame supplied from the luminance/chrominance separator **210** with the luminance component  $Y_{Fn-1}$  of the previous frame supplied from the frame memory **300**. The frame generator **234** generates the insertion frame as the intermediate luminance component by comparing the luminance component of the previous frame with the luminance component of the current frame for each unit of block. Such a frame generator **234**, as shown in FIG. 13, supplies the luminance component  $Y$  of the insertion frame to the mixing unit **240** by inserting the luminance component  $Y$  of the insertion frame between the previous frame  $F_{n-1}$  and the current frame  $F_n$ . For example, the luminance component  $Y'$  supplied from the frame generator **234** to the mixing unit **240** in response to the frame variable signal FVS of the third logic state has a frame frequency of 120 Hz.

The delay unit **220** shown in FIG. 8 generates delayed chrominance components  $UD$  and  $VD$  by delaying the chrominance components  $U$  and  $V$  of a frame unit while the image modulator **230** varies the number of frames in response to the frame variable signal FVS. The delay unit **220** supplies

to the mixing unit **240** the delayed chrominance components UD and VD to synchronize with the modulated luminance component Y'.

The mixing unit **240** generates second data Ro, Go and Bo by mixing the modulated luminance component Y' supplied from the image modulator **230** with the chrominance components UD and VD supplied from the delay unit **220**. The second data Ro, Go and Bo are obtained by the following equations (5) to (7).

$$R_o = Y' + 0.000 \times UD + 1.140 \times VD \quad (5)$$

$$G_o = Y' - 0.396 \times UD - 0.581 \times VD \quad (6)$$

$$B_o = Y' + 2.029 \times UD + 0.000 \times VD \quad (7)$$

The gamma converter **250** performs gamma correction for the second data Ro, Go and Bo supplied from the mixing unit **240** using the following equation (8) to generate modulated data R'G'B'.

$$\begin{aligned} R' &= (R_o)^{1/\lambda} \\ G' &= (G_o)^{1/\lambda} \\ B' &= (B_o)^{1/\lambda} \end{aligned} \quad (8)$$

The gamma converter **250** performs gamma correction for the second data Ro, Go and Bo to the modulated data R'G'B' suitable for the driving circuit of the image display unit **102** using a look-up table, and supplies the resultant data to the timing controller **108**.

FIG. **14** is a block diagram that illustrates a frequency converter shown in FIG. **6**.

Referring to FIG. **14** in connection with FIG. **6**, the frequency converter **112** includes a first selector **370**, a first frequency converter **372**, a second frequency converter **374**, and a second selector **376**.

The first selector **370** supplies the externally supplied reference frame synchronizing signal FS1 to any one of the second selector **376**, the first frequency converter **372**, and the second frequency converter **374** in response to the frame variable signal FVS from the data modulator **110**. The first selector **370** may be a demultiplexer DEMUX. The reference frame synchronizing signal FS1 may have a frequency of 60 Hz. The reference frame synchronizing signal FS1 selected by the first selector **370** will be referred to as a first frame synchronizing signal FS1.

In other words, the first selector **370** supplies the first frame synchronizing signal FS1 to the second selector **376** in response to the frame variable signal FVS of the first logic state, and supplies the first frame synchronizing signal FS1 to the first frequency converter **372** in response to the frame variable signal FVS of the second logic state. The first selector **370** supplies the first frame synchronizing signal FS1 to the second frequency converter **374** in response to the frame variable signal FVS of the third logic state.

The first frequency converter **372** converts the first frame synchronizing signal FS1 supplied from the first selector **370** into a second frame synchronizing signal FS2 and supplies the second frame synchronizing signal FS2 to the second selector **376**. The second frame synchronizing signal FS2 may have a frequency of 90 Hz.

The second frequency converter **374** converts the first frame synchronizing signal FS1 supplied from the first selector **370** into a third frame synchronizing signal FS3 and supplies the third frame synchronizing signal FS3 to the second selector **376**. The third frame synchronizing signal FS3 may have a frequency of 120 Hz.

The second selector **376** supplies the first frame synchronizing signal FS1, supplied from the first selector **370**, to the timing controller **108** in response to the frame variable signal FVS of the first logic state to the timing controller **108**. The second selector **376** supplies the second frame synchronizing signal FS2 supplied from the first frequency converter **372** to the timing controller **108** in response to the frame variable signal FVS of the second logic state. The second selector **376** supplies the third frame synchronizing signal FS3 supplied from the second frequency converter **374** to the timing controller **108** in response to the frame variable signal FVS of the third logic state.

