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(54) **Signal-line driving circuit, display device and electronic equipments**

Signalleitungstreiberschaltung, Bildanzeigevorrichtung und elektronisches Gerät

Circuit de commande de ligne à signal, dispositif d'affichage et équipements électroniques

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EP 2 075 788 B1

Description

CROSS REFERENCES TO RELATED APPLICATIONS

[0001] The present invention contains subject matter related to Japanese Patent Application JP 2007-338996 filed in the Japan Patent Office on December 28, 2007.

BACKGROUND OF THE INVENTION

1. Field of the Invention

[0002] In general, the present invention relates to a signal-line driving circuit employed in an active-matrix display device such as a liquid-crystal display device, the display device and electronic equipments each employing the display device.

2. Description of the Related Art

[0003] An image display device such as a liquid-crystal display device employs a large number of pixel circuits arranged to form a matrix as display cells and controls the light intensity of each of the display cells in accordance with information of an image to be displayed in order to display the image.

[0004] In recent years, the development of liquid-crystal display device and the growth of their performance have been becoming remarkable and it becomes possible to apply the liquid-crystal display device to electronic equipments each used for displaying a video signal input thereto or generated thereby as an image or a video in all fields. Examples of the electronic equipments are a TV set, a portable terminal such as a cellular phone or a PDA (Personal Digital Assistant), a digital camera, a notebook personal computer and a video camera.

[0005] Fig. 1 is a block diagram roughly showing the configuration of an ordinary liquid-crystal display device 1.

[0006] As shown in the block diagram of Fig. 1, the liquid-crystal display device 1 employs an available display section 2, a signal-line driving circuit 3 and a gate-line driving circuit 4. The available display section 2 includes a plurality of pixel circuits arranged on a transparent insulation substrate such as a glass substrate to form a matrix. Each of the pixel circuits has a liquid-crystal cell. Also referred to hereafter as a horizontal driving circuit HDRV or a source driver, the signal-line driving circuit 3 is a circuit for driving signal lines. On the other hand, also referred to hereafter as a vertical driving circuit VDRV or a gate driver, the gate-line driving circuit 4 is a circuit for driving vertical scan lines or gate lines.

[0007] As described above, the available display section 2 includes a plurality of pixel circuits arranged to form a matrix and each of the pixel circuits has a liquid-crystal cell not shown in the block diagram of Fig. 1.

[0008] In addition, the available display section 2 also has the signal lines driven by the signal-line driving circuit

3 and the gate lines (or the vertical scan lines) driven by the gate-line driving circuit 4. The signal lines are laid out to serve as lines each corresponding to one of the columns of the matrix whereas the gate lines (or the vertical scan lines) are laid out to serve as lines each corresponding to one of the rows of the matrix.

[0009] In order to prevent molecules of liquid crystal in the available display section 2 employed in the liquid-crystal display device from deteriorating, it is necessary to apply an AC voltage to the liquid crystal in the display device. An ordinary liquid-crystal display device adopts the so-called polarity inverting operation method of a common constant driving technique or a common inversion driving technique in order to apply the AC voltage also referred to as a common voltage to the liquid crystal.

[0010] In accordance with the common constant driving technique, a voltage appearing on a facing electrode exposed to a pixel electrode is fixed at a constant level whereas, relative to the fixed voltage appearing on the facing electrode, a voltage having a positive polarity and a voltage having a negative polarity are applied to the pixel electrode alternately.

[0011] In accordance with the common inversion driving technique, on the other hand, while the voltage appearing on the facing electrode is being inverted from a high level to a low level and vice versa, relative to the varying voltage appearing on the facing electrode, a voltage having a positive polarity and a voltage having a negative polarity are applied to the pixel electrode alternately.

[0012] To put it in detail, when the voltage appearing on the facing electrode is set at a high level, a voltage having a negative polarity relative to the high level taken as a reference is applied to the pixel electrode. When the voltage appearing on the facing electrode is set at a low level, on the other hand, a voltage having a positive polarity relative to the low level taken as a reference is applied to the pixel electrode.

[0013] An output buffer section employed in the signal-line driving circuit 3 is designed for the polarity inverting operations described above.

[0014] In order to carry out the polarity inverting operations, the signal-line driving circuit 3 makes use of a rail-to-rail output analog buffer circuit in the output buffer section as described in CMOS, Circuit Design, Layout and Simulation, P661, Figures 25 and 49, authored by R. Jacob, Baker Harry, W. Li and David Boyce and/or adopts a configuration utilizing an output selector having switches in the output buffer section as disclosed in Japanese Patent Laid-open No. Hei 10-153986.

[0015] US 6,304,241 B1 refers to a liquid crystal display panel including a driver having pairs of first and second D/A converters, corresponding pairs of first and second polarity changeover switches and plural switching elements. Each of the first D/A converters receives a picture signal and outputs a positive-polarity voltage and each of the second D/A converters receives the picture

signal and outputs a negative-polarity voltage. The first polarity changeover switches are connected to the outputs of the first and second D/A converters and alternately output the positive and negative polarity voltages. The second polarity changeover switches are connected to the outputs of the first and second D/A converters and output reverse polarity voltages. The switching elements are connected between the outputs of the first D/A converters and the first polarity switch and the output of the second D/A converters and the second polarity changeover switch. The switching elements are actuated until the voltages at the outputs of the first and second D/A converters become substantially equal.

[0016] Fig. 2 is a block diagram showing a typical configuration of the existing signal-line driving circuit 3 making use of the output selector.

[0017] The signal-line driving circuit 3 employs a line buffer 31, a level shifter 32, a selector section 33, a buffer amplifier section 34 and an output selector 35. The line buffer 31 is a buffer used for storing driving data obtained as a result of a parallel-into-serial conversion process as data to be used for driving signal lines. The level shifter 32 is a section for converting the level of the driving data stored in the line buffer 31 into a level according to a driving level. The selector section 33 is a section including a plurality of DACs (digital-into-analog converters) each used for converting driving data supplied by the level shifter 32 from digital data into analog data in accordance with received gradation voltages. The buffer amplifier section 34 is a section for amplifying driving data received from the selector section 33 in order to generate a signal voltage having the positive polarity and a signal voltage having the negative polarity. The output selector 35 is a section for selectively asserting a signal voltage having the positive polarity and a signal voltage having the negative polarity on every 2 adjacent signal lines respectively.

[0018] Fig. 3 is a block diagram showing a typical configuration of the buffer amplifier section 34 and the output selector 35 which are employed in the signal-line driving circuit 3 shown in the block diagram of Fig. 2.

[0019] To be more specific, Fig. 3 is a block diagram showing an analog output buffer stage included in the signal-line driving circuit 3 as a stage provided for two adjacent channels CH1 and CH2. In actuality, the number of channels of the analog output buffer stage like the one shown in the block diagram of Fig. 3 is at least 100 and signal lines corresponding to these channels are driven by making use of the driving data.

[0020] The buffer amplifier section 34 shown in the block diagram of Fig. 3 employs a first amplifier circuit 34-1 and a second amplifier circuit 34-2 which are provided for the two adjacent channels CH1 to CH2. The first amplifier circuit 34-1 is a circuit for driving a signal line SGL1 connected to the channel CH1 or a signal line SGL2 connected to the channel CH2 by supplying a signal voltage with the positive polarity to the selected signal line SGL1 or SGL2. On the other hand, the second am-

plifier circuit 34-2 is a circuit for driving the signal line SGL1 or the signal line SGL2 by supplying a signal voltage with the negative polarity to the selected signal line SGL1 or SGL2.

[0021] The first amplifier circuit 34-1 employs an OTA (Operational Transconductance Amplifier) 34-11 linked to a DAC of the immediately preceding stage by adoption of a cascade connection technique and an OAMP (output amplifier) 34-12.

[0022] To put it in detail, the inverting input terminal (or the (-) input terminal) of the OTA 34-11 is connected to an output line of the DAC (DAC_UPPER) whereas the non-inverting input terminal (or the (+) input terminal) of the OTA 34-11 is connected to the output line of the OAMP 34-12. The output line of the OTA 34-11 is connected to the input line of the OTA 34-12.

[0023] By the same token, the second amplifier circuit 34-2 employs an OTA (Operational Transconductance Amplifier) 34-21 linked to a DAC of the immediately preceding stage by adoption of a cascade connection technique and an OAMP (output amplifier) 34-22.

[0024] To put it in detail, the inverting input terminal (or the (-) input terminal) of the OTA 34-21 is connected to an output line of the DAC (DAC_LOWER) whereas the non-inverting input terminal (or the (+) input terminal) of the OTA 34-21 is connected to the output line of the OAMP 34-22. The output line of the OTA 34-21 is connected to the input line of the OTA 34-22.

[0025] The output selector 35 employs a first switch group 35-1 and a second switch group 35-2.

[0026] The first switch group 35-1 has a switch SW11 turned on and off in accordance with control according to a signal STR and a switch SW12 turned on and off in accordance with control according to a signal CRS. The switches SW11 and SW12 are controlled to turn on and off in a mutually complementary manner. That is to say, when the switch SW11 is controlled to turn on, the switch SW12 is controlled to turn off and vice versa.

[0027] A contact point a of the switch SW11 is connected to the output line of the OAMP 34-12 employed in the first amplifier circuit 34-1 whereas a contact point b of the switch SW11 is connected to the signal line SGL1 linked to the channel CH1.

[0028] On the other hand, while a contact point a of the switch SW12 is also connected to the output line of the OAMP 34-12 employed in the first amplifier circuit 34-1, a contact point b of the switch SW12 is connected to the signal line SGL2 linked to the channel CH2.

[0029] By the same token, the second switch group 35-2 has a switch SW21 turned on and off in accordance with control according to a signal STR and a switch SW22 turned on and off in accordance with control according to a signal CRS. The switches SW21 and SW22 are turned on and off in a mutually complementary manner. That is to say, when the switch SW21 is controlled to turn on, the switch SW22 is controlled to turn off and vice versa.

[0030] A contact point a of the switch SW21 is con-

nected to the output line of the OAMP 34-22 employed in the second amplifier circuit 34-2 whereas a contact point b of the switch SW21 is connected to the signal line SGL2 linked to the channel CH2.

[0031] On the other hand, while a contact point a of the switch SW22 is also connected to the output line of the OAMP 34-22 employed in the second amplifier circuit 34-2, a contact point b of the switch SW22 is connected to the signal line SGL1 linked to the channel CH1.

[0032] When the switches SW11 and SW21 are controlled to turn on whereas the switches SW12 and SW22 are controlled to turn off, a positive-polarity signal voltage generated by the first amplifier circuit 34-1 is supplied to the signal line SGL1 whereas a negative-polarity signal voltage generated by the second amplifier circuit 34-2 is supplied to the signal line SGL2.

[0033] When the switches SW12 and SW22 are controlled to turn on whereas the switches SW11 and SW21 are controlled to turn off, on the other hand, the positive-polarity signal voltage generated by the first amplifier circuit 34-1 is supplied to the signal line SGL2 whereas the negative-polarity signal voltage generated by the second amplifier circuit 34-2 is supplied to the signal line SGL1.

BRIEF DESCRIPTION OF THE DRAWINGS

[0034] These and other features of the present embodiment will become clear from the following description of the preferred embodiments given with reference to the accompanying diagrams, in which:

Fig. 1 is a block diagram roughly showing the configuration of an ordinary liquid-crystal display device; Fig. 2 is a block diagram showing a typical configuration of the existing signal-line driving circuit making use of an output selector;

Fig. 3 is a block diagram showing a typical configuration of a buffer amplifier section and an output selector which are employed in the signal-line driving circuit shown in the block diagram of Fig. 2;

Fig. 4 is a block diagram showing a typical configuration of a liquid-crystal display device according to an embodiment of the present invention;

Fig. 5 is a circuit diagram showing a typical concrete configuration of an available display section of the liquid-crystal display device;

Fig. 6 is a block diagram showing a typical configuration of a signal-line driving circuit employed in the liquid-crystal display device according to the embodiment;

Fig. 7 is a block diagram showing a typical configuration of an output buffer section employed in the signal-line driving circuit according to the embodiment;

Fig. 8 is a circuit diagram showing a more concrete typical configuration of the output buffer section shown in the block diagram of Fig. 7;

Figs. 9A, 9B, 9C and 9D are timing diagrams showing a plurality of timing charts referred to in description of operations carried out by the output buffer section employed in the signal-line driving circuit according to the embodiment;

Figs. 10A and 10B are plurality of diagrams referred to in comparing a configuration adopting the output selector method explained earlier by referring to the block diagram of Fig. 3 with the configuration of the output buffer section employed in the signal-line driving circuit according to the embodiment;

Figs. 11A and 11B are plurality of diagrams used for showing differences between the appearances of voltages at a bias stage of the configuration adopting the output selector method and the appearances of voltages at a bias stage of the output buffer section employed in the signal-line driving circuit according to the embodiment and referred to in comparing offset-voltage effects caused by amplifiers provided at the output stage;

Fig. 12 is a diagram showing a squint view of the external appearance of a TV set to which the present embodiment is applied;

Figs. 13A and 13B are plurality of diagrams each showing a squint view of the external appearance of a digital camera to which the present embodiment is applied;

Fig. 14 is a diagram showing a squint view of the external appearance of a notebook personal computer to which the present embodiment is applied;

Fig. 15 is a diagram showing a squint view of the external appearance of a video camera to which the present embodiment is applied; and

Figs. 16A, 16B, 16C, 16D, 16E 16F and 16G are plurality of diagrams each showing the external appearance of a portable terminal such as a cellular phone to which the present embodiment is applied.

SUMMARY OF THE INVENTION

[0035] As described above, in order to carry out polarity inverting operations, the signal-line driving circuit employed in the liquid-crystal display device makes use of a rail-to-rail output analog buffer circuit or utilizes an output selector shown in the block diagrams of Figs. 2 and 3.

[0036] However, the rail-to-rail output analog buffer circuit raises the following problems that the circuit configuration is complex, the power consumption is high and the layout area is large.

[0037] In addition, if the output selector is used, the circuit configuration is complex even though the power consumption can be reduced. However, the output selector raises the following problems.

[0038] In order to reduce the turned-on-state resistance, the sizes of the output selector and the output stage increase. As a result, the layout area also becomes large.

[0039] In addition, a settling process worsens due the turned-on-state resistance of the output selector.

[0040] As described earlier, in actuality, the number of channels CH of the analog output buffer stage like the one shown in the block diagram of Fig. 3 is at least 100. In a number of high-definition applications of the channels CH, reduction of the layout area is strongly demanded. In addition, accompanying recent-year conversions into high-definition screens, there is raised a problem of an increasing operating frequency.

[0041] Addressing the problems described above, inventors of the present invention have innovated a signal-line driving circuit capable of preventing the circuit configuration from becoming complicated, the consumed current from increasing and the characteristics of the signal-line driving circuit from deteriorating and capable of reducing the element size (or the layout area) and innovated a display device employing the signal-line driving circuit as well as electronic equipments each employing the display device.

[0042] In order to achieve the need to realize the innovations described above, in accordance with a first embodiment of the present invention, there is provided a signal-line driving circuit configured to supply a signal voltage having the positive or negative polarity to each of signal lines connected to display cells arranged to form a matrix as cells each to be subjected to a polarity inversion driving operation according to inversion of the polarity of said signal voltage. The signal-line driving circuit employs an output buffer section configured to amplify input data, which is used for driving said signal lines, in order to generate a signal voltage having said positive polarity as well as a signal voltage having said negative polarity and selectively supplying said signal voltage having said positive polarity and said signal voltage having said negative polarity to a pair of first and second ones of said signal lines. The output buffer section employs a positive-polarity operating amplifier configured to amplify said input data in order to generate said signal voltage having said positive polarity, a negative-polarity operating amplifier configured to amplify said input data in order to generate said signal voltage having said negative polarity, and a first output buffer configured to supply a signal voltage having said positive polarity or a signal voltage having said negative polarity to said first signal line. The output buffer section further employs a second output buffer configured to supply a signal voltage having said positive polarity or a signal voltage having said negative polarity to said second signal line, and a switch group including switches placed between the output of said positive-polarity operating amplifier and the input of said first output buffer, and between the output of said negative-polarity operating amplifier and the input of said second output buffer. The switch group further includes switch placed at the feedback input stage of said positive-polarity operating amplifier, and at the feedback input stage of said negative-polarity operating amplifier. There are no switches placed between the output of said first output buffer and said first signal line and between the output of said second output buffer and said second signal

nal line.

[0043] The switch group is operated as follows. In a first mode, the switch group supplies a signal voltage generated by the positive-polarity operating amplifier having the positive polarity to the first output buffer, feeds a signal output by the first output buffer back to the positive-polarity operating amplifier, supplies a signal voltage generated by the negative-polarity operating amplifier having the negative polarity to the second output buffer and feeds a signal output by the second output buffer back to the negative-polarity operating amplifier. In a second mode, on the other hand, the switch group supplies a signal voltage generated by the positive-polarity operating amplifier having the positive polarity to the second output buffer, feeds a signal output by the second output buffer back to the positive-polarity operating amplifier, supplies a signal voltage generated by the negative-polarity operating amplifier having the negative polarity to the first output buffer and feeds a signal output by the first output buffer back to the negative-polarity operating amplifier.

