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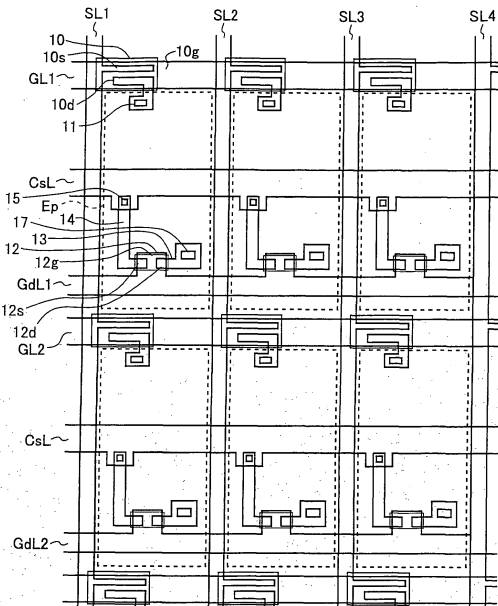
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(54) **ACTIVE MATRIX SUBSTRATE, AND DISPLAY DEVICE HAVING THE SUBSTRATE**

(57) Intended is to enable a display device to make the display impulse and to improve the charging characteristics of pixel capacitances while suppressing the complexity of a drive circuit, etc. and the increase in the operating frequency. In an active matrix substrate of a liquid crystal display device, each pixel electrode (Ep) is connected via a pixel TFT (10) to a source line (SLi) and via a discharge TFT (12) to a storage capacitance line (CsL). For one frame period, liquid crystal capacitances (Clc), as formed by the pixel electrodes (Ep) corresponding to each display line and a common electrode (Ec), are charged, when the pixel TFTs(10) are turned ON by a pixel scanning signal (G(j)) on a pixel gate line (GLj), and are then discharged when the discharging TFTs (12) are turned ON by a discharge scanning signal (Gd(j)) on a discharge gate line (GdLj). The source lines (SLi) are subjected to a 2H dot inversion drive but to a charge share for every horizontal period.

Fig. 1



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Description

Technical Field

[0001] The present invention relates to an active matrix substrate using switching elements such as thin film transistors, and an active matrix-type display device such as a liquid crystal display device provided with the active matrix substrate.

Background Art

[0002] An active matrix substrate is widely used in an active matrix-type display device such as a liquid crystal display device and an EL (Electroluminescence) display device. For example, the main part of the active matrix-type liquid crystal display device is composed of a liquid crystal panel and drive circuits thereof, and the liquid crystal panel typically includes: the active matrix substrate having pixel circuits arranged in a matrix, each of which contains a thin film transistors (hereinafter, simply abbreviated as "TFT") as a switching element and a pixel electrode and the like; an opposite substrate of a transparent insulating substrate such as a glass plate having an opposite electrode and an alignment film which are sequentially deposited over the entire surface thereof; a liquid crystal layer sandwiched between the both substrates; and polarizer plates attached to the outside surfaces of the both substrates, respectively.

[0003] Fig. 23 is a plan view showing a structure of a conventional active matrix substrate 700 used for the liquid crystal display device as described above, and illustrates a pattern configuration in a part corresponding to one pixel. The active matrix substrate 700 includes a plurality of data signal lines 715, a plurality of scanning signal lines 716 intersecting the plurality of data signal lines 715, TFTs 712 as the switching elements, which are formed in the neighborhoods of the intersections of the plurality of data signal lines 715 and the plurality of scanning signal lines 716, respectively, and pixel electrodes 717. The scanning signal line 716 works also as a gate electrode of the TFT 712, a source electrode 719 of the TFT 712 is connected to the data signal line 715, and a drain electrode 708 thereof is connected to the pixel electrode 717 via a drain extraction electrode 707. A hole is formed in an insulating film disposed between the drain extraction electrode 707 and the pixel electrode 717 and thereby a contact hole 710 is formed connecting the drain extraction electrode 707 and the pixel electrode 717. The pixel electrode 717 is a transparent electrode made of ITO (Indium Tin Oxide) or the like and transmits light from the back of a liquid crystal panel including the active matrix substrate 700 (light from a backlight).

[0004] In the active matrix substrate 700, the TFT 712 is turned on by a gate-on voltage of a scanning signal applied to the scanning signal line 716 (conductive state between the source electrode 719 and the drain electrode 708), a data signal applied to the data signal line

715 in this state is written into a pixel capacitance (capacitance formed by the pixel electrode 717 and the opposite electrode) via the source electrode 719, the drain electrode 708, and the drain extraction electrode 707.

5 Note that, in this active matrix substrate 700, a storage capacitance line 718 is formed along the scanning signal line 716 and this storage capacitance line 718 has a function for avoiding self discharge of the liquid crystal layer during an OFF period in the TFT 712, and the like.

10 **[0005]** Meanwhile, in an impulse type display device such as a CRT (Cathode Ray Tube), when focusing on each pixel, an ON period when an image is displayed and an OFF period when an image is not displayed are repeated alternately. For example, when a moving image is displayed, the OFF period is inserted when an image of one screen is rewritten and a residual image of a moving object is not produced in human visual sense. Thereby, the object and the background thereof can be distinguished clearly and the moving image can be viewed without uncomfortable feeling.

20 **[0006]** On the other hand, in a hold type display device such as the liquid crystal display device using the active matrix substrate as described above, luminance of each pixel is determined by a voltage retained in its pixel capacitance, and the retained voltage in each of the pixel capacitances is maintained during one frame period, once rewritten. In such a hold type display device, the voltage to be retained in the pixel capacitance as pixel data is retained, once written, until the pixel capacitance is rewritten next, and thereby the image in each frame is temporally close to the image one frame prior thereto. This produces the residual image of the moving object in the human visual sense when the moving image is displayed. For example, as shown in Fig. 22, a residual image AI is produced as an image OI representing a moving object is dragging a tail (hereinafter, called "tail-dragging residual image").

35 **[0007]** Since such a tail-dragging residual image is produced in the moving image display of the hold type display device such as the active matrix-type liquid crystal display device or the like, conventionally the impulse type display device has been employed typically for a display such as a television or the like which mainly displays the moving image. However, the display such as the television or the like recently has been required strongly to have a lighter weight and a slimmer body and the hold type display device such as the liquid crystal display device which is easily to have the lighter weight and the slimmer body has been employed rapidly for such a display.

Patent document 1: Japanese Patent Application Laid-Open No. H4-309995

Patent document 2: Japanese Patent Application Laid-Open No. H5-119346

Patent document 3: Japanese Patent Application Laid-Open No. 2003-255912

Patent document 4: Japanese Patent Application

Laid-Open No. 2003-66918
 Patent document 5: Japanese Patent Application
 Laid-Open No. H9-243998
 Patent document 6: Japanese Patent Application
 Laid-Open No. 2004-61590

Disclosure of the Invention

Problems to be Solved by the Invention

[0008] As a method for improving the above tail-dragging residual image in the hold type display device such as the active matrix-type liquid crystal display device or the like, there is known a method to cause the display of the liquid crystal display device to be an (pseudo) impulse type, for example, by inserting a period of black display during one frame period (hereinafter, called "black insertion") (e.g., Patent document 4 (Japanese Patent Application Laid-Open No. 2003-66918, which corresponds to US Patent No. 7,161,576)). However, when the impulse type is tried to be realized by the conventional method in the active matrix-type liquid crystal display device which is the hold type display device, a drive circuit and the like becomes complicated for the black insertion, and also an operating frequency of the drive circuit increases and a time to be secured for charging the pixel capacitance becomes shorter.

[0009] For solving such problems, there has been proposed a configuration of the active matrix substrate in the liquid crystal display device or the like to provide two switching elements for each pixel formation portion and to provide a second signal line separately for transmitting black data in addition to a first signal line for transmitting pixel data to the pixel formation portion (e.g., Patent document 3 (Japanese Patent Application Laid-Open No. 2003-255912)). However, in such a configuration, not only a structure of the active matrix substrate becomes complicated, but also a power supply is required for a signal applied to the second signal line for the black display, resulting in a more complicated drive circuit and increased power consumption.

[0010] Further, the active matrix-type liquid crystal display device has been improved recently in a resolution thereof, and there is a trend that a shorter charging time can be secured for writing the pixel data into the pixel capacitance. When the charging time becomes shorter, there arises a problem that right pixel data cannot be written into the pixel capacitance because of charging shortage.

[0011] Meanwhile, in the liquid crystal display device using a dot inversion drive method which inverts a polarity of the data signal every two horizontal periods (hereinafter, called "2H dot inversion drive method"), there is a case of employing a charge sharing method to short between neighboring data signal lines in the polarity inversion of the data signal for reducing power consumption (e.g., Patent document 5 (Japanese Patent Application Laid-Open No. H9-243998)). In this case, the charge

amount of the pixel capacitance may be different between the two lines of a polarity inversion unit and sometimes a horizontal irregularity is viewed in a line. On the other hand, there is proposed a method to make the charging characteristic uniform by setting the data signal to exhibit a certain intermediate potential between a positive polarity and a negative polarity in a blanking period every one horizontal period (Patent document 6 (Japanese Patent Application Laid-Open Publication No. 2004-61590, which corresponds to US Patent Application No. 2004/0017344)).

[0012] However, when it becomes difficult to secure a sufficient charging time or a sufficient charge sharing period because of the progress in the higher resolution or the operating frequency increase for the impulse type drive, even application of such a method cannot satisfactorily eliminates the charge amount difference in the pixel capacitance between the above two lines of the polarity inversion unit and there remains a possibility that the line shape horizontal irregularity is viewed. Further, when each data signal line potential does not reach a source center potential or a common potential because a sufficient charge sharing period cannot be secured, this also becomes a factor to worsen the charging shortage.

Such a point, that the charging shortage becomes a more serious problem as the resolution improvement or the operating frequency increase, is similar for the liquid crystal display device using a 1H dot inversion drive method.

[0013] Meanwhile, Patent document 2 (Japanese Patent Application Laid-Open No. H5-119346) discloses the liquid crystal display device, which is the active matrix-type liquid crystal display device having an auxiliary capacitance electrode extended under the pixel electrode, and includes an auxiliary switching element which has a source connected to the pixel electrode, a gate connected to a gate line (scanning signal line) of the previous stage, and a drain connected to the auxiliary capacitance electrode, wherein the pixel electrode is configured to be charged up to a potential of the auxiliary capacitance electrode (common potential) by a gate signal (scanning signal) of the gate line in the previous stage. This configuration can suppress the charging shortage of the pixel capacitance caused by the increase in the pixel numbers, since the pixel electrode is charged up to the common potential via the auxiliary switching element before the pixel electrode is charged up to a target potential (potential corresponding to a pixel value of a display image). However, the patent publication does not disclose anything about the impulse type display, for example, for eliminating the tail-dragging residual image in the moving image display. Further, in the configuration disclosed in the publication, the pixel electrode is charged up to the common potential via the auxiliary switching element immediately (at most approximately one horizontal period) before the charge up to the target potential of the pixel electrode, and thereby the auxiliary switching element cannot be utilized for the impulse type display.

[0014] Accordingly, an object of the present invention

is to provide an active matrix substrate and a display device having the same, which can realize the impulse type display while suppressing the complication of the drive circuit or the increase in the operating frequency, and also improve the charging characteristic of the pixel capacitance.

Means for Solving the Problems

[0015] A first aspect of the present invention provides an active matrix substrate including:

a plurality of data signal lines;
 a plurality of pixel scanning signal lines intersecting the plurality of data signal lines;
 a pixel switching element provided corresponding to each of intersections of the plurality of data signal lines and the plurality of pixel scanning signal lines, and turned on and off by the pixel scanning signal line passing through the corresponding intersection;
 a pixel electrode connected to the data signal line passing through the intersection corresponding to the pixel switching element via the pixel switching element;
 a storage capacitance line disposed so as to form a predetermined capacitance with each of the pixel electrodes;
 a plurality of discharge scanning signal lines corresponding to the plurality of pixel scanning signal lines, respectively; and
 a discharge switching element provided corresponding to each of the pixel electrodes, and turned on and off by the discharge scanning signal line which corresponds to the pixel scanning signal line for turning on and off the pixel switching element connected to the corresponding pixel electrode,
 each of the pixel electrodes being connected to the storage capacitance line via the corresponding discharge switching element.

[0016] A second aspect of the present invention provides the active matrix substrate according to the first aspect of the present invention, wherein;
 the storage capacitance line has an extended portion extending in a direction along the data signal line; and
 the discharge switching element is a thin film transistor having a drain electrode and a source electrode, the drain electrode being connected to the pixel electrode corresponding to the discharge switching element, and the source electrode being connected to the extended portion via a predetermined source extraction electrode.

[0017] A third aspect of the present invention provides the active matrix substrate according to the second aspect of the present invention, wherein the extended portion and the source extraction electrode form a structure disposed circularly along an edge of the pixel electrode.

[0018] A fourth aspect of the present invention provides the active matrix substrate according to the second

aspect of the present invention, wherein an electrode connected to the source and an electrode connected to the drain of the thin film transistor as the discharge switching element are formed of the same material as that of the data signal line.

[0019] A fifth aspect of the present invention provides the active matrix substrate according to the first aspect of the present invention, wherein the storage capacitance line has a circular portion including a portion which extends in parallel to the data signal line along an edge of the pixel electrode and a portion which extends in parallel to the pixel scanning signal line along the edge of the pixel electrode.

[0020] A sixth aspect of the present invention provides the active matrix substrate according to the first aspect of the present invention, wherein the discharge switching element is disposed so as to overlap an electrode pattern forming the discharge scanning signal line.

[0021] A seventh aspect of the present invention provides the active matrix substrate according to the first aspect of the present invention, wherein the pixel electrode is disposed so as to overlap the discharge scanning signal line.

[0022] An eighth aspect of the present invention provides a normally black mode display device, including;
 an active matrix substrate according to any one of the first to seventh aspects of the present invention;

a common electrode disposed so as to face each of pixel electrodes in the active matrix substrate;

a pixel scanning signal line drive circuit for selectively applying an active signal, which turns on the pixel switching element, to the plurality of pixel scanning signal lines and causing the pixel scanning signal line to which the active signal is applied to be in a selected state such that each of the plurality of pixel scanning signal lines is in the selected state at least once during each frame period;
 a discharge scanning signal line drive circuit for selectively applying an active signal, which turns on the discharge switching element, to the plurality of discharge scanning signal lines and causing the discharge scanning signal line to which the active signal is applied to be in a selected state such that each of the plurality of discharge scanning signal lines is in the selected state during a predetermined period within a period from a first time point when the corresponding pixel scanning signal line changes from the selected state to a non-selected state to a second time point when the corresponding pixel scanning signal line is in the selected state in the next frame period;

a data signal line drive circuit for generating a plurality of data signals to display an image as voltage signals whose polarities are inverted every predetermined number of horizontal periods, and applying the plurality of data signals to the plurality of data signal lines, respectively;

a common potential supply section for providing a predetermined common potential to the common electrode; and

a storage capacitance line potential supply section for

providing a predetermined potential approximately the same as the common potential to the storage capacitance line.

[0023] A ninth aspect of the present invention provides the display device according to the eighth aspect of the present invention, wherein the data signal line drive circuit generates the plurality of data signals so as to invert the voltage polarity thereof every two or more predetermined number of horizontal periods.

[0024] A tenth aspect of the present invention provides the display device according to the eighth aspect of the present invention, wherein the data signal line drive circuit generates the plurality of data signals so as to invert the polarity thereof every predetermined number of the data signal lines, and cuts off the application of the plurality of data signals to the respective plurality of data signal lines and also shorts the plurality of data signal lines with each other, during a predetermined period when the polarities of the plurality of data signals are inverted.

[0025] An eleventh aspect of the present invention provides the display device according to the tenth aspect of the present invention, wherein the data signal line drive circuit cuts off the application of the plurality of data signals to the respective plurality of data signal lines and also shorts the plurality of data signal lines with each other, during a predetermined period every one horizontal period.

[0026] A twelfth aspect of the present invention provides the display device according to the eleventh aspect of the present invention, wherein the data signal line drive circuit provides a predetermined fixed potential to the plurality of data signal lines when the plurality of data signal lines are shorted with each other.

[0027] A thirteenth aspect of the present invention provides the display device according to the twelfth aspect of the present invention, wherein the fixed potential is equal to the predetermined potential.

[0028] A fourteenth aspect of the present invention provides a television receiver, including a display device according to the eighth aspect of the present invention.

[0029] A fifteenth aspect of the present invention provides a drive method of a normally black mode display device provided with an active matrix substrate and a common electrode; the active matrix substrate including a plurality of data signal lines, a plurality of pixel scanning signal lines intersecting the plurality of data signal lines, a pixel switching element which is provided corresponding to each of intersections of the plurality of data signal lines and the plurality of pixel scanning signal lines and turned on and off by the pixel scanning signal line passing through the corresponding intersection, a pixel electrode connected to the data signal line passing through the intersection corresponding to the pixel switching element via the pixel switching element, and a storage capacitance line disposed so as to form a predetermined capacitance with the pixel electrode; the common electrode

being disposed so as to face each of the pixel electrodes in the active matrix substrate; the drive method including:

a pixel scanning signal line drive step of selectively applying an active signal, which turns on the pixel switching element, to the plurality of pixel scanning signal lines and causing the pixel scanning signal line to which the active signal is applied to be in a selected state such that each of the plurality of pixel scanning signal lines is in the selected state at least once during each frame period;

a data signal line drive step of generating a plurality of data signals to display an image as voltage signals whose polarities are inverted every predetermined number of horizontal periods and applying the plurality of data signals to the plurality of data signal lines, respectively;

a common potential supply step of providing a predetermined common potential to the common electrode;

a storage capacitance line potential supply step of providing a predetermined potential approximately equal to the common potential to the storage capacitance line; and

a discharge step of shorting each of the pixel electrodes to the storage capacitance line, wherein the active matrix substrate further includes:

a plurality of discharge scanning signal lines corresponding to the plurality of pixel scanning signal lines, respectively; and

a discharge switching element provided corresponding to each of the pixel electrodes, and turned on and off by the discharge scanning signal line which corresponds to the pixel scanning signal line for turning on and off the pixel switching element connected to the corresponding pixel electrode,

each of the pixel electrodes is connected to the storage capacitance line via the corresponding discharge switching element, and

in the discharge step, an active signal, which turns on the discharge switching element, is selectively applied to the plurality of discharge scanning signal lines and the discharge scanning signal line to which the active signal is applied is caused to be in a selected state, such that each of the plurality of discharge scanning signal lines is in the selected state during a predetermined period within a period from a first time point when the corresponding pixel scanning signal line changes from the selected state to a non-selected state to a second time point when the corresponding pixel scanning signal line is in the selected state in the next frame period.