FIG. **15** is a block diagram that illustrates the data modulator **110** shown in FIG. **6** in accordance with the second embodiment.

Referring to FIG. **15** in connection with FIG. **6**, the data modulator **110** according to the second embodiment includes an inverse gamma converter **200**, a luminance/chrominance separator **210**, a delay unit **220**, an image modulator **430**, a mixing unit **240**, and a gamma converter **250**.

The data modulator **110** according to the second embodiment has the same structure as that of the data modulator according to the first embodiment except for the image modulator **430**.

The image modulator **430** according to the second embodiment, as shown in FIG. **16**, includes a motion detector **232**, a frame generator **234**, and a data filter **236**.

The image modulator **430** has the same structure as that of the image modulator **230** according to the first embodiment shown in FIGS. **9** and **10** except for the data filter **236**.

The data filter **236**, as shown in FIG. **17**, includes a line memory unit **500**, a low pass filter **510**, first and second frame memories **520** and **530**, a block motion detector **540**, a pixel motion detector **550**, a gain value setting unit **560**, a motion filter **570**, and a multiplier **580**.

The line memory unit **500** stores the luminance component Y of at least three horizontal lines using at least three line memories that store the luminance component Y supplied from the frame generator **234** for each unit of one horizontal line, and supplies the luminance component Y of a block unit of  $i \times j$  ( $i$  is a positive number above 3) to the low pass filter **510**.

The low pass filter **510** low pass filters the luminance component Y of a block unit  $i \times j$  supplied from the line memory unit **500** and supplies the low pass filtered luminance component to the motion filter **570**.

The low pass filter **510** enlarges the dispersion size of Gaussian distribution for the luminance component Y of a block unit of  $i \times j$  using the luminance component Y of a block unit of  $i \times j$ . Accordingly, the low pass filtered luminance component Y becomes a soft image by means of the low pass filter **510**.

Each of the first and second frame memories **520** and **530** stores the luminance component Y supplied from the frame generator **234** for each unit of frame.

The block motion detector **540** detects motion sizes X and Y including X-axis displacement and Y-axis displacement for motion of a block unit of  $i \times j$  by comparing the luminance component Y of the current frame  $F_n$  supplied from the frame generator **234** with the luminance component Y of the previous frame  $F_{n-1}$  supplied from the first frame memory **320**.

The pixel motion detector **550** generates a motion signal  $S_m$  of a pixel unit by comparing the luminance component Y of the current frame  $F_n$  supplied from the frame generator **234** with the luminance component Y of the previous frame  $F_{n-1}$  supplied from the first frame memory **320** for each unit of pixel, and supplies the generated motion signal  $S_m$  to the gain value setting unit **560**. The motion signal  $S_m$  becomes the first

logic state (high) if motion exists between the current frame Fn and the previous frame Fn-1. The motion signal Sm becomes the second logic state (low) if not so.

The gain value setting unit 560 sets a gain value G for setting motion speed using the motion sizes X and Y from the block motion detector 540 and the motion signal Sm from the pixel motion detector 550. The gain value setting unit 560 sets motion direction Md using the motion sizes X and Y from the block motion detector 540.

If the motion signal Sm is in the first logic state, the gain value setting unit 560 sets the gain value G in response to the motion sizes X and Y as expressed by the following equation (9) and supplies the set gain value to the multiplier 580. In this case, since the gain value G is determined by X-axis displacement and Y-axis displacement of motion, motion speed increases if the gain value increases.

$$G = \sqrt{X^2 + Y^2} \quad (9)$$

The gain value setting unit 560 detects motion direction Md of a block unit of *ixi* in response to X-axis displacement and Y-axis displacement of motion if the motion signal Sm is in the first logic state and supplies the detection motion direction Md to the motion filter 570. The motion direction of a block unit of *ixi* is determined by any one of eight displacements of a moving image displayed by the previous frame Fn-1 and the current frame Fn, for example, the left side to the right side, upper side to the lower side, left upper corner to the right lower corner, and left lower corner to the right upper corner.

If the motion signal Sm is in the second logic state, the gain value setting unit 560 sets the gain value G at "0" and detects the motion direction Md at "0" so as to supply the resultant value to the multiplier 580.