[0044] In a preferred configuration, the positive-polarity operating amplifier has a differential amplifier including a pair of first-conduction transistors while the negative-polarity operating amplifier has a differential amplifier including a pair of second-conduction transistors and each of the first and second output buffers has an AB-class push-pull operation function.

[0045] In another preferred configuration, each of the first and second output buffers has two input terminals.

[0046] In accordance with a second embodiment of the present invention, there is provided a display device employing a display section having display cells arranged to form a matrix as cells each to be subjected to a polarity inversion driving operation, and a signal-line driving circuit configured to supply said signal voltage having the positive or negative polarity to each of signal lines connected to said display cells in said polarity inversion driving operation. The signal-line driving circuit employs an output buffer section configured to amplify input data, which is used for driving said signal lines, in order to generate a signal voltage having said positive polarity as well as a signal voltage having said negative polarity and selectively supplying said signal voltage having said positive polarity and said signal voltage having said negative polarity to a pair of first and second ones of said signal lines. The output buffer section employs a positive-polarity operating amplifier configured to amplify said input data in order to generate said signal voltage having said positive polarity, a negative-polarity operating amplifier configured to amplify said input data in order to generate said signal voltage having said negative polarity, and a first output buffer configured to supply a signal voltage having said positive polarity or a signal voltage having said negative polarity to said first signal line. The output buffer section further employs a second output buffer configured to supply a signal voltage having said positive polarity or a signal voltage having said negative polarity

to said second signal line, and a switch group including switches placed between the output of said positive-polarity operating amplifier and the input of said first output buffer, and between the output of said negative-polarity operating amplifier and the input of said second output buffer. The switch group further includes switches placed at the feedback input stage of said positive-polarity operating amplifier, and at the feedback input stage of said negative-polarity operating amplifier. There are no switches placed between the output of said first output buffer and said first signal line and between the output of said second output buffer and said second signal line.

[0047] In accordance with a third embodiment of the present invention, there is provided an electronic equipment including a display device employing a display section having display cells arranged to form a matrix as cells each to be subjected to a polarity inversion driving operation, and a signal-line driving circuit configured to supply said signal voltage having the positive or negative polarity to each of signal lines connected to said display cells in said polarity inversion driving operation. The signal-line driving circuit employs an output buffer section configured to amplify input data, which is used for driving said signal lines, in order to generate a signal voltage having said positive polarity as well as a signal voltage having said negative polarity and selectively supplying said signal voltage having said positive polarity and said signal voltage having said negative polarity to a pair of first and second ones of said signal lines. The output buffer section employs a positive-polarity operating amplifier configured to amplify said input data in order to generate said signal voltage having said positive polarity, a negative-polarity operating amplifier configured to amplify said input data in order to generate said signal voltage having said negative polarity, and a first output buffer configured to supply a signal voltage having said positive polarity or a signal voltage having said negative polarity to said first signal line. The output further section further employs a second output buffer configured to supply a signal voltage having said positive polarity or a signal voltage having said negative polarity to said second signal line, and a switch group including switches placed between the output of said positive-polarity operating amplifier and the input of said first output buffer, and between the output of said negative-polarity operating amplifier and the input of said second output buffer. The switch group further includes switches placed at the feedback input stage of said positive-polarity operating amplifier, and at the feedback input stage of said negative-polarity operating amplifier. There are no switches placed between the output of said first output buffer and said first signal line and between the output of said second output buffer and said second signal line.

[0048] In accordance with the present embodiment, for example, in the first mode, a signal voltage generated by the positive-polarity operating amplifier as a signal voltage having the positive polarity is supplied to a first signal line by way of the first output buffer and a signal voltage

generated by the negative-polarity operating amplifier as a signal voltage having the negative polarity is supplied to a second signal line by way of the second output buffer.

[0049] In the second mode, on the other hand, a signal voltage generated by the positive-polarity operating amplifier as a signal voltage having the positive polarity is supplied to the second signal line by way of the second output buffer and a signal voltage generated by the negative-polarity operating amplifier as a signal voltage having the negative polarity is supplied to the first signal line by way of the first output buffer.

[0050] In accordance with the present embodiment, it is possible to prevent the circuit configuration from becoming complicated, the consumed current from increasing and the characteristics of the output buffer section from deteriorating and, hence, possible to reduce the element size (or the layout area).

[0051] In addition, since the present embodiment also exhibits an effect of cancelling an offset of the output-stage amplifier, the present embodiment contributes to improvement of the picture quality.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENT

[0052] A preferred embodiment of the invention is explained in detail by referring to diagrams as follows.

[0053] Fig. 4 is a block diagram showing a typical configuration of a liquid-crystal display device 100 according to an embodiment of the present invention.

[0054] In this case, as an example, the embodiment implements an active-matrix liquid-crystal display device 100 employing liquid-crystal cells each functioning as the electro optical element of every pixel circuit of the active-matrix liquid-crystal display device 100.

[0055] As shown in the block diagram of Fig. 4, the active-matrix liquid-crystal display device 100 employs an available display section (ACDSP) 110, a signal-line driving circuit 120, a gate-line driving circuit 130 and a data processing circuit (DATAPRC) 140. The available display section 110 includes a plurality of pixel circuits arranged on a transparent insulation substrate such as a glass substrate to form a matrix. As described above, each of the pixel circuits has a liquid-crystal cell. Also referred to hereafter as a horizontal driving circuit (HDRV) or a source driver, the signal-line driving circuit 120 is a circuit for driving signal lines. On the other hand, also referred to hereafter as a vertical driving circuit (VDRV) or a gate driver, the gate-line driving circuit 130 is a circuit for driving vertical scan lines or gate lines.

[0056] The following description sequentially explains the configuration and function of every configuration component employed in the active-matrix liquid-crystal display device 100 according to the embodiment.

[0057] As described above, the available display section 110 also referred to hereafter simply as a display section 110 includes a plurality of pixel circuits arranged on a transparent insulation substrate such as a glass

substrate to form a matrix.

[0058] In addition, the available display section 110 also has the signal lines driven by the signal-line driving circuit 120 and the gate lines (or the vertical scan lines) driven by the gate-line driving circuit 130. The signal lines are laid out to serve as lines each corresponding to one of the columns of the matrix whereas the gate lines (or the vertical scan lines) are laid out to serve as lines each corresponding to one of the rows of the matrix.

[0059] Fig. 5 is a circuit diagram showing a typical concrete configuration of the available display section 110.

[0060] In order to make the diagram simple, the typical matrix of the available display section 110 shown in the diagram consists of three rows, i.e., the (n-1)th to (n+1)th rows, and four columns, i.e., the (m-2)th to (m+1)th columns.

[0061] In the available display section 110 shown in the circuit diagram of Fig. 5, gate lines (or vertical scan lines) 111_{n-1} , 111_n and 111_{n+1} are laid out to serve as lines each corresponding to one of the rows of the matrix whereas signal lines (or data lines) 112_{m-2} , 112_{m-1} , 112_m and 112_{m+1} are laid out to serve as lines each corresponding to one of the columns of the matrix. A unit pixel circuit 113 is laid out at intersection portions of these lines.

[0062] A unit pixel circuit 113 is configured to include a TFT (Thin Film Transistor) serving as a pixel transistor, a liquid-crystal cell LC and a signal storage capacitor Cs.

[0063] The liquid-crystal cell LC is actually a capacitor created between a pixel electrode and an other electrode facing the pixel electrode. The pixel electrode is an electrode connected to the drain electrode of the thin-film transistor TFT. Facing the pixel electrode, the other electrode is also referred to as a facing electrode.

[0064] The gate electrode of the thin-film transistor TFT is connected to a gate line (also referred to as a vertical scan line), that is, one of gate lines ..., 111_{n-1} , 111_n , 111_{n+1} , The source electrode of the thin-film transistor TFT is connected to a signal line (also referred to as a data line), that is, one of signal lines ..., 112_{m-2} , 112_{m-1} , 112_m , 112_{m+1} ,

[0065] The pixel electrode of the liquid-crystal cell LC is connected to the drain electrode of the thin-film transistor TFT as described above whereas the facing electrode of the liquid-crystal cell LC is connected to a common line 114. The storage capacitor Cs is connected between the drain electrode of the thin film transistor and the common line 114.

[0066] A common-voltage supply circuit (VCOM circuit) 150 asserts an AC voltage determined in advance on the common line 114 as a common voltage Vcom.

[0067] One end of any particular one of the gate lines (or vertical scan lines) ..., 111_{n-1} , 111_n , 111_{n+1} , of the available display section 110 is connected an output terminal employed in the gate-line driving circuit 130 included in the active-matrix liquid-crystal display device 100 as shown in the block diagram of Fig. 4 as a terminal for a row corresponding to the particular gate line.

[0068] The gate-line driving circuit 130 has a configuration for example including a shift register for sequentially outputting a gate-line select pulse in synchronization with a vertical transfer clock signal VCK not shown in the circuit diagram of Fig. 5 to the gate lines (or vertical scan lines) ..., 111_{n-1} , 111_n , 111_{n+1} ,

[0069] One end of any particular one of the signal lines (or the data lines) ..., 112_{m-2} , 112_{m-1} , 112_m , 112_{m+1} , of the available display section 110 is connected an output terminal employed in the signal-line driving circuit 120 included in the active-matrix liquid-crystal display device 100 as shown in the block diagram of Fig. 4 as a terminal for a column corresponding to the particular signal line 112.

[0070] The signal-line driving circuit 120 has a function to convert driving data from digital data into analog data and amplify the analog data in accordance with a gradation voltage in order to generate a signal voltage having the positive polarity as well as a signal voltage having the negative polarity from the amplified analog data. The level of the driving data processed in this way is converted into a driving level used for driving a signal line 112. In addition, the signal-line driving circuit 120 also has a function to selectively supply the signal voltage having the positive polarity and the signal voltage having the negative polarity to respectively two signal lines 112 adjacent to each other.

[0071] The data processing circuit 140 employed in the active-matrix liquid-crystal display device 100 as shown in the block diagram of Fig. 4 has a level shifter and a serial-into-parallel converter. The level shifter is a section for shifting the level of parallel data received from an external source to a level determined in advance. In order to adjust the phase of data output by the level shifter and lower the frequency of the data, the serial-into-parallel converter carries out a process of conversion from serial data into parallel data. The serial-into-parallel converter outputs the parallel data to the signal-line driving circuit 120.

[0072] The following description concretely explains the configuration and function of the signal-line driving circuit 120 according to the embodiment.

[0073] Fig. 6 is a block diagram showing a typical configuration of the signal-line driving circuit 120 according to the embodiment.

[0074] The signal-line driving circuit 120 shown in the block diagram of Fig. 6 employs a shift register 121, a data latch section 122, a DAC (digital-into-analog converter) 123 and an output buffer section 124.

[0075] It is to be noted that an SPC (Serial-into-Parallel Converter) denoted by reference numeral 141 in the block diagram of Fig. 6 is the serial-into-parallel converter described above as the serial-into-parallel converter included in the data processing circuit 140.

[0076] The shift register 121 is a component for sequentially supplying a shift pulse (or a sampling pulse) from its transfer stages each corresponding to a column of the pixel matrix to the data latch section 122 in syn-

chronization with a horizontal transfer clock signal HCK not shown in the block diagram of Fig. 6.

[0077] The data latch section 122 is a component for sequentially sampling digital driving data received from the serial-into-parallel converter 141 as data to be used for driving the signal lines 112 in synchronization with the sampling pulse received from the shift register 121.

[0078] The DAC 123 is a component for receiving a gradation voltage and converting digital driving data received from the data latch section 122 into analog driving data.

[0079] The output buffer section 124 is a component for amplifying analog driving data received from the DAC 123 in order to generate a signal voltage having the positive polarity as well as a signal voltage having the negative polarity and selectively supplying the signal voltage having the positive polarity and the signal voltage having the negative polarity respectively to a pair of first and second ones of the signal lines 112 which are adjacent to each other.

[0080] Fig. 7 is a block diagram showing a typical configuration of the output buffer section 124 employed in the signal-line driving circuit 120 according to the embodiment.

[0081] In the following description, the output buffer section 124 is denoted by reference numeral 200.

[0082] The output buffer section 200 shown in the block diagram of Fig. 7 employs a positive-polarity OTA (Operational Transconductance Amplifier) 211, a first common output amplifier (OAMP) 212, a negative-polarity OTA 221, a second OAMP 222 and a switch group 230 including a first switch SW231 to an eighth switch SW238. The positive-polarity OTA 211 is a component connected to an output of the DAC 123 (DAC_UPPER) provided at the immediately preceding stage as a component having a function of amplifying input data from the DAC 123 in order to generate a signal voltage with the positive polarity. The first common OAMP 212 is a component provided with an output buffer function as a component for supplying a signal voltage with the positive or negative polarity to a first signal line 112_m connected to channel CH_m where notation m denotes an integer having a typical value of 1 in this case for example. The negative-polarity OTA 221 is a component connected to an output of the DAC 123 (DAC_LOWER) as a component having a function of amplifying input data from the DAC 123 in order to generate a signal voltage with the negative polarity. The second common OAMP 222 is a component provided with an output buffer function as a component for supplying a signal voltage with the positive or negative polarity to a second signal line 112_{m+1} connected to channel CH_{m+1} . For $m = 1$, channel CH_m is channel 1 whereas channel CH_{m+1} is channel 2. The first switch SW231 is a switch connected between the output terminal of the positive-polarity OTA 211 and a first input terminal of the first common OAMP 212. The second switch SW232 is a switch connected between the output terminal of the positive-polarity OTA 211 and a first input terminal of the

second common OAMP 222. The third switch SW233 is a switch connected between the output terminal of the negative-polarity OTA 221 and a second input terminal of the second common OAMP 222. The fourth switch SW234 is a switch connected between the output terminal of the negative-polarity OTA 221 and a second input terminal of the first common OAMP 212. The fifth switch SW235 is a switch connected between the output terminal of the first common OAMP 212 and the non-inverting (+) input terminal of the positive-polarity OTA 211. The sixth switch SW236 is a switch connected between the output terminal of the second common OAMP 222 and the non-inverting (+) input terminal of the positive-polarity OTA 211. The seventh switch SW237 is a switch connected between the output terminal of the second common OAMP 222 and the non-inverting (+) input terminal of the negative-polarity OTA 221. The eighth switch SW238 is a switch connected between the output terminal of the first common OAMP 212 and the non-inverting (+) input terminal of the negative-polarity OTA 221.

[0083] A signal output by the positive-polarity OTA 211 is supplied to the first input terminal of the first common OAMP 212 by way of the first switch SW231 and the first input terminal of the second common OAMP 222 by way of the second switch SW232.

[0084] A signal output by the negative-polarity OTA 221 is supplied the second input terminal of the second common OAMP 222 by way of the third switch SW233 and to the second input terminal of the first common OAMP 212 by way of the fourth switch SW234.

[0085] The inverting (-) input terminal of the positive-polarity OTA 211 is connected to an input terminal TI1 to which an output line (DAC_UPPER) of the DAC 123 provided at the immediately preceding stage is tied. The non-inverting (+) input terminal of the positive-polarity OTA 211 is connected to the output terminal of the first common OAMP 212 through the fifth switch SW235 and connected to the output terminal of the second common OAMP 222 through the sixth switch SW236.

[0086] The inverting (-) input terminal of the negative-polarity OTA 221 is connected to an input terminal TI2 to which an output line (DAC_LOWER) of the DAC 123 provided at the immediately preceding stage is tied. The non-inverting (+) input terminal of the negative-polarity OTA 221 is connected to the output terminal of the second common OAMP 222 through the seventh switch SW237 and connected to the output terminal of the first common OAMP 212 through the eighth switch SW238.

[0087] The output terminal of the first common OAMP 212 is connected to an output terminal TO1 to which a first signal line 112_m of channel 1 for $m = 1$ is tied.

[0088] The output terminal of the second common OAMP 222 is connected to an output terminal TO2 to which a second signal line 112_{m+1} of channel 2 for $m = 1$ is tied.

[0089] The switch group 230 is divided into first and second switch groups. The first switch group consists of the first switch SW231, the third switch SW233, the fifth

switch SW235 and the seventh switch SW237 which are each controlled by a common signal STR to turn on and off.

[0090] On the other hand, the second switch group consists of the second switch SW232, the fourth switch SW234, the sixth switch SW236 and the eighth switch SW238 which are each controlled by an other common signal CRS to turn on and off.

[0091] The first switch SW231, the third switch SW233, the fifth switch SW235 and the seventh switch SW237 which pertain to the first switch group are controlled to turn on and off complementarily to the second switch SW232, the fourth switch SW234, the sixth switch SW236 and the eighth switch SW238 which pertain to the second switch group.