[0030] Other aspects of the present invention are apparent from the description of the above-described as-

pects and the following embodiments of the present invention, and thus description thereof is not given.

Advantages of the Invention

[0031] According to the first aspect of the present invention, each of the pixel electrodes is connected to the storage capacitance line via the discharge switching element which is turned on and off by the discharge scanning signal line disposed corresponding to each of the pixel scanning signal lines in the active matrix substrate. Accordingly, when the active matrix substrate according to this aspect of the present invention is used in the display device such as the liquid crystal display device, in which the pixel capacitances are charged up to a voltage of image data having a polarity inverted every one frame period, each of the pixel capacitances discharges a charge with the inverse polarity by application of the discharge scanning signal to the discharge scanning signal line before the charge up corresponding to the pixel data writing. Thereby, even when the resolution or the like is increased, the charging shortage is suppressed and the display quality is improved. Further, when the display device uses a normally black mode, a black display period can be inserted for each display line by the application of the discharge scanning signal to each of the discharge scanning signal lines, and thereby the display quality of the moving image can be improved by employing the impulse type display while suppressing the complication of the drive circuit or the increase of the operating frequency.

[0032] According to the second aspect of the present invention, the storage capacitance line has the extended portion which extends in parallel to the data signal line, and thereby influence of potential changes in the other electrodes on a pixel electrode potential can be suppressed.

[0033] According to the third aspect of the present invention, the extended portion of the storage capacitance line and the source extraction electrode in the thin film transistor as the discharge switching element constitute a structure disposed circularly along the edge of the pixel electrode, and thereby the influence of the potential change in the data signal line, the pixel scanning signal line, or the discharge scanning signal line on the pixel electrode potential can be suppressed.

[0034] According to the fourth aspect of the present invention, the electrode connected to the source and the electrode connected to the drain in the thin film transistor of the discharge switching element are formed using the same material as that of the data signal line, and thereby these electrodes can be formed in a data signal line formation process and need not be formed separately.

[0035] According to the fifth aspect of the present invention, the storage capacitance line has a portion extending in parallel to the data signal line along the edge of the pixel electrode and a portion extending in parallel to the pixel scanning signal line along the edge of the

pixel electrode, and thereby the influence of the potential change in the data signal line or the pixel scanning signal line on the pixel electrode potential can be suppressed. Accordingly, a parasitic capacitance formed by the pixel electrode and the data signal line and a parasitic capacitance formed by the pixel electrode and the pixel scanning signal line can be reduced, thus improving the display quality.

[0036] According to the sixth aspect of the present invention, the discharge switching element is disposed so as to overlap the electrode pattern forming the discharge scanning signal line, thus enabling a larger aperture ratio.

[0037] According to the seventh aspect of the present invention, it is possible to secure a larger pixel area by disposing the pixel electrode so as to overlap the discharge scanning signal line, thus enabling a larger aperture ratio.

[0038] According to the eighth aspect of the present invention, in the normally-black mode display device, the discharge scanning signal line is provided corresponding to each of the pixel scanning signal lines, wherein each of the pixel scanning signal lines enters the selected state at least once every frame period for writing the image data of the image to be displayed, and each of the discharge scanning signal lines enters the selected state during the predetermined period within the period from the first time point when the pixel scanning signal line corresponding thereto changes from the selected state to the non-selected state to the second time point when the pixel scanning signal line changes to the selected state in the next frame period. Thereby, the black display period is inserted every display line, and it becomes possible to realize the impulse type display by suppressing the complication of the drive circuit, etc. and the increase in the operating frequency, resulting in improvement of the display quality of the moving image. When the display device according to this aspect of the present invention is the display device such as the liquid crystal display device, in which the pixel capacitance is charged with the voltage (as image data) having the polarity inverted every one frame period, each of the pixel capacitances discharges the inverse polarity charge by the application of the discharge scanning signal to the discharge scanning signal line before the charge up corresponding to the pixel data writing. That is, in the liquid crystal display device, since the pixel data writing means to charge up the pixel capacitance, which has been charged up by a certain polarity voltage, by the inverse polarity voltage, the discharging of the pixel capacitance for the black insertion also works for the pre-charging. Thereby, it is possible to improve the display quality, since the charging shortage is suppressed even when the resolution and the like is increased.

[0039] According to the ninth aspect of the present invention, the inverse polarity charge of the pixel capacitance is discharged for the black insertion before the charge up of the pixel capacitance corresponding to the pixel data writing, and thereby the charging shortage is

suppressed in the pixel data writing. Accordingly, while reducing the power consumption of the data signal line drive circuit by inverting the polarity of each of the data signals every two or more predetermined number of horizontal periods, it is possible to prevent the occurrence of the horizontal irregularity by the discharge for the above black insertion.

[0040] According to the tenth aspect of the present invention, the data signal is generated as a voltage signal having the polarity inverted every predetermined number of data signal lines, and charge sharing is carried out among the data signal lines by shorting the data signal lines with each other in the active matrix substrate during a predetermined period when the polarity of the data signal is inverted. Thereby, the power consumption of the data signal line drive circuit is reduced and also each of the data signal lines has the intermediate potential before the charge up of the pixel capacitance corresponding to the pixel data writing. By both of such charge sharing operation and the discharge of the charge in the pixel capacitance for the black insertion, the charging shortage of the pixel capacitance is suppressed securely in the pixel data writing, thus improving the display quality.

[0041] According to the eleventh aspect of the present invention, the data signal is generated as the voltage signal having the polarity inverted every predetermined number of data signal lines, and the data signal lines are shorted with each other on the active matrix substrate during a predetermined period every one horizontal period. Accordingly, the charge is transferred among the data signal lines every one horizontal period and thereby the charging condition can be made uniform while the charging shortage of the pixel capacitance is suppressed in the pixel data writing, even when the polarity of the data signal is inverted every two or more predetermined number of horizontal periods, thus securely preventing the horizontal irregularity from occurring.

[0042] According to the twelfth aspect of the present invention, the fixed potential is applied to the data signal lines when the data signal lines are shorted with each other on the active matrix substrate (charge sharing period), and thereby the potentials become always the same among the data signal lines immediately after the charge sharing period regardless of display gradation. By both of this and the discharge of the pixel capacitance for the black insertion, the charging condition of the pixel capacitance is made uniform in the pixel data writing, and it is possible to securely prevent the horizontal irregularity from occurring when the polarity of the data signal is inverted every two or more predetermined horizontal periods.

[0043] According to the thirteenth aspect of the present invention, the fixed potential equal to the storage capacitance line potential is applied to each of the data signal lines during the charge sharing period, and thereby the potential of the pixel electrode forming the pixel capacitance is made equal to the potential of the data signal line to transmit the data signal for charging the pixel ca-

pacitance, at a start timing of the pixel capacitance charge up for the pixel data writing. Accordingly, the charging condition can be made the same regardless of whether the polarity of the data signal for charging the pixel capacitance is positive or negative, and thereby it is possible to accurately suppress the generation of the horizontal irregularity when the polarity of the data signal is inverted every two or more predetermined number of horizontal periods. Note that, for the fixed potential, it is possible to use the common potential to be provided to the common electrode or a potential corresponding to a middle value (center value) between the minimum and maximum values of the data signal.

[0044] Advantages of the other aspects of the present invention are apparent from the description of the advantages of the above-described aspects and the description of the following embodiments of the present invention, and thus description thereof is not given.

20 Brief Description of the Drawings

[0045]

Fig. 1 is a plan view showing a first pattern configuration example for an active matrix substrate in a liquid crystal display device according to a first embodiment of the present invention.

Fig. 2 is a plan view showing a second pattern configuration example for the active matrix substrate in the first embodiment.

Fig. 3 is a cross-sectional view taken along the line A-A in Fig. 2.

Fig. 4 is a plan view showing a third pattern configuration example for the active matrix substrate in the first embodiment.

Fig. 5 is a block diagram showing a configuration of the liquid crystal display device according to the first embodiment.

Fig. 6 is an equivalent circuit diagram showing an electrical configuration of the active matrix substrate in the first embodiment.

Fig. 7 is a block diagram showing a configuration of a source driver in the liquid crystal display device according to the first embodiment.

Fig. 8 is a circuit diagram showing a first configuration example of an output section of the above source driver.

Fig. 9 consists of signal waveform charts (A) to (G) for explaining the operation of the liquid crystal display device according to the first embodiment.

Fig. 10 is a detailed signal waveform chart for explaining the operation of an active matrix substrate in a conventional liquid crystal display device with a 2H dot inversion drive employing a charge sharing method.

Fig. 11 is a detailed signal waveform chart for explaining the operation of the active matrix substrate in the first embodiment.

Fig. 12 is a circuit diagram showing a second configuration example of the output section of the above source driver.

Fig. 13 is a circuit diagram showing a third configuration example of the output section of the above source driver.

Fig. 14 consists of signal waveform charts (A) to (G) for explaining the operation of a liquid crystal display device according to a second embodiment of the present invention.

Fig. 15 is a detailed signal waveform chart for explaining the operation of the active matrix substrate in the second embodiment.

Fig. 16 is a detailed signal waveform chart for explaining the operation of an active matrix substrate in a variation of the second embodiment.

Fig. 17 is a detailed signal waveform chart for explaining the operation of an active matrix substrate in another embodiment of the present invention.

Fig. 18 is a block diagram showing a configuration of a liquid crystal display device according to another embodiment of the present invention.

Fig. 19 is a block diagram showing a configuration example of a display device for a television receiver using the active matrix substrate according to the present invention.

Fig. 20 is a block diagram showing an entire configuration, including a tuning section, of the television receiver using the active matrix substrate according to the present invention.

Fig. 21 is a perspective exploded view showing a mechanical configuration of the television receiver.

Fig. 22 is a diagram for explaining a problem in moving image display in a hold type display device.

Fig. 23 is a partial plan view showing a pattern configuration of a conventional active matrix substrate.

Description of the Reference Symbols

[0046]

10	Pixel TFT (Pixel switching element)
12	Discharge TFT (Discharge switching element)
12d	Drain electrode
12s	Source electrode
14	Source extraction electrode
16a-16e	Extended portion (of storage capacitance line)
100	Display part
110	Active matrix substrate
120	Opposite substrate
200	Display control circuit
300	Source driver (Data signal line drive circuit)
302	Data signal generation section
304	Output section
410	Pixel gate driver (Pixel scanning signal line drive circuit)

420	Discharge gate driver (Discharge scanning signal line drive circuit)
600	Common electrode/storage capacitance line drive circuit (Common potential supply section, Storage capacitance line potential supply section)
610	Common electrode drive circuit (Common potential supply section)
620	Storage capacitance line drive circuit (Storage capacitance line potential supply section)
C _{lc}	Liquid crystal capacitance
C _{cs}	Storage capacitance
E _p	Pixel electrode
E _c	Common electrode
SW _a	First MOS transistor
SW _b ,	SW _c Second MOS transistor
SL _i	Source line (Data signal line) (i = 1, 2, ..., N)
GL _j	Pixel gate line (Pixel scanning signal line) (j = 1, 2, ..., M)
Gd _{Lj}	Discharge gate line (Discharge scanning signal line) (j = 1, 2, ..., M)
C _{sL}	Storage capacitance line
S(i)	Data signal (i = 1, 2, ..., N)
G(j)	Pixel scanning signal (j = 1, 2, ..., M)
Gd(j)	Discharge scanning signal (j = 1, 2, ..., M)
V _{com}	Common potential (Opposite voltage)
V _{Sdc}	Source center potential (DC level of data signal)
C _{sh}	Charge sharing control signal
P _w	Pixel data write pulse
P _b	Black voltage application pulse
T _{sh}	Charge sharing period

35 Best Modes for Carrying Out the Invention

[0047] hereinafter, embodiments of the present invention will be described with reference to the accompanying drawings.

<1. First embodiment>

< 1.1 Configuration and operation>

[0048] A first embodiment will be described as an example of a liquid crystal display device using an active matrix substrate according to the present invention. Fig. 5 is a block diagram showing a configuration of the liquid crystal display device according to the present embodiment. Fig. 6 is an equivalent circuit diagram showing a first example of a circuit configuration in an active matrix substrate 110 in the present embodiment, and shows an electrical configuration of a part of this active matrix substrate 110 (corresponding to a part of neighboring four pixels) 101.

[0049] This liquid crystal display device includes: an active matrix-type display part 100 using the active matrix substrate 110; a source driver 300 as a data signal line

drive circuit; a pixel gate driver 410 as a pixel scanning signal line drive circuit; a discharge gate driver 420 as a discharge scanning signal line drive circuit; a common electrode/storage capacitance line drive circuit 600 as a common potential supply section and a storage capacitance line potential supply section; and a display control circuit 200 for controlling the source driver 300, the pixel gate driver 410, the discharge gate driver 420, and the common electrode/storage capacitance line drive circuit 600.

[0050] The display part 100 in the above liquid crystal display device includes a pair of electrode substrates sandwiching a liquid crystal layer and a polarizer plate is attached to each of the electrode substrates on the outside surface thereof. One of the above pair of electrode substrates is the active matrix substrate 110. As shown in Fig. 5 and Fig. 6, on an insulating substrate such as glass plate or the like in this active matrix substrate 110, there are formed: a plurality of (M) pixel gate lines GL1 to GLM as pixel scanning signal lines; a plurality of (N) source lines SL1 to SLN intersecting each of the pixel gate lines GL1 to GLM, as data signal lines; a plurality of (M x N) pixel circuits provided corresponding to the intersections of the pixel gate lines GL1 to GLM and the source lines SL1 to SLN, respectively; and a plurality of (M) discharge gate lines GdL1 to GdLM corresponding to the pixel gate lines GL1 to GLM, respectively. Each of the pixel circuits includes a TFT 10 as a switching element having a gate terminal connected to the pixel gate line GLj passing through the corresponding intersection and also having a source terminal connected to the source line SLi passing through the intersection (hereinafter, called "pixel TFT"), and a pixel electrode Ep connected to a drain terminal (electrode) of the pixel TFT 10.

[0051] On the other hand, the other one of the pair of electrode substrates is called an opposite substrate 120, and a common electrode Ec is formed there over the whole surface of an insulating substrate such as glass plate or the like. This common electrode Ec is provided commonly to each of the above plurality of (M x N) pixel circuits. Then, each of the pixel circuits in the active matrix substrate 110 constitutes a pixel formation portion together with the commonly provided common electrode Ec and the liquid crystal layer, and a liquid crystal capacitance Clc is formed by the pixel electrode Ep and the common electrode Ec in this pixel formation portion. Further, to securely retain the voltage of this pixel capacitance, a storage capacitance Ccs is provided in parallel to the liquid crystal capacitance Clc. That is, in the active matrix substrate 110, a storage capacitance line CsL is disposed in parallel to each of the pixel gate lines GLj, and the above storage capacitance Ccs is formed by this storage capacitance line CsL and the pixel electrode Ep which is disposed opposite the storage capacitance line CsL sandwiching an insulating film and the like. Accordingly, a capacitance to retain the written data signal S(i) as the pixel data (hereinafter, this capacitance is called a "pixel capacitance" and denoted by a symbol "Cp") is

composed of the liquid crystal capacitance Clc and the auxiliary capacitance Ccs. That is, if these symbols "Cp", "Clc", and "Ccs" express also capacitance values, $C_p = C_{lc} + C_{cs}$.

[0052] Further in the present embodiment, as shown in Fig. 6, each of the pixel circuits in the active matrix substrate 110 includes a TFT 12 as a discharge switching element (hereinafter, called "discharge TFT") in addition to the pixel TFT 10. Thereby, to each of the intersections of the pixel gate lines GL1 to GLM and the source lines SL1 to SLN, corresponds one of the pixel circuits and also correspond the pixel electrode Ep, the pixel TFT 10, and the discharge TFT 12 included in the pixel circuit. Accordingly, to each of the pixel electrodes Ep, correspond one of the pixel TFTs 10 and one of the discharge TFTs 12. Further, for each of the pixel gate lines GLj, the corresponding discharge gate line GdLj is disposed therealong in the present embodiment. A gate terminal in each of discharge TFTs 12 is connected to the discharge gate line GdLj corresponding to the pixel gate line GLj which is connected to the gate terminal of the corresponding pixel TFT 10, a drain terminal in each of the discharge TFTs 12 is connected to the corresponding pixel electrode Ep, and a source terminal in each of the discharge TFTs 12 is connected to the storage capacitance line CsL disposed so as to form the auxiliary capacitance together with the corresponding pixel electrode Ep. An active signal (voltage to turn on the TFT 12), when applied to any of the discharge gate lines GdLj in the active matrix substrate 110, turns on the discharge TFT 12 connected to the discharge gate line GdLj which is provided with the active signal, and the pixel electrode Ep in each of the pixel circuits, which the discharge gate line GdLj passes through, is electrically connected (shorted) to the storage capacitance line CsL via the discharge TFT 12.