The motion filter 570, as shown in FIG. 18, includes an adder 572, a comparator 574, a Gaussian filter 576, and a sharpness filter 578.

The adder 572 adds a luminance component Yf of a block unit of *ixi* low pass filtered by the low pass filter 510 to a luminance component Yf of a peripheral area excluding a center portion, and supplies the added luminance component Ya to the comparator 574.

The comparator 574 generates a comparing signal Cs by comparing the luminance component Yc of the center portion from the luminance component Yf of a block unit of *ixi* low pass filtered by the low pass filter 510 with the luminance component Ya added from the adder 572, and supplies the generated comparing signal Cs to the Gaussian filter 576 and the sharpness filter 578. The comparing signal Cs becomes the first logic state (high) if the luminance component Yc of the center portion is greater than the luminance component Ya. The comparing signal Cs becomes the second logic state (low) if not so.

If the comparing signal Cs that is supplied from the comparator 574 is in the first logic state, the Gaussian filter 576 filters the luminance component Yf of a block unit of *ixi* low pass filtered by the low pass filter 510 in response to the gain value G supplied from the gain value setting unit 560 to obtain a value of "1" as the sum of the luminance component Yf, and supplies the resultant value to the multiplier 580. The Gaussian filter 576 smoothly filters the luminance component Yf of a block unit of *ixi* so as to minimize overshoot occurring in the luminance component Yf of a block unit of *ixi*.

If the comparing signal Cs supplied from the comparator 574 is in the second logic state, the sharpness filter 576 filters the luminance component Yf of a block unit of *ixi* low pass filtered by the low pass filter 510 in response to the gain value

G supplied from the gain value setting unit 560 and the motion direction Md to obtain a value of "0" as the sum of the luminance component Yf, and supplies the resultant value to the multiplier 580. The sum of the luminance component Ym of a block unit of *ixi* filtered by the sharpness filter 578 has a value of "0" because the luminance component in the center portion has a value greater (+) than the luminance component in the peripheral area while the luminance component in the peripheral area has a value smaller (-) than the luminance component in the center portion. The sharpness filter 578 sharply filters the luminance component Yf of a block unit of *ixi* in response to the gain value G and the motion direction Md so as to generate undershoot in the luminance component Yf of a block unit of *ixi*.

The motion filter 570 filters the luminance component Yf of a block unit of *ixi* low pass filtered by the low pass filter 510 in response to the motion speed of the block motion detector 540 so as to generate undershoot in a boundary between a still image and a moving image and to minimize overshoot.

The multiplier 580 supplies the modulated luminance component Y' to the mixing unit 240 by multiplying the luminance component Ym filtered from the motion filter 570 and the gain value G supplied from the gain value setting unit 560. The size of the undershoot occurs in the boundary between the still image and the moving image is controlled by the gain value G.

If the luminance component Y of the modulated data is sharply filtered, the image of the modulated data shown in FIG. 19A generates undershoot (black portion) and overshoot (white portion) in every boundary between the still image and the moving image as shown in FIG. 19B. Motion blurring occurs in the image of the modulated data due to overshoot occurring in every boundary between the still image and the moving image. In other words, overshoot causes motion blurring using twinkling effect susceptible to the eyes of a human being.

The data filter 236 modulates the luminance component Y to generate clear black lines in the boundaries between the still image and the moving image using only undershoot except for overshoot susceptible to the eyes of the human being. For example, the data filter 236 modulates the luminance component Y of the modulated data, of which the moving image is sharply filtered as shown in FIG. 19C, so as to generate undershoot only in the boundaries between the still image and the moving image as shown in FIG. 19D. As shown in FIG. 20A, the size of undershoot is determined by the motion speed of the moving image as shown in FIG. 20B in the boundaries between the still image and the moving image. In other words, if the moving image is moving at a motion speed above three pixels for each unit of frame, the size of undershoot is increased relatively. If the moving image is moving at a motion speed below three pixels for each unit of frame, the size of undershoot becomes reduced relatively.

According to the second embodiment, the motion of the moving image is detected from the original image in which the number of frames is varied by the frame variable signal FVS, and the luminance component Y is modulated by sharpness filtering in response to the gain value G caused by the detected motion speed and direction Md so as to generate only undershoot in the boundaries between the still image and the moving image. As a result, it is possible to naturally separate the still image from the moving image and obtain a clear moving image, whereby a three-dimensional moving image can be obtained by accommodation effect.