[0092] That is to say, a control system not shown in the block diagram of Fig. 7 controls the common signals STR and CRS in such a way that, when the common signal STR is at a high level, the other common signal CSR is set at a low level but, when the common signal STR is at a low level, the other common signal CSR is set at a high level.

[0093] For example, the first switch SW231, the third switch SW233, the fifth switch SW235 and the seventh switch SW237 which pertain to the first switch group are each controlled to turn on by setting the common signal STR to a high level or turn off by setting the common signal STR to a low level.

[0094] By the same token, the second switch SW232, the fourth switch SW234, the sixth switch SW236 and the eighth switch SW238 which pertain to the second switch group are each controlled to turn on by setting the other common signal CRS to a high level or turn off by setting the other common signal CRS to a low level.

[0095] It is to be noted that, in this embodiment, control to turn on the common signals STR and CRS at the same time is prohibited.

[0096] In this embodiment, a state in which the common signal STR is at a high level and the common signal CRS is at a low level is referred to as a first mode whereas a state in which the common signal CRS is at a high level and the common signal STR is at a low level is referred to as a second mode which is a disjunction mode of the first mode.

[0097] A terminal a of the first switch SW231 is connected to the output terminal of the positive-polarity OTA 211 whereas a terminal b of the first switch SW231 is connected to the first input terminal of the first common OAMP 212.

[0098] A terminal a of the second switch SW232 is connected to the output terminal of the positive-polarity OTA 211 whereas a terminal b of the second switch SW232 is connected to the first input terminal of the second common OAMP 222.

[0099] A terminal a of the third switch SW233 is connected to the output terminal of the negative-polarity OTA 221 whereas a terminal b of the third switch SW233 is connected to the second input terminal of the second

common OAMP 222.

[0100] A terminal a of the fourth switch SW234 is connected to the output terminal of the negative-polarity OTA 221 whereas a terminal b of the fourth switch SW234 is connected to the second input terminal of the first common OAMP 212.

[0101] A terminal a of the fifth switch SW235 is connected to the output terminal of the first common OAMP 212 whereas a terminal b of the fifth switch SW235 is connected to the non-inverting (+) input terminal of the positive-polarity OTA 211.

[0102] A terminal a of the sixth switch SW236 is connected to the output terminal of the second common OAMP 222 whereas a terminal b of the sixth switch SW236 is connected to the non-inverting (+) input terminal of the positive-polarity OTA 211.

[0103] A terminal a of the seventh switch SW237 is connected to the non-inverting (+) terminal of the negative-polarity OTA 221 whereas a terminal b of the seventh switch SW237 is connected to the output terminal of the second common OAMP 222.

[0104] A terminal a of the eighth switch SW238 is connected to the non-inverting (+) terminal of the negative-polarity OTA 221 whereas a terminal b of the eighth switch SW238 is connected to the output terminal of the first common OAMP 212.

[0105] Fig. 8 is a circuit diagram showing a more concrete typical configuration of the output buffer section 200 shown in the block diagram of Fig. 7.

[0106] The positive-polarity OTA 211 employs PMOS (p-channel MOS) transistors PT211 and PT212 each serving as a transistor of a first conduction type, NMOS (n-channel MOS) transistors NT211 and NT212 each serving as a transistor of a second conduction type as well as a current source I211.

[0107] The source electrode of the PMOS transistor PT211 and the source electrode of the PMOS transistor PT212 are connected to a connection point wired to a power-supply electric potential Vdd.

[0108] The drain electrode of the PMOS transistor PT211 and the drain electrode of the NMOS transistor NT211 are connected to a connection point serving as a node ND211. The drain and gate electrodes of the PMOS transistor PT211 are connected to a connection point wired to the gate electrode of the PMOS transistor PT212. That is to say, the node ND211 is connected to the gate electrode of the PMOS transistor PT211 and the gate electrode of the PMOS transistor PT212.

[0109] The drain electrode of the PMOS transistor PT212 and the drain electrode of the NMOS transistor NT212 are connected to a connection point serving as an output node ND212 which is also the output terminal of the positive-polarity OTA 211.

[0110] The source electrode of the NMOS transistor NT211 and the source electrode of the NMOS transistor NT212 are connected to a connection point wired to the drain terminal of the current source I211.

[0111] The gate electrode of the NMOS transistor

NT211 serves as the non-inverting (+) of the positive-polarity OTA 211 whereas the gate electrode of the NMOS transistor NT212 serves as the inverting (-) of the positive-polarity OTA 211.

[0112] Since the non-inverting (+) of the positive-polarity OTA 211 is connected to the terminal b of the fifth switch SW235 and the terminal b of the sixth switch SW236, the gate electrode of the NMOS transistor NT211 is also connected to the terminals b of the fifth switch SW235 and the sixth switch SW236. Since the inverting (-) of the positive-polarity OTA 211 is connected to an input terminal T11 wired to an output terminal (DAC_UPPER) of the DAC 123, the gate electrode of the NMOS transistor NT212 is also connected to the input terminal T11.

[0113] The output node ND212 of the positive-polarity OTA 211 is connected to the terminal a of the first switch SW231 and the terminal a of the second switch SW232.

[0114] A differential amplifier including the NMOS transistor NT211 and the NMOS transistor NT212 in the positive-polarity OTA 211 having such a configuration amplifies the difference between a signal output by the DAC 123 (DAC_UPPER) and a signal output by the first common OAMP 212 or the second common OAMP 222 and outputs a data signal obtained as a result of the amplification to the first input terminal of the first common OAMP 212 by way of the first switch SW231 or the first input terminal of the second common OAMP 222 by way of the second switch SW232.

[0115] The first common OAMP 212 employs a PMOS transistor PT213, an NMOS transistor NT213, current sources I212 and I213 as well as a transfer gate TMG211. The transfer gate TMG211 employs a PMOS transistor PT214 and an NMOS transistor NT214 by connecting the source electrode of the PMOS transistor PT214 to the drain electrode of the NMOS transistor NT214 and connecting the drain electrode of the PMOS transistor PT214 to the source electrode of the NMOS transistor NT214.

[0116] The source electrode of the PMOS transistor PT213 is connected to the power-supply electric potential Vdd whereas the drain electrode of the PMOS transistor PT213 and the drain electrode of the NMOS transistor NT213 are connected to a connection point serving as an output node ND213 which is the output terminal of the first common OAMP 212. The source electrode of the NMOS transistor NT213 is connected to a ground electric potential GND.

[0117] The drain terminal of the current source I212 is connected to the power-supply electric potential Vdd. The source terminal of the current source I212, the gate electrode of the PMOS transistor PT213 and an input/output terminal T211 of the transfer gate TMG211 are connected to a connection point serving as an input node ND214 which is the first input terminal of the first common OAMP 212.

[0118] The source terminal of the current source I213 is connected to the ground electric potential GND. The

drain terminal of the current source I213, the gate electrode of the NMOS transistor NT213 and an input/output terminal T212 of the transfer gate TMG211 are connected to a connection point serving as an input node ND215 which is the second input terminal of the first common OAMP 212.

[0119] A first bias signal BIAS1 is supplied to the gate electrode of the PMOS transistor PT214 employed in the transfer gate TMG211 whereas a second bias signal BIAS2 is supplied to the gate electrode of the NMOS transistor NT214 employed in the transfer gate TMG211. The first and second bias signals BIAS1 and BIAS2 are applied to the gate electrodes as voltages used for setting a DC current flowing to the first common OAMP 212 provided at the output stage.

[0120] The output node ND213 of the first common OAMP 212 is connected to an output terminal TO1 for channel CH1. The input node ND214 of the first common OAMP 212 is connected to the terminal b of the first switch SW231 whereas the input node ND215 of the first common OAMP 212 is connected to the terminal b of the fourth switch SW234.

[0121] The first common OAMP 212 having such a configuration to serve as an output buffer carries out AB-class push-pull operations.

[0122] The negative-polarity OTA 221 employs PMOS (p-channel MOS) transistors PT221 and PT222 each serving as a transistor of a first conduction type, NMOS (n-channel MOS) transistors NT221 and NT222 each serving as a transistor of a second conduction type as well as a current source I221.

[0123] The source electrode of the PMOS transistor PT221 and the source electrode of the PMOS transistor PT222 are connected to a connection point wired to the source terminal of the current source I221 which has its drain terminal connected to the power-supply electric potential Vdd.

[0124] The drain electrode of the PMOS transistor PT221 and the drain electrode of the NMOS transistor NT221 are connected to a connection point serving as a node ND221. The drain and gate electrodes of the NMOS transistor NT221 are connected to a connection point wired to the gate electrode of the NMOS transistor NT222. That is to say, the node ND221 is connected to the gate electrode of the NMOS transistor NT221 and the gate electrode of the NMOS transistor NT222.

[0125] The drain electrode of the PMOS transistor PT222 and the drain electrode of the NMOS transistor NT222 are connected to a connection point serving as an output node ND222 which is also the output terminal of the negative-polarity OTA 221.

[0126] The source electrode of the NMOS transistor NT221 and the source electrode of the NMOS transistor NT222 are connected to a connection point wired to the ground electric potential GND.

[0127] The gate electrode of the PMOS transistor PT221 serves as the non-inverting (+) of the negative-polarity OTA 221 whereas the gate electrode of the PMOS

transistor PT222 serves as the inverting (-) of the negative-polarity OTA 221.

[0128] Since the non-inverting (+) of the negative-polarity OTA 221 is connected to the terminal a of the seventh switch SW237 and the terminal a of the eighth switch SW238, the gate electrode of the PMOS transistor PT221 is also connected to the terminals a of the seventh switch SW237 and the eighth switch SW238. Since the inverting (-) of the negative-polarity OTA 221 is connected to an input terminal T12 wired to an output terminal (DAC_LOWER) of the DAC 123, the gate electrode of the PMOS transistor PT222 is also connected to the input terminal T12.

[0129] The output node ND222 of the negative-polarity OTA 221 is connected to the terminal a of the third switch SW233 and the terminal a of the fourth switch SW234.

[0130] A differential amplifier including the PMOS transistor PT221 and the PMOS transistor PT222 in the negative-polarity OTA 221 having such a configuration amplifies the difference between a signal output by the DAC 123 (DAC_LOWER) and a signal output by the second common OAMP 222 or the first common OAMP 212 and outputs a data signal obtained as a result of the amplification to the second input terminal of the second common OAMP 222 by way of the third switch SW233 or the second input terminal of the first common OAMP 212 by way of the fourth switch SW234.

[0131] The second common OAMP 222 employs a PMOS transistor PT223, an NMOS transistor NT223, current sources I222 and I223 as well as a transfer gate TMG221. The transfer gate TMG221 employs a PMOS transistor PT224 and an NMOS transistor NT224 by connecting the source electrode of the PMOS transistor PT224 to the drain electrode of the NMOS transistor NT224 and connecting the drain electrode of the PMOS transistor PT224 to the source electrode of the NMOS transistor NT224.

[0132] The source electrode of the PMOS transistor PT223 is connected to the power-supply electric potential Vdd whereas the drain electrode of the PMOS transistor PT223 and the drain electrode of the NMOS transistor NT223 are connected to a connection point serving as an output node ND223 which is the output terminal of the second common OAMP 222. The source electrode of the NMOS transistor NT223 is connected to the ground electric potential GND.

[0133] The drain terminal of the current source I222 is connected to the power-supply electric potential Vdd. The source terminal of the current source I222, the gate electrode of the PMOS transistor PT223 and an input/output terminal T221 of the transfer gate TMG221 are connected to a connection point serving as an input node ND224 which is the first input terminal of the second common OAMP 222.

[0134] The source terminal of the current source I223 is connected to the ground electric potential GND. The drain terminal of the current source I223, the gate electrode of the NMOS transistor NT223 and an input/output

terminal T222 of the transfer gate TMG221 are connected to a connection point serving as an input node ND225 which is the second input terminal of the second common OAMP 222.

[0135] A first bias signal BIAS1 is supplied to the gate electrode of the PMOS transistor PT224 employed in the transfer gate TMG221 whereas a second bias signal BIAS2 is supplied to the gate electrode of the NMOS transistor NT224 employed in the transfer gate TMG221.

[0136] The first and second bias signals BIAS1 and BIAS2 are applied to the gate electrodes as voltages used for setting a DC current flowing to the second common OAMP 222 provided at the output stage.

[0137] The output node ND223 of the second common OAMP 222 is connected to an output terminal TO2 for channel CH2. The input node ND224 of the second common OAMP 222 is connected to the terminal b of the second switch SW232 whereas the input node ND225 of the second common OAMP 222 is connected to the terminal b of the third switch SW233.

[0138] The second common OAMP 222 having such a configuration to serve as an output buffer carries out AB-class push-pull operations.

[0139] As described above, in the configuration shown in the circuit diagram of Fig. 8 as a typical configuration of the output buffer section 200, the positive-polarity OTA 211 is configured to employ a differential amplifier including the NMOS transistor NT211 and the NMOS transistor NT212 whereas the negative-polarity OTA 221 is configured to employ a differential amplifier including the PMOS transistor PT221 and the PMOS transistor PT222.

[0140] Since each of the first common OAMP 212 and the second common OAMP 222 each serving as an output-stage buffer carries out an AB-class push-pull operation whereas the outputs of the positive-polarity OTA 211 and the negative-polarity OTA 221 have operating points different from each other, each of the first common OAMP 212 and the second common OAMP 222 is provided with 2 input terminals which are connected to 2 different nodes respectively.

[0141] By referring to the block diagram of Fig. 7 and a timing diagram of Fig. 9, the following description explains operations carried out by the output buffer section 200 employed in the signal-line driving circuit 120 according to the embodiment. As described earlier, the output buffer section 200 is also denoted by reference numeral 124 in this patent specification.

[0142] It is to be noted that Fig. 9 is a timing diagram showing a plurality of timing charts referred to in description of the operations carried out by the output buffer section 200 employed in the signal-line driving circuit 120 according to the embodiment. To be more specific, Fig. 9A shows the timing chart of the switch signal STR whereas Fig. 9B shows the timing chart of the switch signal CRS. Fig. 9C shows the timing charts of the levels of signals DAC_UPPER and DAC_LOWER generated by the DAC 123 whereas Fig. 9D shows the timing charts of the levels of signals output for channels CH1 and CH2.

[0143] The output buffer section 200 adopts a technique different from the output selector method explained earlier by referring to the block diagram of Fig. 3. In accordance with the technique adopted by the output buffer section 200, the first and second input terminals of the first common OAMP 212 and the first and second input terminals of the second common OAMP 222 are connected to the switches SW231 to SW234 provided at a stage immediately preceding the first common OAMP 212 and the second common OAMP 222 in order to switch signals supplied to the first common OAMP 212 and the second common OAMP 222, which are provided at the output stage for channels CH1 and CH2 respectively, in a mutually complementary manner.

[0144] In addition, the switches SW235 to SW238 are provided on feedback paths to the positive-polarity OTA 211 and the negative-polarity OTA 221 as switches for switching signals fed back to the positive-polarity OTA 211 and the negative-polarity OTA 221 in a mutually complementary manner.

[0145] In such a configuration, in the first mode with the switch signal STR set at a high level and the switch signal CRS set at a low level, each of the switches SW231, SW233, SW235 and SW237 pertaining to the first switch group of the switch group 230 is in a turned-on state whereas each of the switches SW232, SW234, SW236 and SW238 pertaining to the second switch group of the switch group 230 is in a turned-off state.

[0146] In these states of the first mode, a positive-polarity signal voltage generated by the positive-polarity OTA 211 is supplied to a first signal line 112_m by way of the first common OAMP 212 whereas a negative-polarity signal voltage generated by the negative-polarity OTA 221 is supplied to a second signal line 112_{m+1} by way of the second common OAMP 222.

[0147] In the second mode with the switch signal STR set at a low level and the switch signal CRS set at a high level, on the other hand, each of the switches SW231, SW233, SW235 and SW237 pertaining to the first switch group of the switch group 230 is in a turned-off state whereas each of the switches SW232, SW234, SW236 and SW238 pertaining to the second switch group of the switch group 230 is in a turned-on state.

[0148] In these states of the second mode, a positive-polarity signal voltage generated by the positive-polarity OTA 211 is supplied to the second signal line 112_{m+1} by way of the second common OAMP 222 whereas a negative-polarity signal voltage generated by the negative-polarity OTA 221 is supplied to the first signal line 112_m by way of the first common OAMP 212.

[0149] By employing the output buffer section 200 provided with such a configuration in accordance with the embodiment, each of the signal voltage generated by the positive-polarity OTA 211 and the signal voltage generated by the negative-polarity OTA 221 can be switched from the positive polarity to the negative polarity and vice versa without providing switches (SW) on the output paths of the output buffer section 200.

[0150] Since no switches (SW) are provided on the output paths of the output buffer section 200, the settling process ends early in comparison with the configuration adopting the output selector method.