[0053] As shown in Fig. 5 and Fig. 6, the pixel electrode Ep in each of the pixel formation portions is provided with a potential according to an image to be displayed from the source driver 300 and the pixel gate driver 410 which operates as will be described hereinafter, and the common electrode Ec is provided with a predetermined potential as a common potential Vcom (this common potential Vcom is also called "opposite voltage" or "common voltage") from the common electrode/storage capacitance line drive circuit 600. Thereby, a voltage according to a potential difference between the pixel electrode Ep and the common electrode Ec is applied to the liquid crystal, and this voltage application controls light transmittance of the liquid crystal layer and performs image display. Note that the polarizer plates are used for the light transmittance control by the voltage application to the liquid crystal layer, and in the liquid crystal display device according to the present embodiment the polarizer plates are disposed so as to realize the normally black mode. Here, the common potential Vcom provided to the common electrode Ec is also provided to the storage capacitance line CsL and the source driver 300, as shown in Fig. 5.

[0054] The display control circuit 200 receives a digital video signal D_v representing an image to be displayed, a horizontal synchronization signal HSY and a vertical synchronization signal VSY corresponding to the digital video signal D_v , and a control signal D_c for controlling display operation, from an external signal source, and, from these signals D_v , HSY, VSY, and D_c , generates and outputs a data start pulse signal SSP, a data clock signal SCK, a charge sharing control signal Csh, a digital image signal DA (signal corresponding to the video signal D_v) representing the image to be displayed, a pixel gate start pulse signal GSP, a pixel gate clock signal GCK, a pixel gate driver output control signal GOE, a discharge gate start pulse signal GSPd, a discharge gate clock signal GCKd, and a discharge gate driver output control signal GOEd, as signals for displaying the image represented by the digital video signal D_v on the display part 100. In more detail, the display control circuit 200 outputs the video signal D_v as the digital image signal DA, after performing timing adjustment or the like on the video signal D_v as needed with an internal memory, generates the data clock signal SCK as a signal composed of a pulse corresponding to each pixel of the image represented by the digital image signal DA, generates the data start pulse signal SSP as a signal exhibiting a high level (H level) during a predetermined period every one horizontal period based on the horizontal synchronization signal HSY, generates the pixel gate start pulse signal GSP and the discharge gate start pulse signal GSPd as signals exhibiting the H level during a predetermined period every one frame period (one vertical scanning period) based on the vertical synchronization signal VSY, generates the pixel gate clock signal GCK and the discharge gate clock signal GCKd based on the horizontal synchronization signal HSY, and generates the charge sharing control signal Csh, the pixel gate driver output control signal GOE, and the discharge gate driver output control signal GOEd based on the horizontal synchronization signal HSY and the control signal D_c .

[0055] Among the signals generated in the display control circuit 200 as described above, the digital image signal DA, the charge sharing control signal Csh, the data start pulse signal SSP, and the data clock signal SCK are inputted into the source driver 300, the pixel gate start pulse signal GSP, the pixel gate clock signal GCK, and the pixel gate driver output control signal GOE are inputted into the pixel gate driver 410, and the discharge gate start pulse signal GSPd, the discharge gate clock signal GCKd, and the discharge gate driver output control signal GOEd are inputted into the discharge gate driver 420.

[0056] The source driver 300 generates data signals $S(1)$ to $S(N)$ as analog voltages corresponding to pixel values in each of the horizontal scanning lines (each display line) of the image represented by the digital image signal DA every one horizontal period (every 1H), based on the digital image signal DA, the data start pulse signal SSP, and the data clock signal SCK, and applies these

data signals $S(1)$ to $S(N)$ to the source lines SL1 to SLN, respectively.

[0057] The present embodiment employs an nH dot inversion drive method, that is the drive method to invert the polarity of the voltage applied to the liquid crystal layer every one frame period and also to output the data signals $S(1)$ to $S(N)$ so as to invert the polarity every n pixel gate lines (n is two or more) and every one source line in each frame. Accordingly, the source driver 300 inverts the polarities of the applied voltages to the source lines SL1 to SLN every source line, and inverts the polarity of the data signal $S(i)$ applied to each of the source lines SLi every n horizontal periods. Here, a base potential in the polarity inversion of the applied voltage to the source line is a DC level (potential corresponding to a direct current component) of the data signals $S(1)$ to $S(N)$, and this DC level is not generally the same as a DC level of the common electrode E_c and is different from the DC level of the common electrode E_c by a pull-in voltage ΔV_d caused by a parasitic capacitance C_{gd} between the gate and drain of the pixel TFT in each of the pixel formation portions. Note that, when the pull-in voltage ΔV_d by the parasitic capacitance C_{gd} is sufficiently smaller than an optical threshold voltage V_{th} of the liquid crystal, the DC level of the data signals $S(1)$ to $S(N)$ can be assumed to be equal to the DC level of the common electrode E_c , and therefore the polarity of the data signals $S(1)$ to $S(N)$, that is, the polarities of the applied voltages to the source lines may be considered to be inverted every n horizontal periods with reference to the potential V_{com} of the common electrode E_c .

[0058] Fig. 7 is a block diagram showing a configuration of the source driver 300 in the present embodiment. This source driver 300 is configured with a data signal generation section 302 and an output section 304. The data signal generation section 302 generates analog voltage signals $d(1)$ to $d(N)$ corresponding to the source lines SL1 to SLN, respectively, from the digital image signal DA according to the data start pulse signal SSP and the data clock signal SCK. The configuration of this data signal generation section 302 is the same as that of a conventional source driver and description thereof will be omitted. The output section 304 performs impedance conversion on the analog voltage signal $d(i)$ generated in the data signal generation section 302 and outputs the converted signal as the data signal $S(i)$ ($i = 1, 2, \dots, N$).

[0059] Further, this source driver 300 employs the charge sharing method in which the neighboring source lines are shorted with each other for reducing power consumption during a predetermined period (period as short as a horizontal blanking period) T_{sh} when the polarity in each of the data signals $S(i)$ ($i = 1, 2, \dots, N$) is inverted. For this purpose, the output section 304 of the source driver 300 is configured as shown in Fig. 8. That is, this output section 304 receives the analog voltage signals $d(1)$ to $d(N)$ generated based on the digital image signal DA, and generates the data signals $S(1)$ to $S(N)$, as image signals to be transmitted over the source lines SL1

to SLN, by performing the impedance conversion on these analog voltage signals $d(1)$ to $d(N)$. The output section 304 includes N output buffers 31 as voltage followers for this impedance conversion. The output terminal of each of the buffers 31 is connected with a first MOS (Metal Oxide Semiconductor) transistor SWa as a switching element, and the data signal $S(i)$ from each of the buffers 31 is outputted from the output terminal of the source driver 300 via the first MOS transistor SWa ($i = 1, 2, \dots, N$). Further, the neighboring output terminals of the source driver 300 are connected to a second MOS transistor SWb as a switching element (thereby, the neighboring source lines are connected with each other by the second MOS transistor SWb). Then, to the gate terminal of the second MOS transistor SWb between these output terminals, the charge sharing control signal Csh is provided from the display control circuit 200, and, to the gate terminal of the first MOS transistor SWa connected to the output terminal of each of the buffers 31, an output signal of an inverter 33, that is, a logically inverted signal of the charge sharing control signal Csh is provided.

[0060] According to the above configuration, when the charge sharing control signal Csh is not active (low level), the first MOS transistor SWa is turned on (conduction state) and the second MOS transistor SWb is turned off (cut-off state), and thereby the data signal from each of the buffers 31 is outputted from the source driver 300 via the first MOS transistor SWa. On the other hand, when the charge sharing control signal Csh is active (high level), the first MOS transistor SWa is turned off (cut-off state) and the second MOS transistor SWb is turned on (conduction state), and thereby the data signal from each of the buffers 31 is not outputted (i.e., the application of the data signals $S(1)$ to $S(N)$ to the source lines SL1 to SLN is cut off) and the neighboring source lines in the display part 100 are shorted with each other via the second MOS transistor SWb.

[0061] The source driver 300 of the present configuration generates the analog voltage signal $d(i)$ as a video signal whose polarity is inverted every n horizontal periods (nH), here, every two horizontal periods ($2H$) for $n = 2$, as shown in (A) of Fig. 9, and the display control circuit 200 generates the charge sharing control signal Csh which exhibits the high level (H level) during such a short period Tsh as one horizontal blanking period when the polarity in each of the analog voltage signals $d(i)$ is inverted, as shown in (B) of Fig. 9. Here, the polarity of the analog voltage signal $d(i)$ is determined with reference to a source center potential (DC level of the data signal $S(i)$) VSdc corresponding to a center value between the minimum and maximum values of the data signal $S(i)$. This point will be the same in the following description, and the polarity of the data signal $S(i)$ is determined in the same manner. Further, the period Tsh, during which the charge sharing control signal Csh shown in (B) of Fig. 9 exhibits the H level, is a period during which the neighboring data signal lines are shorted with each other for

the charge re-distribution and is called a "charge sharing period".

[0062] As described above, the source driver 300 outputs each of the analog voltage signals $d(i)$ as the data signal $S(i)$ when the charge sharing control signal Csh exhibits the low level (L level), and the application of the data signals $S(1)$ to $S(N)$ to the source lines SL1 to SLN is cut off and also the neighboring source lines are shorted with each other when the charge sharing signal Csh exhibits the H level. Since the present configuration employs the nH dot inversion drive method and the voltages of the neighboring source lines have polarities opposite to each other, the voltage of each of the source lines SLi changes toward a certain intermediate potential between the positive polarity data signal voltage and the negative polarity data signal voltage in the charge sharing period Tsh.

[0063] In the present liquid crystal display device, each of the data signals $S(i)$ changes the polarity with reference to the source center potential VSdc which is the DC level of the data signal $S(i)$, and this source center potential VSdc has a value close to the common potential Vcom. Then, as shown in (C) of Fig. 9, each of the data signals $S(i)$ becomes to exhibit a potential equal to the source center potential VSdc in the charge sharing period Tsh by the charge sharing operation. Note that an ideal data signal waveform is shown here, and, in an actual case, the value of each of the data signals $S(i)$, that is, the voltage Vs of each of the source lines SLi changes as shown in Fig. 11 which will be described in the following, and becomes a potential approximately equal to the source center potential VSdc, that is, a potential close to the common potential Vcom at the end timing of each of the charge sharing periods Tsh. Note that a configuration to make the voltage Vs of each of the source lines SLi approximately equal to the DC level VSdc of the data signal $S(i)$, by shorting the neighboring source lines each other when the polarity of the data signal is inverted, is not limited to the configuration shown in Fig. 8.

[0064] The pixel gate driver 410 selects sequentially one of the pixel gate lines GL1 to GLM during approximately one horizontal period in each frame period (each vertical scanning period) of the digital image signal DA in order to write each of the data signals $S(1)$ to $S(N)$ into each of the pixel formation portions (pixel capacitances of the pixel formation portions) according to the pixel gate start pulse signal GSP, the pixel gate clock signal GCK, and the pixel gate driver output control signal GOE. That is, the pixel gate driver 410 applies the scanning signals G(1) to G(M), including pixel data write pulses Pw as shown in (D) and (F) of Fig. 9, to the pixel gate lines GL1 to GLM, respectively. Thereby, the pixel gate line GLj, to which the pulse Pw is applied, enters a selected state and the pixel TFT 10 connected to the pixel gate line GLj in the selected state is turned on (pixel TFT 10 connected to the pixel gate line in an unselected state is turned off). Here, the pixel data write pulse Pw exhibits the H level during the effective scanning period corresponding to a

display period in the horizontal period (1H).

[0065] The discharge gate driver 420 selects sequentially one of the discharge gate lines GdL1 to GdLM during approximately one horizontal period in each frame period in order to discharge the charge accumulated in the pixel capacitance Cp for each of the pixel formation portions according to the discharge gate start pulse signal GSPd, the discharge gate clock signal GCKd, and the discharge gate driver output control signal GOEd. That is, the discharge gate driver 420 applies the discharge scanning signals Gd(1) to Gd(M), including pulses Pb as shown in (E) and (G) of Fig. 9, to the discharge gate lines GdL1 to GdLM, respectively. Thereby, the discharge gate line GdLj, to which the pulse Pb is applied, enters a selected state and the discharge TFT 12 connected to the discharge gate line GdLj in the selected state is turned on (discharge TFT 12 connected to the discharge gate line in an unselected state is turned off). Since the present embodiment performs display in the normally black mode, the discharge in each of the pixel capacitances Cp, by the pulse Pb in each of the discharge scanning signals Gd(j) (j = 1, 2, ..., M), means application of a voltage corresponding to the black display (hereinafter, called "black voltage") to each of the pixel capacitances Cp. Accordingly, this pulse Pb is called a "black voltage application pulse", hereinafter.

[0066] The black voltage application pulse Pb, which is included in the discharge scanning signal Gd(j) applied to each of the discharge gate lines GdLj, corresponds to the pixel data write pulse Pw which is included in the pixel scanning signal G(j) applied to the pixel gate line GLj corresponding to the discharge gate line GdLj. Then, as shown in (D) to (G) of Fig. 9, each of the black voltage application pulses Pb appears being delayed after the pixel data write pulse Pw corresponding thereto, by a predetermined period Tdp. This predetermined period Tdp is a period during which a voltage corresponding to the pixel data is retained in the pixel capacitance Cp, and the period is shorter than one frame period (1V) and is approximately 2/3 frame periods, for example. Hereinafter, this predetermined period Tdp is called a "image display period".

[0067] In each of the pixel formation portions, the potential of the source line SLi, which is connected to the source terminal of the pixel TFT 10, is applied to the pixel electrode Ep via the pixel TFT 10 while the pixel TFT inside the pixel formation portion is in an ON state by the pixel data write pulse Pw. Thereby, the data signal S(i) as the voltage of the source line SLi is written into the pixel capacitance Cp corresponding to the pixel electrode Ep. Then, after the above image display period Tdp has elapsed, the black voltage application pulse Pb is provided to the gate terminal of the discharge TFT 12 inside the pixel formation portion, and thereby the pixel electrode Ep is connected (shorted) to the storage capacitance line CsL via the discharge TFT 12 while the discharge TFT 12 is in an ON state. As a result, the accumulated charge of the pixel capacitance Cp inside the

pixel formation portion is discharged and the pixel capacitance Cp enters a state of the black voltage application.

[0068] Accordingly, each of the pixel formation portions retains the voltage corresponding to the potential of the source line SLi, which is applied to the pixel electrode Ep via the pixel TFT 10 inside the pixel formation portion, in the pixel capacitance Cp during the image display period Tdp, and thereby forms a display pixel based on the digital image signal DA. On the other hand, during a period Tbk from the timing when the black voltage application pulse Pb appears in the discharge scanning signal Gd(j) applied to the gate terminal of the discharge TFT 12 inside the pixel formation portion to the timing when the next pixel data write pulse Pw appears in the pixel scanning signal G(j) applied to the gate terminal of the pixel TFT 10 inside the pixel formation portion (remaining period when the image display period Tdp is eliminated from the one frame period), each of the pixel formation portions forms a black pixel by retaining the black voltage in the pixel capacitance Cp therein. As a result, the display line corresponding to the discharge scanning signal Gd(j) provides the black display during the period Tbk (hereinafter, this period Tbk is called a "black display period"). Accordingly, the present embodiment realizes the impulse type display by inserting the black display period every display line, while suppressing the complication of the drive circuit, etc. and the increase in the operating frequency. Thereby, the tail dragging residual image is suppressed in the moving image and the display quality of the moving image is improved.

[0069] The common electrode/storage capacitance line drive circuit 600 provides the common potential Vcom, which is a predetermined fixed potential, to the common electrode Ec on the opposite substrate 120 and also to the storage capacitance line CsL on the active matrix substrate 110, under the control of the display control circuit 200. This common electrode/storage capacitance line drive circuit 600 functions as a common potential supply section which provides the common potential Vcom to the common electrode Ec and also functions as a storage capacitance line potential supply section which provides the common potential Vcom to the storage capacitance line CsL as a storage capacitance line potential. Note that, while the common potential and the storage capacitance line potential are the same in the present embodiment, these potentials may be different as will be described later.

<1.2 Functions and advantages>

[0070] Fig. 10 is a detailed signal waveform chart showing an operation of an active matrix substrate in a conventional liquid crystal display device using the 2H dot inversion drive employing the charge sharing method. In this conventional liquid crystal display device, the potential Vs of each of the source lines SLi becomes approximately equal to the source center potential VSdc in the charge sharing period Tsh. However, before the pixel

capacitance C_p starts to be charged for the pixel data writing by the pixel data write pulse P_w in the pixel scanning signal $G(j)$, to the pixel capacitance C_p , the potential of the source line SL_i in the previous frame period, that is, the pixel data indicated by the data signal $S(i)$ having the different polarity has been already written. Thereby, at a start timing t_1 of the charging by this pixel data write pulse P_w , the potential $V_p(j, i)$ of the pixel electrode E_p forming the pixel capacitance C_p has a negative polarity potential corresponding to the pixel data written in the previous frame period. Therefore, when one horizontal period becomes shorter for a higher resolution display or the like, the potential $V_p(j, i)$ of the pixel electrode E_p in the first one of the two lines which are a unit of the polarity inversion in the 2H inversion drive does not reach the target potential within the period of the pixel data write pulse P_w (period during which the pixel gate line GL_j is in the selected state) as shown in Fig. 10, causing the charging shortage.

[0071] The potential $V_p(j+1, i)$ of the pixel electrode E_p in the second one of the two lines as the above polarity inversion unit similarly is the negative polarity potential corresponding to the pixel data written in the previous frame period, at a start timing t_2 of the charging by the pixel data write pulse P_w in the pixel scanning signal $G(j+1)$. However, the potential V_s of the source line SL_i is the target potential already at the charging start timing t_2 of this second line pixel capacitance C_p . Therefore, even when the potential $V_p(j+1, i)$ of the pixel electrode E_p in the second line does not reach the target potential within the period of the pixel data write pulse P_w , the potential $V_p(j+1, i)$ is higher than the potential $V_p(j, i)$ of the pixel electrode E_p in the first line, as shown in Fig. 10.

[0072] As described above, in the conventional liquid crystal display device using the 2H dot inversion drive, even if the charge sharing method is employed, a difference of the charge amount is caused between the pixel capacitance in the first line and that in the second line, of the two lines as the unit of polarity inversion when one horizontal period becomes shorter for the higher resolution display or the like. As a result, this difference appears in a luminance difference and sometimes the horizontal irregularity is viewed in a line shape.