FIG. 21 illustrates an apparatus for driving an LCD device according to the third embodiment.

Referring to FIG. 21, the apparatus that drives an LCD device according to the third embodiment includes an image display unit 102 that includes liquid crystal cells formed in each region defined by first to n-th gate lines GL1 to GLn and first to m-th data lines DL1 to DLm. A data driver 104 supplies analog video signals to the data lines DL1 to DLm. A gate driver 106 supplies scan signals to the gate lines GL1 to GLn. A frame varying unit 600 detects a motion vector from externally input source data RGB, generates first modulated data R'G'B' and a frame variable signal FVS for varying the number of frames of an image displayed in the image display unit 102, in response to the motion vector, and modulates the generated first modulated data R'G'B' to second modulated data MR', MG' and MB' for accelerating the response speed of the liquid crystal. A timing controller 108 aligns the second modulated data MR', MG', and MB' from the frame varying unit 600 to supply the aligned data to the data driver 104, generates data control signals DCS that drive the data driver 104, and generates gate control signals GCS that drive the gate driver 106.

The apparatus for driving an LCD device according to the third embodiment has the same structure as that of the apparatus according to the first embodiment excluding the frame varying unit 600 and the timing controller 108.

The frame varying unit 600, according to the third embodiment, includes a data modulator 610 and a frequency converter 112.

The data modulator 610 detects the motion vector from the luminance component of the externally input source data RGB and generates the frame variable signal FVS in response to the detected motion vector. The data modulator 610 generates the first modulated data R'G'B' by modulating the luminance component of the source data RGB to obtain the number of frames corresponding to the frame variable signal FVS. The data modulator 610 modulates the first modulated data R'G'B' to the second modulated data MR', MG', and MB' to accelerate the response speed of the liquid crystal and supplies the second modulated data to the timing controller 108.

The frequency converter 112 generates a frame synchronizing signal FS by varying an externally input reference frame synchronizing signal FS1 in response to the frame variable signal FVS from the data modulator 610, and supplies the generated frame synchronizing signal FS to the timing controller 108. Since the frequency converter 112 is constructed in the same manner as shown in FIG. 14, its description is the same as the description of FIG. 14.

The frame varying unit 600, which includes the data modulator 610 and the frequency converter 112, may be provided inside the timing controller 108.

The timing controller 108 aligns the second modulated data MR', MG', and MB' supplied from the data modulator 610 to a data signal Data that drives the image display unit 102, and supplies the aligned data signal Data to the data driver 104.

The timing controller 108 drives the data driver 104 by generating the data control signals DCS, which include, for example, a source start pulse SSP, a source shift clock SSC, a polarity signal POL, and a source output enable signal SOE, using the frame synchronizing signal FS input from the frequency converter 112. In this case, the frame synchronizing signal FS may be a main clock MCLK, a data enable signal DE, and horizontal and vertical synchronizing signals Hsync and Vsync.

The timing controller 108 drives the gate driver 106 by generating the gate control signals GCS, which include a gate start pulse GSP, a gate shift clock GSC, and a gate output

enable signal GOE, using the frame synchronizing signal FS input from the frequency converter 112.

The data modulator 610 according to the third embodiment, as shown in FIG. 22, includes an inverse gamma converter 200, a luminance/chrominance separator 210, a delay unit 220, an image modulator 630, a mixing unit 240, a gamma converter 250, and an over-driving circuit 660.

The data modulator 610 according to the third embodiment has the same structure as that of the data modulator according to the first embodiment except for the image modulator 630 and the over-driving circuit 660.

The image modulator 630 according to the third embodiment is comprised of the image modulator 230 according to the first embodiment as shown in FIGS. 9 and 10, or the image modulator 430 according to the second embodiment as shown in FIGS. 16 and 17. Therefore, the description of the image modulator 630 according to the third embodiment is the same as the description of the image modulators 230 and 430 according to the first and second embodiments.

The over-driving circuit 660, as shown in FIG. 23, includes a frame memory 662 that stores the first modulated data R'G'B' supplied from the gamma converter 250, a look-up table 664 that generates over-driving data MR, MG and MB that accelerates the response speed of the liquid crystal by comparing the first modulated data R'G'B' of the current frame Fn supplied from the gamma converter 250 with the first modulated data R'G'B' of the previous frame Fn-1 from the frame memory 662, and a mixing unit 666 that mixes the over-driving data MR, MG and MB from the look-up table 664 with the first modulated data R'G'B' of the current frame Fn and supplies the mixed data to the timing controller 108.