[0151] In the configuration according to the embodiment, each of the first common OAMP 212 and the second common OAMP 222 which are provided at the output stage is an amplifier common to the positive-polarity OTA 211 and the negative-polarity OTA 221.

[0152] In the configuration shown in the block diagram of Fig. 7 and the circuit diagram of Fig. 8 as a typical configuration of the output buffer section 200, each of the first common OAMP 212 and the second common OAMP 222 which are placed at the output stage is provided with 2 input terminals. It is to be noted, however, that the scope of the present embodiment is by no means limited to this output buffer section 200 employed in the signal-line driving circuit 120 according to the embodiment. For example, each of the first common OAMP 212 and the second common OAMP 222 can also be provided with 1 input terminal.

[0153] Fig. 10A and 10B are a plurality of diagrams referred to in comparing the configuration adopting the output selector method explained earlier by referring to the block diagram of Fig. 3 with the configuration of the output buffer section 200 employed in the signal-line driving circuit 120 according to the embodiment.

[0154] As shown in the diagram of Fig. 10B, since the switches (SW) are not provided on the output paths of the output buffer section 200, it is not necessary to reduce the turned-on-state resistances of the switches (SW). Thus, the sizes of the switches (SW) can be decreased.

[0155] In addition, since no switches (SW) are connected to each other in series, each of the size of the first common OAMP 212 and the size of the second common OAMP 222 can also be reduced as well.

[0156] Fig. 11A and 11B are a plurality of diagrams used for showing differences between the appearances of voltages at a bias stage of the configuration adopting the output selector method and the appearances of voltages at a bias stage of the output buffer section 200 employed in the signal-line driving circuit 120 according to the embodiment. The diagrams of Fig. 11 are to be referred to in comparing offset-voltage effects caused by the amplifiers provided at the output stage.

[0157] If the voltage offsets caused by the amplifiers provided at the output stage match each other, in the configuration adopting the output selector method, the offset voltages cannot be cancelled as shown in the diagram of Fig. 11A so that the luminance is shifted from the ideal value.

[0158] In the case of the output selector method, an average voltage relative to the center level of a common voltage V_{com} is represented by an expression of $V_a + (\Delta V_1 + \Delta V_2)/2$ indicating that the offset values ΔV_1 and ΔV_2 are not cancelled.

[0159] In accordance with the output buffer section 200 employed in the signal-line driving circuit 120 according

to the embodiment, on the other hand, the polarity inversion causes the offset voltages for 2 frames to cancel each other as shown in the diagram of Fig. 11B.

[0160] In the case of the output buffer section 200 employed in the signal-line driving circuit 120 according to the embodiment, an average voltage relative to the center level of the common voltage V_{com} is V_a which is independent of the offset values $\Delta V1$ and $\Delta V2$.

[0161] As described above, in accordance with this embodiment, the output buffer section 200 employs the positive-polarity OTA 211, the first common OAMP 212, the negative-polarity OTA 221, the second common OAMP 222 and the switch group 230 which together exhibits effects described below. As explained before, the positive-polarity OTA 211 is a component having a function to amplify input data received from the DAC 123 (DAC_UPPER) provided at the stage immediately preceding the output buffer section 200 in order to generate a signal voltage with the positive polarity. The first common OAMP 212 is a component having the function of an output buffer for supplying a signal voltage with the positive or negative polarity to a first signal line 112_m connected to channel CH_m , that is, channel CH1 for $m = 1$ for example. By the same token, the negative-polarity OTA 221 is a component having a function to amplify input data received from the DAC 123 (DAC_LOWER) provided at the stage immediately preceding the output buffer section 200 in order to generate a signal voltage with the negative polarity. The second common OAMP 222 is a component having the function of an output buffer for supplying a signal voltage with the positive or negative polarity to a second signal line 112_{m+1} connected to channel $CH_{(m+1)}$ that is, channel CH2 for $m = 1$. The switch group 230 consists of switches ranging from a first switch SW231 to an eighth switch SW238. The first switch SW231 is a switch connected between the output terminal of the positive-polarity OTA 211 and a first input terminal of the first common OAMP 212. The second switch SW232 is a switch connected between the output terminal of the positive-polarity OTA 211 and a first input terminal of the second common OAMP 222. The third switch SW233 is a switch connected between the output terminal of the negative-polarity OTA 221 and a second input terminal of the second common OAMP 222. The fourth switch SW234 is a switch connected between the output terminal of the negative-polarity OTA 221 and a second input terminal of the first common OAMP 212. The fifth switch SW235 is a switch connected between the output terminal of the first common OAMP 212 and the non-inverting (+) input terminal of the positive-polarity OTA 211. The sixth switch SW236 is a switch connected between the output terminal of the second common OAMP 222 and the non-inverting (+) input terminal of the positive-polarity OTA 211. The seventh switch SW237 is a switch connected between the output terminal of the second common OAMP 222 and the non-inverting (+) input terminal of the negative-polarity OTA 221. The eighth switch SW238 is a switch connected between the output

terminal of the first common OAMP 212 and the non-inverting (+) input terminal of the negative-polarity OTA 221.

[0162] Since the switches (SW) are not provided on the output paths of the output buffer section 200, that is, since the switches (SW) are provided inside the output buffer section 200, the sizes of the switches (SW) can be reduced. Thus, the layout area can be decreased as well.

[0163] In addition, since the switches (SW) are not provided on the output paths of the output buffer section 200, the MOS sizes at the output stages of the output buffer section 200 can be reduced. Thus, the layout area can be decreased as well.

[0164] On top of that, since the switches (SW) are not provided on the output paths of the output buffer section 200, the settling process and, hence, the characteristic of the output buffer section 200 can be improved.

[0165] In addition, the polarity inversion causes the offset voltages caused by the amplifiers at the output stage of the output buffer section 200 for 2 frames to cancel each other. It is thus possible to improve the characteristic of the output buffer section 200 and, hence, the picture quality.

[0166] The embodiment described above implements a liquid-crystal display device of the active matrix type. It is to be noted, however, that the scope of the present invention is by no means limited to the active-matrix liquid-crystal display device according to the embodiment. That is to say, the present invention can also be applied to an other active-matrix display device of the active-matrix type. An example of the other active-matrix display device of the active-matrix type is an active-matrix organic EL (Electro Luminescence) display device employing pixel circuits each including an organic EL element as an electro optical element.

[0167] In addition, it should be understood by those skilled in the art that a variety of modifications, combinations, sub-combinations and alterations may occur, depending on design requirements and other factors as far as they are within the scope of the appended claims or the equivalents thereof.

[0168] The active-matrix liquid-crystal display device according to the present embodiment described above is employed in a variety of electronic equipments shown in diagrams of Figs. 12 to 16 as electronic equipments used in all fields. Examples of the electronic equipments are a digital camera, a notebook personal computer, a portable terminal such as a cellular phone and a video camera. In each of these electronic equipments, the liquid-crystal display device is used for displaying a video signal supplied thereto or generated therein as an image or a video. The following description explains concrete implementations of the electronic equipment to which the present embodiment is applied.

[0169] Fig. 12 is a diagram showing a squint view of the external appearance of a TV set 300 to which the present embodiment is applied. The TV set 300 serving

as a typical implementation of the electronic equipment to which the present embodiment is applied employs a front panel 320 and a video display screen section 310 which is for example a filter glass plate 330. The TV set 300 is constructed by employing the liquid-crystal display device provided by the present embodiment in the TV set as the video display screen section 310.

[0170] Fig. 13A and 13B are a plurality of diagrams each showing a squint view of the external appearance of a digital camera 300A to which the present embodiment is applied. To be more specific, Fig. 13A is a diagram showing a squint view of the external appearance of the digital camera 300A seen from a position on the front side of the digital camera whereas Fig. 13B is a diagram showing a squint view of the external appearance of the digital camera 300A seen from a position on the rear side of the digital camera.

[0171] The digital camera 300A serving as a typical implementation of the electronic equipment to which the present embodiment is applied employs a light emitting section 311 for generating a flash, a display section 312, a menu switch 113 and a shutter button 314. The digital camera 300A is constructed by employing the liquid-crystal display device provided by the present embodiment in the digital camera 300A as the display section 312.

[0172] Fig. 14 is a diagram showing a squint view of the external appearance of a notebook personal computer 300B to which the present embodiment is applied. The notebook personal computer 300B serving as a typical implementation of the electronic equipment to which the present embodiment is applied employs a main body 321 including a keyboard 322 to be operated by the user for entering characters and a display section 323 for displaying an image. The notebook personal computer 300B is constructed by employing the liquid-crystal display device provided by the present embodiment in the personal computer 300B as the display section 323.

[0173] Fig. 15 is a diagram showing a squint view of the external appearance of a video camera 300C to which the present embodiment is applied. The video camera serving 300C as a typical implementation of the electronic equipment to which the present embodiment is applied employs a main body 331, a photographing lens 332, a start/stop switch 333 and a display section 334. Provided on the front face of the video camera, the photographing lens 332 oriented forward is a lens for taking a picture of a subject of photographing. The start/stop switch 333 is a switch to be operated by the user to start or stop a photographing operation. The video camera 300C is constructed by employing the liquid-crystal display device provided by the present embodiment in the video camera 300C as the display section 334.

[0174] Fig. 16 is a plurality of diagrams each showing the external appearance of a portable terminal such as a cellular phone 300D to which the present embodiment is applied. To be more specific, Fig. 16A is a diagram showing the front view of the cellular phone 300D in a state of being already opened. Fig. 16B is a diagram

showing a side of the cellular phone 300D in a state of being already opened. Fig. 16C is a diagram showing the front view of the cellular phone 300D in a state of being already closed. Fig. 16D is a diagram showing the left side of the cellular phone 300D in a state of being already closed. Fig. 16E is a diagram showing the right side of the cellular phone 300D in a state of being already closed. Fig. 16F is a diagram showing the top view of the cellular phone 300D in a state of being already closed. Fig. 16G is a diagram showing the bottom view of the cellular phone 300D in a state of being already closed.

[0175] The cellular phone 300D serving as a typical implementation of the electronic equipment to which the present embodiment is applied employs an upper case 341, a lower case 342, a link section 343 which is a hinge, a display section 344, a display sub-section 345, a picture light 346 and a camera 347. The cellular phone 300D is constructed by employing the liquid-crystal display device provided by the present embodiment in the cellular phone 300D as the display section 344 and/or the display sub-section 345.

Claims

1. A signal-line driving circuit (120) configured to supply a signal voltage having a positive or negative polarity to each of signal lines (112) connected to display cells arranged to form a matrix as cells each to be subjected to a polarity inversion driving operation according to inversion of the polarity of said signal voltage, said signal-line driving circuit (120) comprising:

an output buffer section (200) configured to amplify input data, which is used for driving said signal lines (112), in order to generate a signal voltage having said positive polarity as well as a signal voltage having said negative polarity and selectively supplying said signal voltage having said positive polarity and said signal voltage having said negative polarity to a pair of first and second ones of said signal lines (112m, 112m+1), wherein said output buffer section (200) employs

a positive-polarity operating amplifier (211) comprising an inverting input terminal connected to a first input terminal (T11) of the signal driving circuit, and

a negative-polarity operating amplifier (221) comprising an inverting input terminal connected to a second input terminal (T12) of the signal driving circuit, **characterized by**

an inverting first output buffer (212) configured to supply a signal voltage generated by said positive-polarity operating amplifier (211) and having said positive polarity or a signal voltage generated by said negative-polarity operating amplifier (221) and having said negative polarity to

said first signal line (112m),
 an inverting second output buffer (222) configured to supply the signal voltage having said positive polarity or the signal voltage having said negative polarity to said second signal line (112m+1),
 a first switch (SW231) connected between the output terminal of said positive-polarity operating amplifier (211) and an input of said first output buffer (212),
 a second switch (SW232) connected between the output terminal of said positive-polarity operating amplifier (211) and an input of said second output buffer (222),
 a third switch (SW233) connected between the output terminal of said negative-polarity operating amplifier (221) and an input of said second output buffer (222),
 a fourth switch (SW234) connected between the output terminal of said negative-polarity operating amplifier (221) and an input of said first output buffer (212),
 a fifth switch (SW235) connected between an output terminal of the first output buffer (212) and a non-inverting input terminal of said positive-polarity operating amplifier (211),
 a sixth switch (SW236) connected between an output of the second output buffer (222) and the non-inverting input terminal of said positive-polarity operating amplifier (211),
 a seventh switch (SW237) connected between an output terminal of the second output buffer (222) and a non-inverting input terminal of said negative-polarity operating amplifiers (221), and
 an eighth switch (SW238) connected between an output terminal of the first output buffer (212) and the non-inverting input terminal of said negative-polarity operating amplifier (221), the switches being controllable such that
 a first group of switches comprising the first, third, fifth and seventh switch (SW231, SW233, SW235, SW237) is in a turned-on state when a second group of switches comprising the second, fourth, sixth and eighth switch (SW232, SW234, SW236, SW238) is in a turned-off state and vice versa.

2. The signal-line driving circuit (120) according to claim 1, wherein:

said positive-polarity operating amplifier (211) has a differential amplifier including a pair of first-conduction transistors;
 said negative-polarity operating amplifier (221) has a differential amplifier including a pair of second-conduction transistors; and
 each of said first and second output buffers (212, 222) has an AB-class push-pull operation func-

tion.

3. The signal-line driving circuit according to claim 2, wherein
 each of said first and second output buffers (212, 222) has two input terminals.
4. A display device comprising:
 a display section (110) having display cells arranged to form a matrix as cells each to be subjected to a polarity inversion driving operation; and
 a signal-line driving circuit (120) as claimed in one of the preceding claims, the signal-line driving circuit (120) being configured to supply said signal voltage having the positive or negative polarity to each of signal lines connected to said display cells.
5. An electronic equipment including the display device according to claim 4.
6. A method of operating a signal-line driving circuit (120) configured to supply a signal voltage having a positive or negative polarity to each of signal lines (112) connected to display cells arranged to form a matrix as cells each to be subjected to a polarity inversion driving operation according to inversion of the polarity of said signal voltage, said signal-line driving circuit (120) comprising:

an output buffer section (200) configured to amplify input data, which is used for driving said signal lines (112), in order to generate a signal voltage having said positive polarity as well as a signal voltage having said negative polarity and selectively supplying said signal voltage having said positive polarity and said signal voltage having said negative polarity to a pair of first and second ones of said signal lines (112m, 112m+1), wherein said output buffer section (200) employs

a positive-polarity operating amplifier (211) comprising an inverting input terminal connected to a first input terminal (T11) of the signal driving circuit, and

a negative-polarity operating amplifier (221) comprising an inverting input terminal connected to a second input terminal (T12) of the signal driving circuit,

an inverting first output buffer (212) configured to supply a signal voltage generated by said positive-polarity operating amplifier (211) and having said positive polarity or a signal voltage generated by said negative-polarity operating amplifier (221) and having said negative polarity to said first signal line (112m),

an inverting second output buffer (222) configured to supply the signal voltage having said positive polarity or the signal voltage having said negative polarity to said second signal line (112m+1),
 a first switch (SW231) connected between the output terminal of said positive-polarity operating amplifier (211) and an input of said first output buffer (212),
 a second switch (SW232) connected between the output terminal of said positive-polarity operating amplifier (211) and an input of said second output buffer (222),
 a third switch (SW233) connected between the output terminal of said negative-polarity operating amplifier (221) and an input of said second output buffer (222),
 a fourth switch (SW234) connected between the output terminal of said negative-polarity operating amplifier (221) and an input of said first output buffer (212),
 a fifth switch (SW235) connected between an output terminal of the first output buffer (212) and a non-inverting input terminal of said positive-polarity operating amplifier (211),
 a sixth switch (SW236) connected between an output of the second output buffer (222) and the non-inverting input terminal of said positive-polarity operating amplifier (211),
 a seventh switch (SW237) connected between an output terminal of the second output buffer (222) and a non-inverting input terminal of said negative-polarity operating amplifiers (221), and
 an eighth switch (SW238) connected between an output terminal of the first output buffer (212) and the non-inverting input terminal of said negative-polarity operating amplifier (221), the switches being controllable such that
 a first group of switches comprising the first, third, fifth and seventh switch (SW231, SW233, SW235, SW237) is in a turned-on state when a second group of switches comprising the second, fourth, sixth and eighth switch (SW232, SW234, SW236, SW238) is in a turned-off state and vice versa, wherein
 in a first mode, said switch group (230) supplies a signal voltage generated by said positive-polarity operating amplifier (211) having the positive polarity to said first output buffer (212), feeds a signal output by said first output buffer (212) back to said positive-polarity operating amplifier (211),
 supplies a signal voltage generated by said negative-polarity operating amplifier (221) having the negative polarity to said second output buffer (222), and
 feeds a signal output by said second output buffer (222) back to said negative-polarity operating

amplifier (221) whereas, in a second mode, said switch group
 supplies a signal voltage generated by said positive-polarity operating amplifier (211) having the positive polarity to said second output buffer (222),
 feeds a signal output by said second output buffer (222) back to said positive-polarity operating amplifier (211),
 supplies a signal voltage generated by said negative-polarity operating amplifier (211) having the negative polarity to said first output buffer (212), and
 feeds a signal output by said first output buffer (212) back to said negative-polarity operating amplifier (221).