[0073] On the other hand, the present embodiment carries out the charge sharing operation as same as the above conventional example in the liquid crystal display device with the 2H dot inversion drive method as shown in (A) to (C) of Fig. 9, and additionally shorts each of the pixel electrodes E_p to the storage capacitance line Cs_L via the discharge TFT 12 by the black voltage application pulse P_b as shown in (E) and (G) of Fig. 9. Since the storage capacitance line Cs_L is provided with the common potential V_{com} , each of the pixel capacitances C_p discharges after the image display period T_{dp} of about 2/3 frame periods has elapsed from the charging thereof by the data signal $S(i)$ (potential of the source line SL_i) corresponding to the pixel data. As a result, at the start timing t_1 of the charging by the pixel data write pulse P_w

in the pixel scanning signal $G(j)$ for the pixel capacitance C_p in the first line of the two lines as the unit of polarity inversion, the potential $V_p(j, i)$ of the pixel electrode E_p in the first line is equal to the potential of the storage capacitance line Cs_L , that is, the common potential V_{com} due to the discharge of the pixel capacitance C_p by the black voltage application pulse P_b in the discharge scanning signal $G_d(i)$ before the timing t_1 . Further, the potential of the source line SL_i is approximately the same as the source center potential V_{Sdc} (potential close to the common potential V_{com}) by the charge sharing operation as in the conventional example. Therefore, the potential $V_p(j, i)$ of the pixel electrode E_p in the first line reaches the target potential during the period of the pixel data write pulse P_w as shown in Fig. 11, and the charging shortage is suppressed.

[0074] Further, also at the start timing t_2 of the charging by the pixel data write pulse P_w in the image scanning signal $G(j+1)$ for the pixel capacitance C_p in the second line of the two lines as the unit of polarity inversion, the potential $V_p(j+1, i)$ of the pixel electrode E_p in the second line is the same as the potential of the storage capacitance line Cs_L , that is, the common potential V_{com} due to the discharge of the pixel capacitance C_p by the black voltage application pulse P_b in the discharge scanning signal $G_d(j+1)$ before the timing t_2 . Then, the potential V_s of the source line SL_i is already the target potential at this timing t_2 . Therefore, the potential $V_p(j+1, i)$ of the pixel electrode E_p in the second line also reaches the target potential in the period of the pixel data write pulse P_w as shown in Fig. 11, similarly to the potential $V_p(j, i)$ of the pixel electrode E_p in the first line, and the charging shortage is suppressed.

[0075] In this manner, according to the present embodiment, while the potential V_s of the source line SL_i is different between the first line and the second line of the two lines as the unit of polarity inversion in the 2H inversion drive, at the charge start timings t_1 and t_2 of the pixel capacitance C_p , both of the potentials $V_p(j, i)$ and $V_p(j+1, i)$ of the pixel electrode E_p are the common potential V_{com} . Therefore, the potentials of pixel electrode E_p , $V_p(j, i)$ and $V_p(j+1, i)$ reach the target potential in the two lines, respectively, by the pixel data write pulse P_w , and a substantial difference is not caused between the charge amount of the pixel capacitance in the first line of the two lines and the charge amount of the pixel capacitance in the second line, and the horizontal irregularity in a line shape is not viewed. Further, as described above, by inserting the black display period T_{bk} for each display line utilizing the discharge of the pixel capacitance C_p through the black voltage application pulse P_b , it is possible to realize the impulse type display while suppressing the complication of the drive circuit, etc. and the increase in the operating frequency. Thereby, it is possible to suppress the tail-dragging residual image and to improve the display quality of the moving image. In this manner, the present embodiment can improve the display quality of the moving image by realizing the impulse

type display while suppressing the complication of the drive circuit, etc. and the increase in the operating frequency, and also can realize a higher display quality by improving the charging characteristic of the pixel capacitance (suppression of the charge amount difference and elimination of the charging shortage), in the active matrix-type liquid crystal display device which is the hold type display device.

[0076] Note that the present embodiment employs the charge sharing method assuming the dot inversion drive, but can provide basically the same effect even when not employing the charge sharing method or when not employing the dot inversion drive. However, it is preferable to employ the charge sharing method from the viewpoint of improving the charging characteristic of the pixel capacitance.

<1.3 Pattern configuration of the active matrix substrate>

[0077] Next, a pattern configuration of the active matrix substrate 110 for realizing the liquid crystal display device according to the above embodiment will be described with reference to Fig. 1 to Fig. 4.

[0078] Fig. 1 is a plan view of a first pattern configuration example of the active matrix substrate 110 in the present embodiment, and shows the pattern configuration of a part corresponding to six pixels. The pixel TFT 10 is provided in the neighborhood of the intersection of the pixel gate line GLj and the source line SLi ($i = 1, 2, \dots, N$, and $j = 1, 2, \dots, M$). In this example, the pixel gate line GLj works also as a gate electrode (terminal) 10g of the pixel TFT 10, a source electrode (terminal) 10s of the pixel TFT 10 is connected to the source line SLi, and a drain electrode (terminal) 10d is connected to the pixel electrode Ep via a contact hole 11 provided in an interlayer insulating film.

[0079] Further, the discharge gate line GdLj is disposed so as to extend along the pixel gate line GLj, and the discharge TFT 12 is disposed in the neighborhood of the discharge gate line GdLj. A gate electrode (terminal) 12g of this discharge TFT 12 is connected to the discharge gate line GdLj, a source electrode (terminal) 12s of the discharge TFT 12 is connected to the storage capacitance line CsL via a source extraction electrode 14 and a contact hole 15, and a drain electrode (terminal) 12d thereof is connected to the above pixel electrode Ep via a contact hole 17.

[0080] In the example of Fig. 1, the source extraction electrode 14 and a drain extraction electrode 13, which are connected to the source electrode 12s and the drain electrode 12d of the discharge TFT 12, respectively, do not overlap the discharge gate line GdLj. With this configuration, when the discharge TFT 12 is always in a conduction state caused by a residual film defect or the like in a channel portion of the discharge TFT 12 (in a case of a short fault), it is possible to repair the short fault by cutting the source extraction electrode 14 or the drain extraction electrode 13 through laser irradiation or the

like. Further, in the example of Fig. 1, the pixel electrode Ep overlaps the discharge gate line GdLj and the discharge TFT 12. This pattern configuration makes the pixel area wider and thereby is effective for improving the aperture ratio.

[0081] Note that it is preferable to form the source electrode 12s and the drain electrode 12d of the discharge TFT 12 as well as the electrodes connected to the electrodes 12s and 12d, respectively, using the same material as that of the source line SLi (similar in the other pattern configuration examples). The reason is that this makes it possible to form the source electrode 12s and the drain electrode 12d of the discharge TFT 12 as well as the electrodes connected to the electrodes 12s and 12d, respectively, in a formation process of the source line SLi, and it is not necessary to form these electrodes separately.

[0082] Fig. 2 is a plan view showing a second pattern configuration example of the active matrix substrate 110 according to the present embodiment and shows a pattern configuration of a part corresponding to two pixels. Fig. 3 is a cross-sectional view taken along the line A-A in Fig. 2. A constituent of this second example which is the same as or corresponding to that of the first example is denoted by the same reference symbol and numeral and description of the same part will be omitted below.

[0083] In this second example, the storage capacitance line CsL, disposed so as to pass through the center of the pixel electrode Ep in parallel to the pixel gate line GLj ($j = 1, 2, \dots, M$), has four extended portions 16a to 16d extending along the edge of the pixel electrode Ep in parallel to the source line SLi in each of the pixel circuits. These extended portions 16a to 16d function as shield electrodes for suppressing influence of potential changes in the source lines SL 1 and SL2 on the pixel electrode Ep. Among these four extended portions 16a to 16d, the extended portions 16b and 16c extending toward the discharge gate line GdLj are connected with each other at the ends by the source extraction electrode 14 via the contact holes 15, and the source extraction electrode 14 is connected to the source electrode 12s of the discharge TFT 12.

[0084] That is, as shown in Fig. 3, the extended portions 16b and 16c of the storage capacitance line CsL and a gate insulating film 22 made of silicon nitride (SiNx) or the like are formed sequentially on a glass substrate 20 of a transparent insulating substrate, and the source extraction electrode 14 formed thereon is electrically connected to the extended portion 16b and 16c of the storage capacitance line CsL via the contact holes 15 provided in the gate insulating film 22. On this source extraction electrode 14, an interlayer insulating film 24 made of silicon nitride or the like as a passivation film and an interlayer insulating film 26 made of acryl photo-sensitive resin or the like are formed sequentially, and further the pixel electrode Ep made of ITO (Indium Tin Oxide) or the like is formed thereon as a transparent electrode.

[0085] Thus, in the second example the source extrac-

tion electrode 14, which is a different layer from the discharge gate line GdLj, is formed at a position adjacent to the discharge gate line GdLj. Thereby, it is possible to prevent short between the storage capacitance line CsL and the discharge gate line GdLj. Further, it is possible to reduce a line breaking probability, because a part of the storage capacitance line CsL is formed in a different layer.

[0086] Further in this second example, as shown in Fig. 2, the discharge gate line GdLj is used also for the gate electrode (terminal) of the discharge TFT 12, and this discharge TFT 12 is covered by the discharge gate line GdLj via an insulating layer. Considering yield of producing the active matrix substrate, it is advantageous to use the first example in which the discharge TFT 12 as well as the source extraction electrode and drain extraction electrode thereof do not overlap the discharge gate line GdLj, but this second example has an advantage in terms of the aperture ratio.

[0087] Fig. 4 is a plan view showing a third pattern configuration example of the active matrix substrate 110 according to the present embodiment, and shows a pattern configuration of a part corresponding to two pixels. A constituent of this third example which is the same as or corresponding to that of the first or the second example is denoted by the same reference symbol and numeral and detailed description of the same part will be omitted below.

[0088] In this third example, as in the above second example, the storage capacitance line CsL has four extended portions 16a to 16d extending along the edge of the pixel electrode Ep in parallel to the source line SLi in each of the pixel circuits in order to suppress the influence of the potential changes of the source lines SL1 and SL2 on the potential of the pixel electrode Ep. Among these extended portions 16a to 16d, the extended portions 16a and 16d extending toward the gate line GLj are connected with each other by an electrode 16e extending along the edge of the pixel electrode Ep in parallel to the pixel gate line GLj (hereinafter, called "horizontal extended portion"), and are formed together with this horizontal extended portion 16e as a pattern combined with the storage capacitance line CsL. This horizontal extended portion 16e functions as a shield electrode for suppressing influence of the potential change of the pixel gate line GLj to the potential of the pixel electrode Ep. Further, as in the second example, the extended portions 16b and 16c extending toward the discharge gate line GdLj are connected with each other at the ends by the source extraction electrode 14 via the contact holes 15, and the source extraction electrode 14 is connected to the source electrode 12s of the discharge TFT 12.

[0089] In this third example, a circular structure is constituted by the five extended portions 16a to 16e of the storage capacitance line CsL formed along the edge of the pixel electrode Ep and by the source extraction electrode 14, in this manner. By forming such a circular structure as a part of the storage capacitance line CsL, it is

possible to reduce a parasitic capacitance formed by the pixel electrode Ep and the source line SLi as well as a parasitic capacitance formed by the pixel electrode Ep and the pixel gate line GLj, and to improve the display quality.

[0090] Note that, since the storage capacitance line CsL has the horizontal extended portion 16e as described above, the drain electrode 10d of the pixel TFT 10 is connected to the pixel electrode Ep at the center of the pixel electrode Ep via a drain extraction electrode 18 and the contact hole 11. Further, the drain extraction electrode 18 has a portion 19 facing the storage capacitance line CsL at this connection part, and the storage capacitance Ccs is formed with this portion 19 as a storage capacitance electrode facing the storage capacitance line CsL via an insulating film.

<1.4 Other configuration examples of the source driver>

[0091] As described above, in the present embodiment, the output section 304 of the source driver 300 is configured as shown in Fig. 8 (hereinafter, the configuration shown in Fig. 8 is called a "first configuration example"). In this configuration, the potential Vs of each of the source lines SLi changes toward the intermediate potential close to the source center potential VSdc in the charge sharing period Tsh, but is not secured completely to reach the source center potential VSdc or the common potential Vcom within the charge sharing period Tsh. However, as apparent from Fig. 11, it is preferable, from the viewpoint of improving the charging characteristic, to enable the potential Vs of each of the source lines SLi to reach the common potential Vcom or the source center potential VSdc within the charge sharing period Tsh.

[0092] Fig. 12 is a circuit diagram showing a configuration example (hereinafter, called "second configuration example") of the output section 304 of the source driver 300 for enabling the potential Vs of each of the source lines SLi to reach the common potential Vcom within the charge sharing period Tsh. In this configuration example, a constituent of the output section 304 which is the same as that in the first embodiment is denoted by the same reference symbol and numeral and description thereof will be omitted.

[0093] The output section 304 in the present configuration example is also provided with one second MOS transistor SWc as a switching element for each of the source lines SLi ($i = 1, 2, \dots, N$) as in the first configuration example. However, while the switch circuit is configured in the first configuration example such that the second MOS transistor SWb is inserted one by one between the neighboring source lines, the switch circuit is configured in the present configuration example such that the source driver 300 has an input terminal (hereinafter called "common potential input terminal") for receiving the common potential Vcom from the outside thereof and the second MOS transistor SWc is inserted one by one between the common potential input terminal and each of the source

lines SL_i. That is, in the present configuration example, the output terminal of the source driver to be connected to each of the source lines SL_i is connected to the common potential input terminal via any one of these second MOS transistors SW_c. Then, the charge sharing control signal C_{sh} is provided to all the gate terminals of the second MOS transistors SW_c.

[0094] In the second configuration example as described above, according to the charge sharing control signal C_{sh}, the analog voltage signals d(1) to d(N), which are generated by the data signal generation section 302, are outputted as the data signals S(1) to S(N) and applied to the source lines SL1 to SLN via the buffer 31, respectively, during the (effective scanning) period except for the charge sharing period T_{sh}, and the application of the data signals S(1) to S(N) to the respective source lines SL1 to SLN is cut off and also all the source lines SL1 to SLN are connected to the common potential input terminal during the charge sharing period T_{sh}. Therefore, in the charge sharing period T_{sh}, by the charge sharing operation to short the source lines SL1 to SLN with each other and by the supply of the common potential V_{com} via the common potential input terminal, each of the source lines SL_i has the common potential V_{com}. Accordingly, by using the source driver according to this second configuration example, the potential of each of the source lines SL_i is always the common potential V_{com} in each of the charge sharing periods T_{sh} without depending on the display gradation, and, additionally by the pixel capacitance discharge through the black voltage application pulse P_b, the charging shortage is securely suppressed and the charging characteristic is further improved. As a result, comparing to the case of using the source driver according to the first configuration example, it is possible to securely prevent the horizontal irregularity from occurring in the 2H inversion drive or the like.

[0095] Note that, in the second configuration example as described above, the fixed potential (V_{com}), which is equal to the potential provided to the storage capacitance line Cs_L, is provided to each of the data signal lines during the charge sharing period T_{sh}, and thereby the potential of the pixel electrode E_p forming the pixel capacitance C_p becomes equal to that of the source line SL_i, which is to transfer the data signal S(i) for charging the pixel capacitance C_p at the start timing of the charge up of the pixel capacitance C_p for the pixel data writing. Thereby, the charging condition can be made the same regardless of whether the polarity of the data signal for charging the pixel capacitance is positive or negative. As a result, it is possible to accurately suppress the horizontal irregularity occurrence in the case of the nH dot line inversion drive (n ≥ 2). Note that, for the above fixed potential, the source center potential VS_{dc} may be used instead of the common potential V_{com}, as in a third configuration to be described below.

[0096] Fig. 13 is a circuit diagram showing the third configuration example of the output section 304 in the source driver 300. The output section 304 of this third

configuration example, as in the second configuration example, includes N first MOS transistors SW_a, N second MOS transistors SW_c, and an inverter 33. However, this output section 304 is provided with a charge sharing voltage fixing power supply 35 (hereinafter, simply called "fixed power supply") supplying the source center potential VS_{dc}, instead of the common potential input terminal, and each of the output terminals is connected to the fixed power supply 35 via any one of the N second MOS transistors SW_c. Other configuration is the same as that of the second configuration example.

[0097] In the third configuration example as described above, according to the charge sharing control signal C_{sh}, the analog voltage signals d(1) to d(N), which are generated in the data signal generation section 302, are outputted and applied to the source lines SL1 to SLN, respectively, as the respective data signals S(1) to S(N) via the buffer 31 during the (effective scanning) period except for the charge sharing period T_{sh}, and the application of the data signals S(1) to S(N) to the respective source lines SL1 to SLN is cut off and also all the source lines SL1 to SLN are connected to the fixed power supply 35 during the charge sharing period T_{sh}. Therefore, in the charge sharing period T_{sh}, by the charge sharing operation of shorting the source lines SL1 to SLN with each other and by the supply of the source center potential VS_{dc} from the fixed power supply 35, the potential of each of the source lines SL_i becomes the source center potential VS_{dc}. Thereby, by using the source driver of the third configuration example, the potential of each of the source lines SL_i always becomes the source center potential VS_{dc} in each of the charge sharing periods T_{sh} without depending on the display gradation, and, additionally by the pixel capacitance discharge through the black voltage application pulse P_b, the charging shortage is securely suppressed and the charging characteristic is further improved. As a result, comparing to the case of using the source driver according to the first configuration example, it is possible to securely prevent the horizontal irregularity from occurring in the 2H inversion drive or the like.