The look-up table 664 lists the over-driving data MR, MG and MB that converts a voltage of the first modulated data R'G'B' of the current frame Fn into a higher voltage to obtain the fast response speed of the liquid crystal, thereby adapting to a gray level value of an image moving at the fast speed.

The mixing unit 666 generates the second modulated data MR', MG' and MB' by mixing the first modulated data R'G'B' of the current frame Fn with the over-driving data MR, and supplies the generated second modulated data MR', MG' and MB' to the timing controller 108.

In the apparatus for driving an LCD device according to the third embodiment, as the number of frames is varied by the frame variable signal FVS, the supplied data are modulated to accelerate the response speed of the liquid crystal, whereby motion blurring of the moving image can be removed.

The embodiments described above have the following, as well as other advantages. The frame variable signal is generated by the motion of the image, and the number of frames of the image displayed in the image display unit is varied by the frame variable signal, so that motion blurring of the moving image can be removed.

The image is modulated by filtering in response to the motion direction and speed of the frame image varied by the frame variable signal so as to generate undershoot only in the boundaries between the still image and the moving image. It is possible to naturally separate the still image from the moving image and obtain a clear moving image, whereby a three-dimensional moving image can be obtained by accommodation effect.

It is possible to remove motion blurring using algorithm without changing panel design and hardware and also to obtain a clearer image and a three-dimensional still image having no noise.

It will be apparent to those skilled in the art that various modifications and variations can be made without departing from the spirit or scope of the invention. Thus, it is intended

that the present invention cover the modifications and variations of this invention provided they come within the scope of the appended claims and their equivalents.

The invention claimed is:

1. An apparatus for driving an LCD device, comprising: 5  
 an image display unit that displays an image;  
 a data driver that supplies video signals to the image display unit;  
 a gate driver that supplies scan signals to the image display unit;  
 a frame varying unit that detects a motion vector from 10  
 externally input source data and that generates modulated data and a frame variable signal that varies the number of frames of the image displayed in the image display unit in response to the motion vector; and 15  
 a timing controller that aligns the modulated data to supply the aligned data to the data driver, generates data control signals that drive the data driver, and generates gate control signals that drive the gate driver,  
 wherein the frame varying unit includes: 20  
 a data modulator that generates the frame variable signal by detecting the motion vector from a luminance component of the source data, generates the modulated data to obtain the number of frames corresponding to the frame variable signal, and supplies the generated frame variable 25  
 signal and the generated modulated data to the timing controller; and  
 a frequency converter that generates a frame synchronizing signal that corresponds to the number of frames by varying an externally input reference frame synchronizing 30  
 signal in response to the frame variable signal, and supplies the generated frame synchronizing signal to the timing controller,  
 wherein the data modulator includes:  
 an inverse gamma converter that performs inverse gamma 35  
 correction for the source data for each unit of frame to generate first data including first red, green and blue (hereafter, R/G/B) signals;  
 a luminance/chrominance separator that separates the first data into a luminance component and chrominance components; 40  
 an image modulator that generates the frame variable signal by detecting the motion vector using a luminance component of a previous frame and a luminance component of a current frame supplied from the luminance/ 45  
 chrominance separator, and generates a modulated luminance component as the modulated data in response to the frame variable signal;  
 a delay unit that delays the chrominance components of at least one frame unit while the image modulator gener- 50  
 ates the modulated luminance component and then that output the delayed chrominance components to synchronize with the modulated luminance component;  
 a mixing unit that mixes the modulated luminance component that is supplied from the image modulator with the 55  
 delayed chrominance components supplied from the delay unit to generate second data including second R/G/B signals; and  
 a gamma converter that performs gamma correction for the second data from the mixing unit to generate the modulated data, 60  
 wherein the image modulator includes:  
 a motion detector that detects the motion vector and then that detects the frame variable signal using the motion vector; 65  
 a frame generator that generates the modulated luminance component in response to the frame variable signal; and

a data filter that determines a still image and a moving image between adjacent frames using the modulated luminance component supplied from the frame generator and filters the modulated luminance component to generate undershoot only in a boundary between the still image and the moving image,

wherein the motion detector includes:

a frame memory that stores the luminance component supplied from the luminance/chrominance separator for each unit of frame;

a motion vector generator that generates the motion vector using the luminance component of the current frame supplied from the luminance/chrominance separator and the luminance component of the previous frame supplied from the frame memory; and

a comparator that generates the frame variable signal by comparing the motion vector with first and second reference values set differently from each other,

wherein frame generator generates the modulated luminance component that includes a insertion frame using the luminance components of the previous and current frames, excluding the chrominance components from the luminance/chrominance separator.