Patentansprüche

1. Signalleitungstreiberschaltung (120), die konfiguriert ist, eine Signalspannung mit einer positiven oder negativen Polarität jeweils an Signalleitungen (112) zu liefern, die mit Anzeigezellen verbunden sind, die so angeordnet sind, dass sie eine Matrix bilden, als Zellen, die jeweils einer Polaritätsumkehrtreiberoperation gemäß einer Umkehrung der Polarität der Signalspannung unterzogen werden sollen, wobei die Signalleitungstreiberschaltung (120) Folgendes aufweist:

einen Ausgangspufferabschnitt (200), der konfiguriert ist, Eingangsdaten zu verstärken, die zum Betreiben der Signalleitungen (112) verwendet werden, um eine Signalspannung mit der positiven Polarität sowie eine Signalspannung mit der negativen Polarität zu erzeugen, und die Signalspannung mit der positiven Polarität und die Signalspannung mit der negativen Polarität einem Paar erster und zweiter der Signalleitungen (112m, 112m+1) selektiv zuführt, wobei der Ausgangspufferabschnitt (200) Folgendes einsetzt:

einen Operationsverstärker (211) für die positive Polarität, der einen invertierenden Eingangsanschluss aufweist, der mit einem ersten Eingangsanschluss (T11) der Signalleitungstreiberschaltung verbunden ist, und
 einen Operationsverstärker (221) für die negative Polarität, der einen invertierenden Eingangsanschluss aufweist, der mit einem zweiten Eingangsanschluss (T12) der Signalleitungstreiberschaltung verbunden ist,
gekennzeichnet durch
 einen invertierenden ersten Ausgangspuffer (212), der konfiguriert ist, eine Signal-

spannung, die durch den Operationsverstärker (211) für die positive Polarität erzeugt wird und die positive Polarität aufweist, oder eine Signalspannung, die durch den Operationsverstärker (221) für die negative Polarität erzeugt wird und die negative Polarität aufweist, an die erste Signalleitung (112m) zu liefern, 5
einen invertierenden zweiten Ausgangspuffer (222), der konfiguriert ist, die Signalspannung mit der positiven Polarität oder die Signalspannung mit der negativen Polarität an die zweite Signalleitung (112m+1) zu liefern, 10
einen ersten Schalter (SW231), der zwischen dem Ausgangsanschluss des Operationsverstärkers (211) für die positive Polarität und einem Eingang des ersten Ausgangspuffers (212) geschaltet ist, 15
einen zweiten Schalter (SW232), der zwischen dem Ausgangsanschluss des Operationsverstärkers (211) für die positive Polarität und einem Eingang des zweiten Ausgangspuffers (222) geschaltet ist, 20
einen dritten Schalter (SW233), der zwischen dem Ausgangsanschluss des Operationsverstärkers (221) für die negative Polarität und einem Eingang des zweiten Ausgangspuffers (222) geschaltet ist, 25
einen vierten Schalter (SW234), der zwischen dem Ausgangsanschluss des Operationsverstärkers (221) für die negative Polarität und einem Eingang des ersten Ausgangspuffers (212) geschaltet ist, 30
einen fünften Schalter (SW235), der zwischen einem Ausgangsanschluss des ersten Ausgangspuffers (212) und einem nicht-invertierenden Eingangsanschluss des Operationsverstärkers (211) für die positive Polarität geschaltet ist, 35
einen sechsten Schalter (SW236), der zwischen einem Ausgang des zweiten Ausgangspuffers (222) und dem nicht-invertierenden Eingangsanschluss des Operationsverstärkers (211) für die positive Polarität geschaltet ist, 40
einen siebten Schalter (SW237), der zwischen einem Ausgangsanschluss des zweiten Ausgangspuffers (222) und einem nicht-invertierenden Eingangsanschluss des Operationsverstärkers (221) für die negative Polarität geschaltet ist, und 45
einen achten Schalter (SW238), der zwischen einem Ausgangsanschluss des ersten Ausgangspuffers (212) und dem nicht-invertierenden Eingangsanschluss des Operationsverstärkers (221) für die negative Polarität geschaltet ist, wobei die Schal-

ter so steuerbar sind, dass eine erste Gruppe von Schaltern, die den ersten, dritten, fünften und siebten Schalter (SW231, SW233, SW235, SW237) aufweisen, sich in einem eingeschalteten Zustand befindet, wenn eine zweite Gruppe von Schaltern, die den zweiten, vierten, sechsten und achten Schalter (SW232, SW234, SW236, SW238) aufweisen, sich in einem ausgeschalteten Zustand befindet, und umgekehrt.

2. Signalleitungstreiberschaltung (120) nach Anspruch 1, wobei:

der Operationsverstärker (211) für die positive Polarität einen Differenzverstärker aufweist, der ein Paar Transistoren eines ersten Leitungstyps aufweist;
der Operationsverstärker (221) für die negative Polarität einen Differenzverstärker aufweist, der ein Paar Transistoren eines zweiten Leitungstyps aufweist; und
jeder des ersten und zweiten Ausgangspuffers (212, 222) eine AB-Betriebsgegentaktfunktion aufweist.

3. Signalleitungstreiberschaltung nach Anspruch 2, wobei
jeder des ersten und zweiten Ausgangspuffers (212, 222) zwei Eingangsanschlüsse aufweist.

4. Anzeigevorrichtung, die Folgendes aufweist:

einen Anzeigeabschnitt (110), der Anzeigezellen aufweist, die so angeordnet sind, dass sie eine Matrix bilden, als Zellen, die jeweils einer Polaritätsumkehrtreiberoperation unterzogen werden sollen; und
eine Signalleitungstreiberschaltung (120) nach einem der vorhergehenden Ansprüche, wobei die Signalleitungstreiberschaltung (120) konfiguriert ist, die Signalspannung mit der positiven oder negativen Polarität jeweils an Signalleitungen zu liefern, die mit den Anzeigezellen verbunden sind.

5. Elektronisches Gerät, das die Anzeigevorrichtung nach Anspruch 4 aufweist.

6. Verfahren zum Betreiben einer Signalleitungstreiberschaltung (120), die konfiguriert ist, eine Signalspannung mit einer positiven oder negativen Polarität jeweils an Signalleitungen (112) zu liefern, die mit Anzeigezellen verbunden sind, die so angeordnet sind, dass sie eine Matrix bilden, als Zellen, die jeweils einer Polaritätsumkehrtreiberoperation gemäß einer Umkehrung der Polarität der Signalspannung

unterzogen werden sollen, wobei die Signalleitungstreiberschaltung (120) Folgendes aufweist:

einen Ausgangspufferabschnitt (200), der konfiguriert ist, Eingangsdaten zu verstärken, die zum Betreiben der Signalleitungen (112) verwendet werden, um eine Signalspannung mit der positiven Polarität sowie eine Signalspannung mit der negativen Polarität zu erzeugen, und die Signalspannung mit der positiven Polarität und die Signalspannung mit der negativen Polarität einem Paar erster und zweiter der Signalleitungen (112m, 112m+1) selektiv zuführt, wobei der Ausgangspufferabschnitt (200) Folgendes einsetzt:

einen Operationsverstärker (211) für die positive Polarität, der einen invertierenden Eingangsanschluss aufweist, der mit einem ersten Eingangsanschluss (T11) der Signalleitungstreiberschaltung verbunden ist, und
einen Operationsverstärker (221) für die negative Polarität, der einen invertierenden Eingangsanschluss aufweist, der mit einem zweiten Eingangsanschluss (T12) der Signalleitungstreiberschaltung verbunden ist, einen invertierenden ersten Ausgangspuffer (212), der konfiguriert ist, eine Signalspannung, die durch den Operationsverstärker (211) für die positive Polarität erzeugt wird und die positive Polarität aufweist, oder eine Signalspannung, die durch den Operationsverstärker (221) für die negative Polarität erzeugt wird und die negative Polarität aufweist, an die erste Signalleitung (112m) zu liefern,
einen invertierenden zweiten Ausgangspuffer (222), der konfiguriert ist, die Signalspannung mit der positiven Polarität oder die Signalspannung mit der negativen Polarität an die zweite Signalleitung (112m+1) zu liefern,
einen ersten Schalter (SW231), der zwischen dem Ausgangsanschluss des Operationsverstärkers (211) für die positive Polarität und einem Eingang des ersten Ausgangspuffers (212) geschaltet ist,
einen zweiten Schalter (SW232), der zwischen dem Ausgangsanschluss des Operationsverstärkers (211) für die positive Polarität und einem Eingang des zweiten Ausgangspuffers (222) geschaltet ist,
einen dritten Schalter (SW233), der zwischen dem Ausgangsanschluss des Operationsverstärkers (221) für die negative Polarität und einem Eingang des zweiten Ausgangspuffers (222) geschaltet ist,

einen vierten Schalter (SW234), der zwischen dem Ausgangsanschluss des Operationsverstärkers (221) für die negative Polarität und einem Eingang des ersten Ausgangspuffers (212) geschaltet ist,
einen fünften Schalter (SW235), der zwischen einem Ausgangsanschluss des ersten Ausgangspuffers (212) und einem nicht-invertierenden Eingangsanschluss des Operationsverstärkers (211) für die positive Polarität geschaltet ist,
einen sechsten Schalter (SW236), der zwischen einem Ausgang des zweiten Ausgangspuffers (222) und dem nicht-invertierenden Eingangsanschluss des Operationsverstärkers (211) für die positive Polarität geschaltet ist,
einen siebten Schalter (SW237), der zwischen einem Ausgangsanschluss des zweiten Ausgangspuffers (222) und einem nicht-invertierenden Eingangsanschluss des Operationsverstärkers (221) für die negative Polarität geschaltet ist, und
einen achten Schalter (SW238), der zwischen einem Ausgangsanschluss des ersten Ausgangspuffers (212) und dem nicht-invertierenden Eingangsanschluss des Operationsverstärkers (221) für die negative Polarität geschaltet ist, wobei die Schalter so steuerbar sind, dass
eine erste Gruppe von Schaltern, die den ersten, dritten, fünften und siebten Schalter (SW231, SW233, SW235, SW237) aufweisen, sich in einem eingeschalteten Zustand befindet, wenn eine zweite Gruppe von Schaltern, die den zweiten, vierten, sechsten und achten Schalter (SW232, SW234, SW236, SW238) aufweisen, sich in einem ausgeschalteten Zustand befindet und umgekehrt, wobei
in einer ersten Betriebsart, die Schaltergruppe (230)
eine Signalspannung, die durch den Operationsverstärker (211) für die positive Polarität erzeugt wird, die die positive Polarität aufweist, an den ersten Ausgangspuffer (212) liefert,
ein Signal, das durch den ersten Ausgangspuffer (212) ausgegeben wird, zum Operationsverstärker (211) für die positive Polarität zurückführt,
eine Signalspannung, die durch den Operationsverstärker (221) für die negative Polarität erzeugt wird, die die negative Polarität aufweist, an den zweiten Ausgangspuffer (222) liefert, und
ein Signal, das durch den zweiten Ausgangspuffer (222) ausgegeben wird, zum

Operationsverstärker (221) für die negative Polarität zurückführt, während in einer zweiten Betriebsart, die Schaltergruppe eine Signalspannung, die durch den Operationsverstärker (211) für die positive Polarität erzeugt wird, die die positive Polarität aufweist, an den zweiten Ausgangspuffer (222) liefert, 5
ein Signal, das durch den zweiten Ausgangspuffer (222) ausgegeben wird, zum Operationsverstärker (211) für die positive Polarität zurückführt, 10
eine Signalspannung, die durch den Operationsverstärker (221) für die negative Polarität erzeugt wird, die die negative Polarität aufweist, an den ersten Ausgangspuffer (212) liefert, und 15
ein Signal, das durch den ersten Ausgangspuffer (212) ausgegeben wird, zum Operationsverstärker (221) für die negative Polarität zurückführt. 20

Revendications

1. Circuit de commande (120) de lignes de signal, conçu pour fournir une tension de signal présentant une polarité positive ou négative à chacune d'une pluralité de lignes de signal (112) reliées à des cellules d'affichage agencées de manière à former une matrice de cellules destinées à être chacune soumises à une opération de commande d'inversion de polarité en fonction de l'inversion de la polarité de ladite tension de signal, ledit circuit de commande (120) de lignes de signal comprenant : 25
une partie tampon de sortie (200) conçue pour amplifier des données d'entrée, utilisées pour commander lesdites lignes de signal (112), dans le but de générer une tension de signal présentant ladite polarité positive ainsi qu'une tension de signal présentant ladite polarité négative et fournir de façon sélective ladite tension de signal présentant ladite polarité positive et ladite tension de signal présentant ladite polarité négative à une paire constituée de première et deuxième (112m, 112m+1) desdites lignes de signal, ladite partie tampon de sortie (200) utilisant 30
un amplificateur opérationnel de polarité positive (211) comprenant une borne d'entrée inverseuse reliée à une première borne d'entrée (TI1) du circuit de commande de lignes de signal, et un amplificateur opérationnel de polarité négative (221) comprenant une borne d'entrée inverseuse reliée à une deuxième borne d'entrée (TI2) du circuit de commande de lignes de signal, 35
caractérisé par 40
45
50
55

un premier tampon de sortie inverseur (212) conçu pour fournir une tension de signal générée par ledit amplificateur opérationnel de polarité positive (211) et présentant ladite polarité positive ou une tension de signal générée par ledit amplificateur opérationnel de polarité négative (221) et présentant ladite polarité négative à ladite première ligne de signal (112m), un deuxième tampon de sortie inverseur (222) conçu pour fournir la tension de signal présentant ladite polarité positive ou la tension de signal présentant ladite polarité négative à ladite deuxième ligne de signal (112m+1), un premier commutateur (SW231) relié entre la borne de sortie dudit amplificateur opérationnel de polarité positive (211) et une entrée dudit premier tampon de sortie (212), un deuxième commutateur (SW232) relié entre la borne de sortie dudit amplificateur opérationnel de polarité positive (211) et une entrée dudit deuxième tampon de sortie (222), un troisième commutateur (SW233) relié entre la borne de sortie dudit amplificateur opérationnel de polarité négative (221) et une entrée dudit deuxième tampon de sortie (222), un quatrième commutateur (SW234) relié entre la borne de sortie dudit amplificateur opérationnel de polarité négative (221) et une entrée dudit premier tampon de sortie (212), un cinquième commutateur (SW235) relié entre une borne de sortie du premier tampon de sortie (212) et une borne d'entrée non inverseuse dudit amplificateur opérationnel de polarité positive (211), un sixième commutateur (SW236) relié entre une sortie du deuxième tampon de sortie (222) et la borne d'entrée non inverseuse dudit amplificateur opérationnel de polarité positive (211), un septième commutateur (SW237) relié entre une borne de sortie du deuxième tampon de sortie (222) et une borne d'entrée non inverseuse dudit amplificateur opérationnel de polarité négative (221), et un huitième commutateur (SW238) relié entre une borne de sortie du premier tampon de sortie (212) et la borne d'entrée non inverseuse dudit amplificateur opérationnel de polarité négative (221), les commutateurs étant aptes à être commandés de manière à ce que un premier groupe de commutateurs comprenant les premier, troisième, cinquième et septième commutateurs (SW231, SW233, SW235, SW237) occupent une position fermée lorsqu'un deuxième groupe de commutateurs comprenant les deuxième, quatrième, sixième et huitième commutateurs (SW232, SW234, SW236, SW238) occupent une position ouverte, et inversement.

2. Circuit de commande (120) de lignes de signal selon la revendication 1, dans lequel :

ledit amplificateur opérationnel de polarité positive (211) comprend un amplificateur différentiel comportant une paire de transistors d'un premier type de conduction ;
ledit amplificateur opérationnel de polarité négative (221) comprend un amplificateur différentiel comportant une paire de transistors d'un deuxième type de conduction ; et
chacun desdits premier et deuxième tampons de sortie (212, 222) comprend une fonction de fonctionnement push-pull en classe AB.

3. Circuit de commande de lignes de signal selon la revendication 2, dans lequel
chacun desdits premier et deuxième tampons de sortie (212, 222) comprend deux bornes d'entrée.

4. Dispositif d'affichage, comprenant :

une partie affichage (110) comprenant des cellules d'affichage agencées de manière à former une matrice de cellules destinées à être chacune soumises à une opération de commande d'inversion de polarité ; et
un circuit de commande (120) de lignes de signal selon l'une quelconque des revendications précédentes, le circuit de commande (120) de lignes de signal étant conçu pour fournir ladite tension de signal présentant la polarité positive ou négative à chacune d'une pluralité de lignes de signal reliées auxdites cellules d'affichage.