<2. Second embodiment>

[0098] Next, another example of the liquid crystal display device using the active matrix substrate according to the present invention will be described as a second embodiment. The liquid crystal display device according to the present embodiment has the same configuration as that of the above first embodiment, except for a point that the charge sharing control signal C_{sh} generated in the display control circuit is different, and the same or corresponding part is denoted by the same reference symbol and detailed description thereof will be omitted. Note that, hereinafter, the output section 304 of the source driver 300 is configured as shown in Fig. 12 (second configuration example), and each of the source lines SL_i (i = 1, 2, ..., N) is provided with the common potential

Vcom when the charge sharing control signal Csh exhibits the H level, that is, during the charge sharing period Tsh.

[0099] Fig. 14 is a signal waveform chart for illustrating the operation of the liquid crystal display device according to the present embodiment. This liquid crystal display device also employs the 2H dot inversion drive method as in the first embodiment, and the data signal generation section 302 in the source driver 300 generates the analog voltage signals d(1) to d(N) as shown in (A) of Fig. 14 as a video signal. These analog voltage signals d(1) to d(N) are provided to the output section 304 of the source driver 300 (refer to Fig. 7).

[0100] The display control circuit 200 of the present embodiment generates a signal as shown in (B) of Fig. 14 as the charge sharing control signal Csh. This charge sharing control signal Csh exhibits the H level during a period corresponding to the horizontal blanking period of the image display every one horizontal period, and, in this point, is different from the charge sharing control signal Csh of the first embodiment, which exhibits the H level only when the polarity of the analog voltage signal d(i) or the data signal S(i) is inverted ((B) of Fig. 9).

[0101] The output section 304 of the source driver 300 is configured as shown in Fig. 12, and thereby generates the data signals S(1) to S(N) as shown in (C) of Fig. 14 from the above analog voltage signals d(1) to d(N), respectively, and applies these data signals S(1) to S(N) to the source lines SL1 to SLN, respectively, according to such a charge sharing control signal Csh. In this manner, the present embodiment provides the charge sharing period Tsh every one horizontal period and shorts the respective source lines SLi with each other and also provides the common potential Vcom to each of the source lines SLi during the charge sharing period Tsh ($i = 1, 2, \dots, N$). Note that, in the period except for the charge sharing period Tsh, the above analog voltage signals d(1) to d(N) are applied to the source lines SL 1 to SLN as the data signals S(1) to S(N), respectively.

[0102] The pixel gate driver 410, as in the first embodiment, generates the pixel scanning signals G(1) to G(M) as shown in (D) and (F) of Fig. 14, and applies these pixel scanning signals G(1) to G(M) to the pixel gate lines GL(1) to GL(M), respectively. The discharge gate driver 420, also as in the first embodiment, generates the discharge scanning signals Gd(1) to Gd(M) as shown in (E) and (G) of Fig. 14, and applies these discharge scanning signals Gd(1) to Gd(M) to the discharge gate lines GdL(1) to GdL(M), respectively.

[0103] By the above operation of the liquid crystal display device, as in the first embodiment, the black display period is inserted every display line and thereby the impulse type display is realized while suppressing the complication of the drive circuit, etc. and the increase in the operating frequency, resulting in the suppression of the tail-dragging residual image in the moving image and the improvement of the display quality in the moving image. In addition, this embodiment further improves the charg-

ing characteristic of the pixel capacitance Cp by the operation of the output section 304 in the source driver 300 according to the charge sharing control signal Csh. This will be described below in detail.

[0104] Fig. 15 a detailed signal waveform chart showing the operation of the active matrix substrate in the liquid crystal display device according to the present embodiment. As shown in Fig. 15, in the present embodiment, the charge sharing period Tsh is provided every one horizontal period, and the potential of each of the source lines SLi becomes the common potential Vcom at the end timing of the charge sharing period Tsh. Further, as in the first embodiment, each of the pixel capacitances Cp discharges by the black voltage application pulse Pb before the charging is started by the pixel data write pulse Pw, and thereby the potential Vp of the pixel electrode Ep also becomes the common potential Vcom. In this manner, for the pixel capacitance Cp in the first line and the pixel capacitance Cp in the second line, of the two lines as the unit of polarity inversion in the 2H inversion drive, at the charging start timings thereof t1 and t2, both of the potentials of the source line SLi become the common potential Vcom, and both of the potentials Vp(i, j) and Vp(j+1, i) of the pixel electrode Ep become the common potential Vcom. Accordingly, the charging shortage of the pixel capacitance Cp is further suppressed. In addition, the condition (source line potential and pixel electrode potential) at the charging start timing of the pixel capacitance Cp is the same between the first line and the second line, and thereby the difference in the charge amount between the pixel capacitance in the first line of the two lines and the pixel capacitance in the second line thereof is securely eliminated. Further, as described above, the black display period Tbk is inserted for each of the display lines by the discharge of the pixel capacitance Cp through the black voltage application pulse Pb, and thereby the impulse type display is realized while suppressing the complication of the drive circuit, etc. and the increase in the operating frequency. In this manner, the present embodiment can improve the display quality of the moving image by realizing the impulse type display while suppressing the complication of the drive circuit, etc. and the increase in the operating frequency, and can realize a higher display quality by further improving the charging characteristic of the pixel capacitance, in the active matrix-type liquid crystal display device which is the hold type display device.

[0105] while the above second embodiment provides the common potential Vcom to each of the source lines SLi during the charge sharing period Tsh using the configuration of the output section 304 in the source driver 300, as shown in Fig. 12, instead, each of the source lines SLi may be provided with the source center potential VSdc during the charge sharing period Tsh by use of the configuration of the output section 304 as shown in Fig. 13. In this case, the potential Vs of the source line SLi and the potentials Vp(j, i) and Vp(j+1, i) of the pixel electrode Ep in the active matrix substrate 110 change as

shown in Fig. 16 when the pixel capacitance is charged.

[0106] That is, for the pixel capacitance C_p in the first line and the pixel capacitance C_p in the second line, of the two lines as the unit of polarity inversion in the 2H inversion drive, at the charging start timings thereof t_1 and t_2 , both of the potentials in the source line SLi become the source center potential V_{Sdc} which is the fixed potential, and both of the potentials $V_p(j, i)$ and $V_p(j+1, i)$ of the pixel electrode E_p become the common potential V_{com} . Accordingly, between the charging start timings t_1 and t_2 , the potential of the source line SLi and the potential of the pixel electrode E_p are slightly different but approximately the same potential, and the condition (source line potential and pixel electrode potential) at the charging start timing of the pixel capacitance C_p is the same between the first line and the second line. Accordingly, also in the case of the above configuration of the output section 304 in the source driver 300 as shown in Fig. 12, it is possible to obtain the same advantage as in the above second embodiment.

[0107] Further, in the above second embodiment, the output section 304 of the source driver 300 may have the configuration shown in Fig. 8 (first configuration example) as in the first embodiment. In terms of improving the charging characteristic (suppressing the charging shortage and homogenizing the charging condition), the configuration shown in Fig. 12 or Fig. 13 is more preferable than this configuration, but the above second embodiment, also when employing this configuration, provides an advantage that the charging condition is homogenized and the horizontal irregularity is prevented from occurring, compared to the above first embodiment.

<3. Other embodiments and variations>

[0108] While the above first and second embodiments employ the 2H dot inversion drive method, the active matrix substrate according to the present invention can be used in the liquid crystal display device employing the 1H dot inversion drive method, and also can be used in the liquid crystal display device not employing the dot inversion drive method but employing a line inversion drive method. For example, when the active matrix substrate according to the present invention is employed in the liquid crystal display device employing the 1H dot inversion method, the potential V_s of the source line SLi and the potentials $V_p(j, i)$ and $V_p(j+1, i)$ of the pixel electrode E_p in the active matrix substrate change as shown in Fig. 17 when the pixel capacitance C_p is charged. Here, in the example shown in Fig. 17, each of the source lines SLi is provided with the source center potential V_{Sdc} as a fixed potential during the charge sharing period T_{sh} .

[0109] In this case, at the charging start timing of the pixel capacitance C_p in each of the lines, the potential of the source line SLi becomes uniformly the source center potential V_{Sdc} as the fixed potential, and both of the potentials $V_p(j, i)$ and $V_p(j+1, i)$ of the pixel electrode E_p become the common potential V_{com} . Accordingly, even

when one horizontal period (time to be secured for charging) becomes shorter for realizing the higher resolution or the like, the charging shortage of the pixel capacitance C_p is suppressed and the horizontal irregularity in a line shape, which is caused by the difference in the charge amount of the pixel capacitance between the different lines, is also suppressed. Thereby, it is possible to improve the display quality of the moving image by realizing the impulse type display while suppressing the complication of the drive circuit, etc. and the increase in the operating frequency, and also to realize higher display quality by improving the charging characteristic of the pixel capacitance (suppressing the charge amount difference and eliminating the charging shortage), as in the above first and second embodiments.

[0110] Further, while the above first and second embodiments provide the common potential V_{com} to the storage capacitance line CsL in the active matrix substrate 110 by the common electrode / storage capacitance line drive circuit 600 (Fig. 5), a common electrode / storage capacitance line drive circuit 610 and a storage capacitance line drive circuit 620 may be provided separately instead of the common electrode / storage capacitance line drive circuit 600 as shown in Fig. 18, and the storage capacitance line CsL may be provided with a potential other than the common potential V_{com} (but, potential close to the common potential V_{com}). In the example shown in Fig. 18, the storage capacitance line CsL is provided with the source center potential V_{Sdc} by the storage capacitance line drive circuit 620. Note that, in this example, the source center potential V_{Sdc} is also provided to the source driver 300, and the potential V_p of each of the source lines SLi becomes the source center potential V_{Sdc} during the charge sharing period T_{sh} (refer to Fig. 16). Also in such a configuration, it is possible to obtain substantially the same advantage as in the above second embodiment.

<4. Application to a television receiver>

[0111] Next, an example will be described for application of the active matrix substrate according to the present invention to a television receiver. Fig. 19 is a block diagram showing a configuration of a display device 800 for the television receiver. The display device 800 is provided with a Y/C separation circuit 80, a video-chroma circuit 81, an A/D converter 82, a liquid crystal controller 83, a liquid crystal panel 84, a backlight drive circuit 85, a backlight 86, a microcomputer 87, and a gradation circuit 88.

[0112] The liquid crystal panel 84 includes a display part using the active matrix substrate according to the present invention as well as a source driver, a pixel gate driver, a discharge gate driver, and a common electrode / storage capacitance line drive circuit for driving the display part, and a specific configuration thereof may be any one described in each of the embodiments or each of the variations of the present invention (refer to Figs. 5 to 9 and Figs. 11 to 18).

[0113] In the display device 800 with the above con-

figuration, first a composite color video signal Scv is inputted from outside into the Y/C separation circuit 80 as a television signal and separated there into a luminance signal and color signals. These luminance signal and color signals are converted into analog RGB signals corresponding to the light's three primary colors in the videochroma circuit 81, and further these analog RGB signals are converted into digital RGB signals in the A/D converter 82. These digital RGB signals are inputted into the liquid crystal controller 83. Further, in the Y/C separation circuit 80, horizontal and vertical synchronization signals are extracted from the composite color video signal Scv inputted from outside and these synchronization signals are also inputted into the liquid crystal controller 83 via the microcomputer 87.

[0114] To the liquid crystal panel 84, the digital RGB signals are inputted from the liquid crystal controller 83 together with timing signals based on the above synchronization signals at a predetermined timing. Further, in the gradation circuit 88, gradation voltages of each of the three primary colors R, G, and B for color display are generated and the gradation voltages are supplied also to the liquid crystal panel 84. In the liquid crystal panel 84, drive signals (data signal, pixel scanning signal, discharge scanning signal, etc.) are generated by an internal source driver, a pixel gate driver, a discharge gate driver, etc. based on these RGB signals, timing signal and gradation voltages, and a color image is displayed on an internal display part (using the active matrix substrate) based on these drive signals. Note that, for displaying the image on the liquid crystal panel 84, it is necessary to illuminate the liquid crystal panel 84 with light from the back, and in the display device 800, the backlight 86 is driven by the backlight drive circuit 85 under the control of the microcomputer 87 and thereby illuminates the back surface of the liquid crystal panel 84 with light.

[0115] Control of the entire system including the above processing is performed by the microcomputer 87. Note that, for the video signal (composite color video signal) inputted from outside, not only a video signal from television broadcasting but also a video signal obtained by shooting with a camera, a video signal supplied via the internet line, or the like can be used, and image display of the various video signals is possible in the display device 800.

[0116] When an image of the television broadcasting is displayed on the display device 800 with the above configuration, a tuner part 90 is connected to the display device 800 as shown in Fig. 20. The tuner part 90 extracts a channel signal to be received from receiving waves (high frequency wave signal) received by an antenna (not shown in the drawing), converts the channel signal into an intermediate frequency signal, and takes out the composite color video signal Scv as a television signal by subjecting the intermediate frequency signal to detection. The composite color video signal Scv is inputted into the display device 800 as described above and an image according to the composite color video signal Scv is dis-

played on the display device 800.

[0117] Fig. 21 is a perspective exploded view showing an example of a mechanical configuration when the display device with the above configuration is applied to the television receiver. In the example shown in Fig. 21, the television receiver includes a first chassis 801 and a second chassis 806 in addition to the above display device 800 as constituents thereof, and is configured such that the first chassis 801 and the second chassis 806 sandwich the display device 800 so as to wrap the same. The first chassis 801 has an opening 801a formed for transmitting an image displayed on the display device 800. Further, the second chassis 806 is one to cover the back side of the display device 800 and provided with an operation circuit 805 for operating the display device 800 and also a supporting member 808 attached on the lower side.

[0118] In the television receiver as described above, the active matrix substrate, source driver, pixel gate driver, discharge gate driver, etc. in the liquid crystal panel 84 have the same configurations as those of the foregoing first or second embodiments or the variations thereof, and thereby it is possible to improve the display quality of the moving image by realizing the impulse type display while suppressing the complication of the drive circuit, etc. and the increase in the operating frequency in the display device, and also to realize a higher display quality by improving the charging characteristic of the pixel capacitance.

Industrial Applicability

[0119] The present invention is applied to the active matrix substrate or the display device provided with the same, and is particularly suitable for the active matrix-type liquid crystal display device and the active matrix substrate used for the same.

Claims

1. An active matrix substrate, comprising:

- a plurality of data signal lines;
- a plurality of pixel scanning signal lines intersecting the plurality of data signal lines;
- a pixel switching element provided corresponding to each of intersections of the plurality of data signal lines and the plurality of pixel scanning signal lines, and turned on and off by the pixel scanning signal line passing through the corresponding intersection;
- a pixel electrode connected to the data signal line passing through the intersection corresponding to the pixel switching element via the pixel switching element;
- a storage capacitance line disposed so as to form a predetermined capacitance with each of

- the pixel electrodes;
 a plurality of discharge scanning signal lines corresponding to the plurality of pixel scanning signal lines, respectively; and
 a discharge switching element provided corresponding to each of the pixel electrodes, and turned on and off by the discharge scanning signal line which corresponds to the pixel scanning signal line for turning on and off the pixel switching element connected to the corresponding pixel electrode,
 each of the pixel electrodes being connected to the storage capacitance line via the corresponding discharge switching element.
2. The active matrix substrate according to Claim 1, wherein;
 the storage capacitance line has an extended portion extending in a direction along the data signal line; and the discharge switching element is a thin film transistor having a drain electrode and a source electrode,
 the drain electrode being connected to the pixel electrode corresponding to the discharge switching element, and
 the source electrode being connected to the extended portion via a predetermined source extraction electrode.
 3. The active matrix substrate according to Claim 2, wherein the extended portion and the source extraction electrode form a structure disposed circularly along an edge of the pixel electrode.
 4. The active matrix substrate according to Claim 2, wherein an electrode connected to the source and an electrode connected to the drain of the thin film transistor as the discharge switching element are formed of the same material as that of the data signal line.
 5. The active matrix substrate according to Claim 1, wherein the storage capacitance line has a circular portion including a portion which extends in parallel to the data signal line along an edge of the pixel electrode and a portion which extends in parallel to the pixel scanning signal line along the edge of the pixel electrode.
 6. The active matrix substrate according to Claim 1, wherein the discharge switching element is disposed so as to overlap an electrode pattern forming the discharge scanning signal line.
 7. The active matrix substrate according to Claim 1, wherein the pixel electrode is disposed so as to overlap the discharge scanning signal line.
 8. A normally black mode display device, comprising;
 an active matrix substrate according to any one of Claims 1 to 7;
 a common electrode disposed so as to face each of pixel electrodes in the active matrix substrate;
 a pixel scanning signal line drive circuit for selectively applying an active signal, which turns on the pixel switching element, to the plurality of pixel scanning signal lines and causing the pixel scanning signal line to which the active signal is applied to be in a selected state such that each of the plurality of pixel scanning signal lines is in the selected state at least once during each frame period;
 a discharge scanning signal line drive circuit for selectively applying an active signal, which turns on the discharge switching element, to the plurality of discharge scanning signal lines and causing the discharge scanning signal line to which the active signal is applied to be in a selected state such that each of the plurality of discharge scanning signal lines is in the selected state during a predetermined period within a period from a first time point when the corresponding pixel scanning signal line changes from the selected state to a non-selected state to a second time point when the corresponding pixel scanning signal line is in the selected state in the next frame period;
 a data signal line drive circuit for generating a plurality of data signals to display an image as voltage signals whose polarities are inverted every predetermined number of horizontal periods, and applying the plurality of data signals to the plurality of data signal lines, respectively;
 a common potential supply section for providing a predetermined common potential to the common electrode; and
 a storage capacitance line potential supply section for providing a predetermined potential approximately the same as the common potential to the storage capacitance line.
 9. The display device according to Claim 8, wherein the data signal line drive circuit generates the plurality of data signals so as to invert the voltage polarity thereof every two or more predetermined number of horizontal periods.
 10. The display device according to Claim 8, wherein the data signal line drive circuit generates the plurality of data signals so as to invert the polarity thereof every predetermined number of the data signal lines, and cuts off the application of the plurality of data signals to the respective plurality of data signal lines and also shorts the plurality of data signal lines with each other, during a predetermined period when the polarities of the plurality of data signals are inverted.