2. The apparatus as in claim 1, wherein the comparator generates a frame variable signal of a first logic state if the motion vector is smaller than the first reference value, generates a frame variable signal of a second logic state if the motion vector is between the first and second reference values, and generates a frame variable signal of a third logic state if the motion vector is greater than the second reference value.

3. The apparatus as in claim 2, wherein the frame generator generates the luminance component of the modulated data to have any one of frame frequencies of 60 Hz, 90 Hz and 120 Hz in response to the frame variable signal.

4. The apparatus as in claim 3, wherein the frame generator generates the luminance component of the modulated data having a frame frequency of 60 Hz by bypassing the luminance component of the current frame in response to the frame variable signal of the first logic state, generates the luminance component of the modulated data having a frame frequency of 90 Hz by using the luminance component of the current frame and the luminance component of the previous frame in response to the frame variable signal of the second logic state, and generates the luminance component of the modulated data having a frame frequency of 120 Hz by using the luminance component of the current frame and the luminance component of the previous frame in response to the frame variable signal of the third logic state.

5. The apparatus as in claim 2, wherein the frequency converter includes:

a first selector that selects the reference frame synchronizing signal as a first-frame synchronizing signal in response to the frame variable signals of the first to third logic states;

a first frequency converter that generates a second frame synchronizing signal by converting a frequency of the first frame synchronizing signal output from the first selector in response to the frame variable signal of the second logic state;

a second frequency converter that generates a third frame synchronizing signal by converting the frequency of the first frame synchronizing signal output from the first selector in response to the frame variable signal of the third logic state; and

a second selector that selects the first to third frame synchronizing signals as the frame synchronizing signal in

17

response to the frame variable signals of the first to third logic states and supplies the selected signals to the timing controller.

6. The apparatus as in claim 5, wherein the reference frame synchronizing signal and the first frame synchronizing signal have a frame frequency of 60 Hz, the second frame synchronizing signal has a frame frequency of 90 Hz, and the third frame synchronizing signal has a frame frequency of 120 Hz.

7. The apparatus as in claim 6, wherein the second selector supplies the first frame synchronizing signal to the timing controller in response to the frame variable signal of the first logic state, supplies the second frame synchronizing signal to the timing controller in response to the frame variable signal of the second logic state, and supplies the third frame synchronizing signal to the timing controller in response to the frame variable signal of the third logic state.

8. The apparatus as in claim 1, wherein the data filter includes:

a line memory unit that stores the luminance component of the modulated data supplied from the frame generator for each unit of at least three horizontal lines;

a low pass filter low pass that filters a luminance component of a block unit of  $i \times i$  (where  $i$  is a positive number above 3) supplied from the line memory unit;

a first and second frame memories that store the luminance component of the modulated data supplied from the frame generator for each unit of frame;

a block motion detector that detects motion size of a block unit of  $i \times i$  by comparing the luminance component of the current frame of the modulated data supplied from the frame generator with the luminance component of the previous frame supplied from the first frame memory;

a pixel motion detector that generates a motion signal of a pixel unit by comparing the luminance component of the current frame with the luminance component of the previous frame supplied from the second frame memory;

a gain value setting unit that sets a gain value and the motion direction that changes strength of undershoot in response to the motion size and the motion signal;

a motion filter that filters the luminance component of the block unit of  $i \times i$  low pass filtered by the low pass filter in response to the gain value and the motion direction from the gain value setting unit to minimize overshoot and generate undershoot; and

a multiplier that multiplies the luminance component filtered by the motion filter and the gain value and supplies the multiplied result to the mixing unit.