5. Matériel électronique, comportant le dispositif d'affichage selon la revendication 4.

6. Procédé pour faire fonctionner un circuit de commande (120) de lignes de signal conçu pour fournir une tension de signal présentant une polarité positive ou négative à chacune d'une pluralité de lignes de signal (112) reliées à des cellules d'affichage agencées de manière à former une matrice de cellules destinées à être chacune soumises à une opération de commande d'inversion de polarité en fonction de l'inversion de la polarité de ladite tension de signal, ledit circuit de commande (120) de lignes de signal comprenant :

une partie tampon de sortie (200) conçue pour amplifier des données d'entrée, utilisées pour commander lesdites lignes de signal (112), dans le but de générer une tension de signal présentant ladite polarité positive ainsi qu'une tension de signal présentant ladite polarité négative et fournir de façon sélective ladite tension de signal présentant ladite polarité positive et ladite ten-

sion de signal présentant ladite polarité négative à une paire constituée de première et deuxième (112m, 112m+1) desdites lignes de signal, ladite partie tampon de sortie (200) utilisant un amplificateur opérationnel de polarité positive (211) comprenant une borne d'entrée inverseuse reliée à une première borne d'entrée (T11) du circuit de commande de lignes de signal, et un amplificateur opérationnel de polarité négative (221) comprenant une borne d'entrée inverseuse reliée à une deuxième borne d'entrée (T12) du circuit de commande de lignes de signal,
un premier tampon de sortie inverseur (212) conçu pour fournir une tension de signal générée par ledit amplificateur opérationnel de polarité positive (211) et présentant ladite polarité positive ou une tension de signal générée par ledit amplificateur opérationnel de polarité négative (221) et présentant ladite polarité négative à ladite première ligne de signal (112m),
un deuxième tampon de sortie inverseur (222) conçu pour fournir la tension de signal présentant ladite polarité positive ou la tension de signal présentant ladite polarité négative à ladite deuxième ligne de signal (112m+1),
un premier commutateur (SW231) relié entre la borne de sortie dudit amplificateur opérationnel de polarité positive (211) et une entrée dudit premier tampon de sortie (212),
un deuxième commutateur (SW232) relié entre la borne de sortie dudit amplificateur opérationnel de polarité positive (211) et une entrée dudit deuxième tampon de sortie (222),
un troisième commutateur (SW233) relié entre la borne de sortie dudit amplificateur opérationnel de polarité négative (221) et une entrée dudit deuxième tampon de sortie (222),
un quatrième commutateur (SW234) relié entre la borne de sortie dudit amplificateur opérationnel de polarité négative (221) et une entrée dudit premier tampon de sortie (212),
un cinquième commutateur (SW235) relié entre une borne de sortie du premier tampon de sortie (212) et une borne d'entrée non inverseuse dudit amplificateur opérationnel de polarité positive (211),
un sixième commutateur (SW236) relié entre une sortie du deuxième tampon de sortie (222) et la borne d'entrée non inverseuse dudit amplificateur opérationnel de polarité positive (211),
un septième commutateur (SW237) relié entre une borne de sortie du deuxième tampon de sortie (222) et une borne d'entrée non inverseuse dudit amplificateur opérationnel de polarité négative (221), et
un huitième commutateur (SW238) relié entre une borne de sortie du premier tampon de sortie

(212) et la borne d'entrée non inverseuse dudit amplificateur opérationnel de polarité négative (221), les commutateurs étant aptes à être commandés de manière à ce que

un premier groupe de commutateurs comprenant les premier, troisième, cinquième et septième commutateurs (SW231, SW233, SW235, SW237) occupent une position fermée lorsqu'un deuxième groupe de commutateurs comprenant les deuxième, quatrième, sixième et huitième commutateurs (SW232, SW234, SW236, SW238) occupent une position ouverte, et inversement,

dans un premier mode, lequel groupe (230) de commutateurs

fournit une tension de signal générée par ledit amplificateur opérationnel de polarité positive (211) présentant la polarité positive audit premier tampon de sortie (212),

ramène un signal fourni en sortie par ledit premier tampon de sortie (212) à l'entrée dudit amplificateur opérationnel de polarité positive (211),

fournit une tension de signal générée par ledit amplificateur opérationnel de polarité négative (221) présentant la polarité négative audit deuxième tampon de sortie (222), et

ramène un signal fourni en sortie par ledit deuxième tampon de sortie (222) à l'entrée dudit amplificateur opérationnel de polarité négative (221),

alors que dans un deuxième mode, lequel groupe de commutateurs

fournit une tension de signal générée par ledit amplificateur opérationnel de polarité positive (211) présentant la polarité positive audit deuxième tampon de sortie (222),

ramène un signal fourni en sortie par ledit deuxième tampon de sortie (222) à l'entrée dudit amplificateur opérationnel de polarité positive (211),

fournit une tension de signal générée par ledit amplificateur opérationnel de polarité négative (221) présentant la polarité négative audit premier tampon de sortie (212), et

ramène un signal fourni en sortie par ledit premier tampon de sortie (212) à l'entrée dudit amplificateur opérationnel de polarité négative (221).