11. The display device according to Claim 10, wherein the data signal line drive circuit cuts off the application of the plurality of data signals to the respective plurality of data signal lines and also shorts the plurality of data signal lines with each other, during a predetermined period every one horizontal period. 5
12. The display device according to Claim 11, wherein the data signal line drive circuit provides a predetermined fixed potential to the plurality of data signal lines when the plurality of data signal lines are shorted with each other. 10
13. The display device according to Claim 12, wherein the fixed potential is equal to the predetermined potential. 15
14. The display device according to Claim 12, wherein the storage capacitance line potential supply section provides the common potential to the storage capacitance line as the predetermined potential. 20
15. The display device according to Claim 12, wherein the fixed potential is a potential corresponding to a center value between a minimum value and a maximum value of the data signal. 25
16. The display device according to Claim 11, wherein the storage capacitance line potential supply section provides the potential corresponding to a center value between a minimum value and a maximum value of the data signal to the storage capacitance line as the predetermined potential. 30
17. A television receiver, comprising a display device according to Claim 8. 35
18. A drive method of a normally black mode display device provided with an active matrix substrate and a common electrode; the active matrix substrate including a plurality of data signal lines, a plurality of pixel scanning signal lines intersecting the plurality of data signal lines, a pixel switching element which is provided corresponding to each of intersections of the plurality of data signal lines and the plurality of pixel scanning signal lines and turned on and off by the pixel scanning signal line passing through the corresponding intersection, a pixel electrode connected to the data signal line passing through the intersection corresponding to the pixel switching element via the pixel switching element, and a storage capacitance line disposed so as to form a predetermined capacitance with the pixel electrode; the common electrode being disposed so as to face each of the pixel electrodes in the active matrix substrate; the drive method comprising: 40

a pixel scanning signal line drive step of selec-

tively applying an active signal, which turns on the pixel switching element, to the plurality of pixel scanning signal lines and causing the pixel scanning signal line to which the active signal is applied to be in a selected state such that each of the plurality of pixel scanning signal lines is in the selected state at least once during each frame period;

a data signal line drive step of generating a plurality of data signals to display an image as voltage signals whose polarities are inverted every predetermined number of horizontal periods and applying the plurality of data signals to the plurality of data signal lines, respectively;

a common potential supply step of providing a predetermined common potential to the common electrode;

a storage capacitance line potential supply step of providing a predetermined potential approximately equal to the common potential to the storage capacitance line; and

a discharge step of shorting each of the pixel electrodes to the storage capacitance line, wherein the active matrix substrate further includes:

a plurality of discharge scanning signal lines corresponding to the plurality of pixel scanning signal lines, respectively; and

a discharge switching element provided corresponding to each of the pixel electrodes, and turned on and off by the discharge scanning signal line which corresponds to the pixel scanning signal line for turning on and off the pixel switching element connected to the corresponding pixel electrode,

each of the pixel electrodes is connected to the storage capacitance line via the corresponding discharge switching element, and

in the discharge step, an active signal, which turns on the discharge switching element, is selectively applied to the plurality of discharge scanning signal lines and the discharge scanning signal line to which the active signal is applied is caused to be in a selected state, such that each of the plurality of discharge scanning signal lines is in the selected state during a predetermined period within a period from a first time point when the corresponding pixel scanning signal line changes from the selected state to a non-selected state to a second time point when the corresponding pixel scanning signal line is in the selected state in the next frame period.

19. The drive method according to Claim 18, wherein in

the data signal line drive step the plurality of data signals are generated so as to invert voltage polarity thereof every two or more predetermined number of horizontal periods.

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- 20.** The drive method according to Claim 18, wherein, in the data signal line drive step, the plurality of data signals are generated so as to invert the polarity thereof every predetermined number of data signal lines, and the application of the plurality of data signals to the respective plurality of data signal lines is cut off and the plurality of data signal lines are also shorted with each other during a predetermined period when the polarities of the plurality of data signals are inverted.

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- 21.** The drive method according to Claim 20, wherein, in the data signal line drive step, the application of the plurality of data signals to the respective plurality of data signal lines is cut off and the plurality of data signal lines are also shorted with each other during a predetermined period every one horizontal period.