9. The apparatus as in claim 8, wherein the motion filter includes:

an adder that adds the luminance component of the block unit of  $i \times i$ , low pass filtered by the low pass filter, to a luminance component of a peripheral area excluding a center portion;

a comparator that generates a comparing signal by comparing the luminance component of the center portion with the luminance component added by the adder;

a first filter that filters the luminance component of the block unit of  $i \times i$  in response to the gain value to obtain a value of "1" as the sum of the luminance component to minimize the overshoot, and supplies the resultant value to the multiplier; and

a second filter that filters the luminance component of the block unit of  $i \times i$  in response to the gain value and the motion direction to obtain a value of "0" as the sum of the luminance component to generate the undershoot, and supplies the resultant value to the multiplier.

18

10. The apparatus as in claim 1, wherein the data modulator further includes an over-driving circuit modulates the data output from the gamma converter to an over-driving data for accelerating response speed of a liquid crystal.

11. The apparatus as in claim 10, wherein the over-driving circuit includes:

a frame memory that stores the data supplied from the gamma converter for each unit of frame; and

a look-up table that generates the over-driving data using the data of the current frame supplied from the gamma converter and the data of the previous frame supplied from the frame memory.

12. The apparatus as in claim 11, wherein the over-driving circuit further includes a mixing unit that mixes the over-driving data from the look-up table with the data of the current frame and supplies the mixed result to the timing controller.

13. A method for driving an LCD device having an image display unit displaying an image, the method comprising:

detecting a motion vector from externally input source data of the image;

generating modulated data and a frame variable signal that varies the number of frames of the image displayed in the image display unit in response to the motion vector;

generating the modulated data to obtain the number of frames corresponding to the frame variable signal;

generating a frame synchronizing signal by varying an externally input reference frame synchronizing signal in response to the frame variable signal to correspond to the number of frames;

generating data and gate control signals using the frame synchronizing signal;

supplying scan signals to the image display unit using the gate control signals; and

converting the modulated data into analog video signals using the data control signals and supplying the analog video signals to the image display unit to synchronize with the scan signals,

wherein the act of generating the modulated data includes:

performing inverse gamma correction for the source data for each frame to generate first data including first R/G/B signals;

separating the first data into a luminance component and chrominance components;

generating the frame variable signal by detecting the motion vector using a luminance component of a previous frame and a luminance component of a current frame separated from the first data;

generating a modulated luminance component as the modulated data in response to the frame variable signal;

determining a still image and a moving image between adjacent frames using the luminance component of the modulated data;

filtering the modulated luminance component to generate undershoot only in a boundary between the still image and the moving image;

delaying the chrominance components of at least one frame unit while generating the modulated luminance component and then outputting the delayed chrominance components to synchronize with the modulated luminance component;

mixing the modulated luminance component with the delayed chrominance components to generate second data including second R/G/B signals; and

performing gamma correction for the second data to generate the modulated data,

wherein the act of generating the frame variable signal includes:

storing the luminance component separated from the first data for each unit of frame using a frame memory;  
 generating the motion vector using the luminance component of the current frame separated from the first data and the luminance component of the previous frame supplied from the frame memory; and  
 generating the frame variable signal by comparing the motion vector with first and second reference values set differently from each other using a comparator;  
 wherein the modulated luminance component includes an insertion frame using the luminance components of the previous and current frames, excluding the chrominance components.

**14.** The method as in claim **13**, wherein the act of generating the frame variable signal using the comparator includes generating a frame variable signal of a first logic state if the motion vector is smaller than the first reference value, generating a frame variable signal of a second logic state if the motion vector is between the first and second reference values, and generating a frame variable signal of a third logic state if the motion vector is greater than the second reference value.

**15.** The method as in claim **14**, wherein the act of generating the luminance component of the modulated data includes generating the luminance component of the modulated data having a frame frequency of 60 Hz by bypassing the luminance component of the current frame in response to the frame variable signal of the first logic state, generating the luminance component of the modulated data having a frame frequency of 90 Hz by using the luminance component of the current frame and the luminance component of the previous frame in response to the frame variable signal of the second logic state, and generating the luminance component of the modulated data having a frame frequency of 120 Hz by using the luminance component of the current frame and the luminance component of the previous frame in response to the frame variable signal of the third logic state.

**16.** The method as in claim **14**, wherein the act of generating the frame synchronizing signal includes:

- selecting the reference frame synchronizing signal as a first frame synchronizing signal in response to the frame variable signals of the first to third logic states;
- generating a second frame synchronizing signal by converting a frequency of the first frame synchronizing signal in response to the frame variable signal of the second logic state;
- generating a third frame synchronizing signal by converting the frequency of the first frame synchronizing signal in response to the frame variable signal of the third logic state; and
- selecting the first to third frame synchronizing signals as the frame synchronizing signal in response to the frame variable signals of the first to third logic states.