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FIG. 1

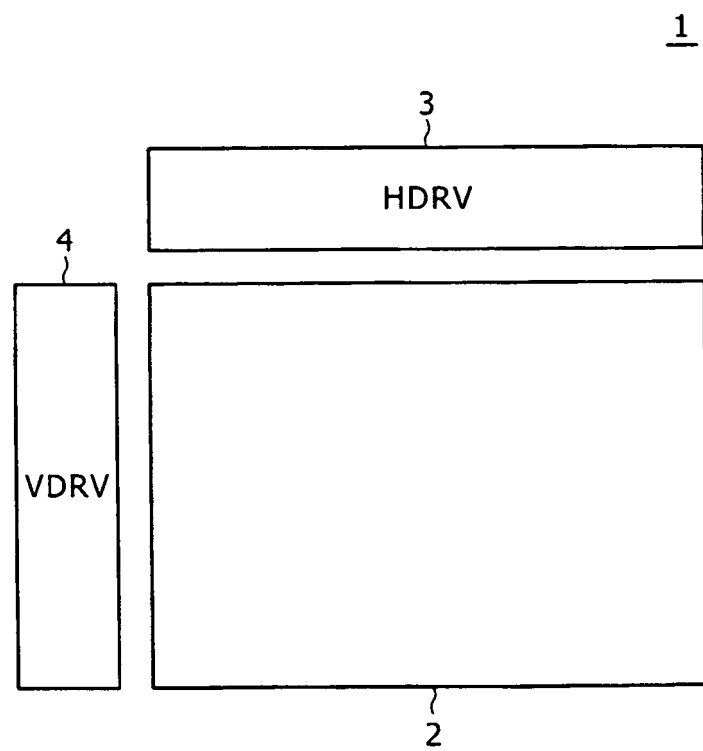


FIG. 2

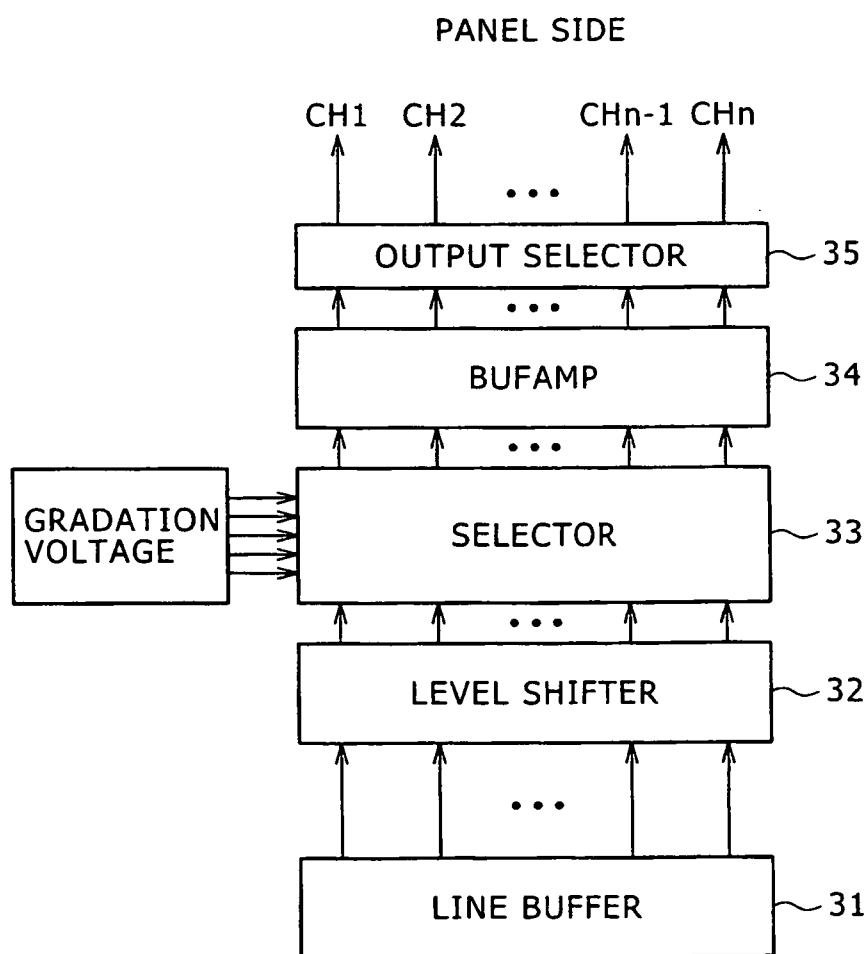


FIG. 3

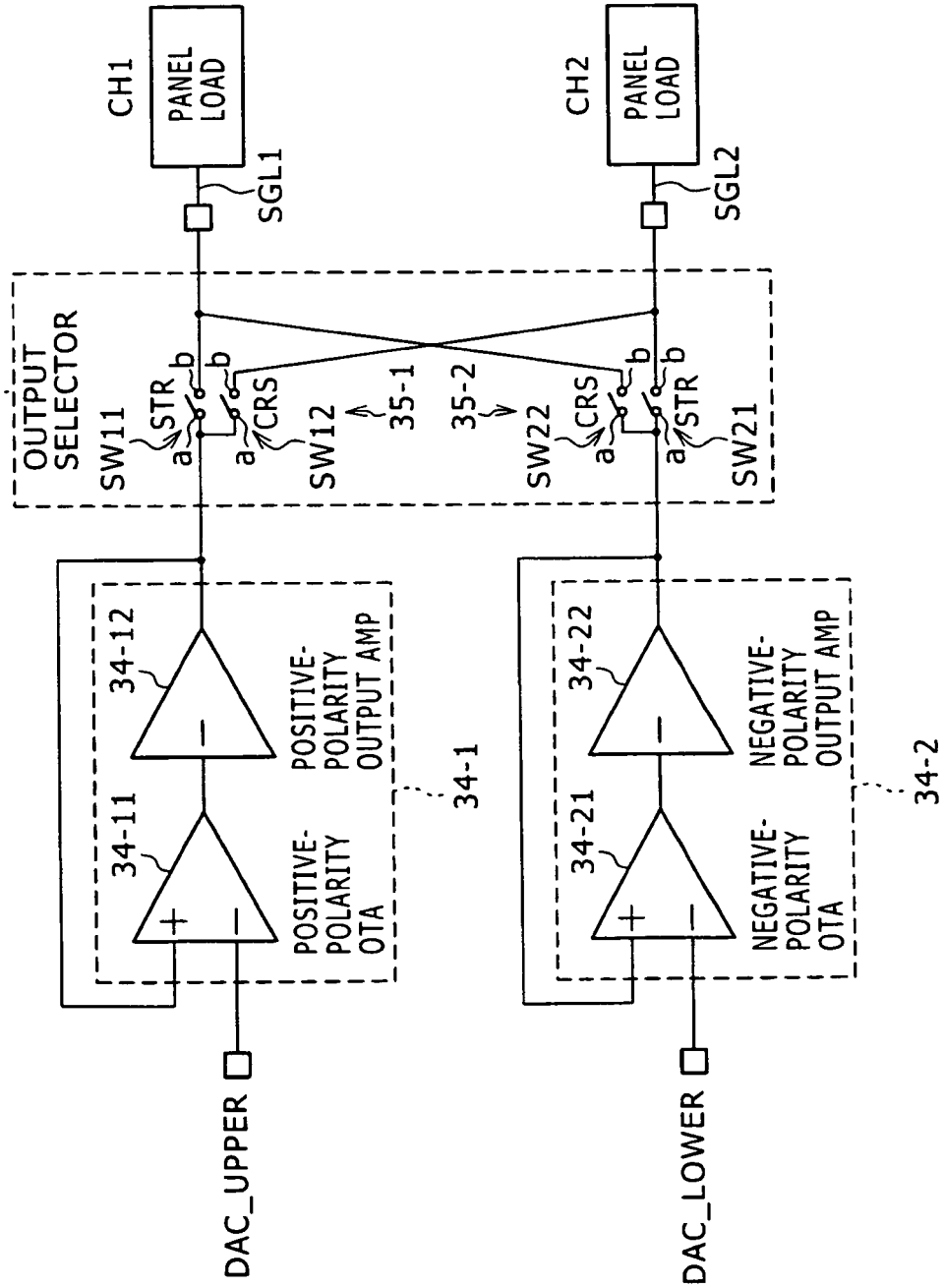


FIG. 4

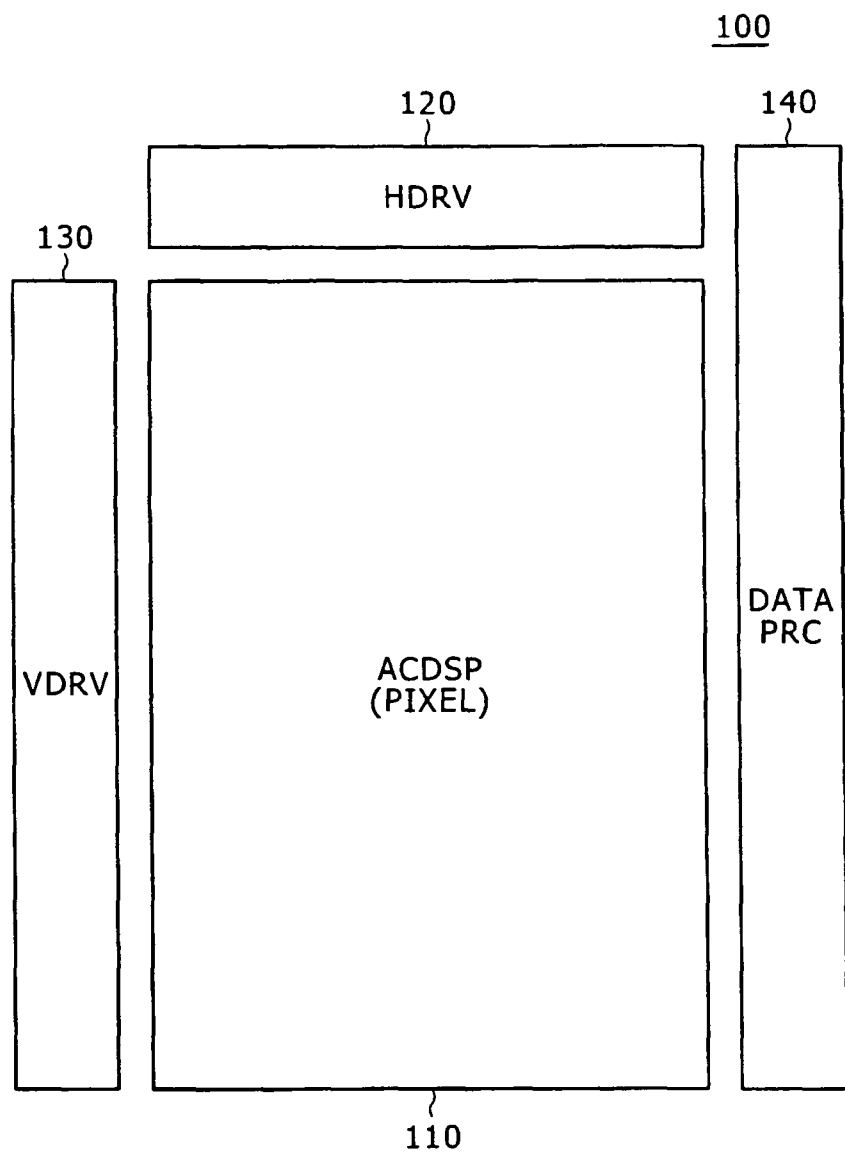


FIG. 5

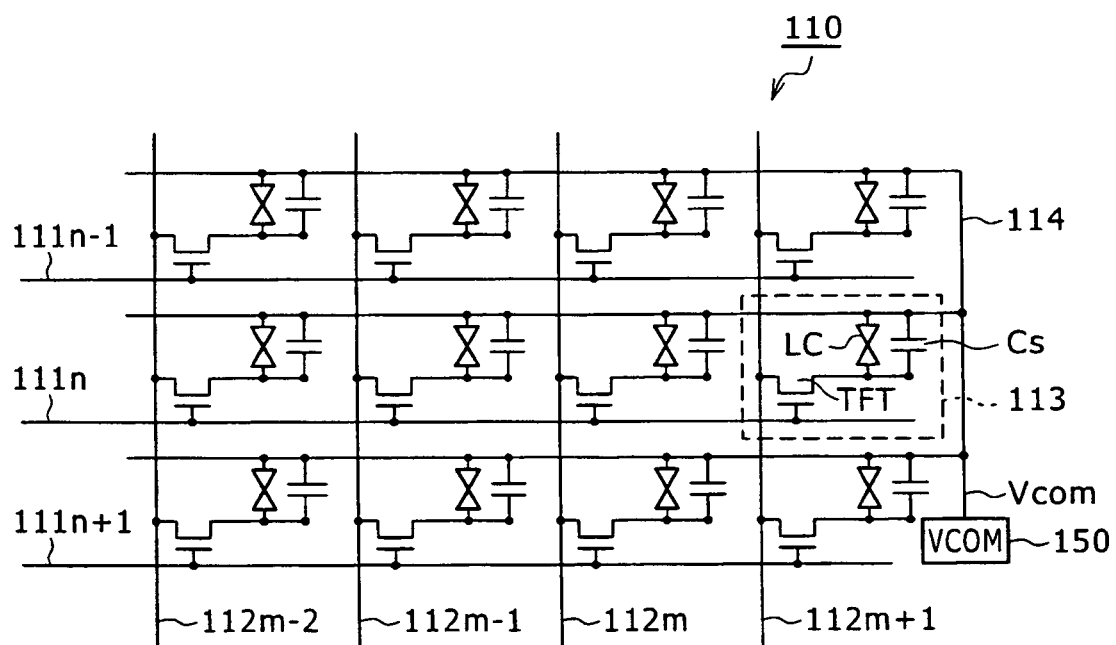


FIG. 6

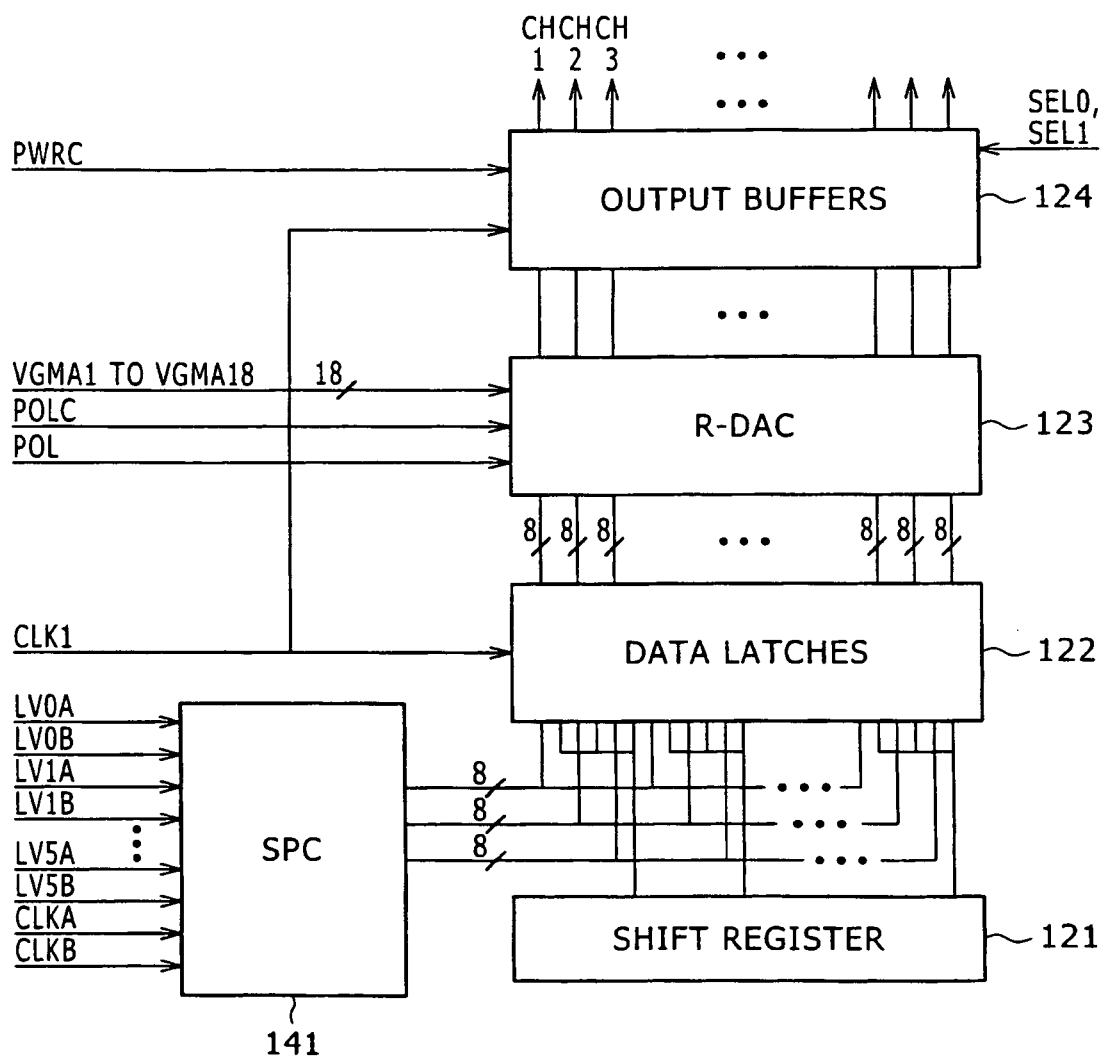


FIG. 7

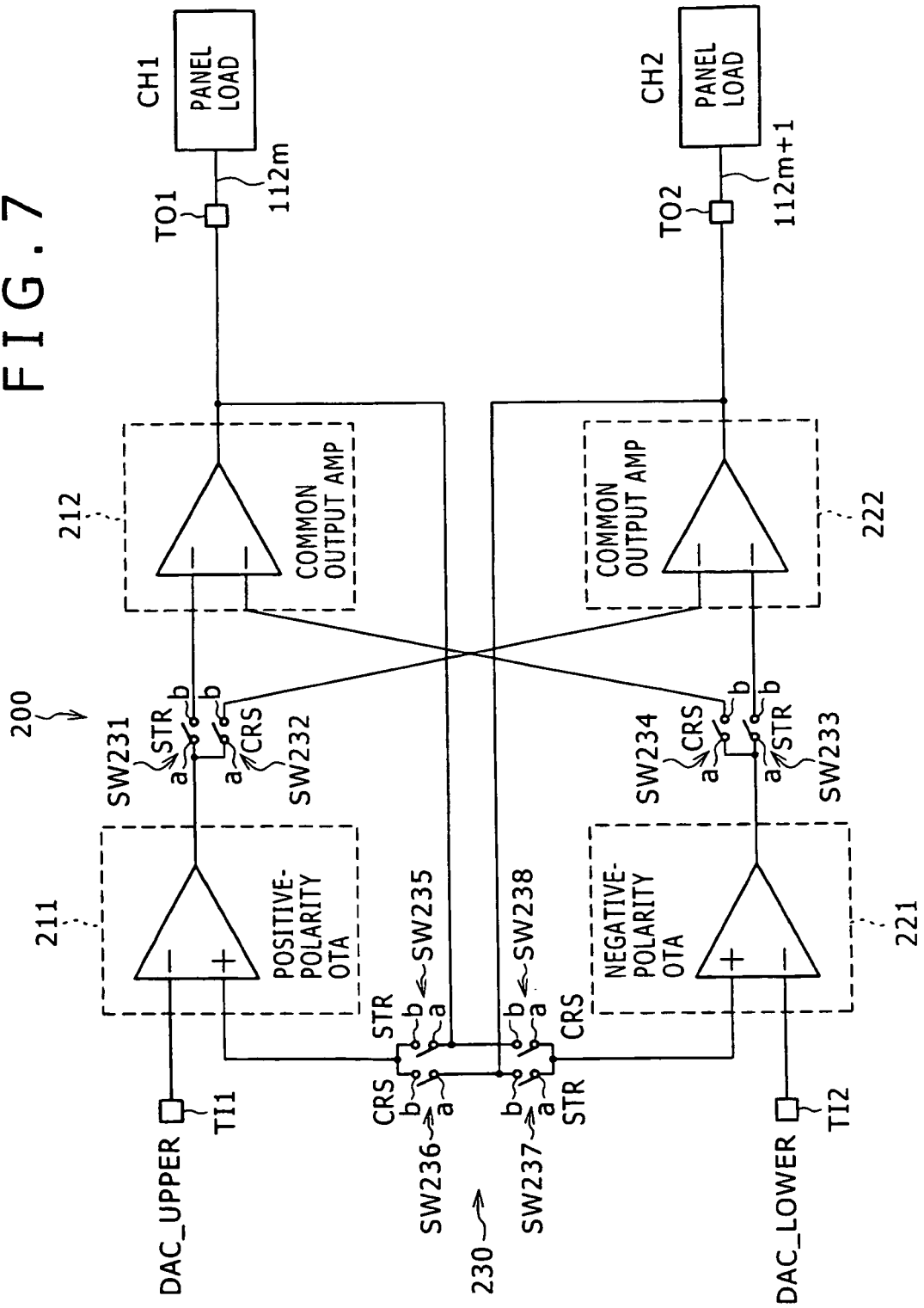
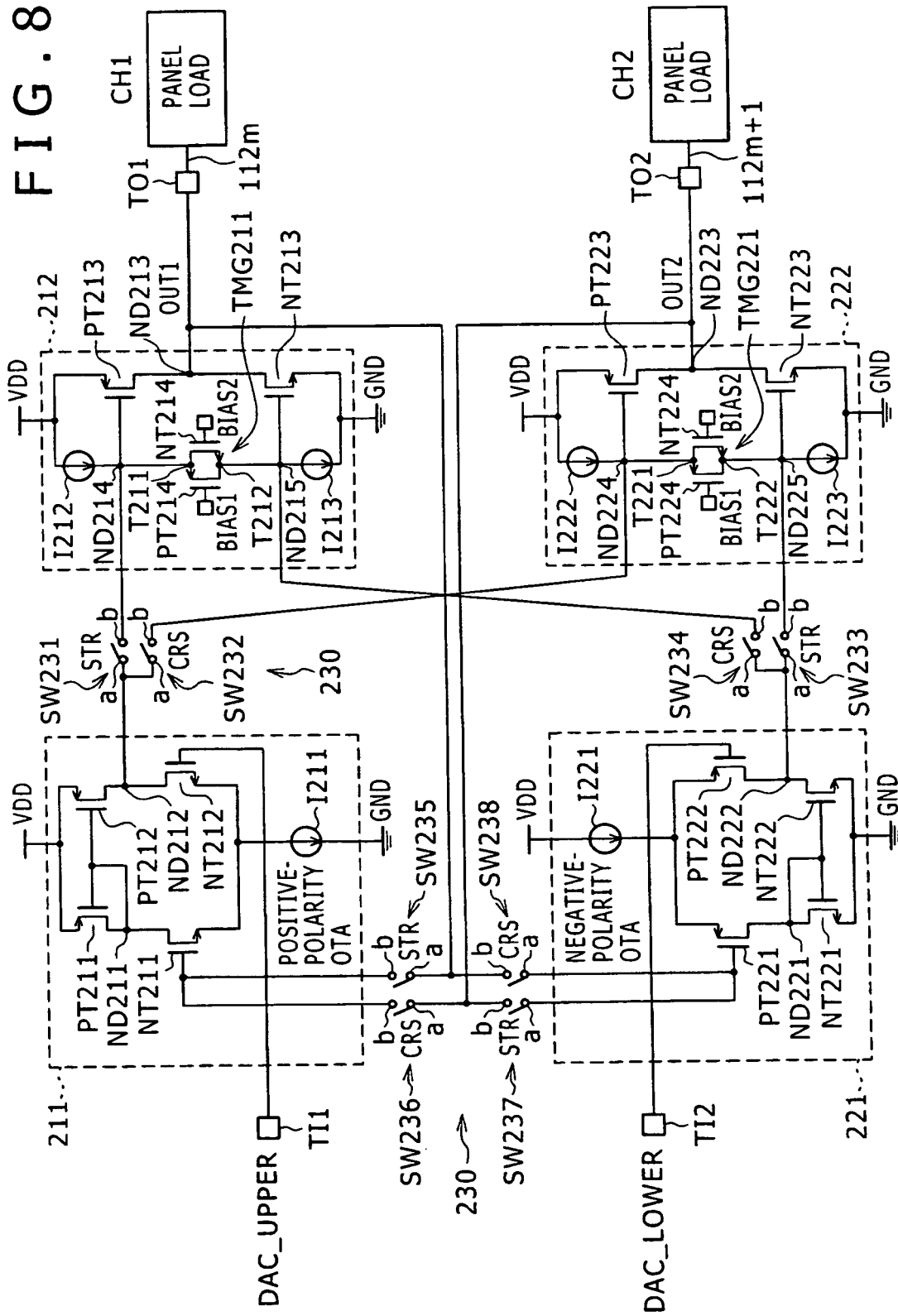


FIG. 8



TIMING CHARTS

FIG. 9A

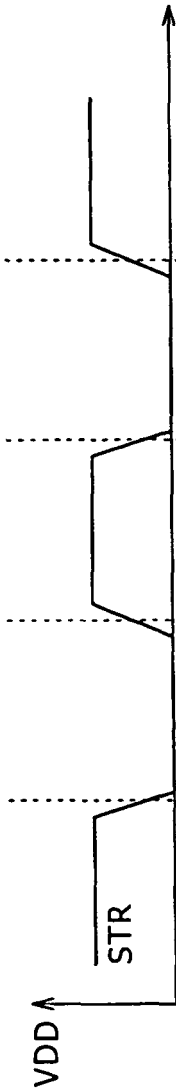


FIG. 9B



FIG. 9C

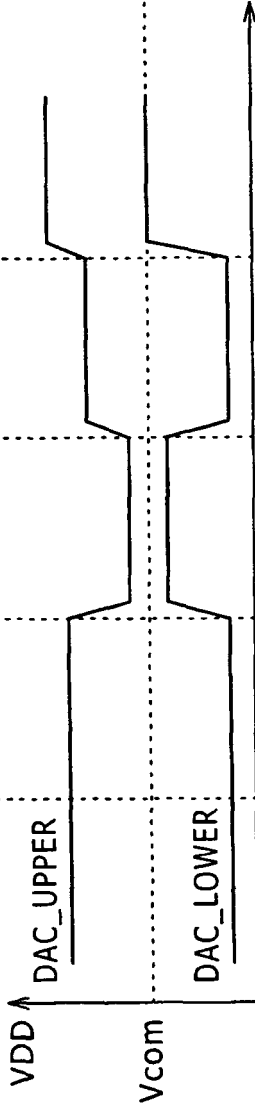
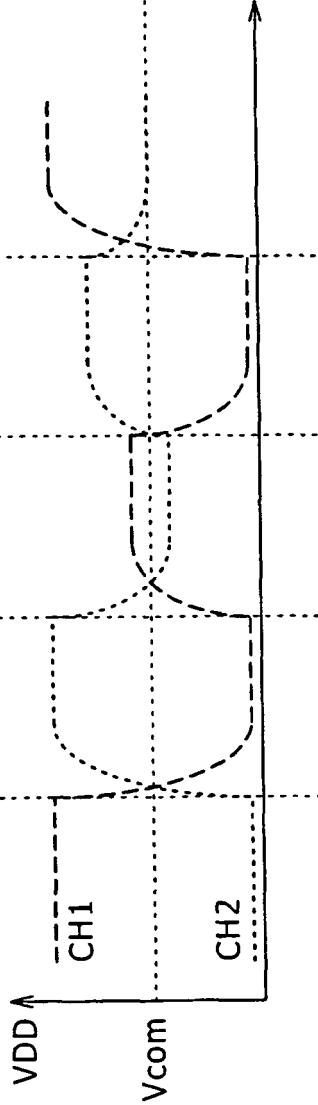