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Fig. 1

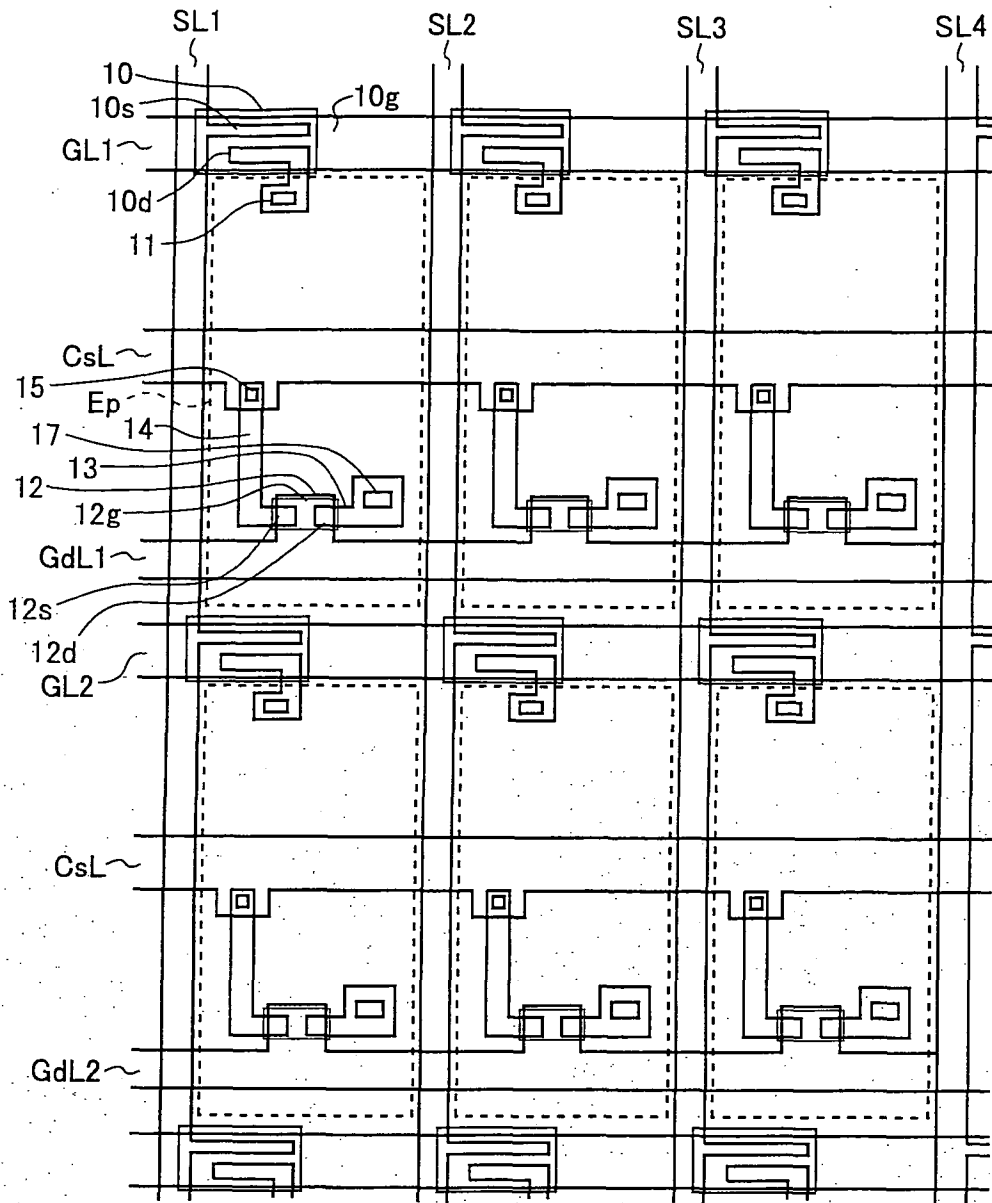


Fig.2

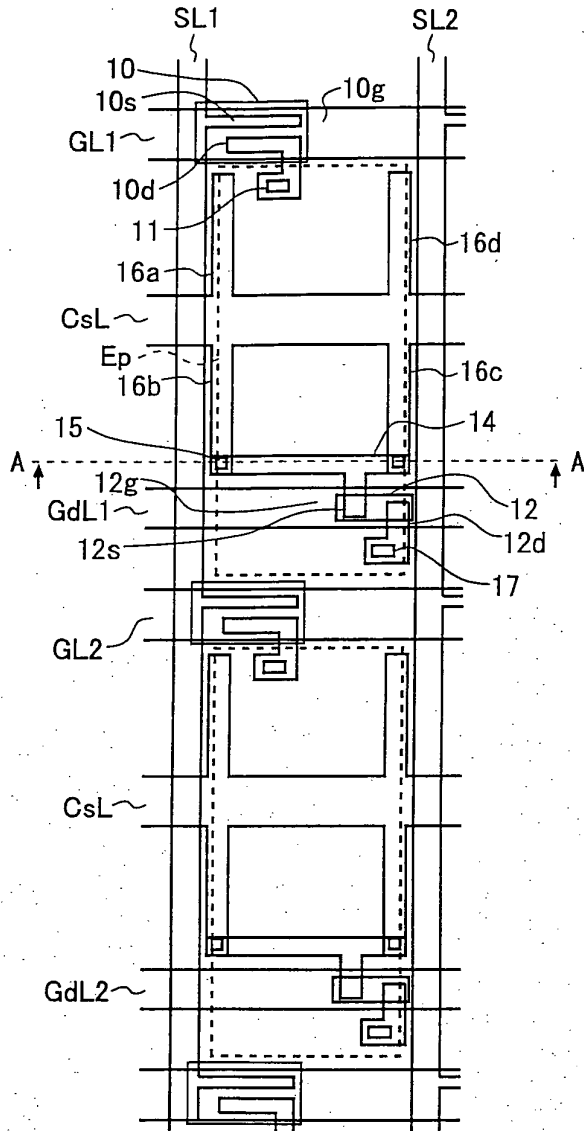


Fig.3

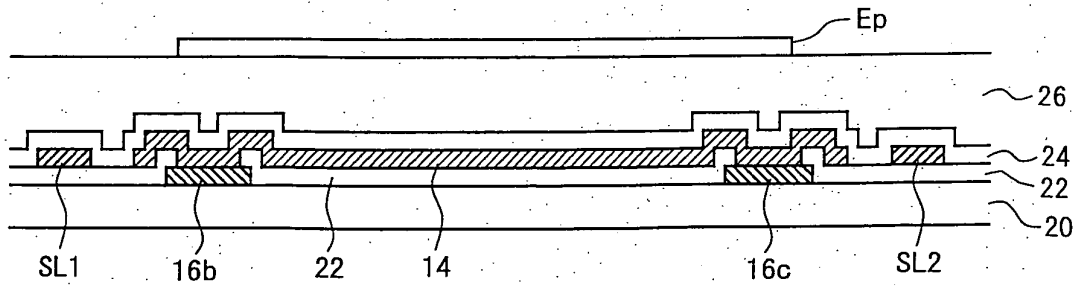


Fig.4

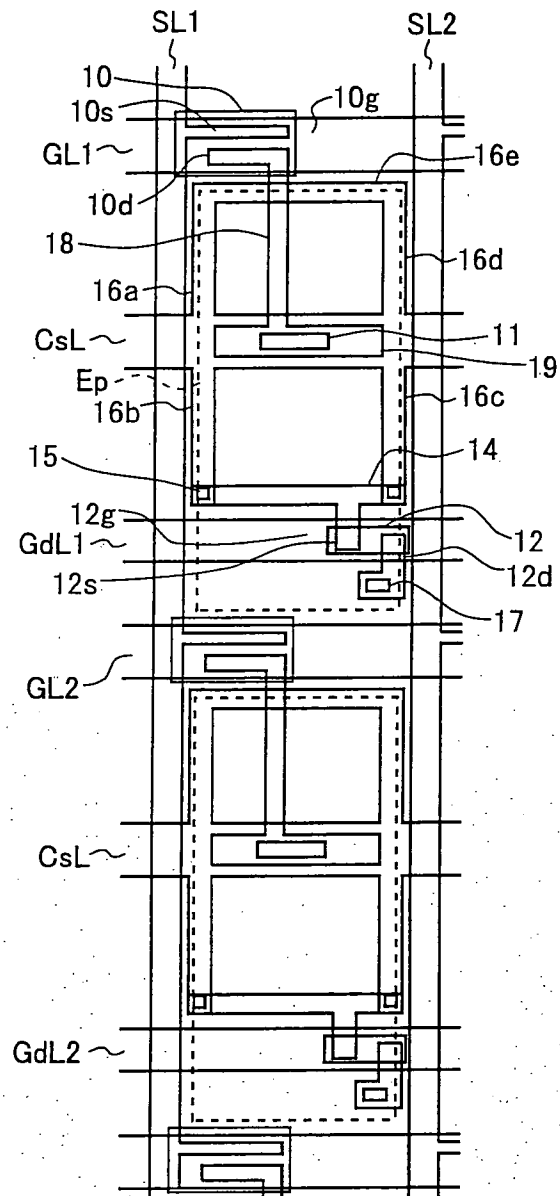


Fig.5

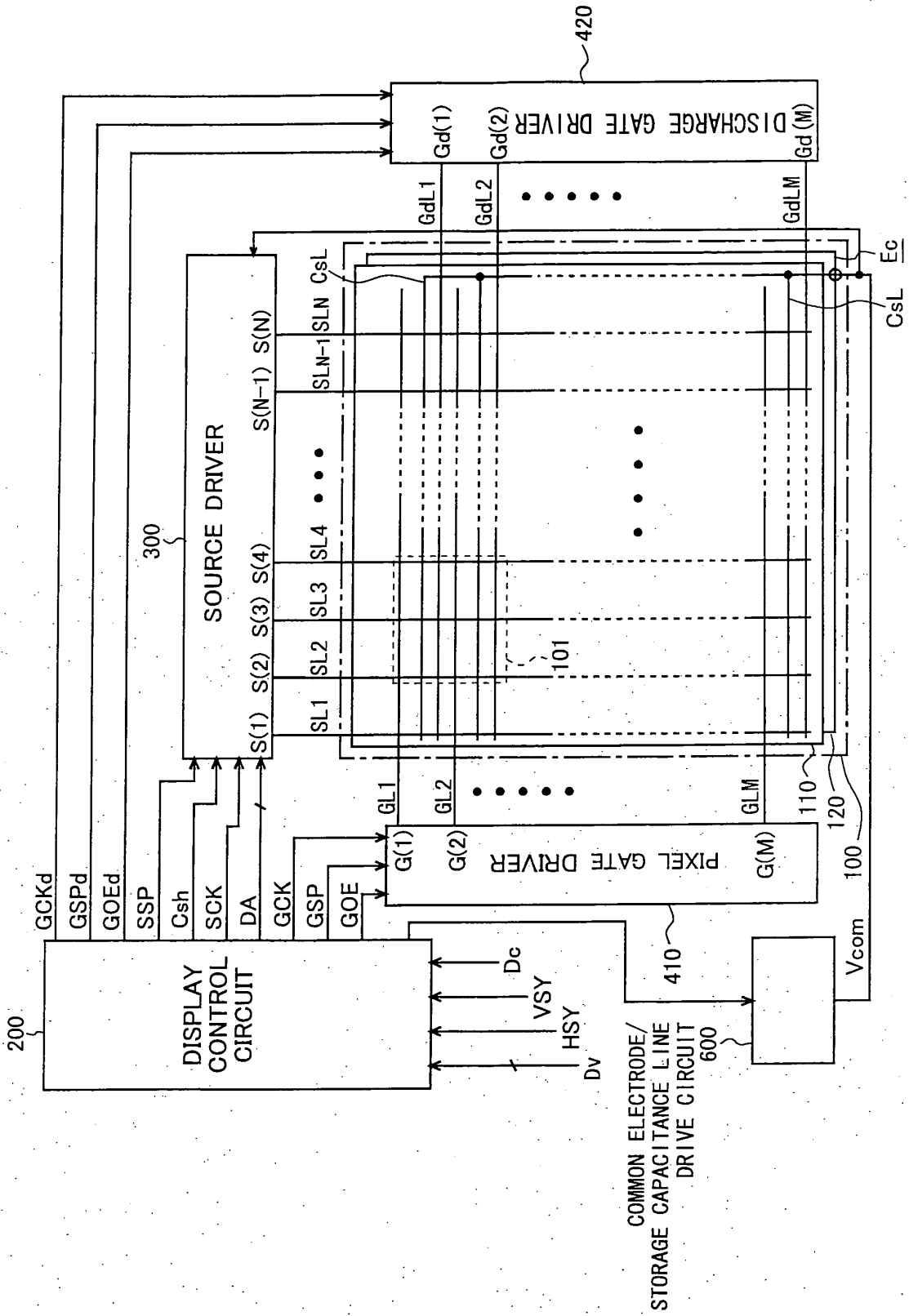


Fig. 7

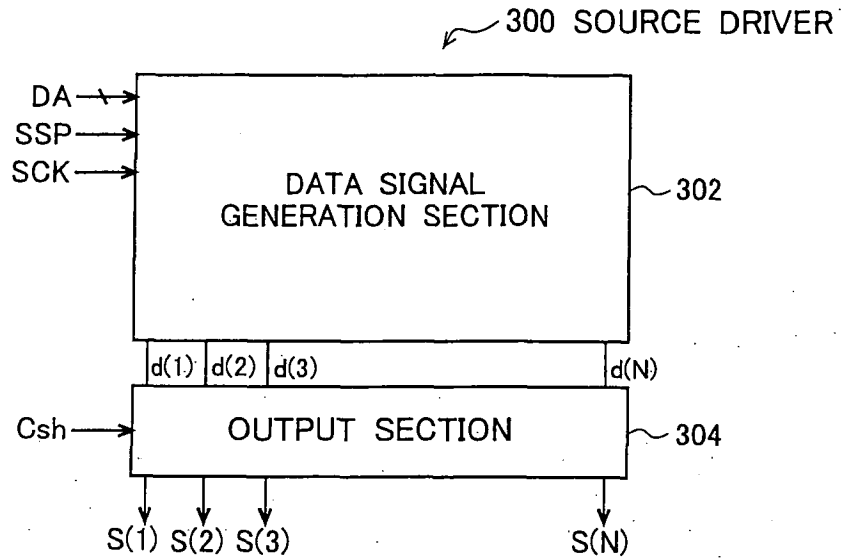


Fig. 8

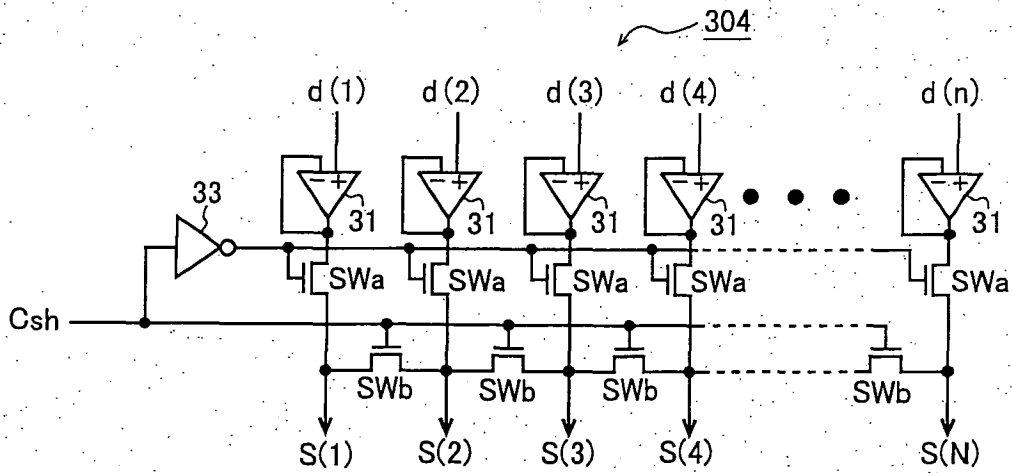


Fig.9

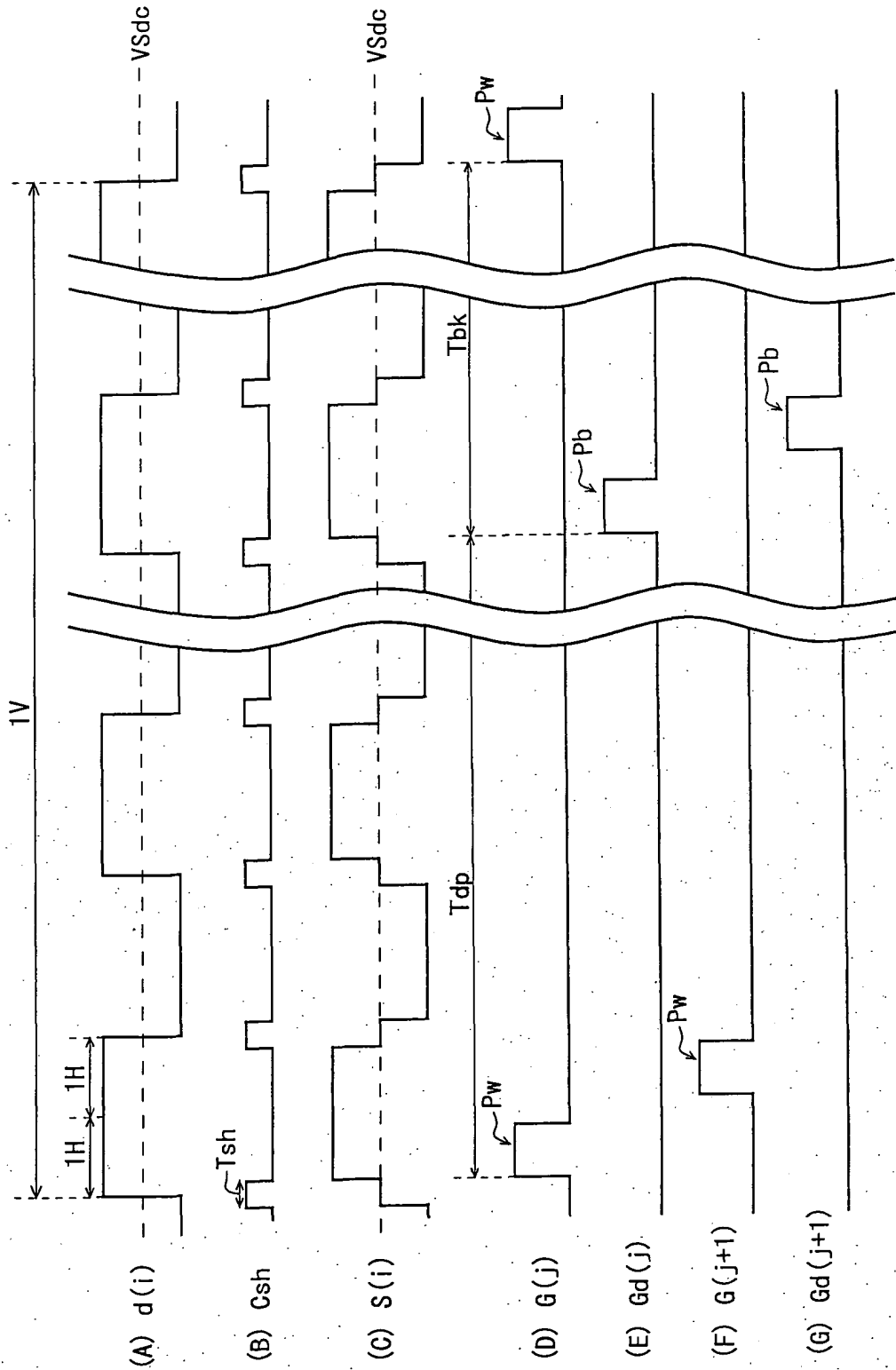


Fig. 10

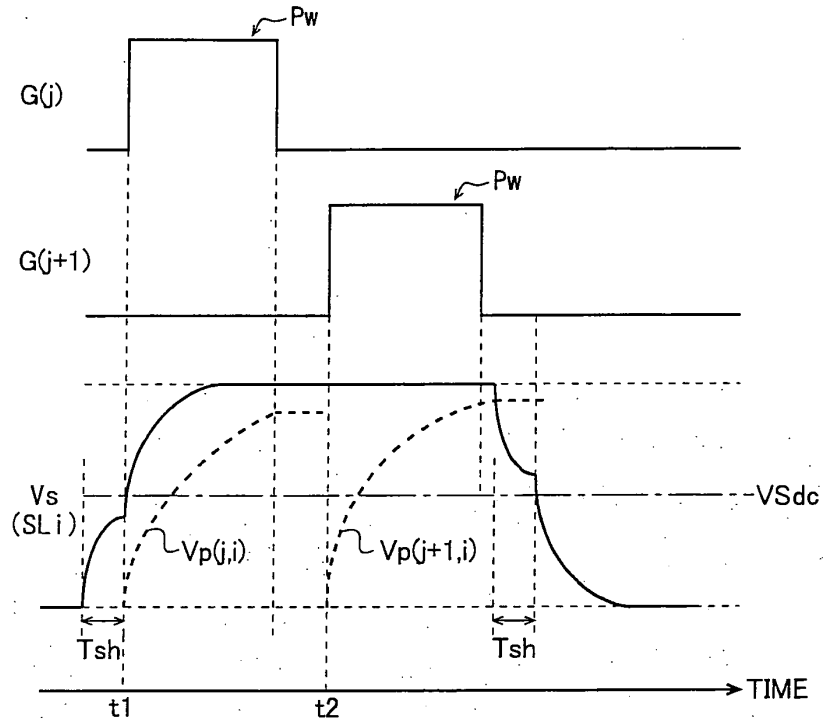


Fig. 11

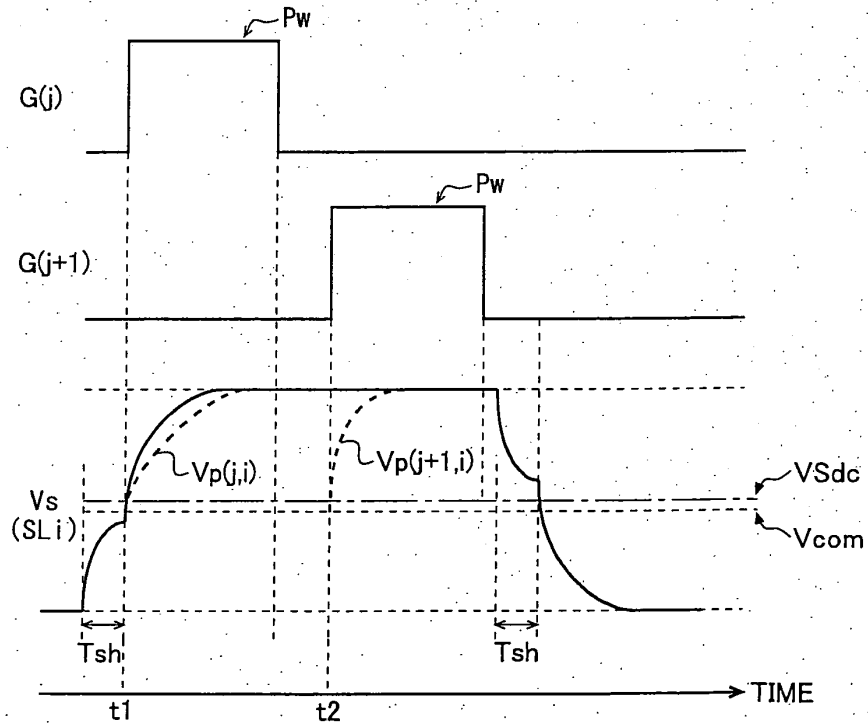


Fig. 12

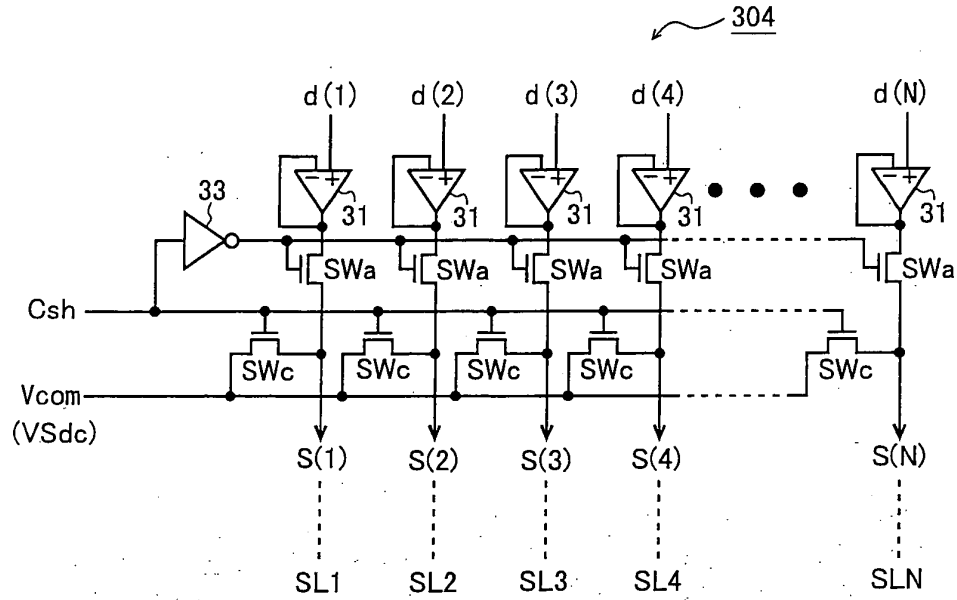


Fig. 13

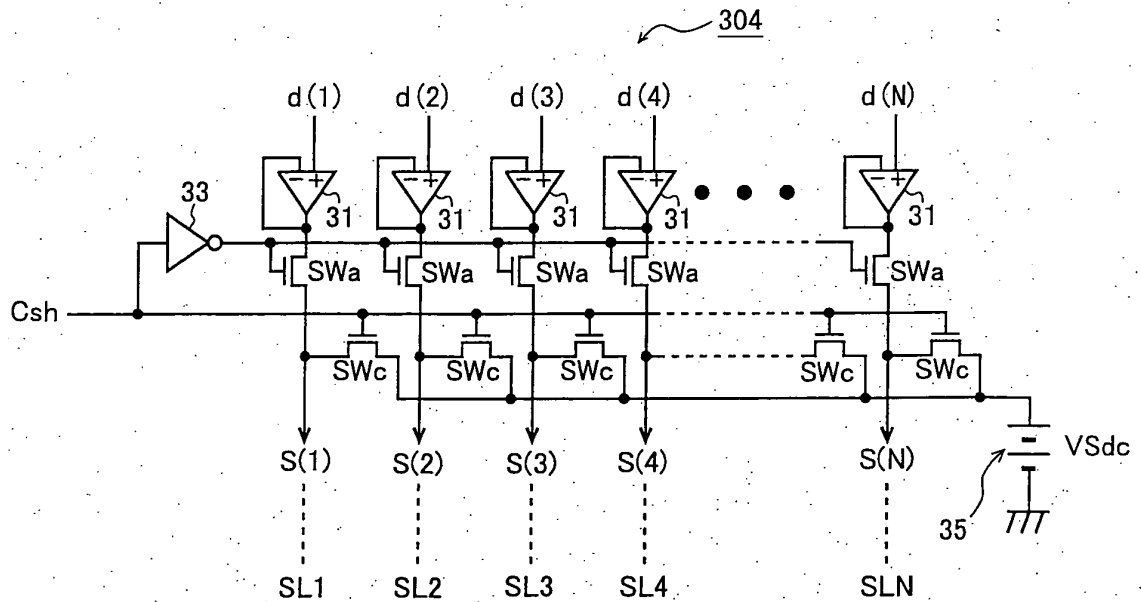


Fig. 14

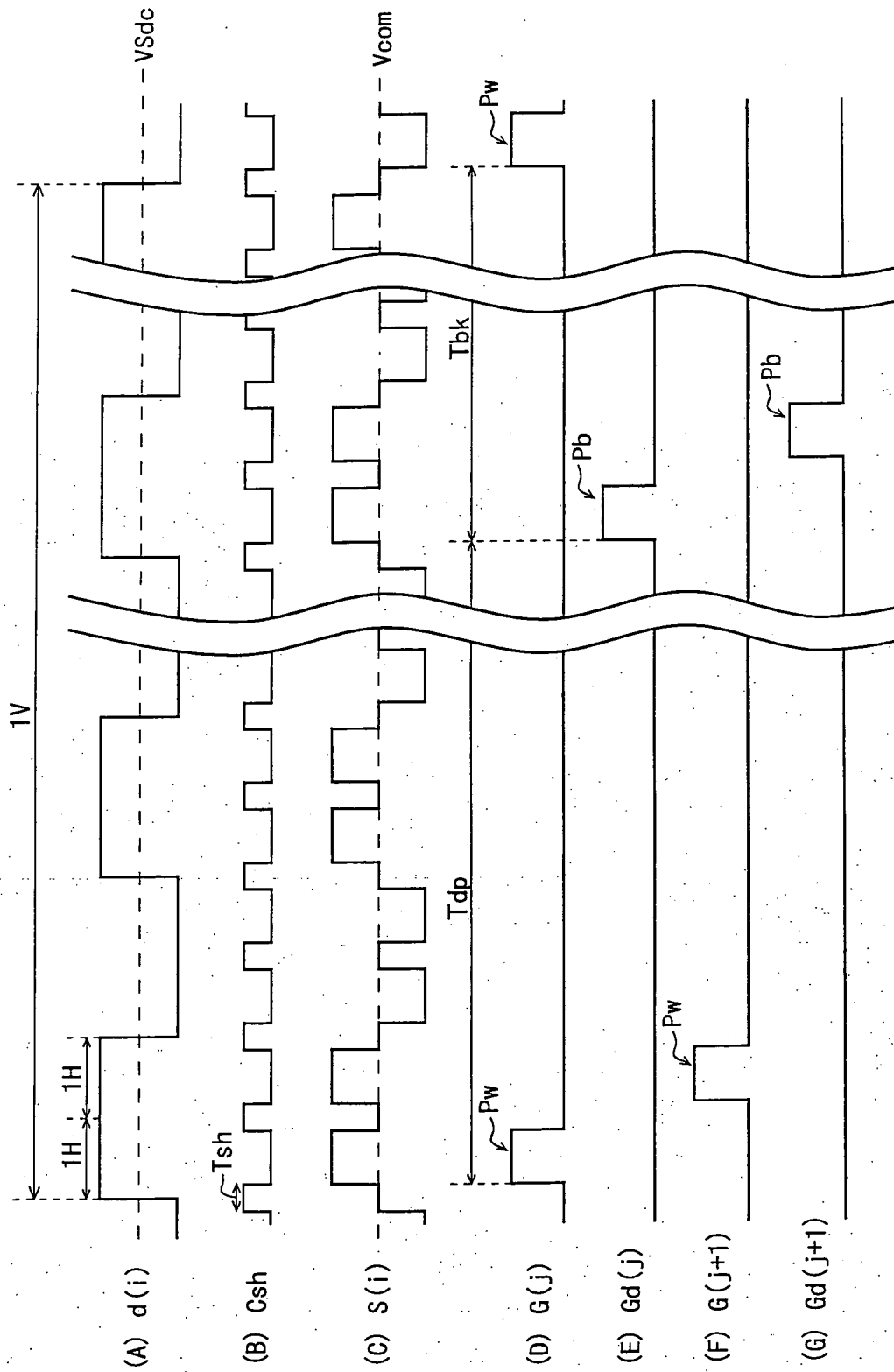


Fig. 17

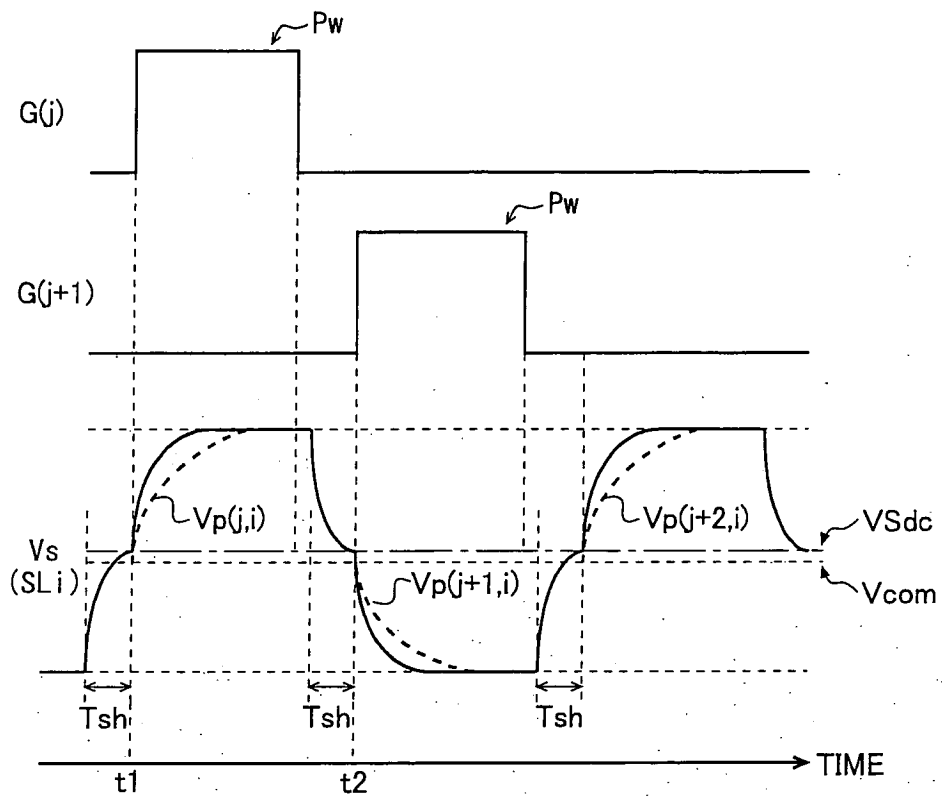


Fig. 18

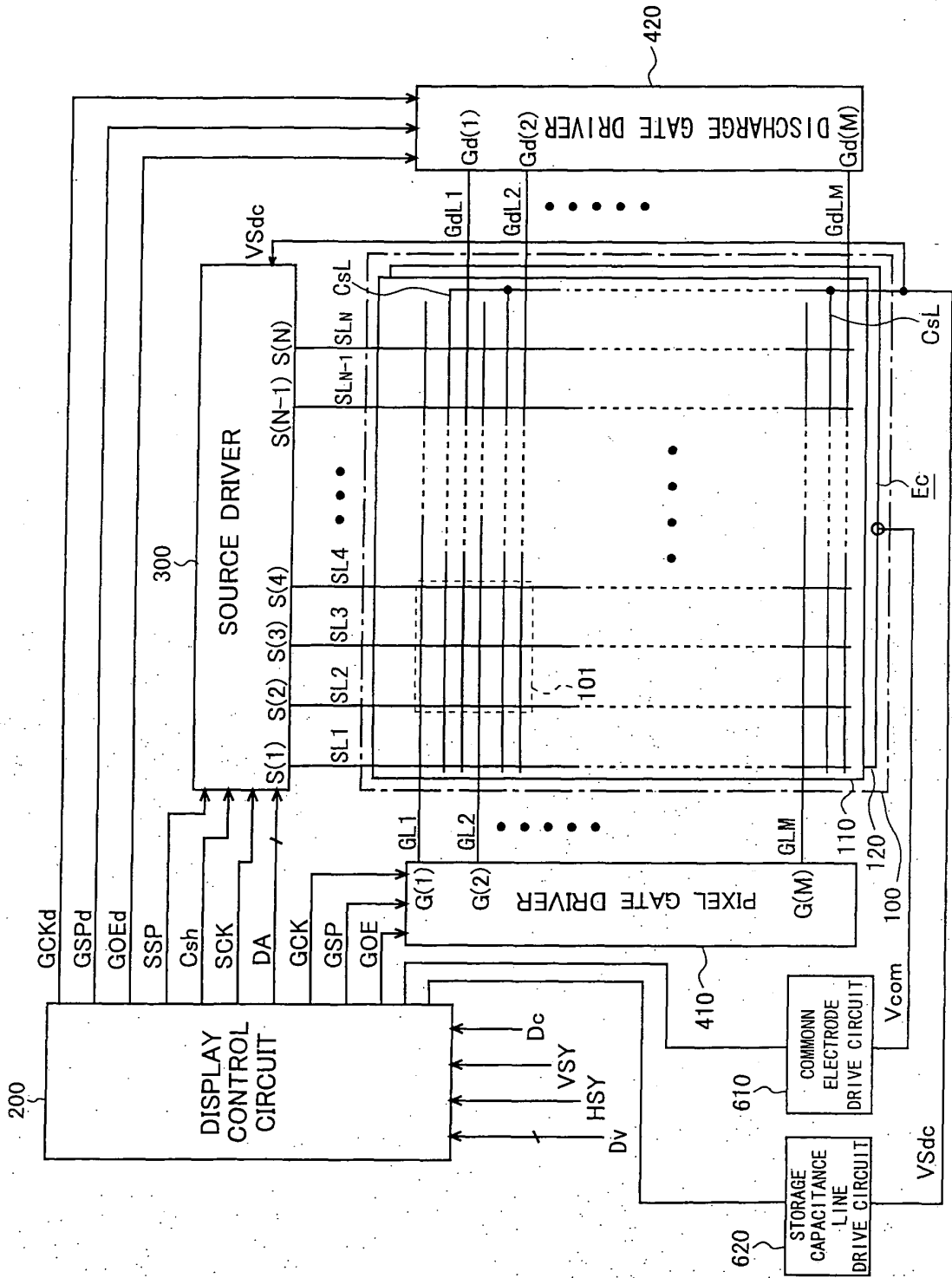


Fig. 19

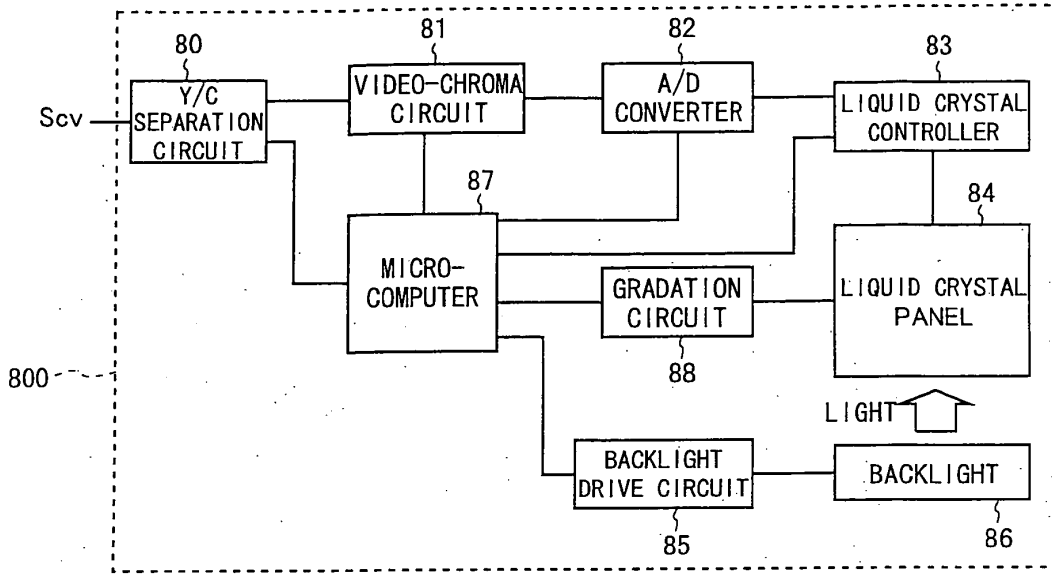


Fig. 20

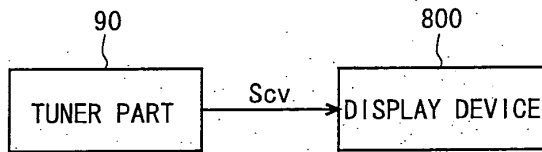


Fig. 21

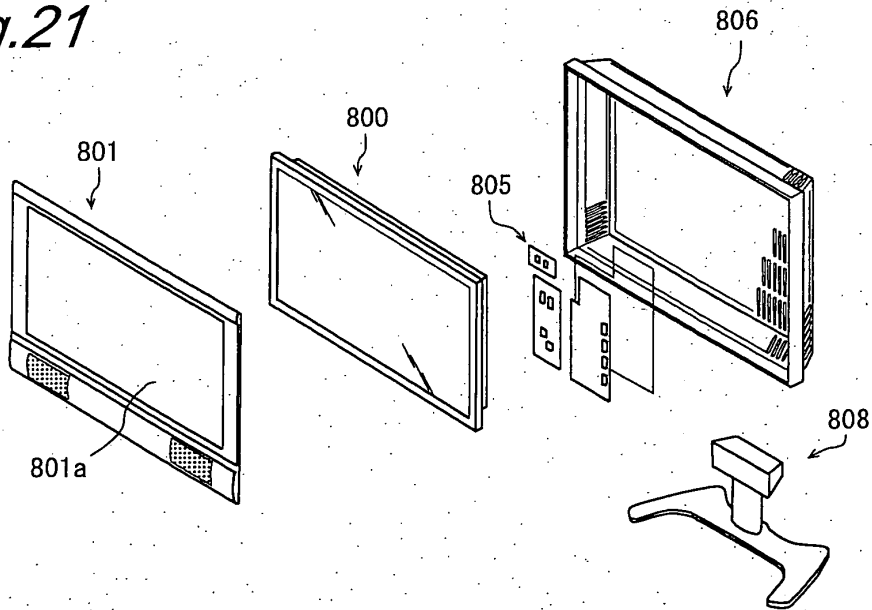


Fig.22

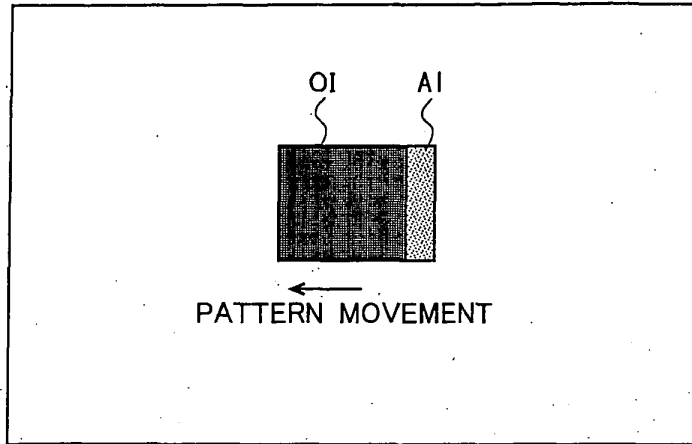
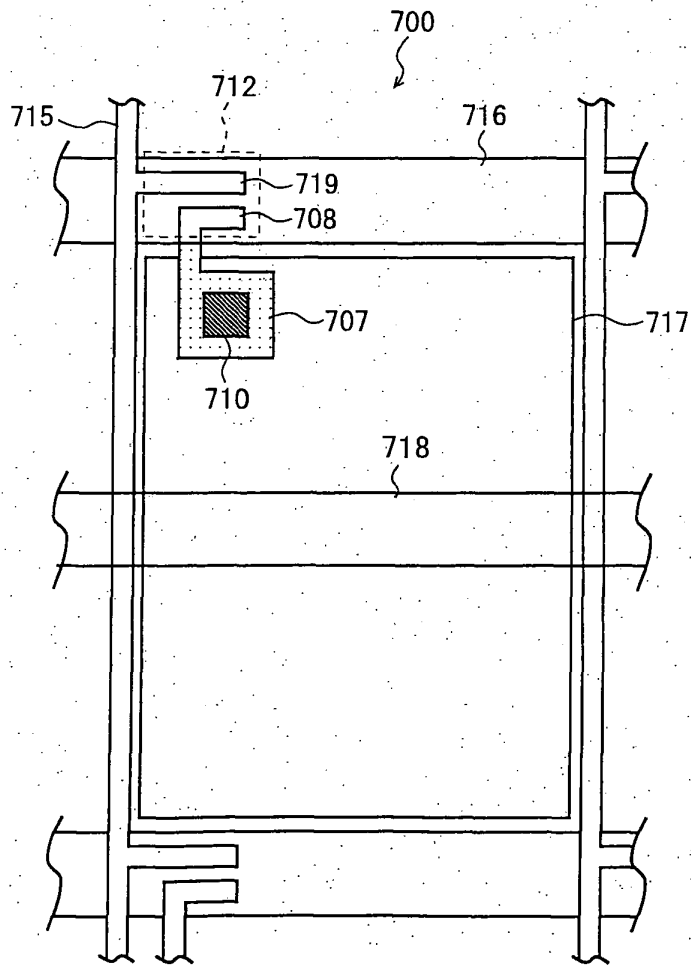


Fig.23



INTERNATIONAL SEARCH REPORT

International application No.

PCT/JP2007/062113

A. CLASSIFICATION OF SUBJECT MATTER G09G3/36(2006.01) i, G02F1/133(2006.01) i, G09G3/20(2006.01) i		
According to International Patent Classification (IPC) or to both national classification and IPC		
B. FIELDS SEARCHED		
Minimum documentation searched (classification system followed by classification symbols) G09G3/36, G02F1/133, G09G3/20		
Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched Jitsuyo Shinan Koho 1922-1996 Jitsuyo Shinan Toroku Koho 1996-2007 Kokai Jitsuyo Shinan Koho 1971-2007 Toroku Jitsuyo Shinan Koho 1994-2007		
Electronic data base consulted during the international search (name of data base and, where practicable, search terms used)		
C. DOCUMENTS CONSIDERED TO BE RELEVANT		
Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X Y	JP 11-326957 A (Toshiba Corp.), 26 November, 1999 (26.11.99), Par. Nos. [0022], [0023], [0228] to [0245]; Figs. 20, 27, 28 & US 006046790 A1	1-8, 18 9-17, 19-21
X	JP 2001-331156 A (Hitachi, Ltd.), 30 November, 2001 (30.11.01), Par. Nos. [0035] to [0039]; Figs. 12, 13 & US 2002/0047818 A1 & TW 000508557 B	1, 18
Y	JP 11-352462 A (NEC Corp.), 24 December, 1999 (24.12.99), Fig. 5 (Family: none)	9-17, 19-21
<input checked="" type="checkbox"/> Further documents are listed in the continuation of Box C. <input type="checkbox"/> See patent family annex.		
* Special categories of cited documents:		