**17.** The method as in claim **16**, wherein the reference frame synchronizing signal and the first frame synchronizing signal have a frame frequency of 60 Hz, the second frame synchronizing signal has a frame frequency of 90 Hz, and the third frame synchronizing signal has a frame frequency of 120 Hz.

**18.** The method as in claim **17**, wherein the act of selecting the frame synchronizing signal includes selecting the first frame synchronizing signal in response to the frame variable signal of the first logic state, selecting the second frame synchronizing signal in response to the frame variable signal of the second logic state, and selecting the third frame synchronizing signal in response to the frame variable signal of the third logic state.

**19.** The method as in claim **13**, wherein the act of filtering the modulated luminance component includes:

- storing the modulated luminance component for each unit of at least three horizontal lines;
- low pass filtering the modulated luminance component of a block unit of  $i \times i$  ( $i$  is a positive number above 3);
- storing the modulated luminance component for each unit of frame in first and second frame memories;
- detecting motion size of a block unit of  $i \times i$  by comparing the modulated luminance component of the current frame with the modulated luminance component of the previous frame;
- generating a motion signal of a pixel unit by comparing the modulated luminance component of the current frame with the modulated luminance component of the previous frame;
- setting a gain value and the motion direction for controlling strength of the undershoot in response to the motion size and the motion signal;
- filtering the low pass filtered luminance component of the block unit of  $i \times i$  in response to the gain value and the motion direction to minimize overshoot and generate the undershoot; and
- multiplying the filtered luminance component and the gain value using a multiplier to generate the luminance component of the modulated data.

**20.** The method as in claim **19**, wherein the act of filtering the luminance component to minimize the overshoot and generate the undershoot includes:

- adding the low pass filtered luminance component of the block unit of  $i \times i$  to a luminance component of a peripheral area excluding a center portion;
- generating a comparing signal by comparing the luminance component of the center portion with the added luminance component;
- filtering the luminance component of the block unit of  $i \times i$  in response to the gain value to obtain a value of "1" as the sum of the luminance component so as to minimize the overshoot, and supplying the resultant value to the multiplier; and
- filtering the luminance component of the block unit of  $i \times i$  in response to the gain value and the motion direction to obtain a value of "0" as the sum of the luminance component so as to generate the undershoot, and supplying the resultant value to the multiplier.

**21.** The method as in claim **13**, wherein the act of generating the modulated data further includes modulating the gamma correction data to an over-driving data for accelerating response speed of a liquid crystal.

**22.** The method as in claim **21**, wherein the act of modulating the gamma correction data to the over-driving data includes:

- storing the gamma correction data in the frame memory for each unit of frame; and
- generating the over-driving data using the gamma correction data of the current frame and the gamma correction data of the previous frame supplied from the frame memory.

**23.** The method as in claim **22**, wherein the act of modulating the gamma correction data to the over-driving data further includes mixing the over-driving data with the gamma correction data of the current frame.

专利名称(译)	用于驱动液晶显示装置的装置和方法		
公开(公告)号	<a href="#">US7898513</a>	公开(公告)日	2011-03-01
申请号	US12/874883	申请日	2010-09-02
[标]申请(专利权)人(译)	乐金显示有限公司		
申请(专利权)人(译)	LG DISPLAY CO., LTD.		
当前申请(专利权)人(译)	LG DISPLAY CO., LTD.		
[标]发明人	KONG NAM YONG		
发明人	KONG, NAM YONG		
IPC分类号	G09G3/36		
CPC分类号	G09G3/2092 G09G3/3648 G09G2320/0252 G09G2320/0261 G09G2320/0276 G09G2320/106 G09G2340/0435 G09G2340/16 G09G2360/18		
代理机构(译)	BRINKS霍费尔GILSON & LIONE		
助理审查员(译)	周, 洪		
优先权	1020050119558 2005-12-08 KR		
其他公开文献	US20110018908A1		
外部链接	<a href="#">Espacenet</a> <a href="#">USPTO</a>		

摘要(译)

提供了一种用于驱动LCD装置的设备和方法。用于驱动LCD装置的设备包括：图像显示单元，显示图像；以及驱动电路，响应于图像的运动，改变在图像显示单元中显示的图像的帧数。