FIG. 9D



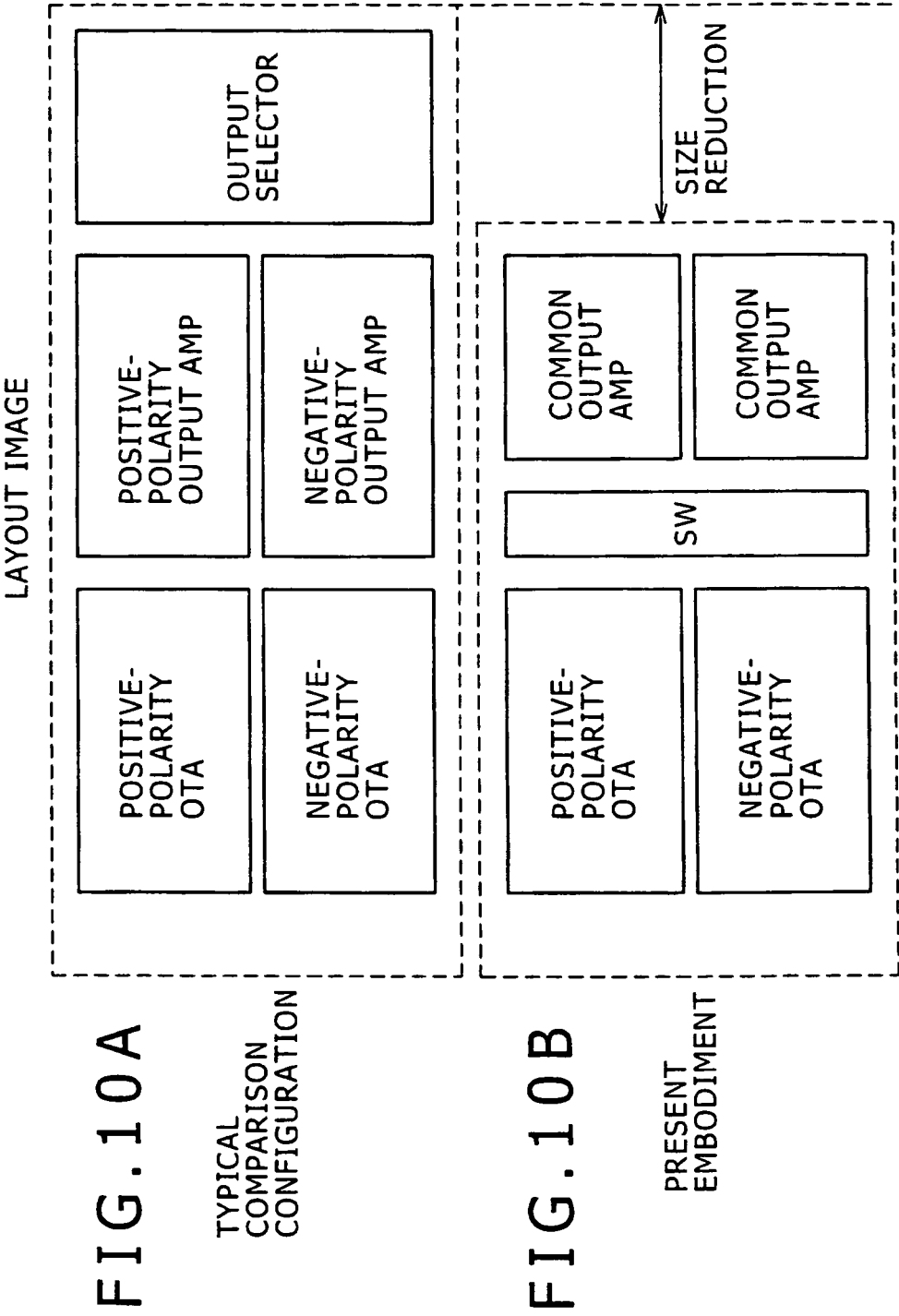
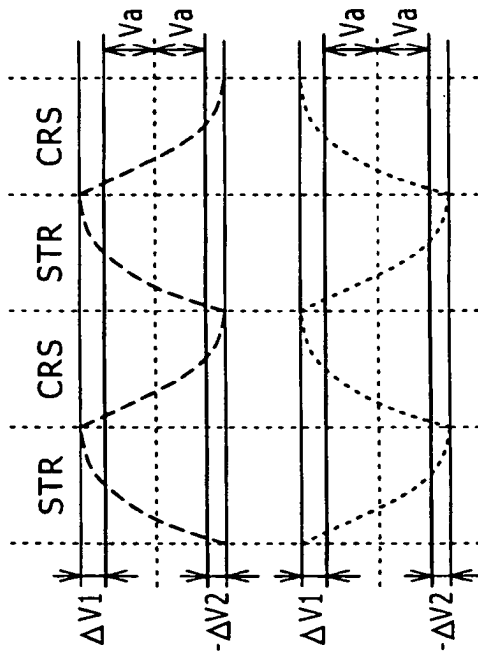
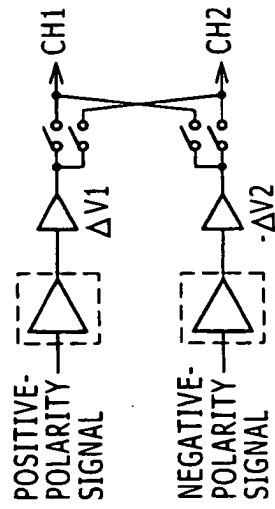


FIG. 11A

COMPARISON CONFIGURATION

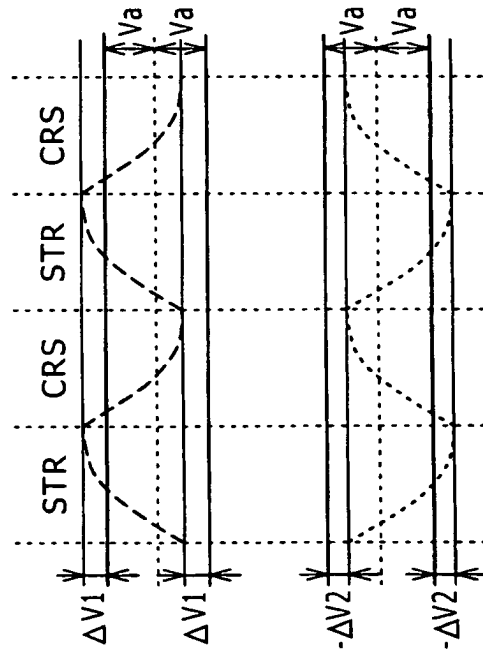
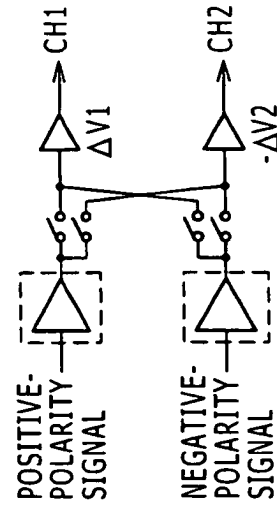


COMPARISON CONFIGURATION
THE OFFSET VOLTAGES CANNOT
BE CANCELLED SO THAT THE
LUMINANCE IS SHIFTED FROM
THE IDEAL VALUE.

AN AVERAGE VOLTAGE RELATIVE
TO THE CENTER LEVEL OF V_{com}
IS $V_a + (\Delta V1 + \Delta V2)/2$.

FIG. 11B

THIS EMBODIMENT



THIS EMBODIMENT
THE POLARITY INVERSION
CAUSES THE OFFSET VOLTAGES
FOR 2 FRAMES TO CANCEL EACH
OTHER.

AN AVERAGE VOLTAGE RELATIVE
TO THE CENTER LEVEL OF V_{com}
IS V_a INDEPENDENT OF THE OFFSET
VALUES $\Delta V1$ AND $\Delta V2$.

FIG.12

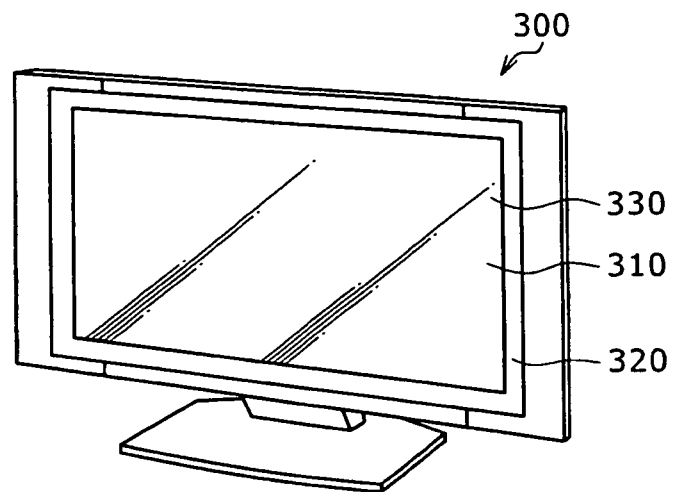


FIG.13A

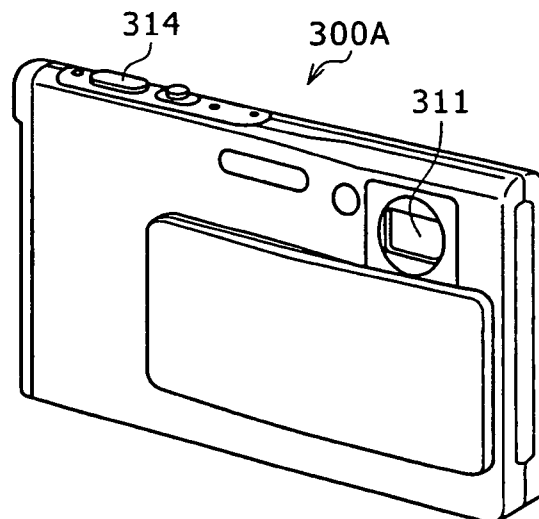


FIG.13B

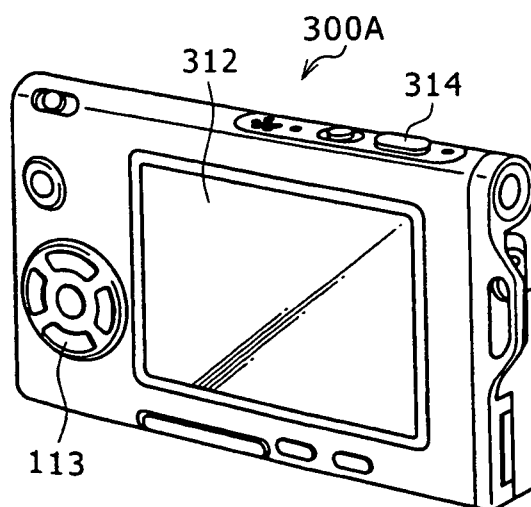


FIG. 14

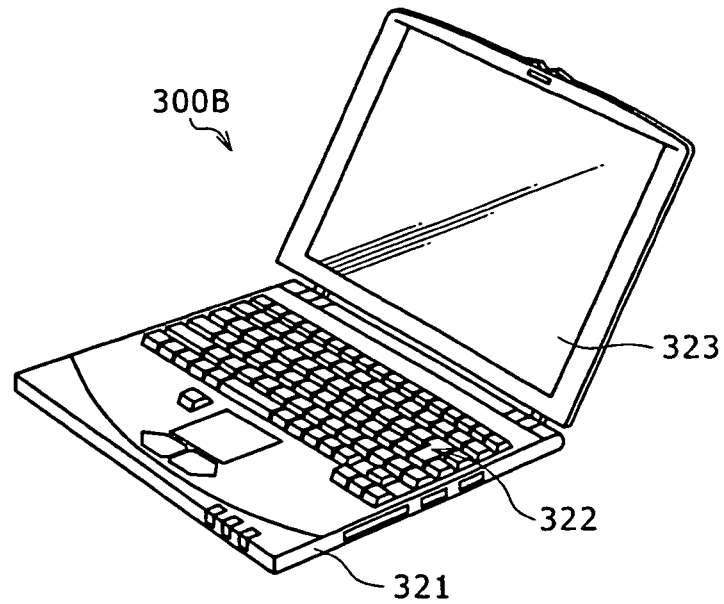


FIG. 15

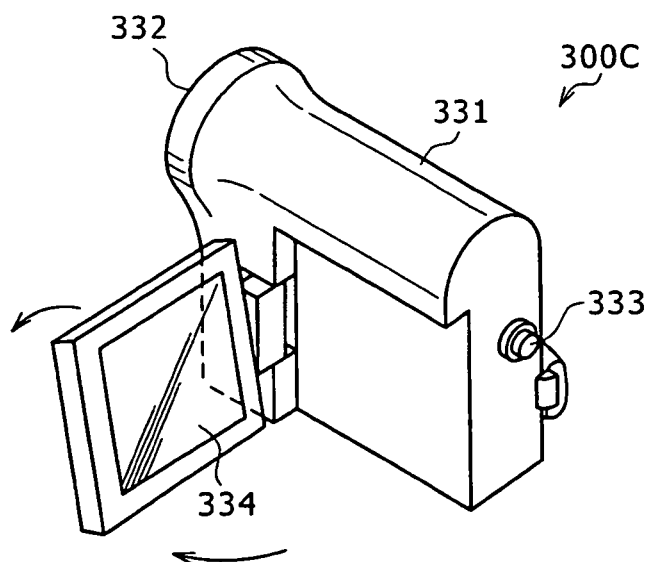


FIG. 16A

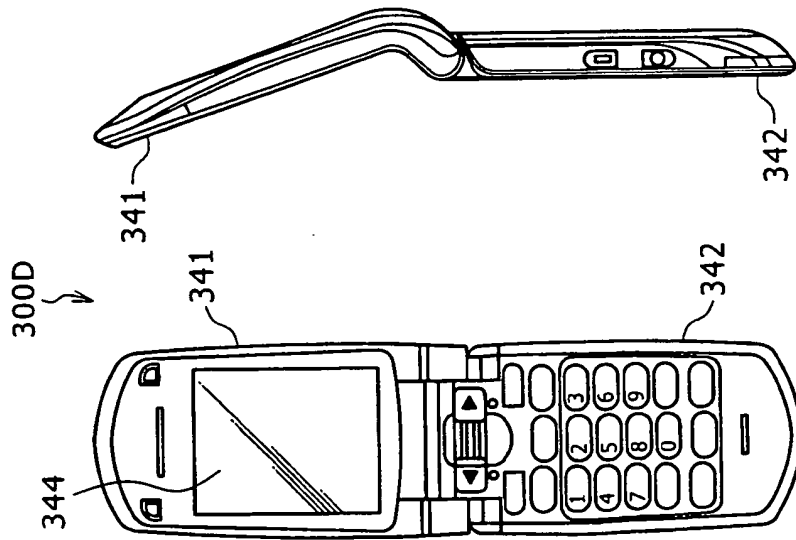


FIG. 16F

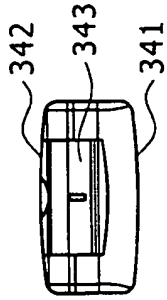


FIG. 16D

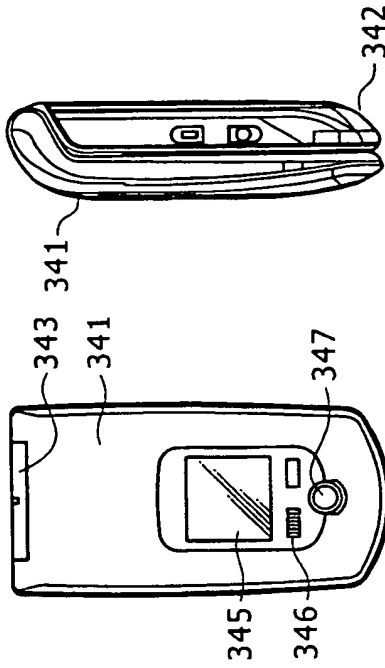
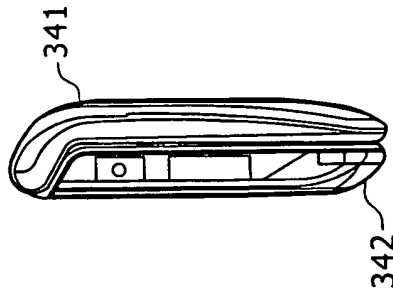
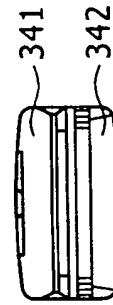


FIG. 16G



REFERENCES CITED IN THE DESCRIPTION

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Patent documents cited in the description

- JP 2007338996 A [0001]
- JP HEI10153986 B [0014]
- US 6304241 B1 [0015]

专利名称(译)	信号线驱动电路，显示装置和电子设备		
公开(公告)号	EP2075788B1	公开(公告)日	2013-06-19
申请号	EP2008022293	申请日	2008-12-22
[标]申请(专利权)人(译)	索尼公司		
申请(专利权)人(译)	索尼公司		
当前申请(专利权)人(译)	索尼公司		
[标]发明人	SUZUKI TOSHIO		
发明人	SUZUKI, TOSHIO		
IPC分类号	G09G3/36		
CPC分类号	G09G3/3688 G09G3/3614 G09G2310/0281 G09G2310/0297		
优先权	2007338996 2007-12-28 JP		
其他公开文献	EP2075788A3 EP2075788A2		
外部链接	Espacenet		

摘要(译)

一种信号线驱动电路，被配置为向连接到显示单元（例如液晶显示器）的每条信号线提供具有正极性或负极性的信号电压，布置成形成矩阵。根据信号电压的极性的反转，对每个单元进行极性反转驱动操作。信号线驱动电路包括输出缓冲部分，其中输出缓冲部分采用正极性运算放大器，负极性运算放大器，第一输出缓冲器，第二输出缓冲器和开关组。