"A"	document defining the general state of the art which is not considered to be of particular relevance	"T" later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention
"E"	earlier application or patent but published on or after the international filing date	"X" document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone
"L"	document which may throw doubts on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)	"Y" document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art
"O"	document referring to an oral disclosure, use, exhibition or other means	"&" document member of the same patent family
"P"	document published prior to the international filing date but later than the priority date claimed	
Date of the actual completion of the international search 09 August, 2007 (09.08.07)	Date of mailing of the international search report 21 August, 2007 (21.08.07)	
Name and mailing address of the ISA/ Japanese Patent Office	Authorized officer	
Facsimile No.	Telephone No.	

Form PCT/ISA/210 (second sheet) (April 2005)

INTERNATIONAL SEARCH REPORT

International application No.
PCT/JP2007/062113

C (Continuation). DOCUMENTS CONSIDERED TO BE RELEVANT		
Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
Y	JP 2005-201974 A (Oki Electric Industry Co., Ltd.), 28 July, 2005 (28.07.05), Par. Nos. [0067] to [0070]; Figs. 6 to 8 & US 2005/0151714 A1	10-17, 20, 21

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REFERENCES CITED IN THE DESCRIPTION

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Patent documents cited in the description

- JP H4309995 B [0007]
- JP H5119346 B [0007] [0013]
- JP 2003255912 A [0007] [0009]
- JP 2003066918 A [0007] [0008]
- JP H9243998 B [0007] [0011]
- JP 2004061590 A [0007] [0011]
- US 7161576 B [0008]
- US 20040017344 A [0011]

专利名称(译)	有源矩阵基板和具有该基板的显示装置		
公开(公告)号	EP2053589A4	公开(公告)日	2011-01-12
申请号	EP2007745368	申请日	2007-06-15
[标]申请(专利权)人(译)	夏普株式会社		
申请(专利权)人(译)	夏普株式会社		
当前申请(专利权)人(译)	夏普株式会社		
[标]发明人	TSUBATA TOSHIHIDE		
发明人	TSUBATA, TOSHIHIDE		
IPC分类号	G09G3/36 G02F1/133 G09G3/20		
CPC分类号	G09G3/3648 G02F1/13624 G09G3/3614 G09G2310/0248 G09G2330/023		
优先权	2006298440 2006-11-02 JP		
其他公开文献	EP2053589A1		
外部链接	Espacenet		

摘要(译)

旨在使显示装置能够产生显示脉冲并改善像素电容的充电特性，同时抑制驱动电路等的复杂性和工作频率的增加。在液晶显示装置的有源矩阵基板中，每个像素电极 (Ep) 经由像素TFT (10) 连接到源极线 (SLi) 并且经由放电TFT (12) 连接到存储电容线 (CsL)。。对于一个帧周期，当像素TFT (10) 被像素接通时，由对应于每个显示线的像素电极 (Ep) 和公共电极 (Ec) 形成的液晶电容 (C_{lc}) 被充电。在像素栅极线 (GLj) 上扫描信号 (G(j))，然后当放电栅极线 (GdLj) 上的放电扫描信号 (Gd(j)) 导通放电TFT (12) 时，放电。源极线 (SLi) 经历2H点反转驱动但是每个水平周期的电荷份额。